

CAN207 Continuous and Discrete Time Signals and Systems

Lecture 18 Sampling and Reconstruction

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Room SC340

Content

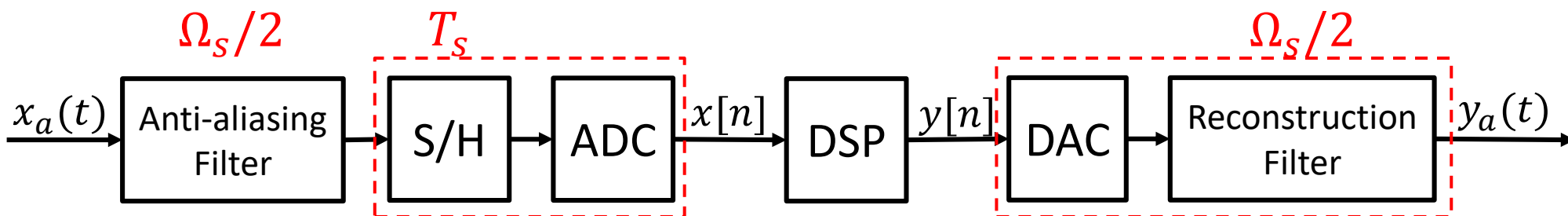
- 1. Sampling
 - 1.0. What and Why?
 - 1.1. Sampling in Time domain (TD)
 - 1.2. Sampling in Frequency domain (FD)
 - 1.3. Nyquist Theorem
- 2. Reconstruction
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 - 2.2 Reconstruction theory
 - 2.3 Reconstruction in FD – filtering
 - 2.4 Reconstruction in TD – interpolation
 - 2.5 Realisation

1.0 Digital Processing of CT Signals

- Most signals in nature are continuous in time
=> Need a way for “digital processing of continuous-time signals” => **SAMPLING!**

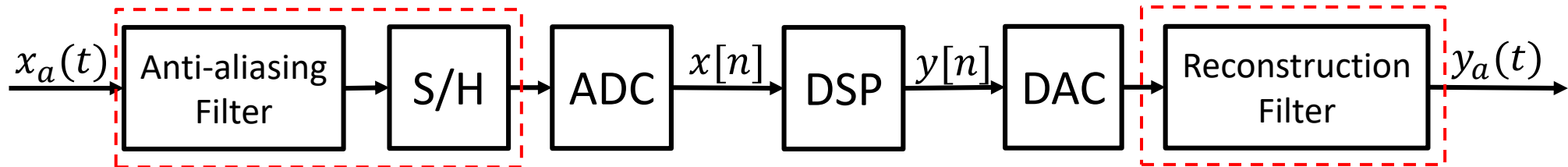


(a) Ideal data flow for the digital processing of continuous-time signals



(b) Practical data flow for the digital processing of continuous-time signals

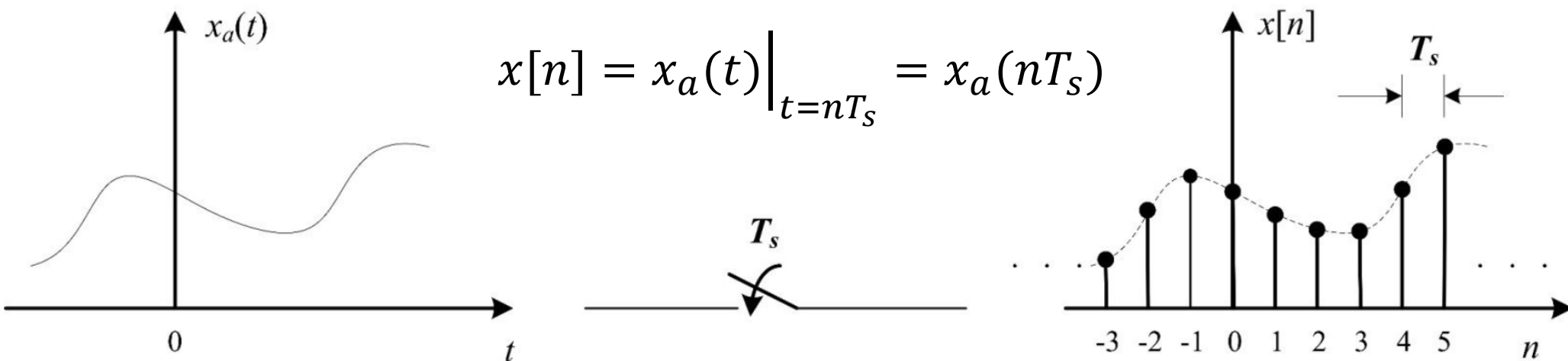
1.0 Analog -> Digital -> Analog



- Conversion of the continuous-time signal into a discrete-time signal
 - Anti-aliasing filter – to prevent potentially detrimental effects of sampling
 - Sample & Hold – discrete in time and keep the sampling values for a while to allow the A/D converter to do its job
 - *Analog to Digital Converter (A/D) – conversion in amplitude*
- Processing of the discrete-time signal
 - Digital Signal Processing – Filter, digital processor
- Conversion of the processed discrete-time signal back into a cont.-time signal
 - *Digital to analog converter (D/A) – to obtain the continuous signal*
 - Reconstruction / smoothing filter -smooth out the signal from the D/A

1.1 Sampling in Time domain (TD)

- A discrete-time sequence is developed by uniformly sampling the continuous-time signal $x_a(t)$



- The time variable - time t is related to the discrete time variable n only at discrete-time instants t_n

$$t_n = nT_s = \frac{n}{F_s} = \frac{2\pi n}{\Omega_s} \begin{cases} T_s = 1/F_s \text{ (Sampling period, second, second/sample)} \\ F_s = 1/T_s \text{ (Sampling frequency, Hz, cycles/second)} \\ \Omega_s = 2\pi/T_s \text{ (Sampling angular frequency, radian/second)} \end{cases} \quad 5$$

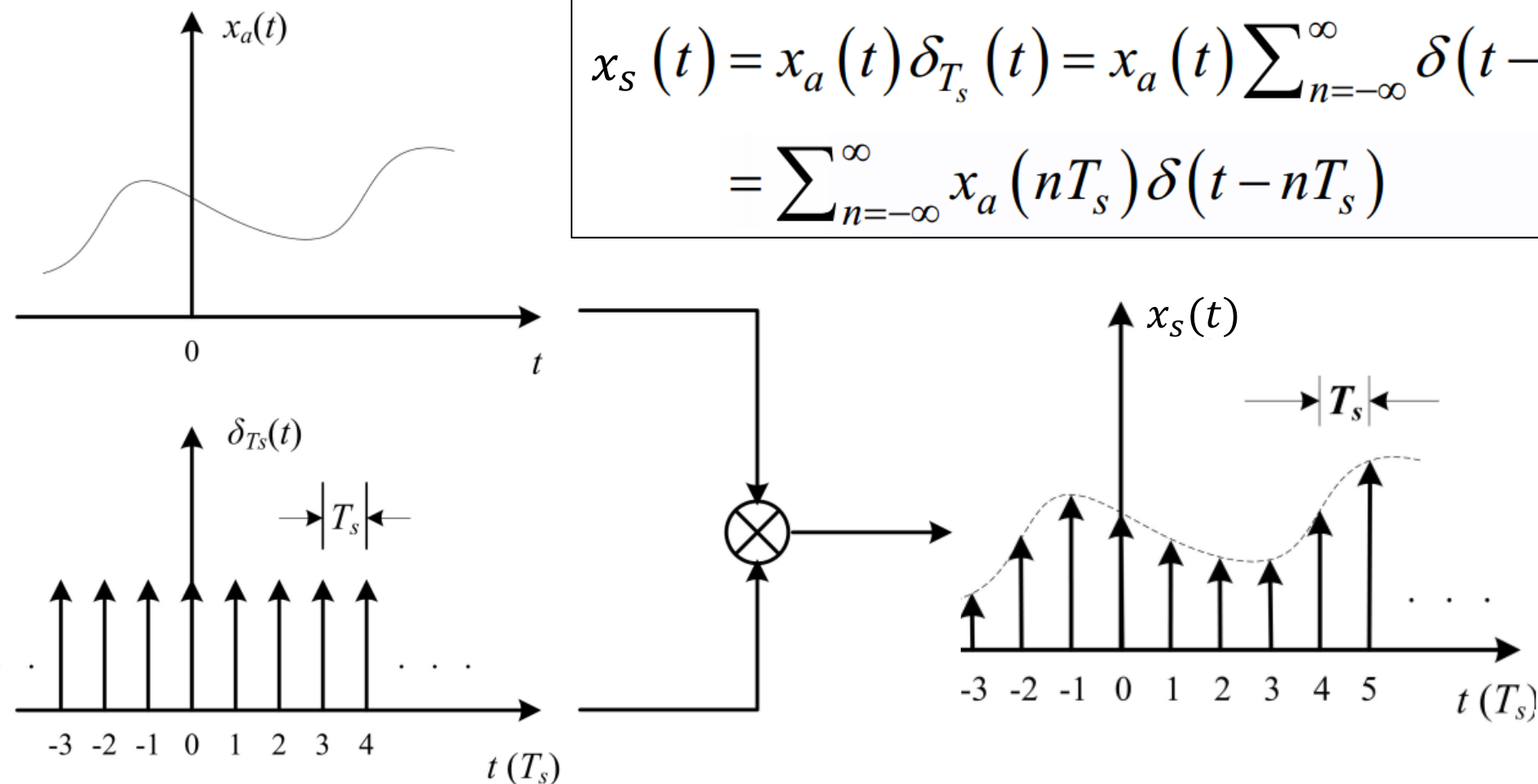
1.1 Sampling in Time domain (TD)

- Consider $x_a(t) = A\cos(\Omega_0 t + \phi)$
- Now $x[n] = A\cos(\Omega_0 nT_s + \phi)$
$$= A\cos\left(\frac{2\pi\Omega_0}{\Omega_s} n + \phi\right) = A\cos(\omega_0 n + \phi) = x_a(nT)$$
- Where $\omega_0 = \frac{2\pi\Omega_0}{\Omega_s} = \Omega_0 T_s$
- ω_0 is the (normalized) digital angular frequency of the signal
 - Unit: radians/sample
- Ω_0 is the analog angular frequency of signal
 - Unit: radians/second
- Ω_s is the sampling analog angular frequency
 - Unit: radians/second

1.1 Sampling in Time domain (TD)

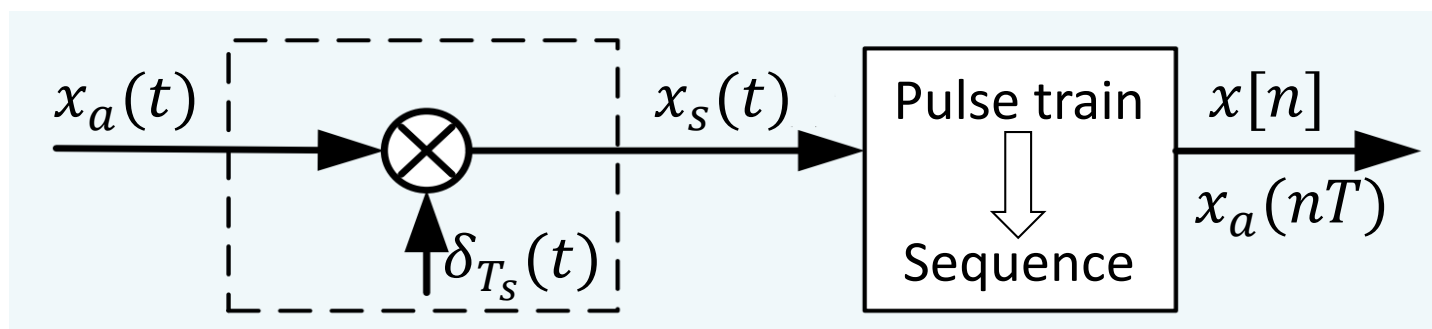
- In mathematics, the periodic sampling is modelled as the multiplication of continuous signal and impulse train

$$\begin{aligned}x_s(t) &= x_a(t) \delta_{T_s}(t) = x_a(t) \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \\ &= \sum_{n=-\infty}^{\infty} x_a(nT_s) \delta(t - nT_s)\end{aligned}$$

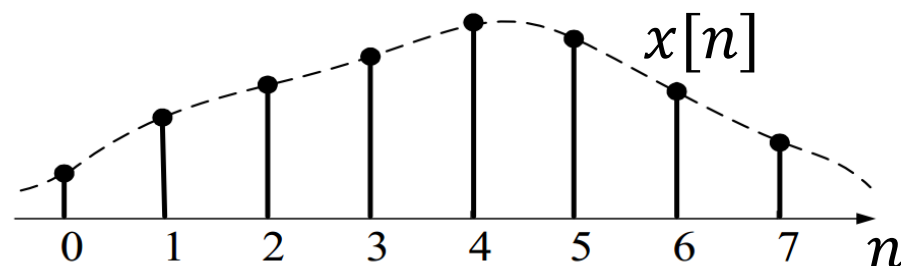
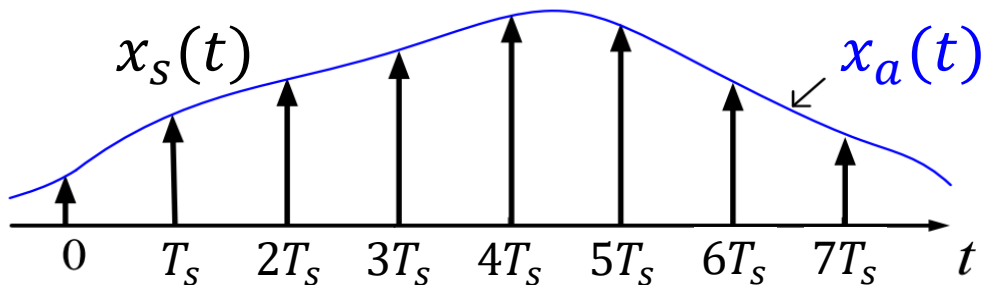


1.1 Sampling in Time domain (TD)

- The system to convert the continuous-time (CT) signal $x_a(t)$ to a discrete-time (DT) signal $x[n]$ is shown:

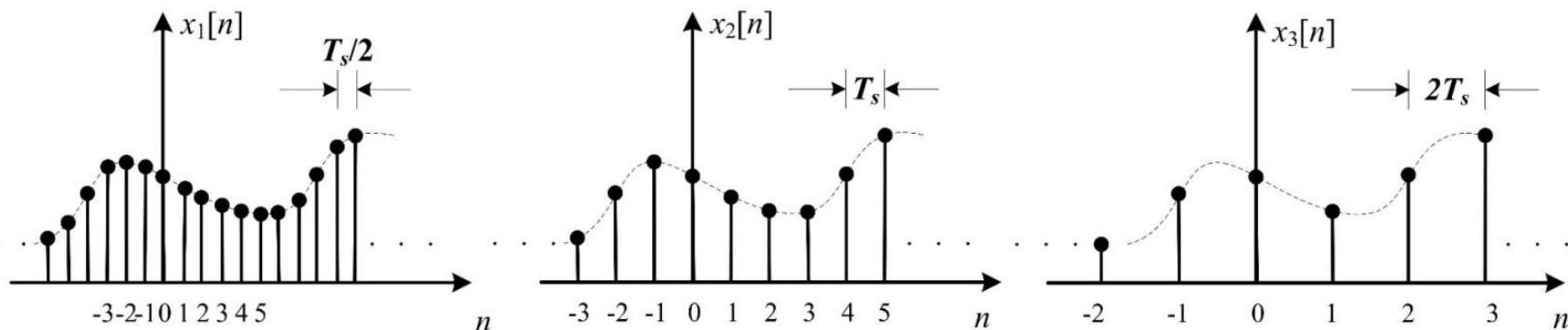


$$x_s(t) = x_a(t)\delta_{T_s}(t) = \sum_{n=-\infty}^{\infty} x_a(nT_s)\delta(t - nT_s)$$



1.1 How signal changed after sampling

- In time domain, continuous \rightarrow discrete



- Different sampling rates, different details
 - More samples = higher sampling rate/frequency = more detail
= more information kept = more resource occupation
 - Less samples = lower sampling rate/frequency = less details
= more information loss = less resource occupation
- How to choose the sampling period/rate?



Recall Lect. 8 p18

- $\tilde{x}(t)$ is periodic
- $x(t)$ represents one period

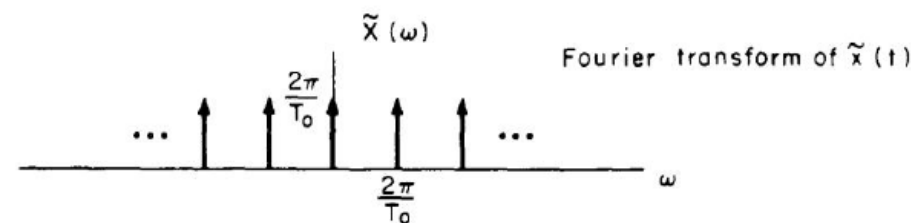
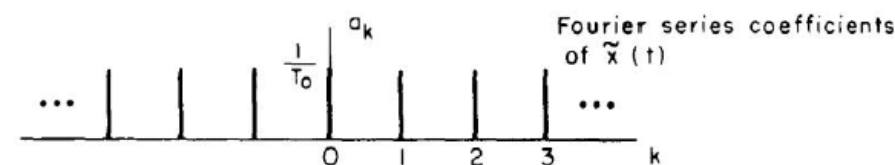
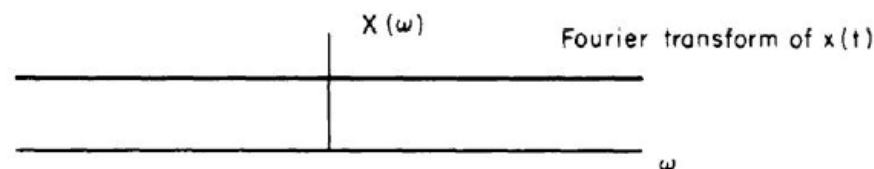
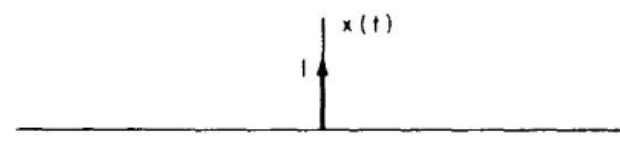
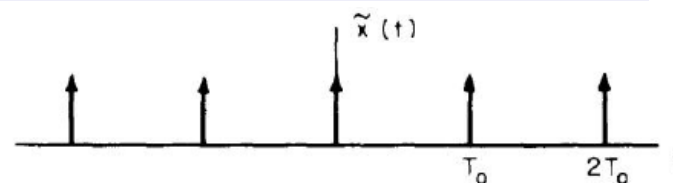
- Fourier series of $\tilde{x}(t)$

$$a_k = \frac{1}{T} X(\omega) |_{\omega=k\omega_0}$$

- Fourier transform of $\tilde{x}(t)$

$$\tilde{X}(\omega) \triangleq \sum_{k=-\infty}^{+\infty} 2\pi a_k \delta(\omega - k\omega_0)$$

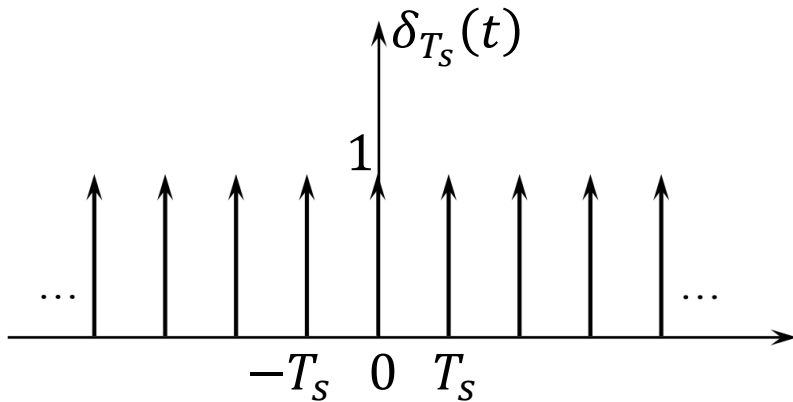
$$\tilde{X}(\omega) = \omega_0 \sum_{k=-\infty}^{+\infty} X(\omega) \delta(\omega - k\omega_0)$$



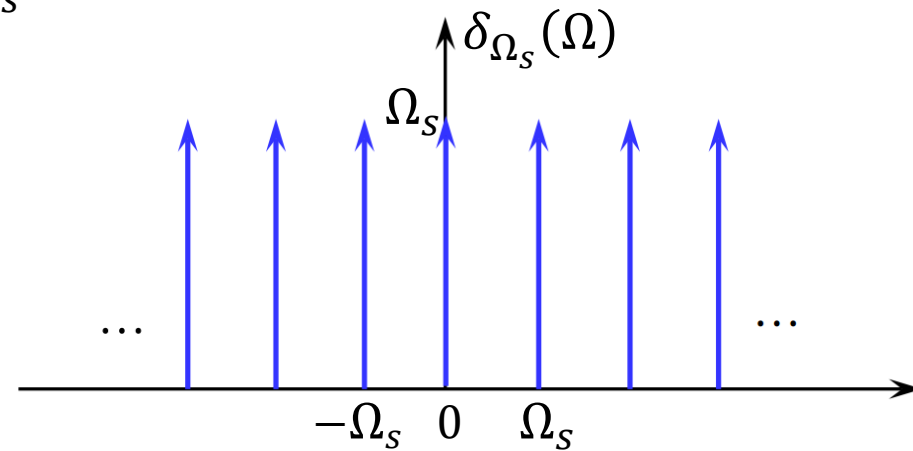
1.2 Frequency domain analyses

- Review: CTFT of a pulse train $\delta_{T_s}(t)$

$$\delta_{T_s}(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT_s) \xleftrightarrow[\Omega_s = 2\pi f_s = \frac{2\pi}{T_s}]{\text{CTFT}} \delta_{\Omega_s}(\Omega) = \Omega_s \sum_{k=-\infty}^{\infty} \delta(\Omega - k\Omega_s)$$



TD: Time domain



FD: Frequency domain

Note: in the following slides, Ω is used to denote the analog frequency for the Continuous-Time Fourier Transform (CTFT), in order to distinguish it from the digital frequency ω used for the Discrete-Time Fourier Transform (DTFT).

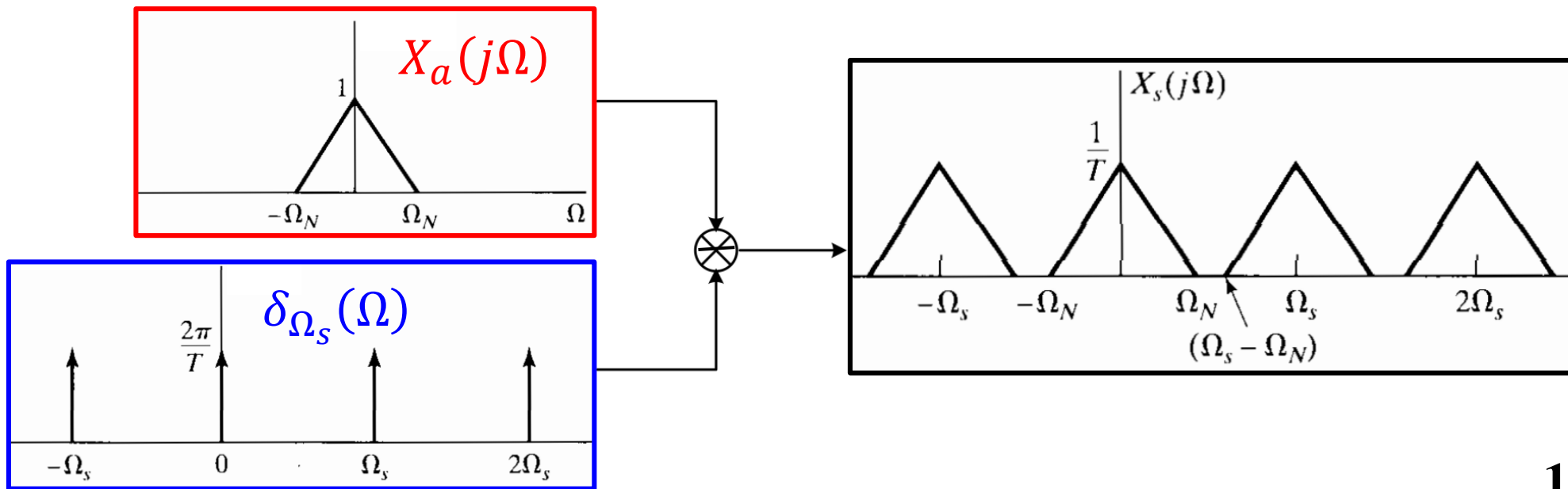
1.2 Sampling in Frequency domain (FD)

- In TD: multiplication between $x_a(t)$ and $\delta_{T_s}(t)$

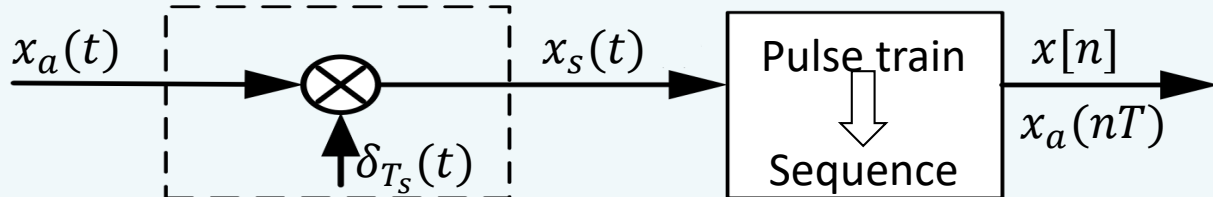
$$x_s(t) = x_a(t) \cdot \delta_{T_s}(t)$$

- In FD: convolution between $X_a(j\Omega)$ and $\delta_{\Omega_s}(\Omega)$

$$X_s(j\Omega) = \frac{1}{2\pi} X_a(j\Omega) * \Omega_s \sum_{k=-\infty}^{\infty} \delta(\Omega - k\Omega_s) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X_a[j(\Omega - k\Omega_s)]$$



1.2 Sampling in FD



- An alternative expression of $X_s(j\Omega)$ is:

$$x_s(t) = \sum_{n=-\infty}^{\infty} x_a(nT_s) \delta(t - nT_s)$$

CTFT \longleftrightarrow

$$X_s(j\Omega) = \sum_{n=-\infty}^{\infty} x_a(nT_s) e^{-j\Omega T_s n}$$

Since:

$$\left. \begin{aligned} x[n] &= x_a(nT_s) \\ X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \end{aligned} \right\} X_s(j\Omega) = X(e^{j\omega}) \Big|_{\omega=\Omega T_s} = X(e^{j\Omega T_s})$$

Recall

$$X_s(j\Omega) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X_a[j(\Omega - k\Omega_s)]$$

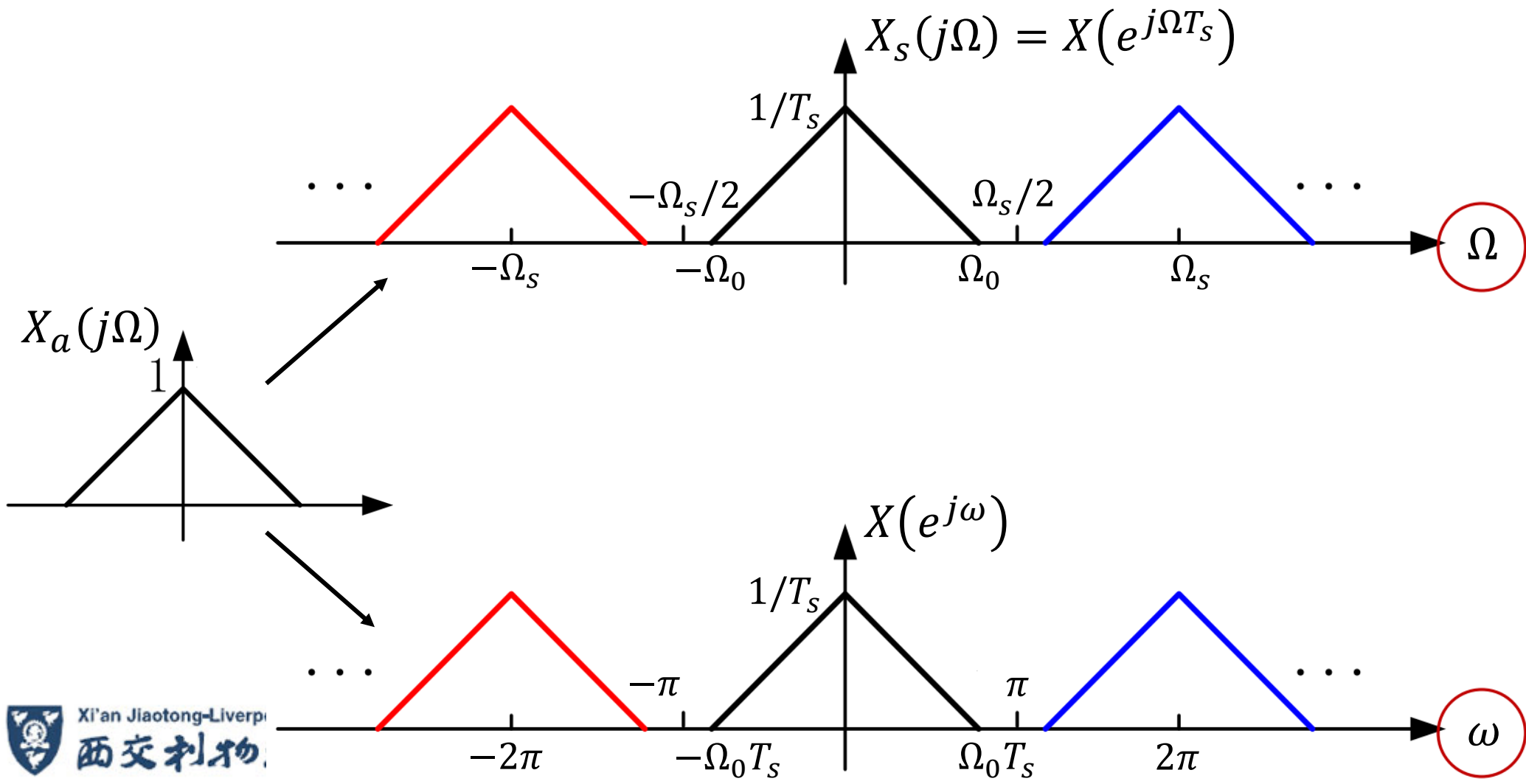
$$X(e^{j\Omega T_s}) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X_a[j(\Omega - k\Omega_s)]$$

$$X(e^{j\omega}) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X_a \left[j \left(\frac{\omega}{T_s} - \frac{2\pi k}{T_s} \right) \right]$$



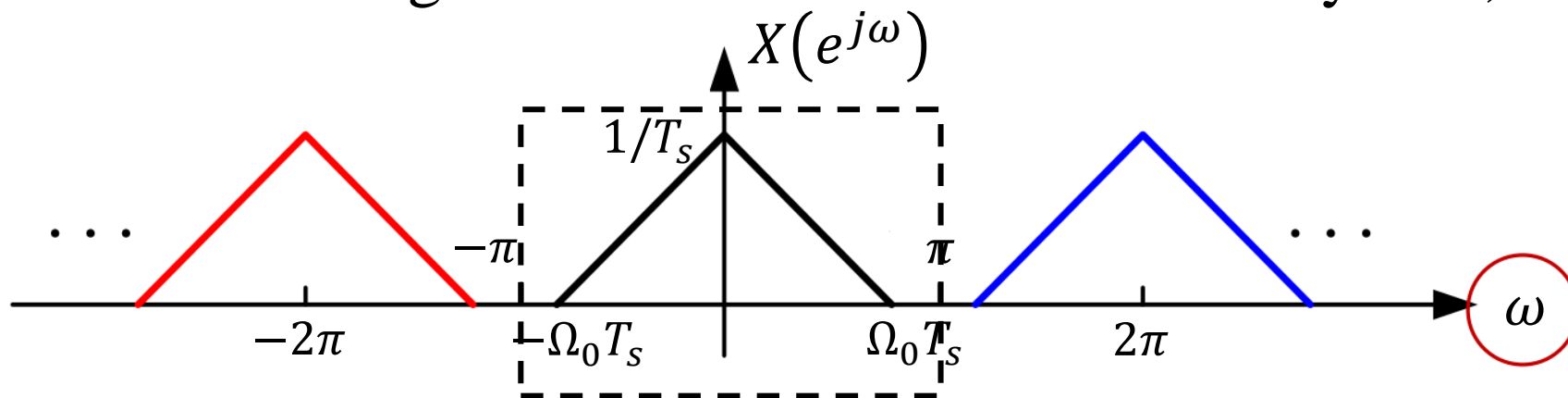
1.2 Sampling in Frequency domain (FD)

- Discretization in TD \Rightarrow Periodicity in FD



1.3 Nyquist-Shannon Theorem

- The spectrum of the sampled signal contains all the information of the original CT signal
 - So the CT signal can be recovered without any loss;



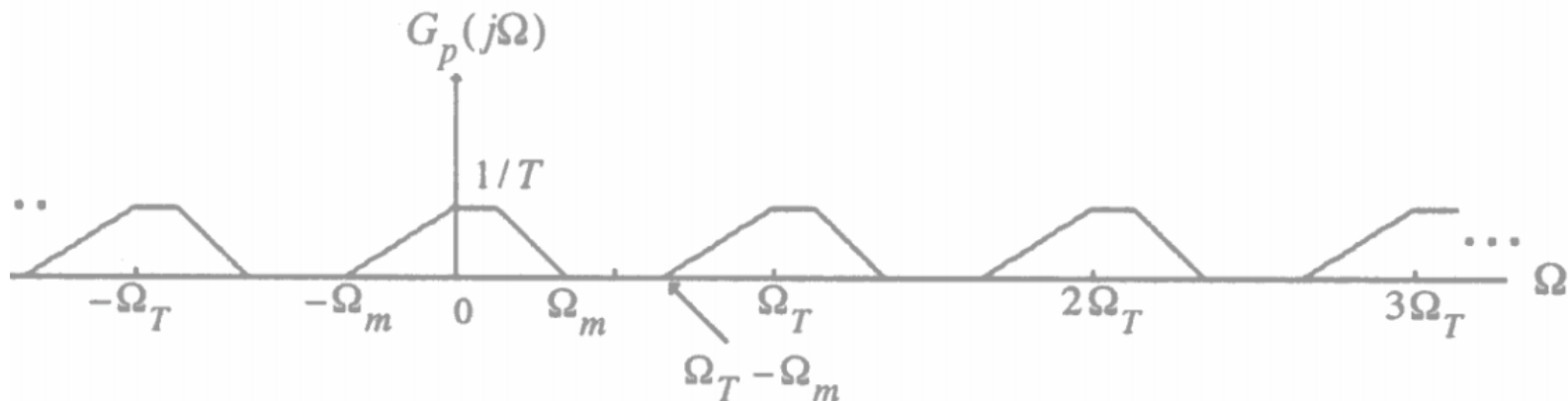
- But a condition needs to be satisfied:

$$\Omega_0 T_s \leq \pi \iff 2\Omega_0 \leq \Omega_s$$

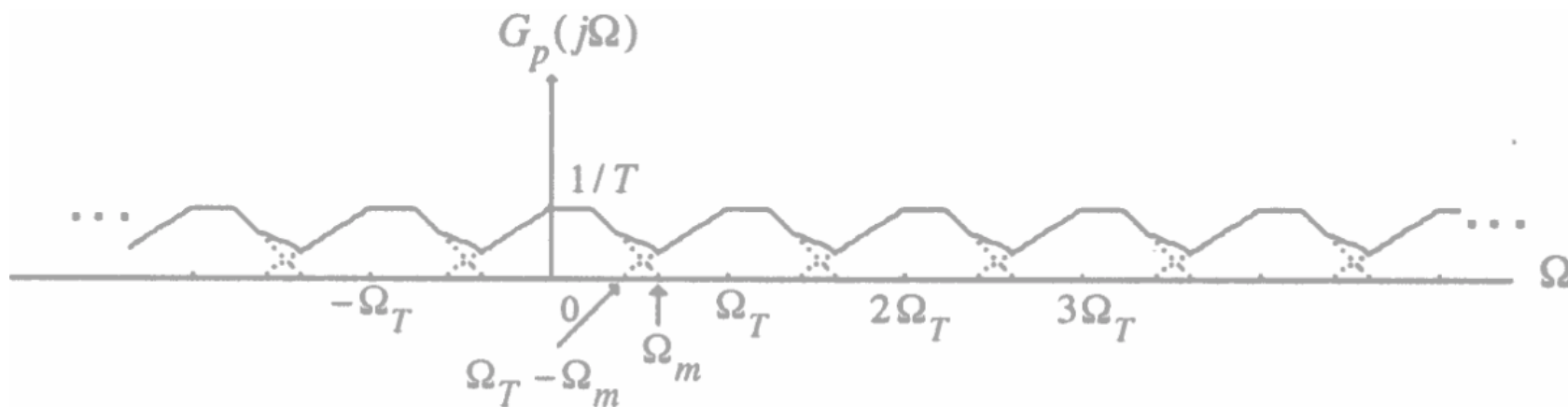
- The Nyquist-Shannon theorem!

1.3 Over-sampling and under-sampling

- Over-sampling: $2\Omega_m < \Omega_T$



- Down-sampling: $2\Omega_m > \Omega_T$



Wrap-up of SAMPLING

- What is sampling process?
 - The first step to convert a continuous-time signal to a discrete-time signal;
- In time domain: multiplication the CT signal to a pulse train, then convert the modulated pulse train to sequence;
- In frequency domain: copy and shift (create infinite replica) the spectrum of the CT signal;
 - The CT signal can be recovered from the sampled signal if Nyquist theorem is satisfied, i.e., $2\Omega_0 \leq \Omega_s$
- Nyquist theorem

Example

- A continuous-time signal $x_a(t)$ is the linear combination of the components with 300Hz, 1.2kHz and 3.5kHz frequencies. Sampling $x_a(t)$ with a sampling frequency 2 kHz gets a sequence $x[n]$. Sending $x[n]$ through an ideal lowpass filter with the cutoff frequency of 900Hz get a continuous-time output $y_a(t)$.
- What are the frequency components in $y_a(t)$?
- What is the critical sampling frequency?

Quiz 1

- The signal $x(t) = \sin(\pi t) + 4\sin(3\pi t)\cos(2\pi t)$, where t is in ms, is sampled at a rate of 3 kHz.
- Find the frequency components in $x(t)$.
- Find the Nyquist rate of the signal.

2.1 Interpolation

- What is interpolation?
 - In the mathematical field of numerical analysis, interpolation is a method of *constructing new data points* within the range of a discrete set of known data points.
 - In this module, interpolation is a procedure whereby we convert a discrete-time (DT) sequence $x[n]$ to a continuous-time (CT) function $x(t)$.
 - Requirement: for the CT function $x(t)$, its values at multiples of T_s should be equal to the corresponding points of the DT sequence $x[n]$:

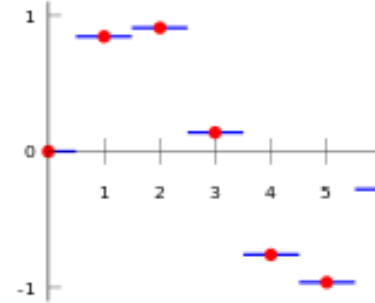
$$x(t) \Big|_{t=nT_s} = x[n]$$

The interpolation problem now reduces to “filling the gap” between these instants.

2.1 Interpolation methods

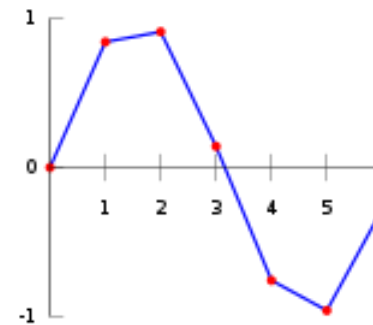
- 1. Zero-order/Local Interpolation

$$I_0(t) = \text{rect}(t)$$

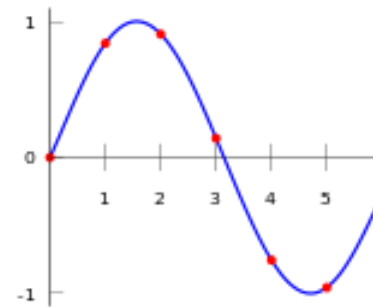


- 2. First-order/Linear Interpolation

$$y = y_1 + (x - x_1) \frac{y_2 - y_1}{x_2 - x_1}$$

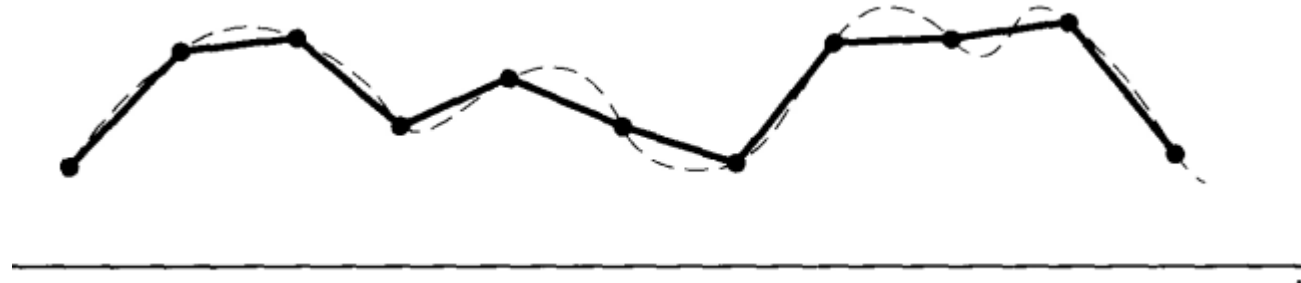


- 3. Higher-order/Polynomial Interpolation



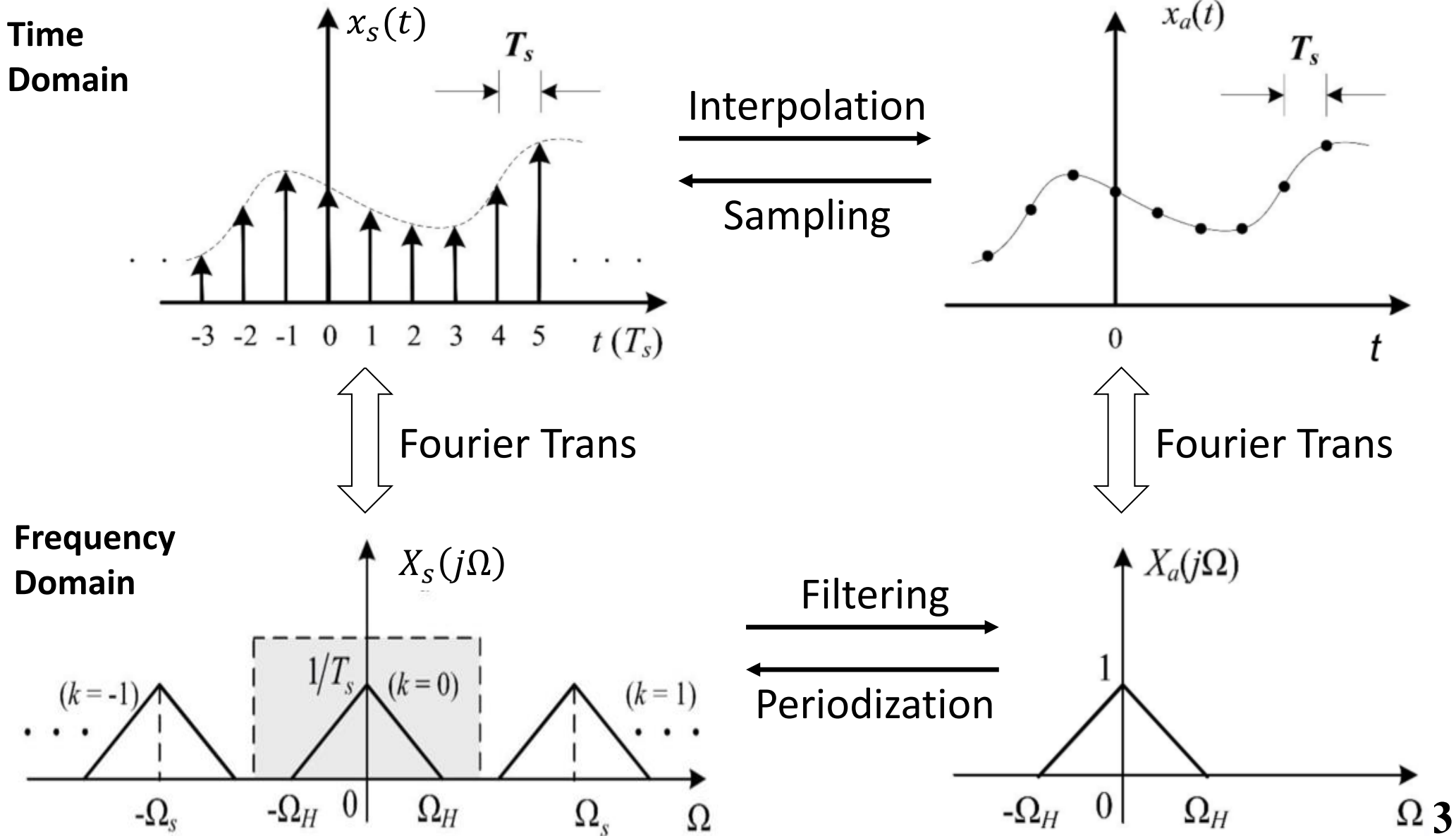
2.1 Reconstruction with interpolation

- Interpolation, the fitting of a continuous signal to a set of sample values, is a commonly used procedure for reconstructing a function, either approximately or exactly, from samples.
 - One simple interpolation procedure is the zero-order hold.
 - Another useful form of interpolation is *linear interpolation*, whereby adjacent sample points are connected by a straight line as shown below.



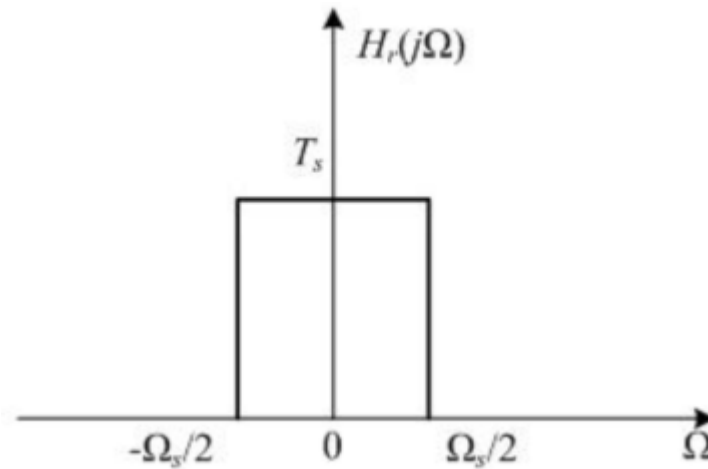
- In more complicated interpolation formulas, sample points may be connected by higher order polynomials or other mathematical functions, like $\text{sinc}()$.

2.2 Reconstruction Theory



2.3 Reconstruction filter in Frequency domain

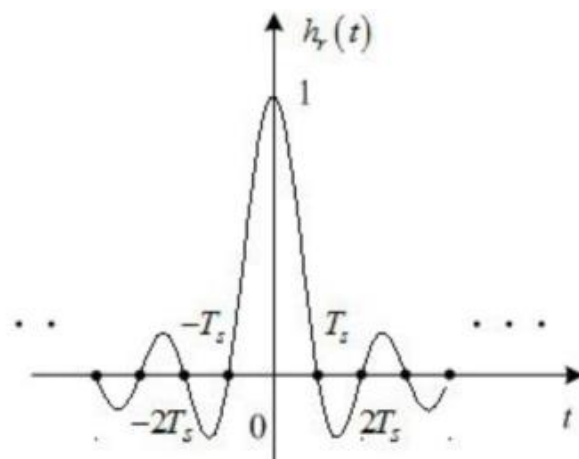
- Reconstruction or smoothing filter is used to eliminate all the replicas of the spectrum outside the baseband
- Ideal lowpass filter
 - Frequency domain $H_r(j\Omega) = \begin{cases} T_s, & |\Omega| < \Omega_c \\ 0, & |\Omega| > \Omega_c \end{cases}$



2.4 Reconstruction filter in time domain

- This lowpass filter in time domain is a “sinc” function:
 - Time domain

$$\begin{aligned} h_r(t) &= \frac{1}{2\pi} \int_{-\infty}^{\infty} H_r(j\Omega) e^{j\Omega t} d\Omega = \frac{T_s}{2\pi} \int_{-\Omega_c}^{\Omega_c} e^{j\Omega t} d\Omega \\ &= \frac{\sin(\Omega_s t/2)}{\Omega_s t/2} = \frac{\sin(\pi t/T_s)}{\pi t/T_s} = \text{sinc}\left(\frac{t}{T_s}\right) \end{aligned}$$



- Multiply with $H_r(j\Omega)$ (in FD) is equivalent to convolve with $h_r(t)$ (in TD), the recovered signal $x_r(t) = x_s(t) * h_r(t)$

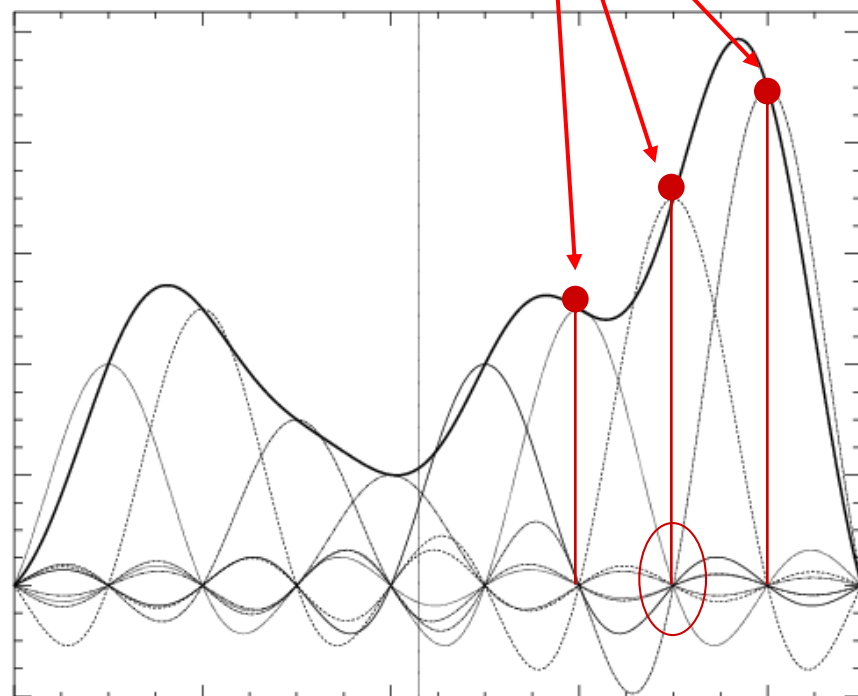
- Impulse train $x_s(t)$: $x_s(t) = \sum_{n=-\infty}^{\infty} x_a(nT_s) \delta(t - nT_s)$

2.4 Reconstruction filter in time domain

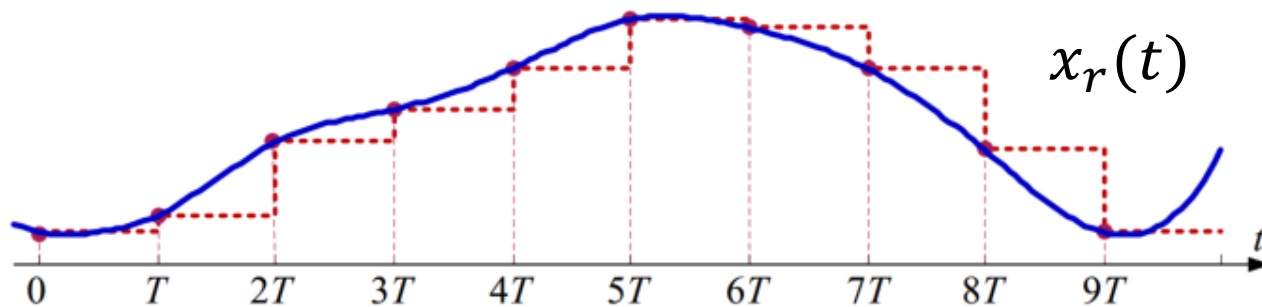
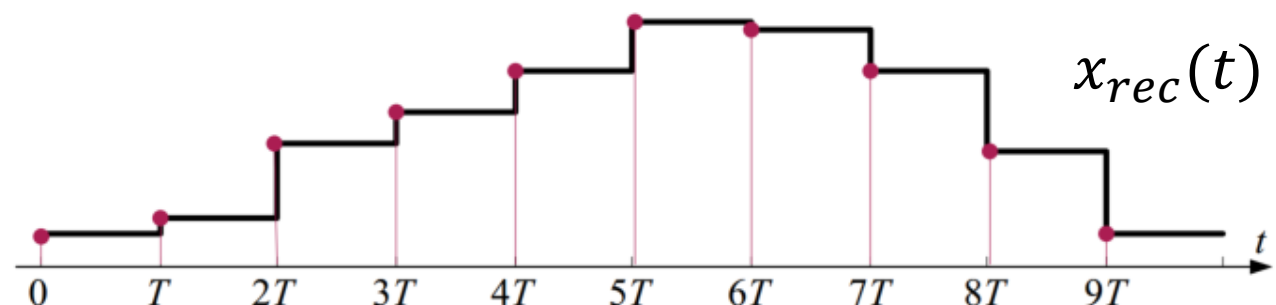
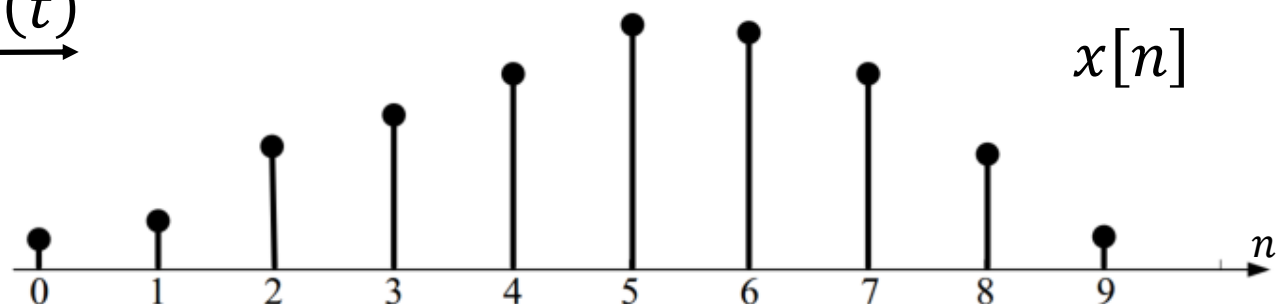
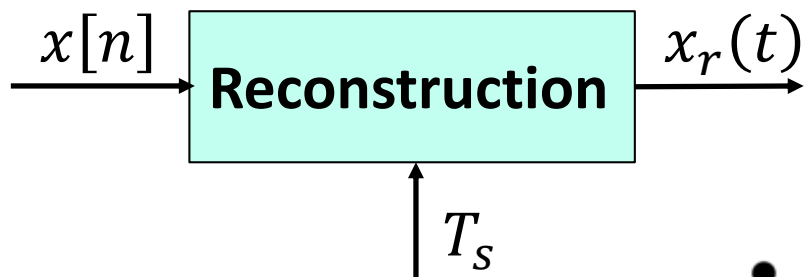
- Convolution between the discretized signal $x_s(t)$ and the reconstruction lowpass filter $h_r(t)$:

$$x_r(t) = x_s(t) * h_r(t) = \sum_{n=-\infty}^{\infty} x_a(nT_s)h_r(t - nT_s) = \sum_{n=-\infty}^{\infty} x[n]\text{sinc}(t - nT_s)$$

- The values are interpolated as a linear combination of the time-shifted sinc functions
- The amplitudes are scaled according to the sample values at the center locations of the sinc (the interpolation functions)

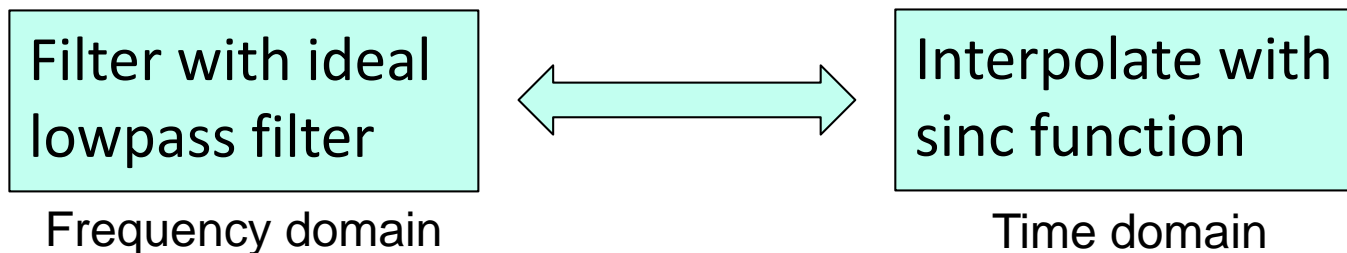


2.5 Realization



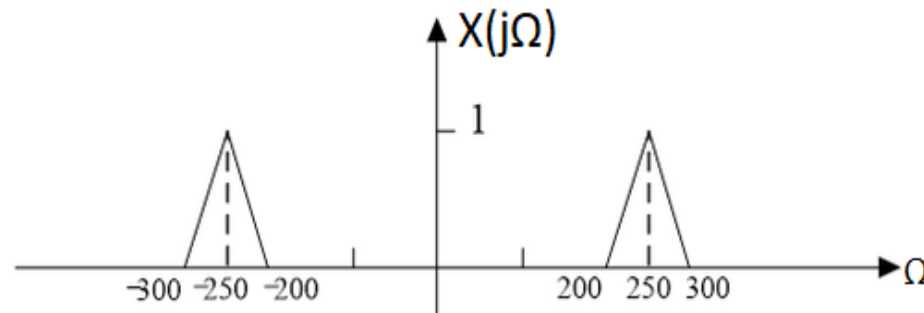
Wrap-up of RECONSTRUCTION

- Continuous-time signal can be reconstructed from the discrete-time sequence;
- Reconstruction can be realized as
 - In time domain: interpolation;
 - In frequency domain: filtering.
- Ideal reconstruction:



Quiz 2

- The spectrum $X(j\Omega)$ of a continuous-time signal $x(t)$ is shown below. $x(t)$ is sampled with the sampling angular frequency Ω_s , and get a discrete-time sequence $x[k]$, whose spectrum is $X(j\omega)$.



- a) If $\Omega_s = 200$ rad/s, sketch $X(j\omega)$ with all the labels for horizontal and vertical axes;
- b) If $\Omega_c = 200$ rad/s, can $x(t)$ be reconstructed from $x[k]$? Explain the reason.

Next ...

- Z-transform
- Discrete Fourier Transform