SBVS152C - DECEMBER 2010 - REVISED OCTOBER 2011

## 16-Channel, Constant-Current LED Driver with 4-Channel Grouped Delay

Check for Samples: TLC59282

#### **FEATURES**

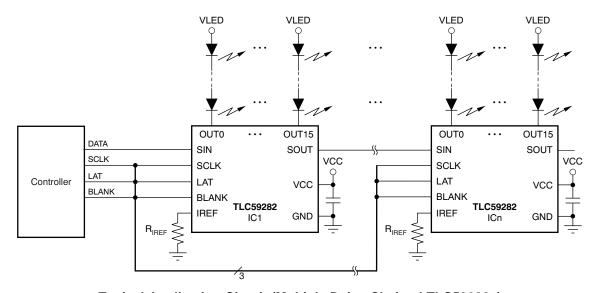
- 16 Channels, Constant-Current Sink Output with On/Off Control
- Capability (Constant-Current Sink): 35 mA (V<sub>CC</sub> ≤ 3.6 V), 45 mA (V<sub>CC</sub> > 3.6 V)
- LED Power-Supply Voltage up to 17 V
- V<sub>CC</sub> = 3 V to 5.5 V
- Constant-Current Accuracy:
  - Channel-to-Channel = ±0.6% (typ),±2% (max)
  - Device-to-Device =  $\pm 1\%$  (typ),  $\pm 3\%$  (max)
- Low Saturation Voltage: 0.31 V at 20 mA (typ)
  - $T_A = +25$ °C, One Channel On
- · CMOS Logic Level I/O
- Data Transfer Rate: 35 MHz
- BLANK Pulse Width: 30 ns
- Four-Channel Grouped Delay for
  - Noise Reduction
- Operating Temperature: –40°C to +85°C

## **APPLICATIONS**

- Video Displays
- Message Boards
- Illumination

## **DESCRIPTION**

The TLC59282 is a 16-channel, constant-current sink driver. Each channel can be individually controlled via a simple serial communications protocol that is compatible with 3.3 V or 5 V CMOS logic levels, depending on the operating VCC. Once the serial data buffer is loaded, a rising edge on LATCH transfers the data to the LEDx outputs. The BLANK pin can be used to turn off all OUTn outputs during power-on and output data latching to prevent unwanted image displays during these times. The constant-current value of all 16 channels is set by a single external resistor. Multiple TLC59282s can be cascaded together to control additional LEDs from the same processor.



Typical Application Circuit (Multiple Daisy-Chained TLC59282s)

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.





This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

## PACKAGE/ORDERING INFORMATION(1)

PRODUCT	PACKAGE-LEAD	ORDERING NUMBER	TRANSPORT MEDIA, QUANTITY
TLC59282	SSOP-24/QSOP-24	TLC59282DBQR	Tape and Reel, 2500
1LC59282	330P-24/Q30P-24	TLC59282DBQ	Tube, 50
TI 050202	OEN 24	TLC59282RGER	Tape and Reel, 3000
TLC59282	QFN-24	TLC59282RGE	Tape and Reel, 250

<sup>(1)</sup> For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the device product folder at www.ti.com.

## **ABSOLUTE MAXIMUM RATINGS**(1)(2)

Over operating free-air temperature range, unless otherwise noted.

	PAR	AMETER	TLC59282	UNIT
$V_{CC}$	Supply voltage		-0.3 to +6	V
l <sub>OUT</sub>	Output current (dc)	OUT0 to OUT15	50	mA
$V_{IN}$	Input voltage range	SIN, SCLK, LAT, BLANK, IREF	-0.3 to V <sub>CC</sub> + 0.3	V
.,	Output voltage renge	SOUT	-0.3 to V <sub>CC</sub> + 0.3	V
V <sub>OUT</sub>	Output voltage range	OUT0 to OUT15	-0.3 to +18	V
$T_{J(MAX)}$	Operating junction temperature	re	+150	°C
T <sub>STG</sub>	Storage temperature range		–55 to +150	°C
	ECD roting	Human body model (HBM)	4000	V
	ESD rating	Charged device model (CDM)	1000	V

<sup>(1)</sup> Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not supported.

#### THERMAL INFORMATION

		TLC	59282	
	THERMAL METRIC <sup>(1)</sup>	DBQ	RGE	UNITS
		24 PINS	24 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance	73.2	46.8	
$\theta_{\text{JCtop}}$	Junction-to-case (top) thermal resistance	44.6	48.6	
$\theta_{JB}$	Junction-to-board thermal resistance	38.9	23.0	°C // //
ΨЈТ	Junction-to-top characterization parameter	12.3	1.2	°C/W
ΨЈВ	Junction-to-board characterization parameter	39.7	22.9	
$\theta_{JCbot}$	Junction-to-case (bottom) thermal resistance	n/a	6.3	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

<sup>(2)</sup> All voltage values are with respect to network ground terminal.

## **RECOMMENDED OPERATING CONDITIONS**

At  $T_A = -40$ °C to +85°C, unless otherwise noted.

			Т	LC59282		
	PARAMETER	TEST CONDITIONS	MIN	NOM	MAX	UNIT
DC Characte	eristics: V <sub>CC</sub> = 3 V to 5.5 V					
V <sub>CC</sub>	Supply voltage		3		5.5	V
Vo	Voltage applied to output	OUT0 to OUT15			17	V
V <sub>IH</sub>	High-level input voltage		0.7 × V <sub>CC</sub>		V <sub>CC</sub>	V
V <sub>IL</sub>	Low-level input voltage		GND		0.3 × V <sub>CC</sub>	V
ОН	High-level output current	SOUT			-1	mA
OL	Low-level output current	SOUT			1	mA
		OUT0 to OUT15, 3 V ≤ V <sub>CC</sub> < 3.6 V	2		35	mA
I <sub>OLC</sub>	Constant output sink current	OUT0 to OUT15, 3.6 V ≤ V <sub>CC</sub> < 5.5 V	2		45	mA
ГА	Operating free-air temperature range		-40		+85	°C
ГЈ	Operating junction temperature range		-40		+125	°C
AC Characte	eristics: V <sub>CC</sub> = 3 V to 5.5 V		·			
: CLK (SCLK)	Data shift clock frequency	SCLK			35	MHz
T <sub>WH0</sub>		SCLK	10			ns
Γ <sub>WL0</sub>		SCLK	10			ns
Γ <sub>WH1</sub>	Pulse duration	LAT	20			ns
Γ <sub>WH2</sub>		BLANK	60			ns
Γ <sub>WL2</sub>		BLANK	30			ns
r <sub>suo</sub>	Contract time of	SIN-SCLK↑	4			ns
Γ <sub>SU1</sub>	Setup time	LAT↓-SCLK↑	10			ns
Γ <sub>H0</sub>	Hald Care	SIN-SCLK↑	4			ns
T <sub>H1</sub>	Hold time	LAT↓-SCLK↑	10			ns



#### **ELECTRICAL CHARACTERISTICS**

At  $V_{CC} = 3$  V to 5.5 V and  $T_A = -40$ °C to +85°C. Typical values at  $V_{CC} = 3.3$  V and  $T_A = +25$ °C, unless otherwise noted.

			1	LC59282		
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	I <sub>OH</sub> = -1 mA at SOUT	V <sub>CC</sub> - 0.4		V <sub>CC</sub>	V
V <sub>OL</sub>	Low-level output voltage	I <sub>OL</sub> = 1 mA at SOUT			0.4	V
I <sub>IN</sub>	Input current	V <sub>IN</sub> = V <sub>CC</sub> or GND at SIN and SCLK	-1		1	μΑ
I <sub>CC0</sub>		SIN/SCLK/LAT = low, BLANK = high, $V_{OUTn} = 1 \text{ V}$ , $R_{IREF} = \text{open}$		0.1	1	mA
I <sub>CC1</sub>	County compact (A/ )	SIN/SCLK/LAT = low, BLANK = high, $V_{OUTn} = 1 \text{ V}$ , $R_{IREF} = 3 \text{ k}\Omega \text{ (}I_{OUT} = 16.8 \text{ mA target)}$		4.5	6	mA
I <sub>CC2</sub>	Supply current (V <sub>CC</sub> )	All OUTn = ON, SIN/SCLK/LAT/BLANK = low, $V_{OUTn}$ = 1 V, $R_{IREF}$ = 3 k $\Omega$		7		
ссз		All OUTn = ON, SIN/SCLK/LAT/BLANK = low, $V_{OUTn}$ = 1 V, $R_{IREF}$ = 1.5 k $\Omega$ ( $I_{OUT}$ = 33.6mA target)		16	34	mA
locc	Constant output current	All OUTn = ON, $V_{OUTn}$ = $V_{OUTfix}$ = 1 V, $R_{IREF}$ = 1.5 k $\Omega$ at OUT0 to OUT15 (see Figure 6), $T_A$ = +25°C	32.1	33.7	35.3	mA
l <sub>oLKG</sub>	Output leakage current	OUTn = OFF, $V_{OUTn} = V_{OUTfix} = 17 \text{ V}$ , BLANK = high, $R_{IREF} = 1.5 \text{ k}\Omega$ at OUT0 to OUT15 (see Figure 6)			0.1	μA
ΔI <sub>OLC0</sub>	Constant-current error (channel-to-channel) (1)	All OUTn = ON, $V_{OUTn}$ = $V_{OUTfix}$ = 1 V, $R_{IREF}$ = 1.5 $k\Omega$ at OUT0 to OUT15		±0.6	±2	%
∆I <sub>OLC1</sub>	Constant-current error (device-to-device) <sup>(2)</sup>	All OUTn = ON, $V_{OUTn}$ = $V_{OUTfix}$ = 1 V, $R_{IREF}$ = 1.5 k $\Omega$ at OUT0 to OUT15, $T_A$ = +25°C		±1	±3	%
∆I <sub>OLC2</sub>	Line regulation (3)	All OUTn = ON, V <sub>OUTn</sub> = V <sub>OUTfix</sub> = 1 V, R <sub>IREF</sub> = 1.5 k $\Omega$ at OUT0 to OUT15, V <sub>CC</sub> = 3 V to 5.5 V		±0.5	±1	%/V
∆I <sub>OLC3</sub>	Load regulation <sup>(4)</sup>	All OUTn = ON, $V_{OUTn}$ = 1 V to 3V, $V_{OUTfix}$ = 1 V, $R_{IREF}$ = 1.5 k $\Omega$		±1	±3	%/V
V <sub>IREF</sub>	Reference voltage output	$R_{IREF} = 1.5 \text{ k}\Omega, T_A = +25^{\circ}\text{C}$	1.18	1.205	1.23	V
R <sub>PUP</sub>	Pull-up resistor	BLANK	250	500	750	kΩ
R <sub>PDWN</sub>	Pull-down resistor	LAT	250	500	750	kΩ

(1) The deviation of each output from the average of OUT0-OUT15 constant-current. Deviation is calculated by the formula:

$$\Delta \text{ (\%)} = \left[ \frac{I_{\text{OUTn}}}{\frac{(I_{\text{OUT0}} + I_{\text{OUT1}} + \dots + I_{\text{OUT14}} + I_{\text{OUT15}})}{16}} - 1 \right] \times 100$$

The deviation of the OUT0-OUT15 constant-current average from the ideal constant-current value. Deviation is calculated by the following formula:

$$\Delta \text{ (\%) = } \left[ \begin{array}{c} \frac{(I_{\text{OUT0}} + I_{\text{OUT1}} + \dots I_{\text{OUT15}})}{16} - \text{(Ideal Output Current)} \\ \hline & \text{Ideal Output Current} \end{array} \right] \times 100$$

Ideal current is calculated by the formula: 
$$I_{OUT(IDEAL)} = 41.9 \times \left[ \frac{1.205}{R_{IREF}} \right]$$

(3) Line regulation is calculated by this equation: 
$$\Delta \ (\%/V) = \left\{ \frac{(I_{OUTn} \ at \ V_{CC} = 5.5 \ V) - (I_{OUTn} \ at \ V_{CC} = 3 \ V)}{(I_{OUTn} \ at \ V_{CC} = 3 \ V)} \right\} \times \frac{100}{5.5 \ V - 3 \ V}$$

(4) Load regulation is calculated by the equation:  

$$\Delta \ (\%/V) = \left( \frac{(I_{OUTn} \text{ at } V_{OUTn} = 3 \text{ V}) - (I_{OUTn} \text{ at } V_{OUTn} = 1 \text{ V})}{(I_{OUTn} \text{ at } V_{OUTn} = 1 \text{ V})} \right) \times \frac{100}{3 \text{ V} - 1 \text{ V}}$$

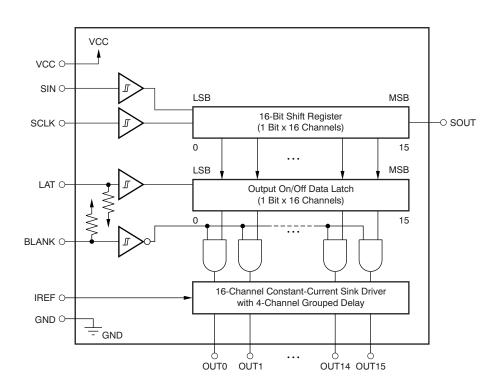
## **SWITCHING CHARACTERISTICS**

At  $V_{CC}$  = 3 V to 5.5 V,  $T_A$  = -40°C to +85°C,  $C_L$  = 15 pF,  $R_L$  = 130  $\Omega$ ,  $R_{IREF}$  = 1.5 k $\Omega$ , and  $V_{LED}$  = 5.5 V. Typical values at  $V_{CC}$  = 3.3 V and  $T_A$  = +25°C, unless otherwise noted.

			TL	C59282		
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t <sub>R0</sub>	Dies time	SOUT (see Figure 5)		5	12	ns
t <sub>R1</sub>	Rise time	OUTn (see Figure 4)		10	30	ns
t <sub>F0</sub>	Fall times	SOUT (see Figure 5)		5	12	ns
t <sub>F1</sub>	Fall time	OUTn (see Figure 4)		10	30	ns
t <sub>D0</sub>		SCLK↑ to SOUT↑↓		8	20	ns
t <sub>D1</sub>	-	LAT↑ or BLANK↑↓ to OUT0/OUT7/OUT8/OUT15 on/off		18	36	ns
t <sub>D2</sub>		LAT↑ or BLANK↑↓ to OUT1/OUT6/OUT9/OUT14 on/off		38	69	ns
t <sub>D3</sub>	Propagation delay time	LAT↑ or BLANK↑↓ to OUT2/OUT5/OUT10/OUT13 on/off		58	102	ns
t <sub>D4</sub>		LAT↑ or BLANK↑↓ to OUT3/OUT4/OUT11/OUT12 on/off		78	135	ns
t <sub>ON_ERR</sub>	Output on-time error <sup>(1)</sup>	On/off latch data = all '1', 30 ns BLANK low level one-shot pulse input	-15		15	ns

<sup>(1)</sup> Output on-time error  $(t_{ON\_ERR})$  is calculated by the formula:  $t_{ON\_ERR}$  (ns) =  $t_{OUT\_ON}$  – BLANK low level one-shot pulse width  $(T_{WL2})$ .  $t_{OUT\_ON}$  indicates the actual on-time of the constant-current output.

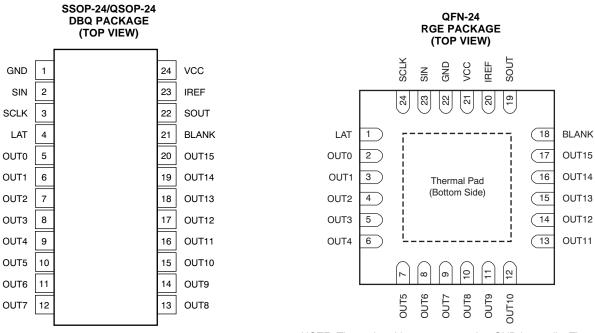
## **FUNCTIONAL BLOCK DIAGRAM**



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## **DEVICE INFORMATION**



NOTE: Thermal pad is not connected to GND internally. The thermal pad must be connected to GND via the PCB pattern.

## **TERMINAL FUNCTIONS**

TERMINAL							
NAME	DBQ	RGE	I/O	DESCRIPTION			
SIN	2	23	ı	Serial data input for driver on/off control; Schmitt buffer input. When SIN is high, data '1' are written into the LSB of the 16-bit shift register at the SCLK rising edge.			
SCLK	3	24	I	Serial data shift clock; Schmitt buffer input. All data in the 16-bit shift register are shifted toward the MSB by 1-bit synchronization of SCLK.			
LAT	4	1	1	Level triggered latch; Schmitt buffer input. The data in the 16-bit shift register continue to transfer to the output on/off data latch while LAT is high. Therefore, if the data in the 16-bit shift register are changed when LAT is high, the data in the data latch are also changed. The data in the data latch are held when LAT is low. This pin is internally pulled down to GND with a 500 k $\Omega$ (typ) resistor.			
BLANK	21	18	I	slank, all outputs; Schmitt buffer input. When BLANK is high, all constant-current outputs OUT0–OUT15) are forced off. When BLANK is low, all constant-current outputs are controlled ne data in the output on/off data latch. This pin is internally pulled up to $V_{CC}$ with a 500 k $\Omega$ (typesistor.			
IREF	23	20	I/O	Constant-current value setting, OUT0–OUT15 sink constant-current is set to desired value by connection to an external resistor between IREF and GND.			
SOUT	22	19	0	Serial data output. This output is connected to the MSB of the 16-bit shift register. SOUT data changes at the rising edge of SCLK.			
OUT0	5	2	0	Constant-current output. Each output can be tied together with others to increase the constant-current. Different voltages can be applied to each output.			
OUT1	6	3	0	Constant-current output			
OUT2	7	4	0	Constant-current output			
OUT3	8	5	0	Constant-current output			
OUT4	9	6	0	Constant-current output			
OUT5	10	7	0	Constant-current output			
OUT6	11	8	0	Constant-current output			
OUT7	12	9	0	Constant-current output			
OUT8	13	10	0	Constant-current output			
OUT9	14	11	0	Constant-current output			
OUT10	15	12	0	Constant-current output			
OUT11	16	13	0	Constant-current output			
OUT12	17	14	0	Constant-current output			
OUT13	18	15	0	Constant-current output			
OUT14	19	16	0	Constant-current output			
OUT15	20	17	0	Constant-current output			
VCC	24	21	_	Power-supply voltage			
GND	1	22	_	Power ground			



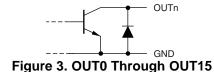
#### PARAMETER MEASUREMENT INFORMATION

## PIN EQUIVALENT INPUT AND OUTPUT SCHEMATIC DIAGRAMS

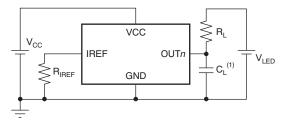


Figure 1. SIN, SCLK, LAT, BLANK

Figure 2. SOUT

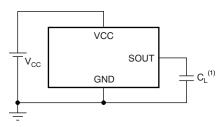


## **TEST CIRCUITS**



(1) C<sub>1</sub> includes measurement probe and jig capacitance.

Figure 4. Rise Time and Fall Time Test Circuit for OUTn



(1) C<sub>L</sub> includes measurement probe and jig capacitance.

Figure 5. Rise Time and Fall Time Test Circuit for SOUT

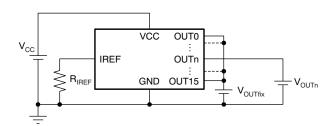
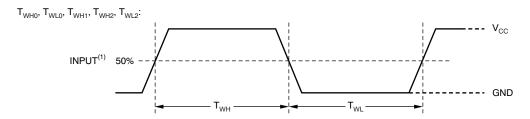
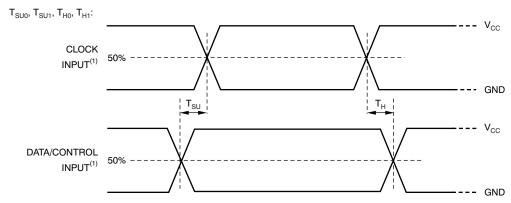


Figure 6. Constant-Current Test Circuit for OUTn



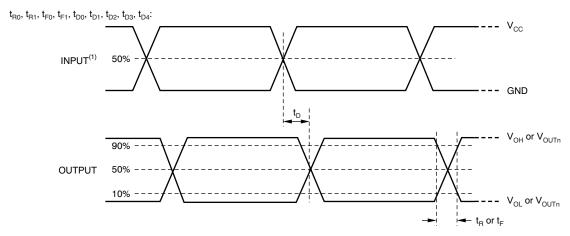
## **TIMING DIAGRAMS**





(1) Input pulse rise and fall time is 1 ns to 3 ns.

Figure 7. Input Timing

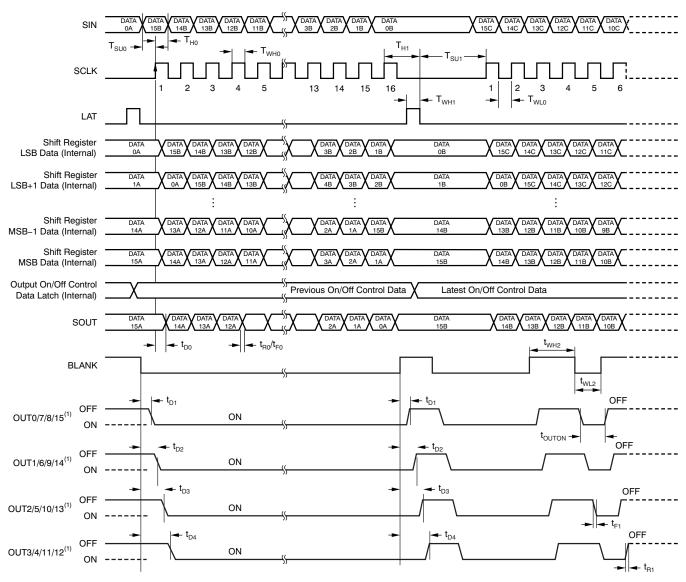


(1) Input pulse rise and fall time is 1 ns to 3 ns.

Figure 8. Output Timing

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- (1) Output on/off data = FFFFh.
- (2)  $t_{ON\_ERR} = t_{OUTON} T_{WL2}$ .

Figure 9. Timing Diagram



#### TYPICAL CHARACTERISTICS

At  $V_{CC} = 3.3 \text{ V}$  and  $T_A = +25^{\circ}\text{C}$ , unless otherwise noted.

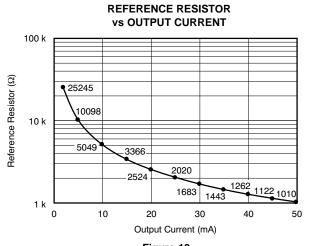


Figure 10.

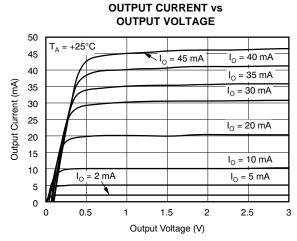


Figure 11.

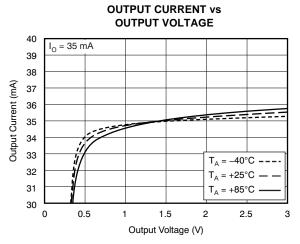


Figure 12.

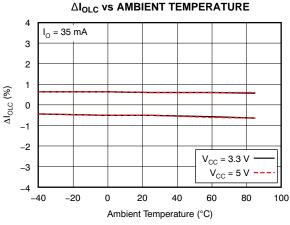


Figure 13.

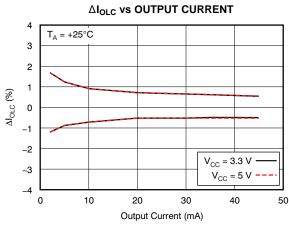


Figure 14.

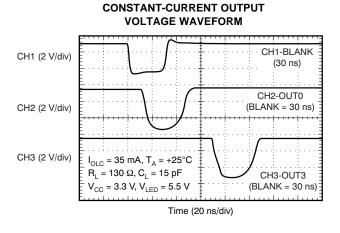


Figure 15.



#### **DETAILED DESCRIPTION**

#### SETTING FOR THE CONSTANT SINK CURRENT VALUE

The constant-current values are determined by an external resistor ( $R_{IREF}$ ) placed between IREF and GND. The resistor ( $R_{IREF}$ ) value is calculated by Equation 1.

$$R_{IREF} (k\Omega) = \frac{V_{IREF} (V)}{I_{OLC} (mA)} \times 41.9$$

Where:

$$V_{IREF}$$
 = the internal reference voltage on the IREF pin (typically 1.205 V) (1)

 $I_{OLC}$  must be set in the range of 2 mA to 35 mA when  $V_{CC}$  is less than 3.6 V. Also, when  $V_{CC}$  is equal to 3.6 V or greater,  $I_{OLC}$  must be set in the range of 2 mA to 45 mA. The constant sink current characteristic for the external resistor value is shown in Figure 10. Table 1 describes the constant-current output versus external resistor value.

Table 1. Constant-Current Output versus External Resistor Value

I <sub>OLC</sub> (mA, Typical)	$R_{IREF}$ (k $\Omega$ )
45 (V <sub>CC</sub> > 3.6 V only)	1.12
40 (V <sub>CC</sub> > 3.6 V only)	1.26
35	1.44
30	1.68
25	2.02
20	2.52
15	3.37
10	5.05
5	10.1
2	25.2

#### CONSTANT-CURRENT DRIVER ON/OFF CONTROL

When BLANK is low, the corresponding output is turned on if the data in the on/off control data latch are '1' and remains off if the data are '0'. When BLANK is high, all outputs are forced off. This control is shown in Table 2.

Table 2. On/Off Control Data Truth Table

OUTPUT ON/OFF DATA	CONSTANT-CURRENT OUTPUT STATUS
0	Off
1	On

When the IC is initially powered on, the data in the 16-bit shift register and output on/off data latch are not set to the respective default value. Therefore, the output on/off data must be written to the data latch before turning the constant-current output on. BLANK should be at a high level when powered on because the constant-current may be turned on as a result of random data in the output on/off data latch.

The output on/off data corresponding to any unconnected OUTn outputs should be set to '0' before turning on the remaining outputs. Otherwise, the supply current (I<sub>CC</sub>) increases while the LEDs are on.

#### **REGISTER CONFIGURATION**

The TLC59282 has a 16-bit shift register and an output on/off data latch. Both the shift register and data latch are 16 bits long and are used to turn the constant-current outputs on and off. Figure 16 shows the shift register and data latch configuration. The data at the SIN pin are shifted in to the LSB of the 16-bit shift register at the rising edge of the SCLK pin; SOUT data change at the rising edge of SCLK.

The output on/off data in the 16-bit shift register continue to transfer to the output on/off data latch while LAT is high. Therefore, if the data in the 16-bit shift register are changed when LAT is high, the data in the data latch are also changed. The data in the data latch are held when LAT is low. When the IC initially powers on, the data in the output on/off shift register and latch are not set to the default values; on/off control data must be written to the on/off control data latch before turning the constant-current output on. BLANK should be high when the IC is powered on because the constant-current may be turned on at that time as a result of random values in the on/off data latch. All constant-current outputs are forced off when BLANK is high. The OUTn on/off are controlled by the data in the output on/off data latch. The timing diagram and truth table for writing data are shown in Figure 17 and Table 3.

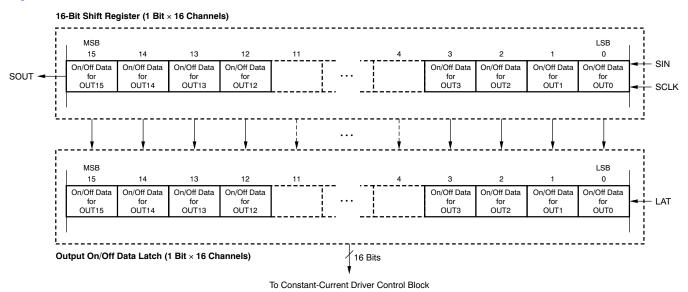


Figure 16. 16-Bit Shift Register and Output On/Off Data Latch Configuration



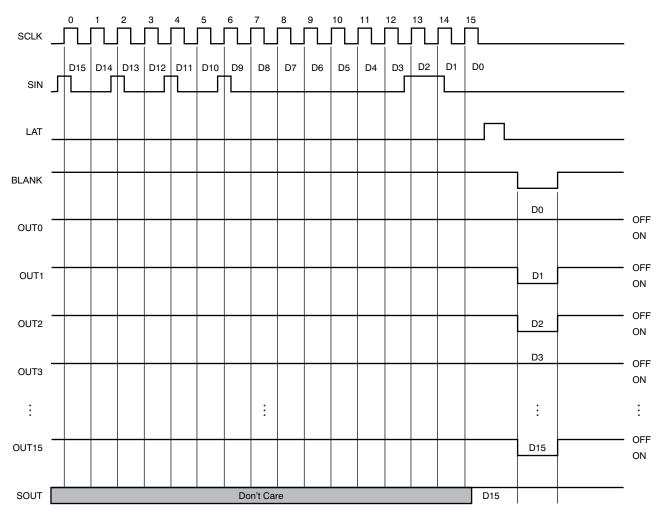


Figure 17. Operation Timing Diagram

**SCLK** LAT **BLANK** SIN OUT0...OUT7...OUT15 SOUT High Dn Dn...Dn - 7...Dn - 15 Dn - 15 Low 1 Low Low Dn + 1 No change Dn - 14 1 Dn + 2 Dn + 2...Dn - 5...Dn - 13 Dn - 13 High Low 1 Dn + 3 Dn + 2...Dn - 5...Dn - 13 Dn - 13 Low Dn + 3Dn - 13 High Off

**Table 3. Truth Table in Operation** 

## **NOISE REDUCTION**

Large surge currents may flow through the IC and the board if all 16 outputs turn on or off simultaneously. These large current surges could induce detrimental noise and electromagnetic interference (EMI) into other circuits. The TLC59282 independently turns on or off the outputs for each color group with a 20 ns (typ) delay time; see Figure 9. The output current sinks are grouped into four groups. The first group that is turned on/off are OUT0/7/8/15; the second group that is turned on/off are OUT1/6/9/14; the third group that is turned on/off are OUT2/5/10/13; and the fourth group is OUT3/4/11/12. Both turn-on and turn-off are delayed. However, the state of each output is controlled by the data in the output on-off data latch and BLANK level.

## **REVISION HISTORY**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	hanges from Revision B (July 2011) to Revision C	Page
•	Added Low Saturation Voltage Features bullet	1
CI	hanges from Revision A (December 2010) to Revision B	Page
•	Changed Constant-Current Accuracy Features bullet	1
•	Added RGE package information to Package/Ordering Information table	2
•	Added RGE package to Thermal Information table	2
•	Changed Input current parameter test conditions in Electrical Characteristics table	4
•	Added RGE pin out and footnote to Device Information section	6
•	Added RGE information to Terminal Functions table	7
•	Deleted Figure 11, POWER DISSIPATION RATE vs FREE-AIR TEMPERATURE	11





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#### **PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
TLC59282DBQ	ACTIVE	SSOP	DBQ	24	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TLC59282DBQR	ACTIVE	SSOP	DBQ	24	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TLC59282RGER	ACTIVE	VQFN	RGE	24	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TLC59282RGET	ACTIVE	VQFN	RGE	24	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free** (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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## PACKAGE MATERIALS INFORMATION

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## TAPE AND REEL INFORMATION

## **REEL DIMENSIONS**





## **TAPE DIMENSIONS**



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

## TAPE AND REEL INFORMATION

\*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC59282DBQR	SSOP	DBQ	24	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC59282RGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
TLC59282RGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

**PACKAGE MATERIALS INFORMATION** 

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\*All dimensions are nominal

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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC59282DBQR	SSOP	DBQ	24	2500	367.0	367.0	38.0
TLC59282RGER	VQFN	RGE	24	3000	367.0	367.0	35.0
TLC59282RGET	VQFN	RGE	24	250	210.0	185.0	35.0

DBQ (R-PDSO-G24)

## PLASTIC SMALL-OUTLINE PACKAGE



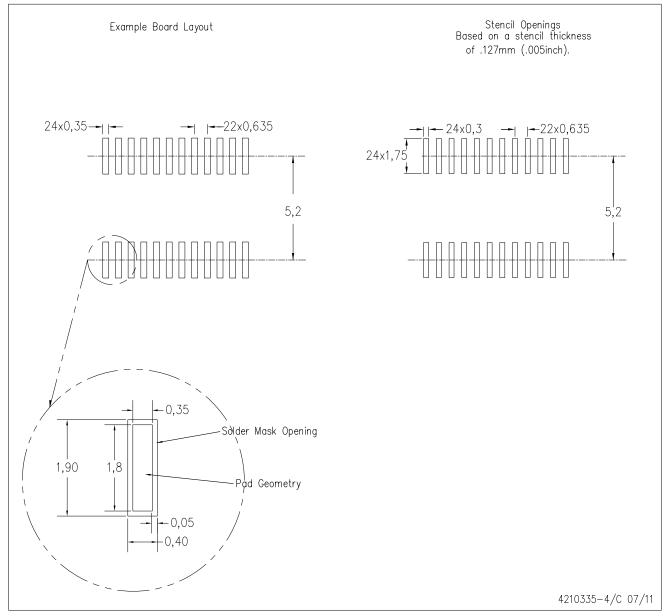
NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15) per side.
- D. Falls within JEDEC MO-137 variation AE.



DBQ (R-PDSO-G24)

# PLASTIC SMALL OUTLINE PACKAGE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.





NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.

- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-Leads (QFN) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- F. Falls within JEDEC MO-220.



## RGE (S-PVQFN-N24)

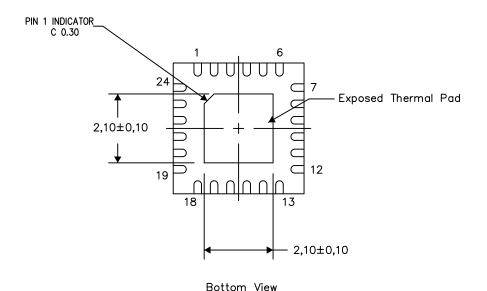
## PLASTIC QUAD FLATPACK NO-LEAD

## THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Exposed Thermal Pad Dimensions

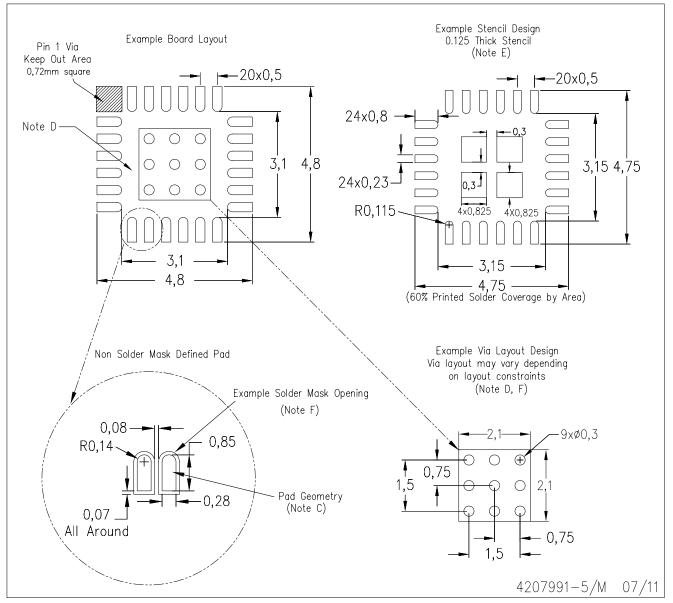
4206344-6/AA 04/12

NOTES: A. All linear dimensions are in millimeters



## RGE (S-PVQFN-N24)

## PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- S: A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="https://www.ti.com">www.ti.com</a>.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
  - F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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