

CAN207 Continuous and Discrete Time Signals and Systems

Lecture 15

DTFT Properties and Applications

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Content

- 1. Definitions
 - Frequency mapping
 - Existence of DTFT (convergence)
- 2. DTFT properties
 - Periodicity, Linearity, ...
 - Convolution property
 - Modulation property
 - Duality

1.0 DTFT Definition

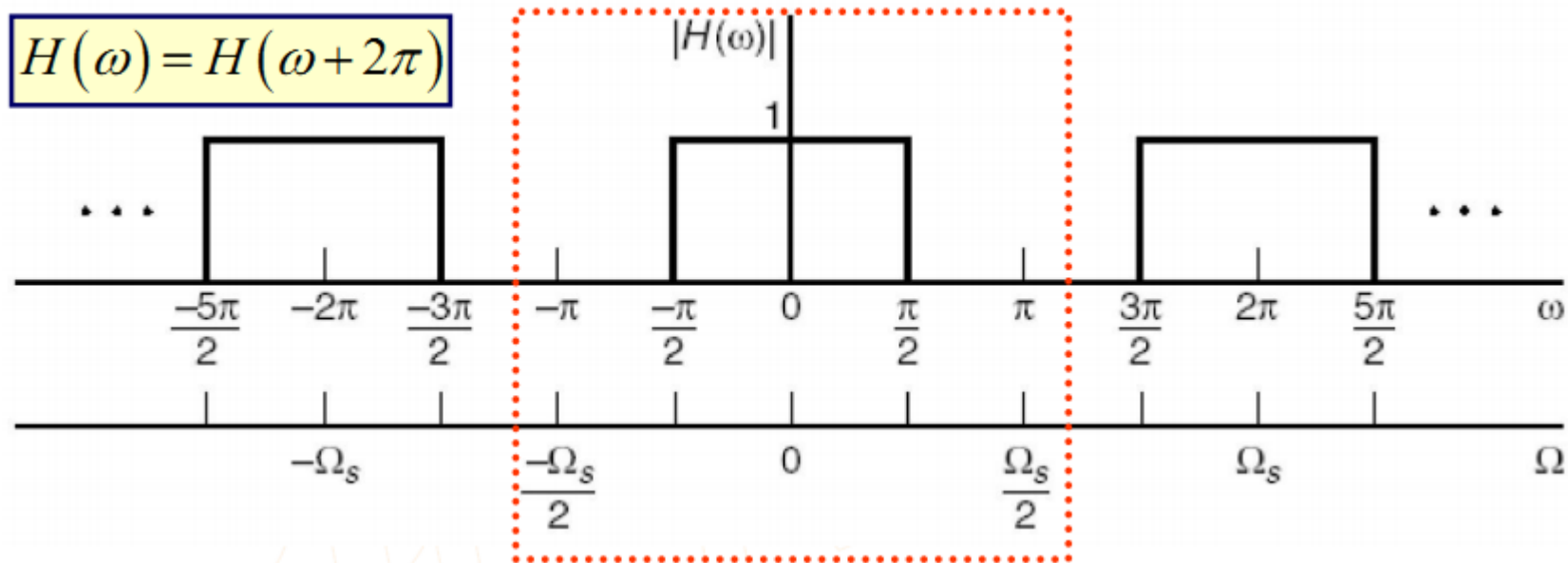
- The discrete-time Fourier transform (DTFT) $X(e^{j\omega})$ of a sequence $x[n]$ is defined by:

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

- DTFT $X(e^{j\omega})$ of a sequence $x[n]$ is a continuous function of ω
- Inverse Discrete-Time Fourier Transform - the Fourier coefficients $\{x[n]\}$ can be computed from $X(e^{j\omega})$ using

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

1.1 Frequency Mapping



- The discrete frequency 2π corresponds to the sampling frequency Ω_s used to sample the original continuous signal $x(t)$ to obtain $x[n]$.

No worries, this will be elaborated in details in the “sampling” lecture later...



1.2 Existence of DTFT

- In the case of **finite-length** sequences, the sum defining the DTFT has a finite number of terms, thus the DTFT always exists.
- In the **general** case, where one or both of the limits on the sum in the definition are infinite, the DTFT sum may diverge (become infinite).
- A **sufficient condition** for the existence of the DTFT of a sequence $x[n]$ is

$$|X(e^{j\omega})| \leq \sum_{n=-\infty}^{\infty} |x[n]| < \infty.$$

Proof:

$$\begin{aligned} |X(e^{j\omega})| &= \left| \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} \right|, \\ &\leq \sum_{n=-\infty}^{\infty} |x[n]e^{-j\omega n}|, \\ &= \sum_{n=-\infty}^{\infty} |x[n]| |e^{-j\omega n}|, \\ &= \sum_{n=-\infty}^{\infty} |x[n]|. \end{aligned}$$



1.2 Existence of DTFT

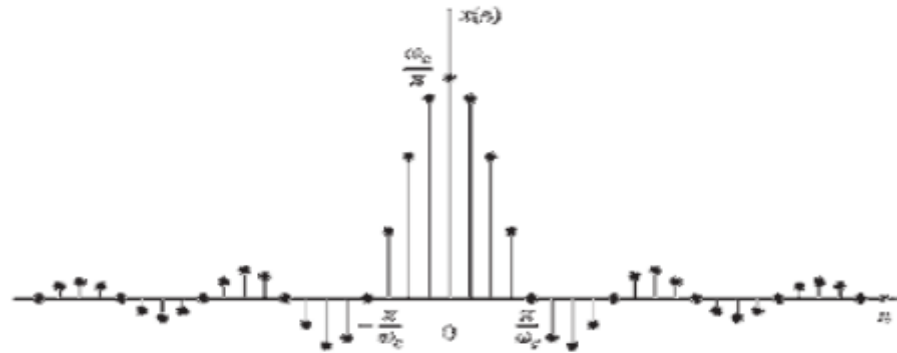
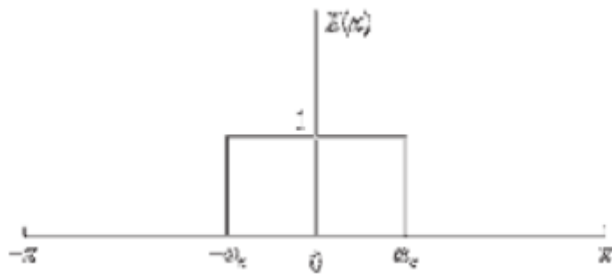
- Example: Find the DTFT of $x[n] = r^n e^{j\omega_0 n} u[n]$, and determine the condition to guarantee its existence.

$$\begin{aligned} X(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} r^n e^{j\omega_0 n} u[n] e^{-j\omega n}, \\ &= \sum_{n=0}^{\infty} r^n e^{j\omega_0 n} e^{-j\omega n} = \sum_{n=0}^{\infty} \left[r e^{-j(\omega - \omega_0)} \right]^n \\ &= \frac{1}{1 - r e^{-j(\omega - \omega_0)}}, \quad |r e^{-j(\omega - \omega_0)}| < 1 \\ &= \frac{1}{1 - r e^{-j(\omega - \omega_0)}}, \quad |r| < 1. \end{aligned}$$



1.2 Special case

$$X(\omega) = \begin{cases} 1, & |\omega| \leq \omega_c \\ 0, & \omega_c \leq |\omega| \leq \pi \end{cases} \xleftrightarrow{\mathcal{F}^{-1}} x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega = \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} e^{j\omega n} d\omega = \frac{\sin \omega_c n}{\pi n} \quad (\text{with } n \neq 0)$$



- The sequence $\{x(n)\}$ is not absolutely summable $\sum_{n=-\infty}^{\infty} |x(n)| \nless \infty$
- But it is mean-square convergent $\sum_{n=-\infty}^{\infty} |x(n)|^2 < \infty$



2.1 DTFT Property - Periodicity

- The DTFT of a discrete sequence is periodic with the period 2π , that is

$$X(\omega) = X(\omega + 2\pi k) \quad \text{for any integer } k$$

- The periodicity of DTFT can be easily verified from the definition:

$$\begin{aligned} X(\omega + 2\pi k) &= \sum_{n=-\infty}^{\infty} x[n] e^{-j(\omega + 2\pi k)n} \\ &= \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} e^{-j(2\pi k)n} = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} = X(\omega) \end{aligned}$$

2.2 DTFT Property - Linearity

- Linearity

- Given $x_1[n]$ and $X_1(\omega)$ form a DTFT pair, and $x_2[n]$ and $X_2(\omega)$ form another DTFT pair i.e.

$$x_1[n] \xleftrightarrow{\text{DTFT}} X_1(\omega)$$

$$x_2[n] \xleftrightarrow{\text{DTFT}} X_2(\omega)$$

- We can show that

$$ax_1[n] + bx_2[n] \xleftrightarrow{\text{DTFT}} aX_1(\omega) + bX_2(\omega)$$

- Example:

$$x[n] = 0.8^n u[n] + 2(-0.5)^n u[n]$$



2.3 DTFT Property - Time-reversal

- Time-reversal: A reversal of the time domain variable causes a reversal of the frequency variable

$$x[-n] \xleftrightarrow{\text{DTFT}} X(-\omega)$$

2.4 DTFT Properties - Conjugate symmetry

- Conjugation

$$x^*[n] \xleftrightarrow{\text{DTFT}} X^*(-\omega) \qquad x^*[-n] \xleftrightarrow{\text{DTFT}} X^*(\omega)$$

- Conjugate Symmetry

– 1. If $x[n]$ is real: $X(\omega) = X^*(-\omega)$

$$|X(\omega)| = |X(-\omega)|$$

$$\varphi(\omega) = -\varphi(-\omega)$$

$$X_R(\omega) = X_R(-\omega)$$

$$X_I(\omega) = -X_I(-\omega)$$

– 2. If $x[n] = x_{\text{even}}[n] + x_{\text{odd}}[n]$

$$x_{\text{even}}[n] \xleftrightarrow{\text{DTFT}} X_{\text{real}}(\omega)$$

$$x_{\text{odd}}[n] \xleftrightarrow{\text{DTFT}} jX_{\text{imag}}(\omega)$$

2.5 DTFT Properties - Shifting

- Time Domain Shifting (TD Delay) \Rightarrow FD Phase Shift

$$x[n - M] \xleftrightarrow{\text{DTFT}} e^{-j\omega M} X(\omega)$$

– Note that the magnitude spectrum is unchanged by time shift.

- Frequency Domain Shifting \Rightarrow TD Phase Shift

$$e^{j\omega_0 n} x[n] \xleftrightarrow{\text{DTFT}} X(\omega - \omega_0)$$

- Example: Find DTFT of $x[n] = A \cos(\omega_0 n + \varphi) \alpha^n u[n]$, with $|\alpha| < 1$.

2.6 DTFT Properties - Differencing

- Differencing in Time

$$x[n] - x[n - 1] \xleftrightarrow{\text{DTFT}} (1 - e^{-j\omega})X(\omega)$$

- Differentiation in Frequency

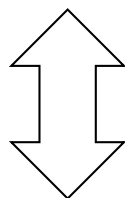
$$nx[n] \xleftrightarrow{\text{DTFT}} j \frac{dX(\omega)}{d\omega}$$



2.7 DTFT Properties - Parseval Theorem

- Parseval Theorem: The energy of the signal, whether computed in TD or FD, is the same!

$$\sum_{n=-\infty}^{\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(\omega)|^2 d\omega$$



Energy density spectrum
of the signal

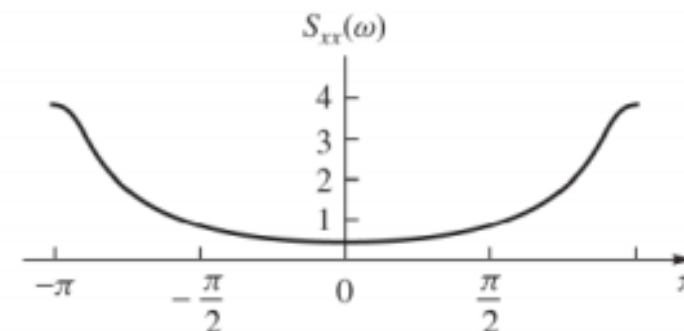
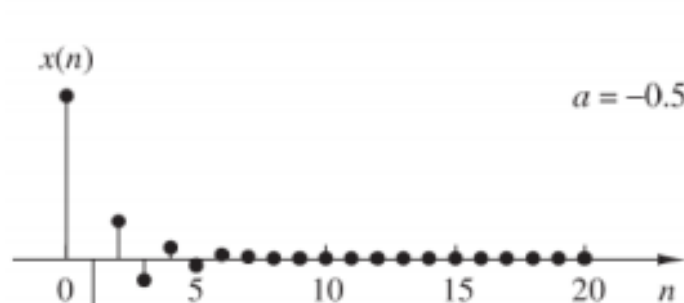
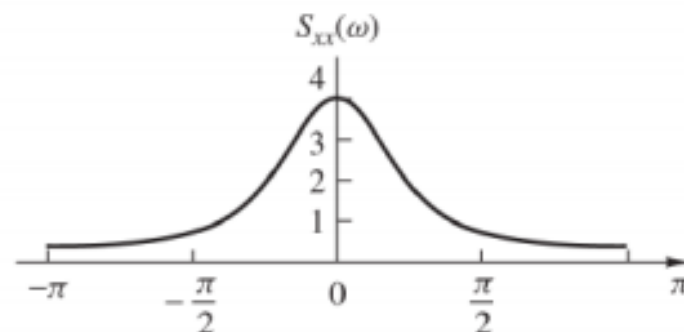
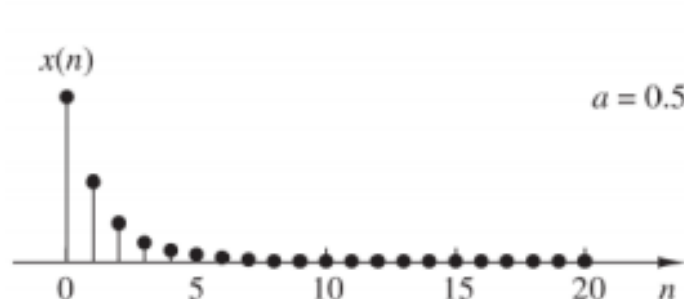
$$\sum_{n=-\infty}^{\infty} x[n] \cdot x^*[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) \cdot X^*(\omega) d\omega$$

2.7 DTFT Properties - Parseval Theorem

- Example - Determine and sketch the energy density spectrum of the signal

$$x(n) = a^n u(n), \quad -1 < a < 1$$

- Result:

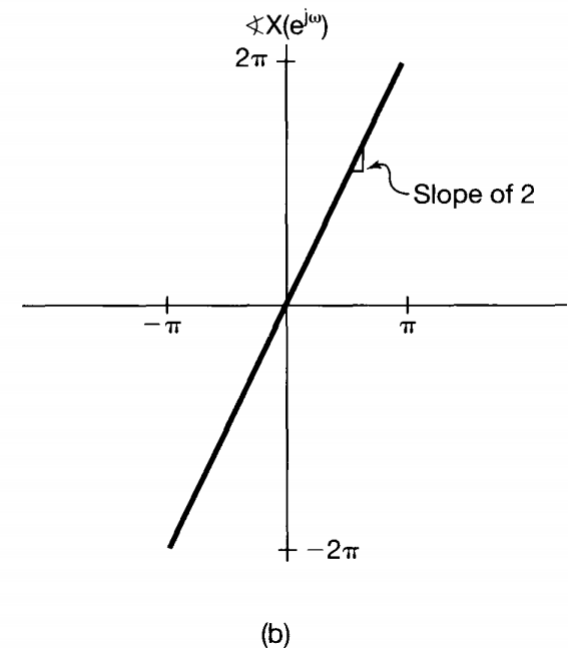
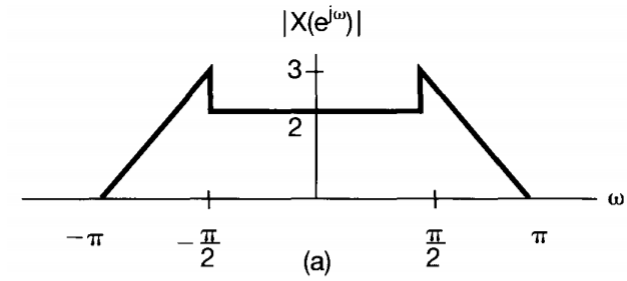


(a)

(b)

Quiz 1

- Example: Consider the sequence $x[n]$ whose Fourier transform $X(e^{j\omega}) = X(\omega)$ is depicted for $-\pi \leq \omega \leq \pi$.
- Determine whether or not, in the time domain, $x[n]$ is periodic, real, even, and/or of finite energy.



2.8 DTFT Properties - Convolution Property

- Convolution in TD = multiplication in FD

$$y[n] = x[n] * h[n] \xleftrightarrow{\text{DTFT}} X(\omega) \cdot H(\omega) = Y(\omega)$$

- The convolution property illustrates the system's response to input, in TD, the output is the result of convolution, and in FD, the output is the result of multiplication;
- By designing the $H(\omega)$ carefully, we can pass certain frequency components, then make $|H(\omega)| \cong 1$; stop certain frequency components, i.e. $|H(\omega)| \cong 0$. This is the concept of ***FILTERING***

2.8 DTFT Properties - Convolution Property

- Example: Consider an LTI system with impulse response

$$h[n] = a^n u[n], \quad |a| < 1$$

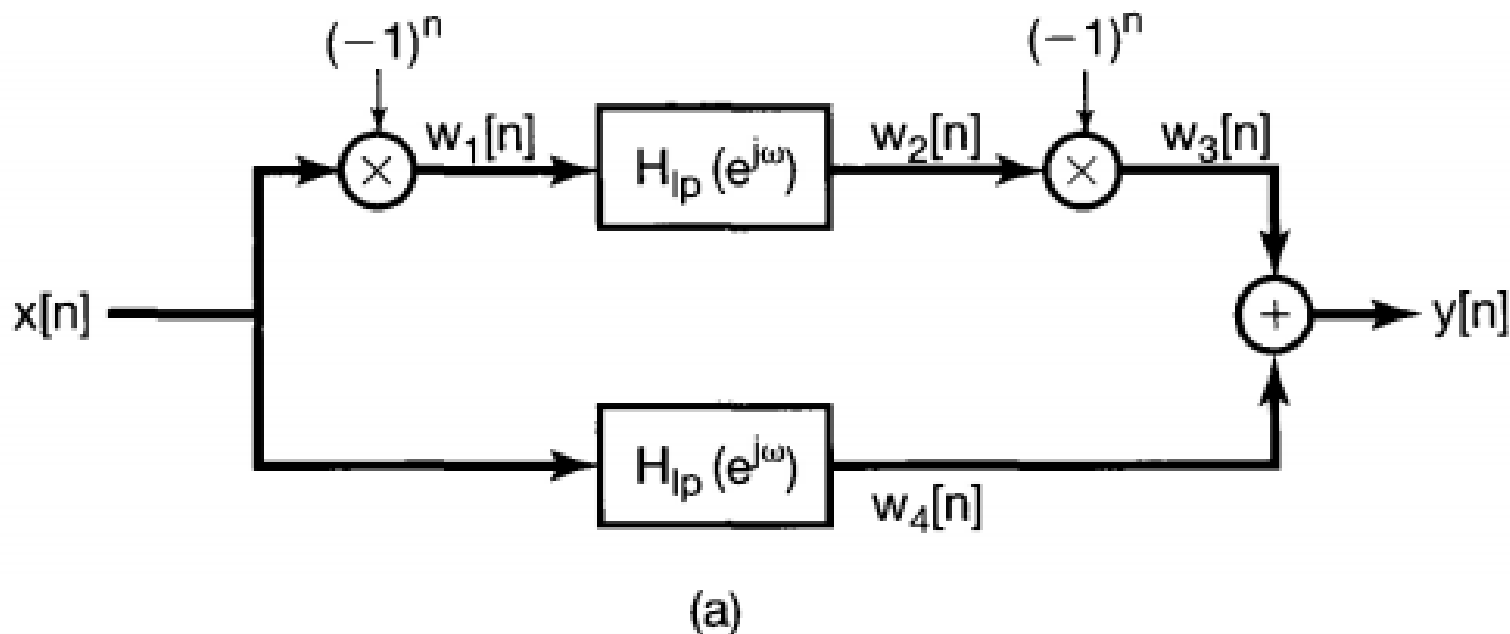
- Suppose that the input to this system is

$$x[n] = \beta^n u[n], \quad |\beta| < 1$$

- Find the output $y[n]$.

2.8 DTFT Properties - Convolution Property

- Example: Consider the system shown in below figure with input $x[n]$ and output $y[n]$. The LTI systems with frequency response $H_{lp}(\omega)$ are ideal lowpass filters with cutoff frequency $\pi/4$ and unity gain in the passband.
- Find the overall frequency response of this system.



2.8 DTFT Properties - Multiplication Property

- Multiplication in TD = convolution integral in FD

$$x[n] \cdot h[n] \xleftrightarrow{\text{DTFT}} \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\gamma) \cdot H(\omega - \gamma) d\gamma$$

- $h[n]$ can be considered as either system impulse response or another signal;
- This property is also called the modulation property, since it involves the modulation of one signal $x[n]$ with the other $h[n]$;

2.8 DTFT Properties - Multiplication Property

- Example: Find the Fourier transform $X(e^{j\omega})$ of a signal $x[n] = x_1[n] x_2[n]$ where:

$$x_1[n] = \frac{\sin(3\pi n/4)}{\pi n}$$

$$x_2[n] = \frac{\sin(\pi n/2)}{\pi n}$$

2.9 Duality - DTFS (Optional)

- Since the Fourier series coefficients a_k of a periodic signal $x[n]$ are themselves a periodic sequence, we can expand the sequence a_k in a Fourier series.
- The **duality property for DTFS** implies that the Fourier series coefficients for the periodic sequence a_k are the values of $(1/N)x[-n]$ (i.e., are proportional to the values of the original signal, reversed in time).

- Example:

$$\begin{array}{ccc}
 x[n] = \begin{cases} \frac{1}{9} \frac{\sin(5\pi n/9)}{\sin(\pi n/9)}, & n \neq \text{multiple of } 9 \\ \frac{5}{9}, & n = \text{multiple of } 9 \end{cases} & \xleftrightarrow{\text{DTFS}} & a_k = \begin{cases} 1/9, & |k| \leq 2 \\ 0, & 2 < |k| \leq 4 \end{cases} \\
 & & \text{with period } N=9
 \end{array}$$

$$\begin{array}{ccc}
 g[n] = \begin{cases} 1, & |n| \leq 2 \\ 0, & 2 < |n| \leq 4 \end{cases} & \xleftrightarrow{\text{DTFS}} & b_k = \begin{cases} \frac{1}{9} \frac{\sin(5\pi k/9)}{\sin(\pi k/9)}, & k \neq \text{multiple of } 9 \\ \frac{5}{9}, & k = \text{multiple of } 9 \end{cases} \\
 \text{with period } N=9 & &
 \end{array}$$

The diagram illustrates the duality property of the Discrete-Time Fourier Series (DTFS). It shows two pairs of signals and their DTFS coefficients. The top pair shows a periodic sequence $x[n]$ and its DTFS coefficients a_k . The bottom pair shows a periodic sequence $g[n]$ and its DTFS coefficients b_k . Arrows labeled "DTFS" indicate the transformation between each signal and its coefficients. A double-headed arrow connects a_k and b_k , representing the duality property where the coefficients of one signal are the samples of the other signal.

2.9 Duality - DTFT (Optional)

- Recall: Duality for CTFT

- For a transform pair

$$x(t) \xleftrightarrow{\mathcal{F}} X(\omega)$$

- implies that

$$X(t) \xleftrightarrow{\mathcal{F}} 2\pi x(-\omega)$$

- Time domain and frequency domain are symmetric.
 - This property suggests if signal A's frequency spectrum is signal B, then signal B's frequency spectrum takes a form similar to signal A.
- Using linear frequency f instead of angular frequency ω , there is:

$$X(t) \xleftrightarrow{\mathcal{F}} x(-f)$$

- for DTFT

- the duality is between the DTFT and CTFS

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega,$$

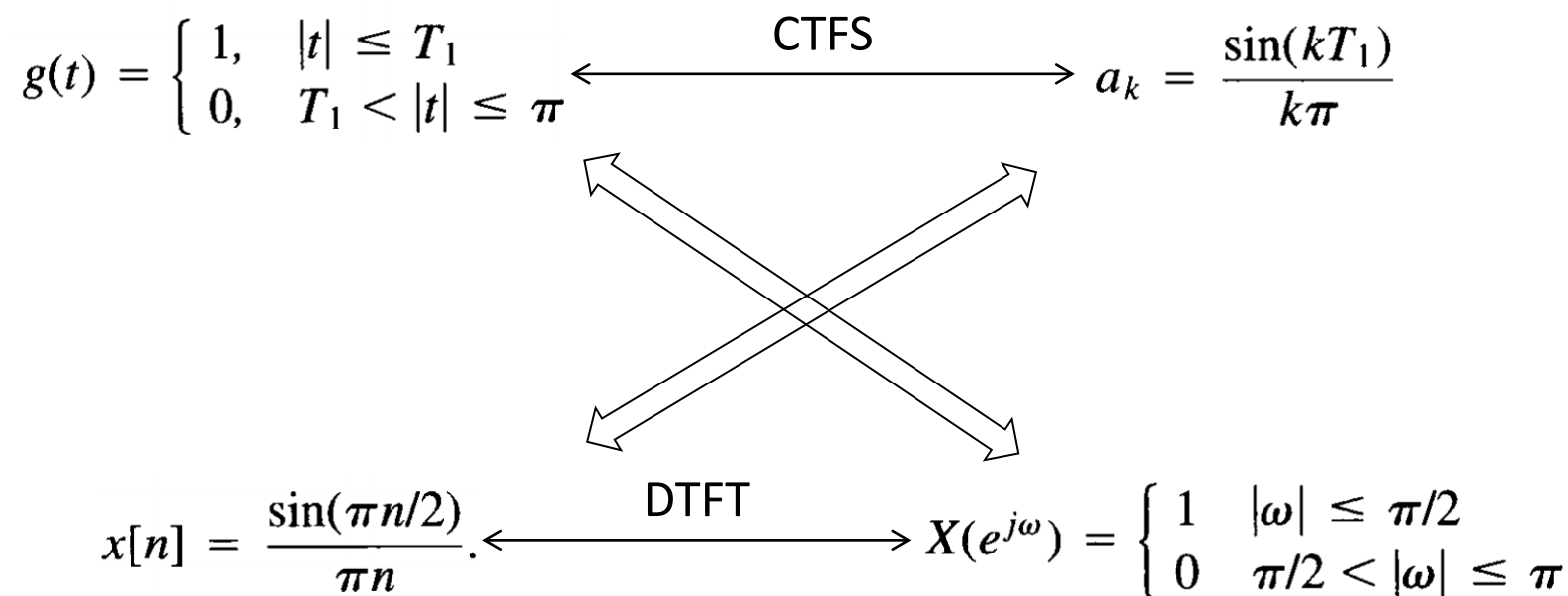
$$X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n},$$

$$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t},$$

$$a_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt.$$

2.9 Duality - DTFT (Optional)

- Example:



2.9 Duality - Summary (Optional)

- Summary

TABLE 5.3 SUMMARY OF FOURIER SERIES AND TRANSFORM EXPRESSIONS

	Continuous time			Discrete time	
	Time domain	Frequency domain		Time domain	Frequency domain
Fourier Series	$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}$ <p>continuous time periodic in time</p>	$a_k = \frac{1}{T_0} \int_{T_0} x(t) e^{-jk\omega_0 t} dt$ <p>discrete frequency aperiodic in frequency</p>		$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n}$ <p>discrete time periodic in time</p>	$a_k = \frac{1}{N} \sum_{k=\langle N \rangle} x[n] e^{-jk(2\pi/N)n}$ <p>discrete frequency periodic in frequency</p>
Fourier Transform	$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega$ <p>continuous time aperiodic in time</p>	$X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt$ <p>continuous frequency aperiodic in frequency</p>		$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$ <p>discrete time aperiodic in time</p>	$X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} x[n] e^{-j\omega n}$ <p>continuous frequency periodic in frequency</p>

duality

duality

duality

Quiz 2

- If $X(e^{j\omega})$ is the DTFT of the sequence $x[n] = \{3, 1, -4, 0, -5, 2, 1; -4 \leq n \leq 2\}$. Calculate the values of following expressions without calculating the DTFT:
 - (a) $X(e^{j0})$;
 - (b) $X(e^{j\pi})$;
 - (c) $\int_{-\pi}^{\pi} X(e^{j\omega}) d\omega$;
 - (d) $\int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$;
 - (e) $\int_{-\pi}^{\pi} \left| \frac{dX(e^{j\omega})}{d\omega} \right|^2 d\omega$.

Next ...

- DTFT pairs
- Inverse DTFT
- DTFT of LTID systems
- Concept of filtering