

① → cálculo

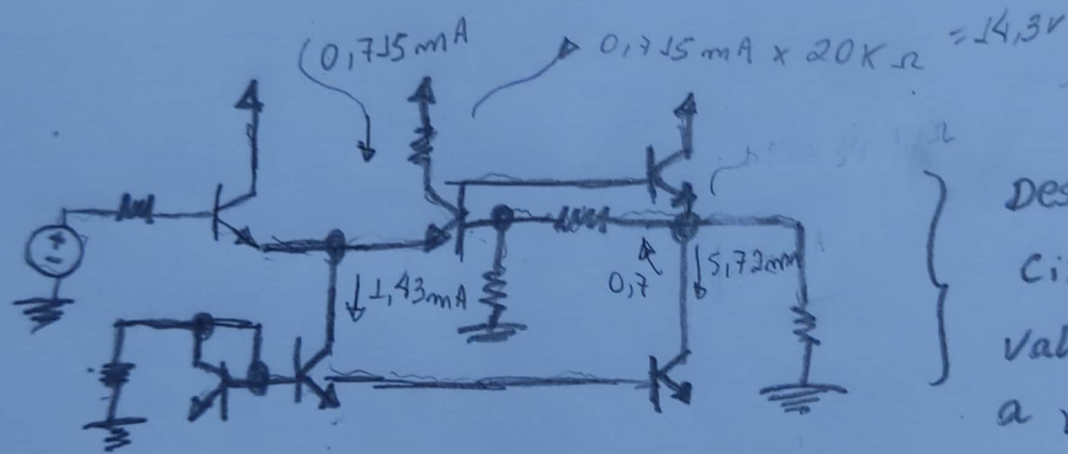
∴ calculando a corrente do circuito

corrente 1

$$\left\{ \frac{15 - 0,7}{10K} = 1,43 \cdot 10^{-3} = 1,43 \text{ mA} \right\}$$

corrente 2

$$\left(\frac{15 - 0,7}{30K} \right) \cdot 4 = 5,72 \text{ mA}$$



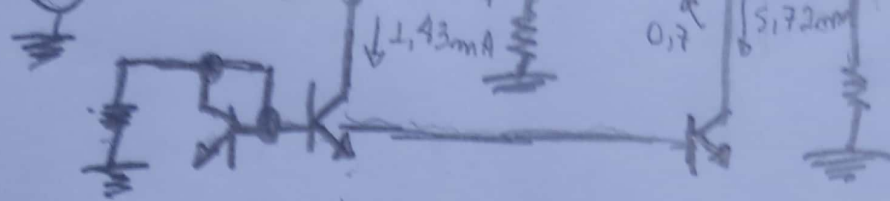
Desenhando o circuito sem os valores p/ ilustrar a resolução!

$$I_C = \frac{1,43 \text{ mA}}{2} = 0,715 \text{ mA}$$

∴ calculando R_E

→ valor nominal do diodo / Tensão Térmica

$$R_E = \frac{V_T}{I_E} \left\{ \begin{array}{l} R_E = \frac{25 \text{ mV}}{0,715 \text{ mA}} = 34,96 \Omega \\ R_E = 25 \text{ mV} \end{array} \right\}$$



Circuito sem os valores p/ ilustrar a resolução!

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∴ calculando R_E

→ Valor nominal da diodo / Tensão Térmica

$$R_E = \frac{V_T}{I_C} \quad \left\{ \quad R_E = \frac{25 \text{ mV}}{0,715 \text{ mA}} = 34,96 \, \Omega \quad \right. \quad \text{e} \quad R_E = \frac{25 \text{ mV}}{I_C}$$

$$I_{C1} = 0,715 \text{ mA}, \quad I_{C2} = 5,72$$

porque divide por 2!

$$R_E = \frac{25 \text{ mV}}{5,72 \text{ mA}} = 4,37 \, \Omega$$

... continuando o cálculo de R_o

$$R_o = \frac{33,91 \cdot 7,076K}{33,91 + 7,076K} \approx 33,75 \Omega = R_o$$

... Ganho

$$A_o = \frac{R_{in}}{R_{in} + 10K} \approx 0,83 = A_o$$

$$A = A_o \cdot A_1 \rightarrow \text{fator} \quad A_o = 1 + A \cdot \beta$$

$$R_{in2} [10K \parallel 24,2K + 4,37] (\beta + 1) = 4,8M$$

$$r_c = 20 \parallel R_{in2} \approx 20K \Omega$$

$$R_{bb} = 2 \cdot 34,96 \cdot 677$$

$$R_{bb} = 88K$$

$$V_o = \frac{V_{bb}}{2r_e} \cdot R_c' = \frac{V_o}{V_{bb}} = \frac{R_c'}{2r_e}$$

$$A_1 = \frac{R_c'}{2r_e} \cdot \left(\frac{R_{bb}}{R_{bb} + 2K} \right) \approx 280$$

$$R_{inF} = 192K \Omega$$

$$1 + 19,106 = 20,106$$

$$A_F = \frac{A}{\text{fator}} \approx 11,6$$

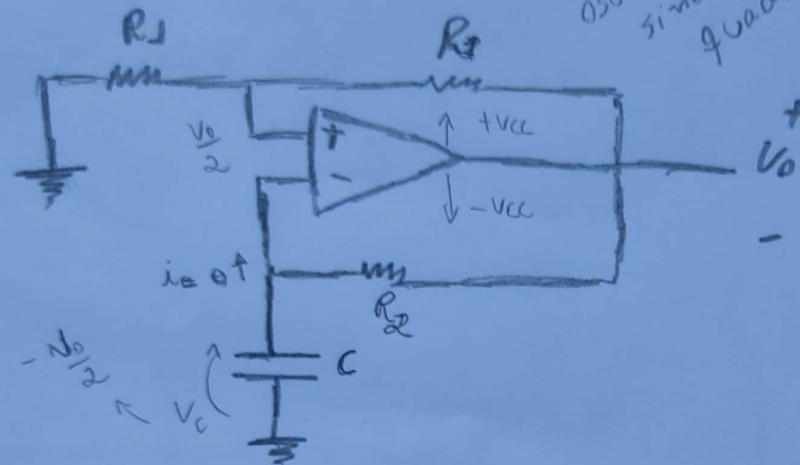
$$R_{oF} = \frac{33,75}{20,106} = 1,68 \Omega$$

①. Vantagens e Desvantagens da Alimentação Negativa

- Vantagens
- Estabilização no ganho
 - Redução da distorção não linear
 - Redução do efeito da ruído
 - ↳ Controle das impedâncias de entrada e saída
 - ↳ Extensão da largura de banda

- Desvantagens
- Diminuição do ganho
 - ↳ Tendência à oscilações

②.



Circuito de
Carga e descarga
de um capacitor

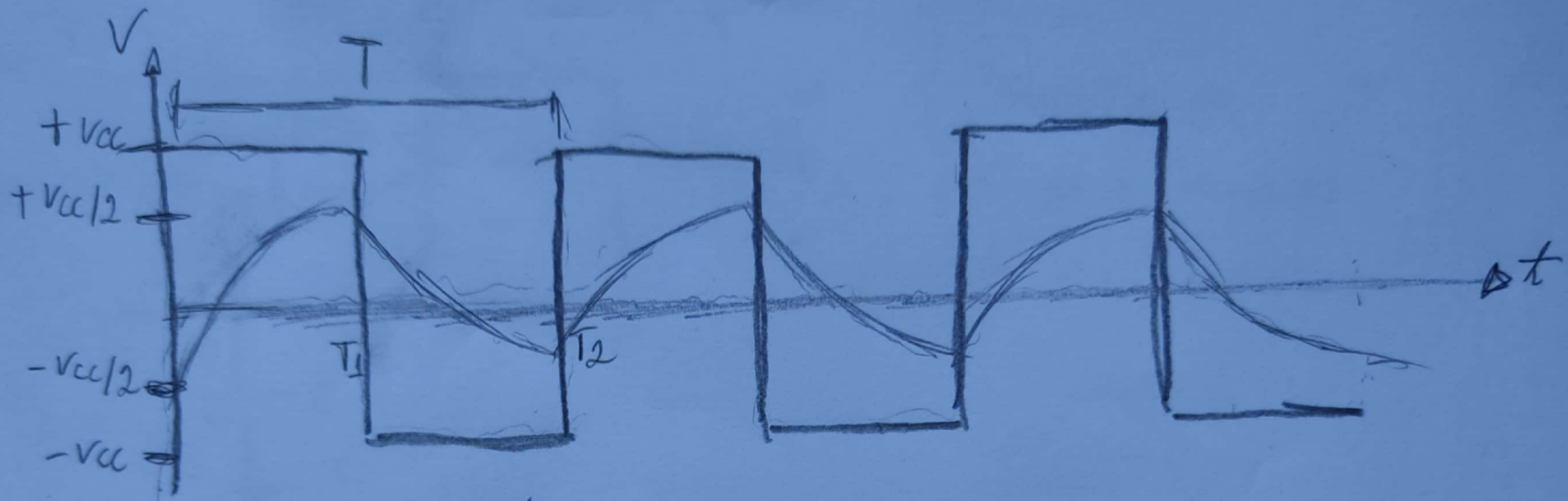
Nesse circuito, temos que saber condições iniciais, necessárias.

Ex: $V_0 = V_{CC}$

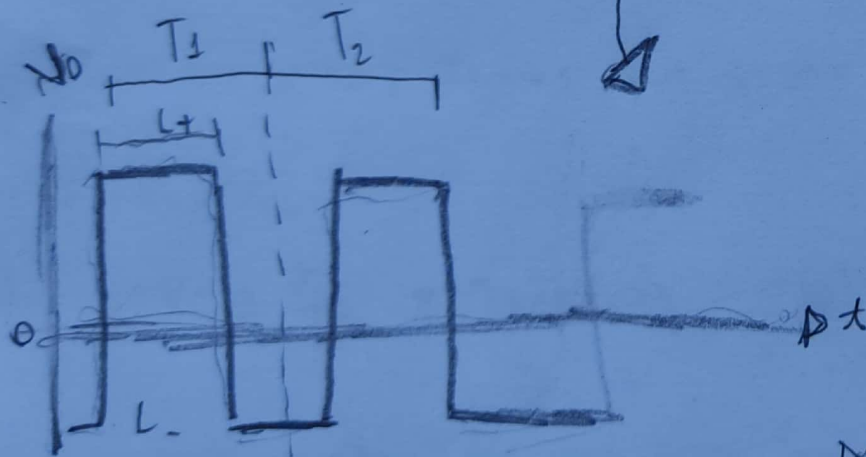
$V_C = \frac{V_{CC}}{2}$ a tensão do capacitor

O funcionamento de um capacitor se dá, com um circuito RC alimentado por uma tensão de saída, até que a tensão no capacitor (V_-) seja maior $V_+ = V_{CC}/2$, $V_0 = [V(+) - V(-)] \cdot A = -V_{CC}$, quando isso acontece, o capacitor começa a descarregar, até que V_0 se torne igual a V_{CC} [$V_0 = V_{CC}$], e o processo se repete.

2 ... continuando a questões 2.



#



$\phi_{LT} = V_{TH}$

ϕ_{LT}

$L+$

Para T_1

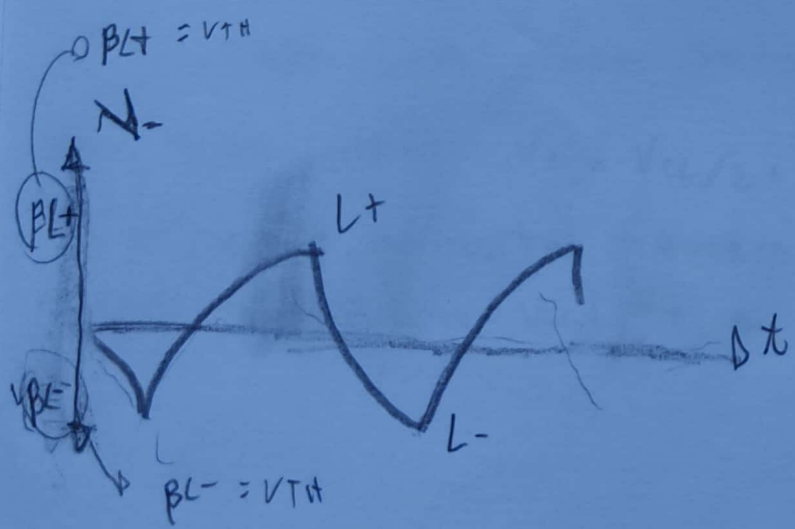
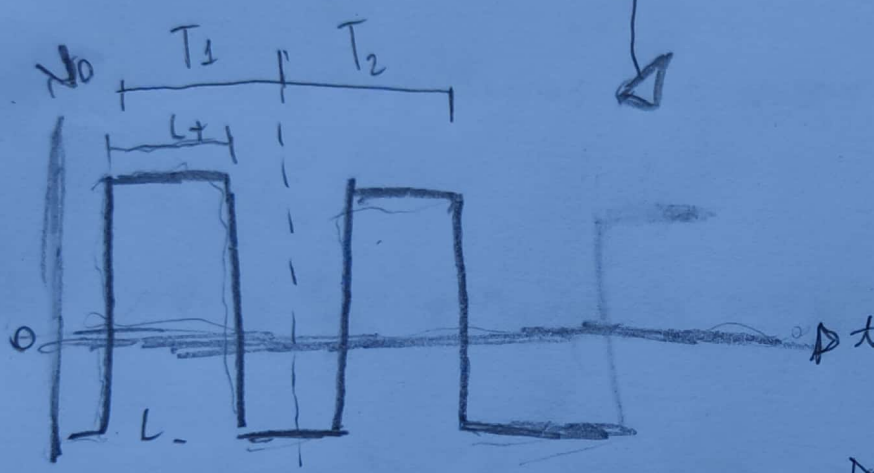
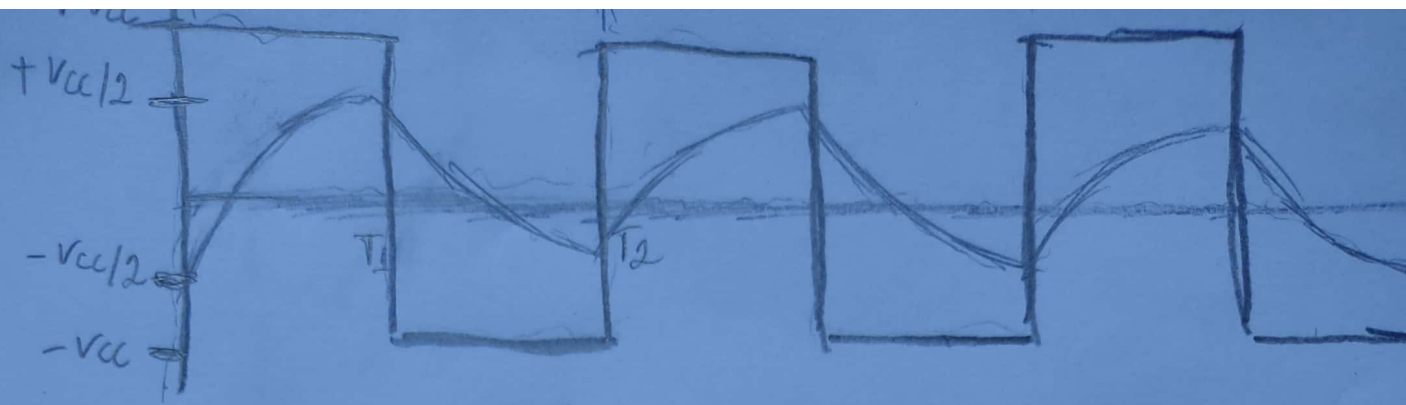
$$V_{C_1}(t) = V_F + (V_C - V_F) \cdot e^{-t/RC}$$

$$\Rightarrow V_0 + \left(\frac{V_0}{2} - V_0 \right) \cdot e^{-t/RC}$$

$$\Rightarrow \frac{V_0}{2} = V_0 - \frac{3}{2} V_0 \cdot e^{-t_1/RC}$$

$$\Rightarrow \frac{1}{2} = \frac{3}{2} \cdot e^{-t_1/RC}$$

$$\Rightarrow e^{-t_1/RC} = 1/3$$



Para T_1

$$V_{C_1}(t) = V_f + (V_c - V_f) \cdot e^{-t/RC}$$

$$\Rightarrow V_0 + \left(\frac{V_0}{2} - V_f \right) \cdot e^{-t/RC}$$

$$\Rightarrow \frac{V_0}{2} = V_f - \frac{V_0}{2} \cdot e^{-T_1/RC}$$

$$\Rightarrow \frac{1}{2} = \frac{1}{2} \cdot e^{-T_1/RC}$$

$$\Rightarrow e^{-T_1/RC} = 1$$

$$T_1 = -\ln(1)$$

$$T_1 = RC \ln(3)$$

Para T_2

→ Utilizar Lógica

$$\therefore V_{C_2}(t) = -V_0 + \left(\frac{V_0}{2} - (-V_0) \right) \cdot e^{-t/RC}$$

$$T_2 = RC \ln(3)$$

Para T_1

$$V_{C_1}(t) = V_f + (V_c - V_f) \cdot e^{-t/RC}$$

$$\Rightarrow V_0 + \left(\frac{V_0}{2} - V_0 \right) \cdot e^{-t/RC}$$

$$\Rightarrow \frac{V_0}{2} = V_0 - \frac{3}{2} V_0 \cdot e^{-t_1/RC}$$

$$\Rightarrow -\frac{1}{2} = -\frac{3}{2} \cdot e^{-t_1/RC}$$

$$\Rightarrow e^{-t_1/RC} = 1/3$$

$$T_1 = -\ln\left(\frac{1}{3}\right) \cdot RC$$

$$T_1 = RC \ln(3)$$

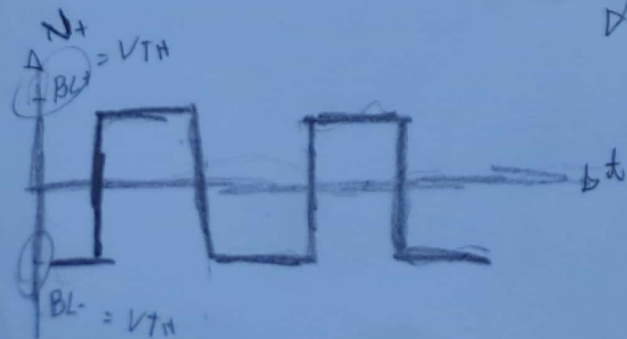
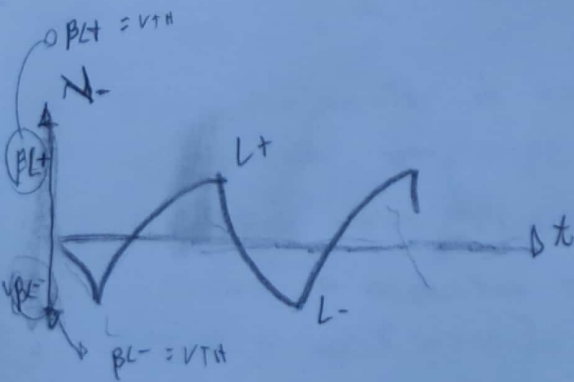
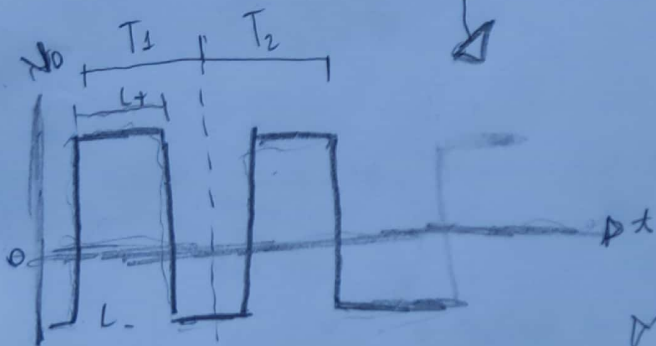
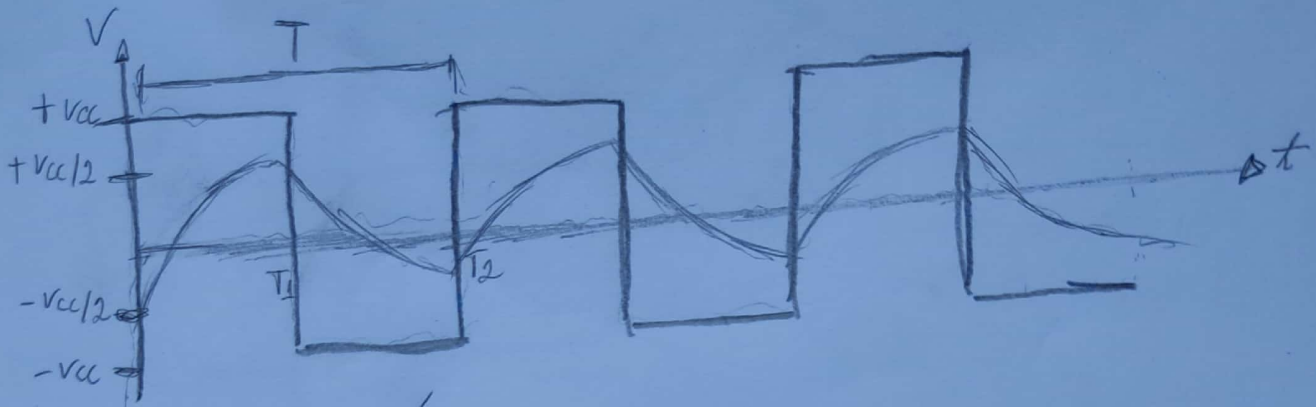
Para T_2

→ utilizando a mesma lógica de T_1

$$\therefore V_{C_2}(t) = -V_0 + \left(\frac{V_0}{2} + V_0 \right) \cdot e^{-t/RC}$$

$$T_2 = RC \ln(3)$$

... continuando a questão 2.



Para T_1

$$V_{C_1}(t) = V_f + (V_c - V_f) \cdot e^{-t/RC}$$

$$\Rightarrow V_0 + \left(\frac{V_0}{2} - V_0 \right) \cdot e^{-t/RC}$$

$$\Rightarrow \frac{V_0}{2} = V_0 - \frac{3}{2} V_0 \cdot e^{-t_1/RC}$$

$$\Rightarrow \frac{1}{2} = \frac{3}{2} \cdot e^{-t_1/RC}$$

$$\Rightarrow e^{-t_1/RC} = 1/3$$

$$T_1 = -\ln\left(\frac{1}{3}\right) \cdot RC$$

$$T_1 = RC \ln(3)$$

Para T_2

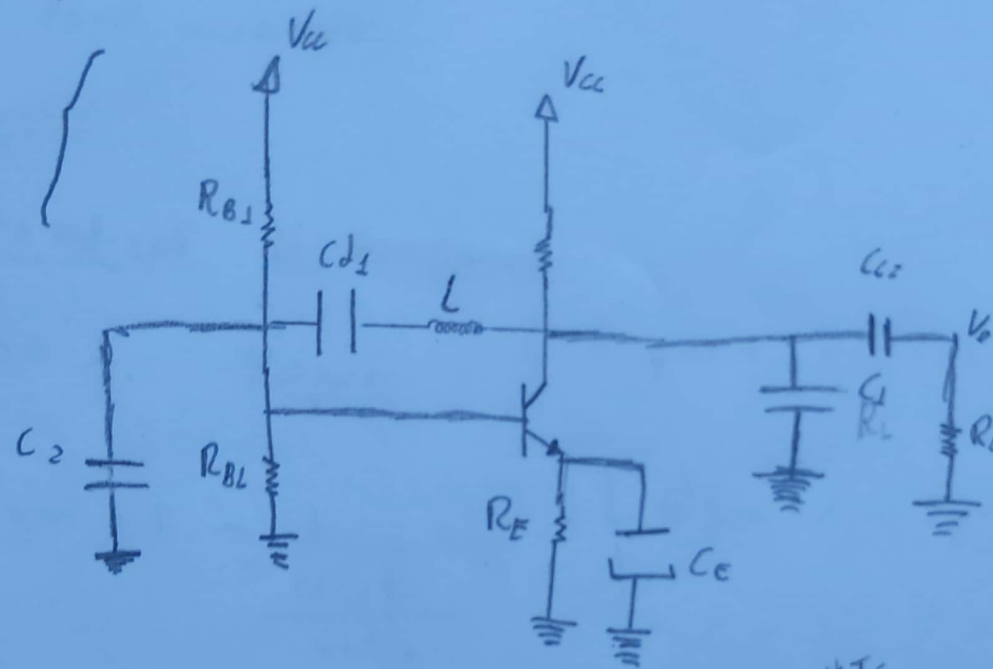
→ utilizando a mesma lógica de T_1

$$\therefore V_{C_2}(t) = -V_0 + \left(\frac{V_0}{2} + V_0 \right) \cdot e^{-t/RC}$$

$$T_2 = RC \ln(3)$$

③

Desenhando Novamente



Dados

$$R_{B1} = 498 \text{ K}\Omega$$

$$R_{B2} = 93 \text{ K}\Omega$$

$$R_E = 1 \text{ K}\Omega$$

$$R_C = 4,7 \text{ K}\Omega$$

$$R_L = 4,7 \text{ K}\Omega$$

$$\left. \begin{array}{l} C_{d1} \\ C_{c1} \\ C_{c2} \\ C_e \end{array} \right\} = 10 \mu\text{F}$$

$$\left\{ \begin{array}{l} C_1 = 15 \text{ nF} \\ C_2 = 455 \text{ nF} \end{array} \right.$$

$$L_1 = 10 \mu\text{H}$$

$$\beta = 676$$

$$V_B = 1,44 \text{ V}$$

$$V_E = 0,792 \text{ V}$$

$$V_C = 6,285$$

$$g_m = \frac{I_C}{V_T} ; I_C = \frac{V_{CC} - V_C}{R_C}$$

$$I_C = \frac{10 - 6,285}{4700} = 0,8 \text{ mA}$$

$$g_m = \frac{0,8 \text{ mA}}{2,5 \text{ mV}} \approx 0,316 \text{ A/V}$$

$\frac{KT}{q}$

$$V_{CC} = 10 \mu\text{V}$$

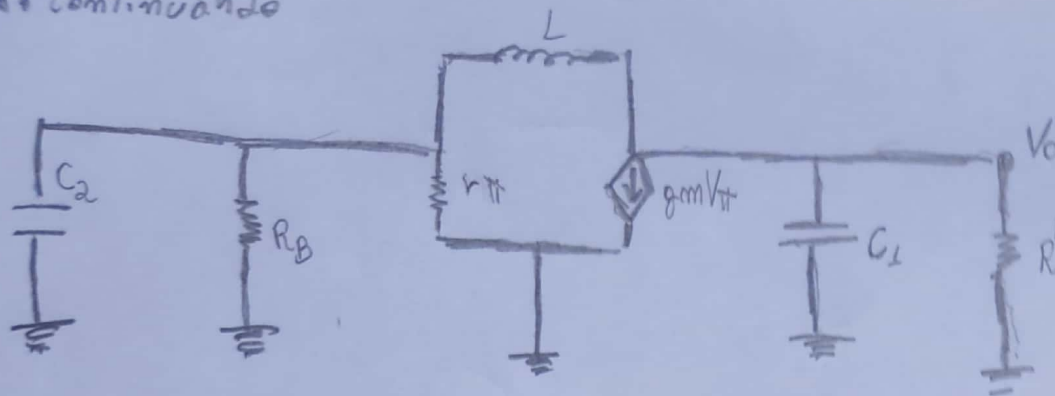
$$V_T \approx 2,5 \text{ mV}$$

$$V_C = 6,285 \text{ V}$$

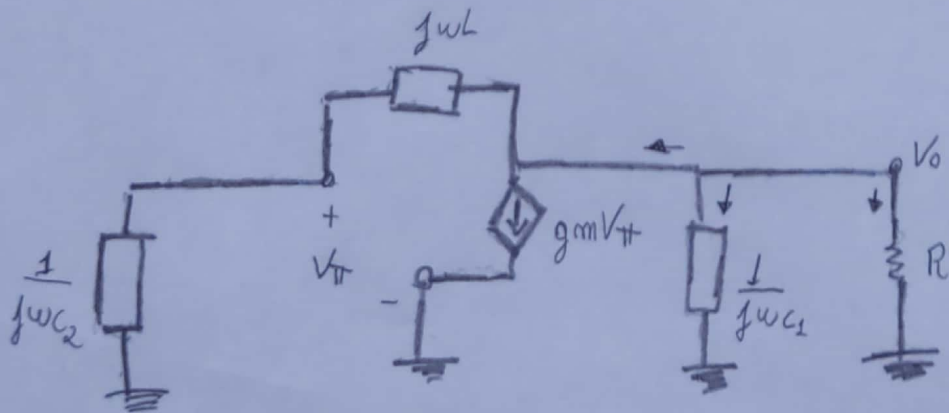
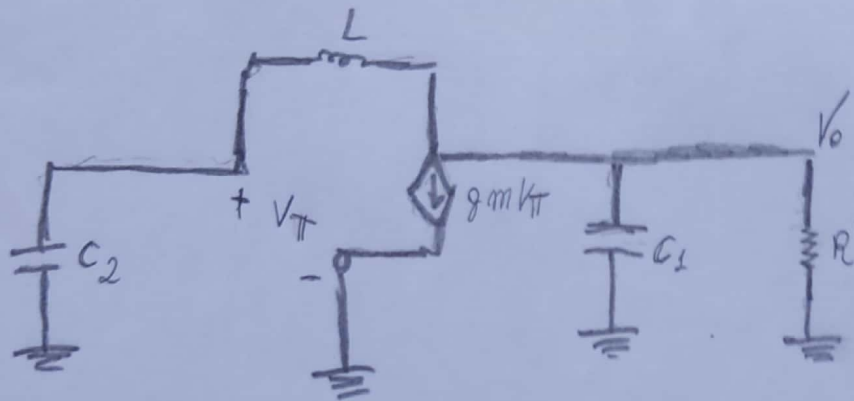
$$I_C = 0,8 \text{ mA}$$

$$g_m = 0,316 \text{ V}$$

... Continuando



Impedância de $C_2 \ll V_{TT} // R_B$



$$\left. \begin{aligned} V_{TT} j\omega C_2 + g_m V_{TT} + V_O \left(j\omega C_1 + \frac{1}{R'} \right) &= 0 \\ V_O &= V_{TT} + V_{TT} j\omega C_2 \\ \omega L &= V_{TT} (1 - \omega^2 L C_2) \end{aligned} \right\}$$

$$\therefore V_{TT} j\omega C_2 + g_m V_{TT} + V_{TT} (1 - \omega^2 L C_2) \cdot \left(j\omega C_1 + \frac{1}{R'} \right) = 0$$

$$\rightarrow \left(g_m + \frac{1 - \omega^2 L C_2}{R'} \right) + j \left(\omega (C_1 + C_2) - \omega^3 L C_1 C_2 \right)$$

... continuando

condição de Oscilação

$$g_m + \frac{1}{R'} - \frac{\omega^2 C_2 L}{R'} = 0 \quad \left. \vphantom{g_m + \frac{1}{R'} - \frac{\omega^2 C_2 L}{R'} = 0} \right\} R' g_m + 1 - \omega^2 C_2 L = 0$$

$$\rightarrow R' g_m + 1 - \frac{1}{\frac{C_1 C_2 L}{C_1 + C_2}} = 0 \quad \left. \vphantom{R' g_m + 1 - \frac{1}{\frac{C_1 C_2 L}{C_1 + C_2}} = 0} \right\} R' g_m + 1 - \frac{C_1 + C_2}{C_1} = 0$$

$$\rightarrow R' g_m - \frac{C_2}{C_1} = 0 \quad \left. \vphantom{R' g_m - \frac{C_2}{C_1} = 0} \right\} \frac{C_2}{C_1} = R' g_m \quad \rightarrow \boxed{g_m R' \gg \frac{C_2}{C_1}}$$

$$\rightarrow g_m R' = 73,3 \quad \text{e} \quad \frac{C_2}{C_1} = 29,667, \quad \text{Logo} \rightarrow \boxed{\text{OSCI LA}}$$

Frequência da Oscilação

$$j\omega C_2 + g_m + (1 - \omega^2 C_2 L) \cdot \left(j\omega C + \frac{1}{R'} \right) = 0$$

$$\rightarrow j\omega C_2 + g_m + j\omega C + \frac{1}{R'} - j\omega^3 C_1 C_2 L - \frac{\omega^2 C_2 L}{R'} = 0$$

$$\left. \begin{aligned} \omega_0 (C_1 + C_2) - \omega_0^3 C_1 C_2 L &= 0 \end{aligned} \right\} \omega_0^2 = \frac{C_1 + C_2}{C_1 C_2 L} = \frac{1}{\frac{C_1 C_2 L}{C_1 + C_2}} = \frac{1}{C_{eq} L}$$

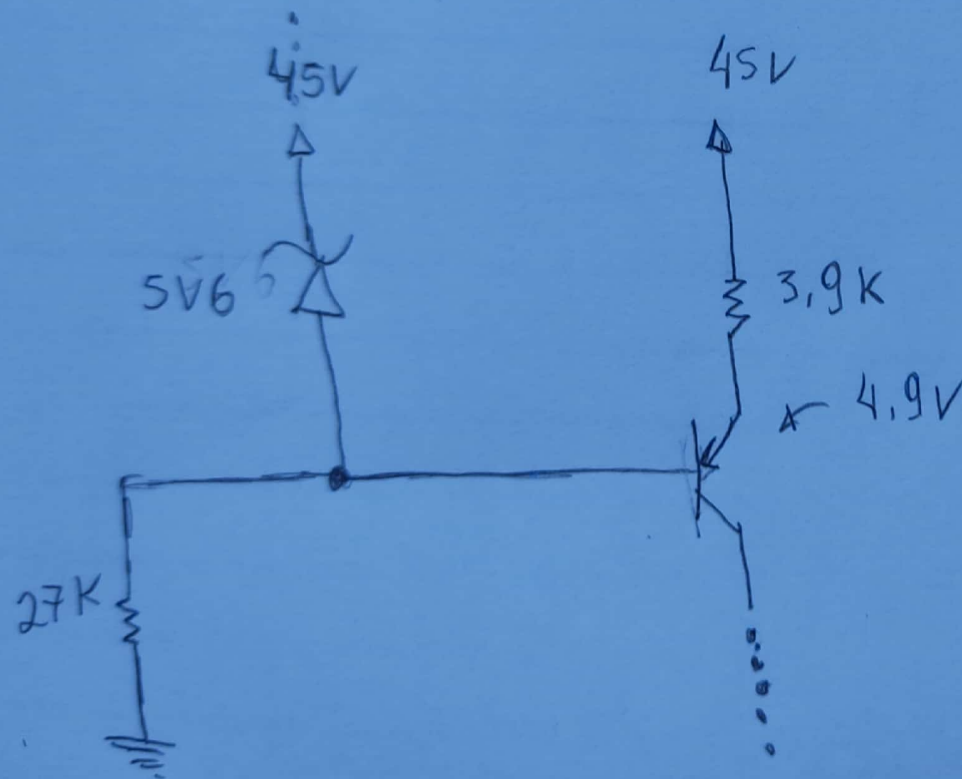
$$\omega_0^2 = \frac{1}{C_{eq} L} \quad \rightarrow \quad \boxed{\omega_0 = \frac{1}{\sqrt{C_{eq} L}}}$$

$$\rightarrow \omega_0 = 2,6 \text{ Mrad/s}$$

$$\boxed{\omega_0 = 417,8 \text{ KHz}}$$

④.

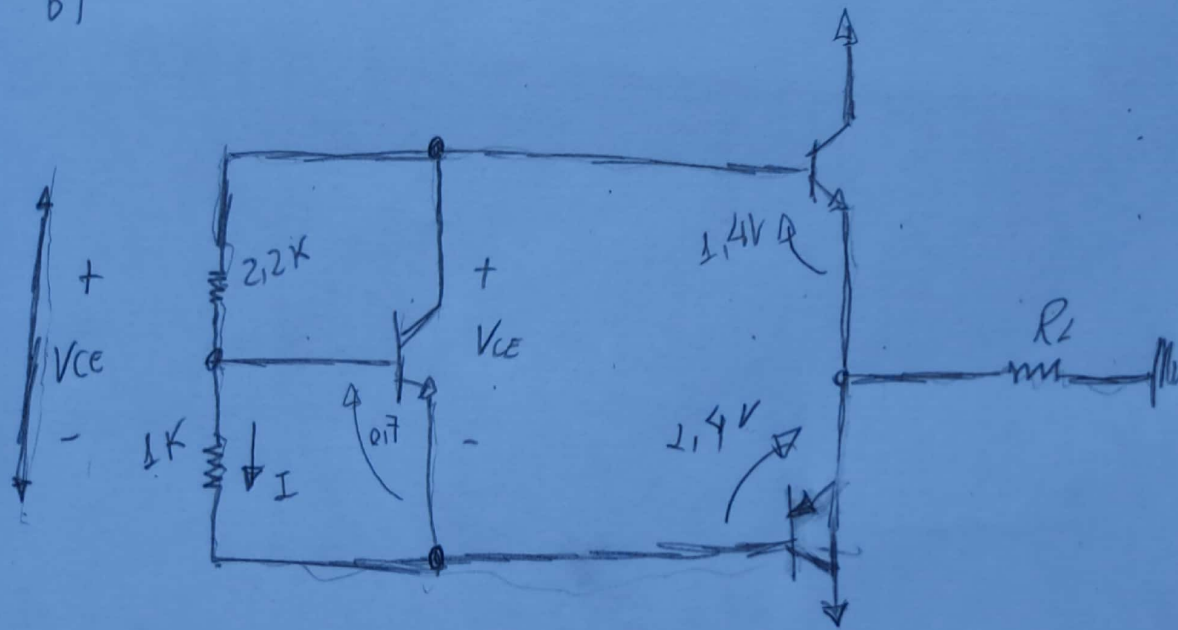
a) ... coletor 2



$$I_c = \frac{(5,6 - 0,7)}{3900} = 1,26 \text{ mA}$$

$$I_c = 1,26 \text{ mA}$$

b)



$$I = \frac{0,7}{1000} = 0,7 \text{ mA}$$

$$V_{ce} = 3 \cdot 200 \cdot 0,7 \cdot 10^{-3}$$

$$V_{ce} = 2,24 \text{ V}$$

O Transistor "T14" realiza a função de multiplicador de Tensão, fornecendo uma tensão fixa afim de pré-polarizar os Transistores de potência.