

Circuit Theory and Electronics Fundamentals

Lecture 15: The Bipolar Junction Transistor

- The transistor
- The Bipolar Junction Transistor (BJT):
 - Operation regions
 - Large signal model (DC)
 - Small signal model (AC)
 - Spice model

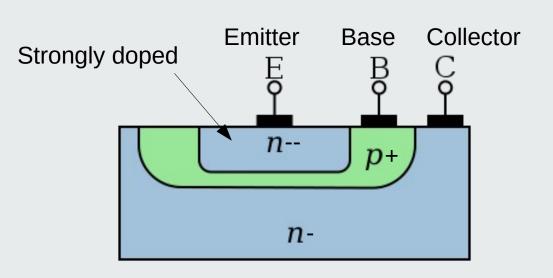


The transistor

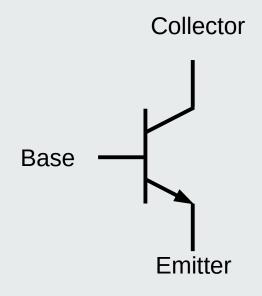
- The transistor is the most widespread semiconductor device
- The transistor is used in both analogue and digital electronic circuits
 - In analogue circuits it is mostly used as a controlled voltage or current amplifier
 - In digital circuits it is used as controlled switch useful for logic operations
- The main transistor types are
 - Bipolar Junction Transistors (BJT)
 - Field effect transistors
- In this lecture we focus on BJTs



- The diode is a <u>pn junction</u>
- The transistor is a <u>npn</u> or <u>pnp junction</u>
- Let's focus on <u>npn</u> first



npn junction cross section

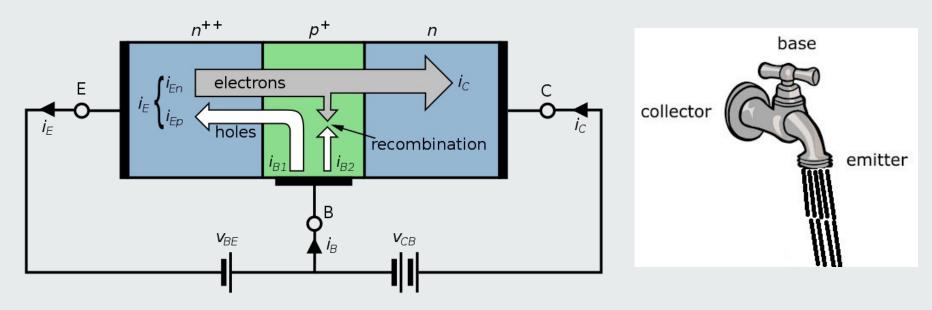


npn transistor circuit symbol



How the BJT works

- The npn BJT is not two back to back diodes!
- The the two pn junctions interact with each other!



Base-emitter junction forwardly biased Bias voltage is $V_{\rm BE}$

Base-collector junction reversely biased Bias voltage is V_{CB}

$$I_E = I_B + I_C(KCL)$$

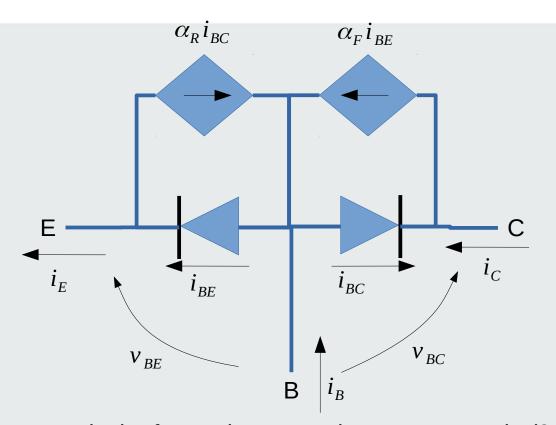


The BJT operation regions

- The bipolar is a highly non-linear device with four different operation regions:
 - 1) Forward active region: base-emitter junction forwardly biased and base-collector junction reversely biased. In this region $I_C = \beta_F I_B$
 - 2) Reverse active region: base-emitter junction reversely biased and base-collector junction forwardly biased. In this region $-I_E = \beta_R I_B$
 - 3) Saturation region: both base-emitter and base-collector junctions forwardly biased
 - 4) Cut-off region: both base-emitter and base-collector junctions reversely biased



The Ebers-Moll bipolar npn transistor model



$$i_{B} = \left[\frac{I_{S}}{\beta_{F}} \left(e^{\frac{v_{BE}}{V_{T}}} - 1 \right) + \frac{I_{S}}{\beta_{R}} \left(e^{\frac{v_{BC}}{V_{T}}} - 1 \right) \right]$$

$$i_{C} = \left[I_{S} \left(e^{\frac{v_{BE}}{V_{T}}} - e^{\frac{v_{BC}}{V_{T}}} \right) - \frac{I_{S}}{\beta_{R}} \left(e^{\frac{v_{BC}}{V_{T}}} - 1 \right) \right]$$

$$i_{E} = \left[I_{S} \left(e^{\frac{v_{BE}}{V_{T}}} - e^{\frac{v_{BC}}{V_{T}}} \right) + \frac{I_{S}}{\beta_{F}} \left(e^{\frac{v_{BE}}{V_{T}}} - 1 \right) \right]$$

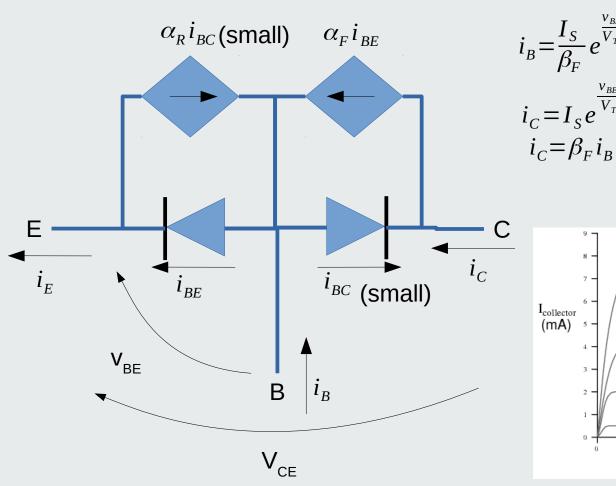
$$\beta_F = \frac{\alpha_F}{1 - \alpha_F}$$

 $\alpha_{_{\! R}}$ is the forward common base current gain (0.98 to 0.998) $\alpha_{_{\! R}}$ is the reverse common base current gain (0 to 0.952) $\beta_{_{\! R}}$ is the forward common emitter current gain (20 to 500) $\beta_{_{\! R}}$ is the reverse common emitter current gain (0 to 20)

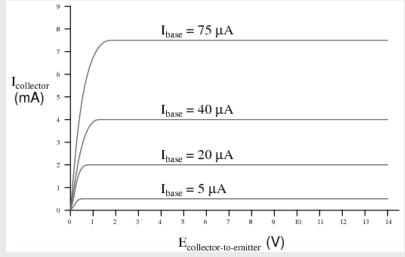
$$\beta_{R} = \frac{\alpha_{R}}{1 - \alpha_{R}}$$



Forward active region (Ebbers-Moll model)

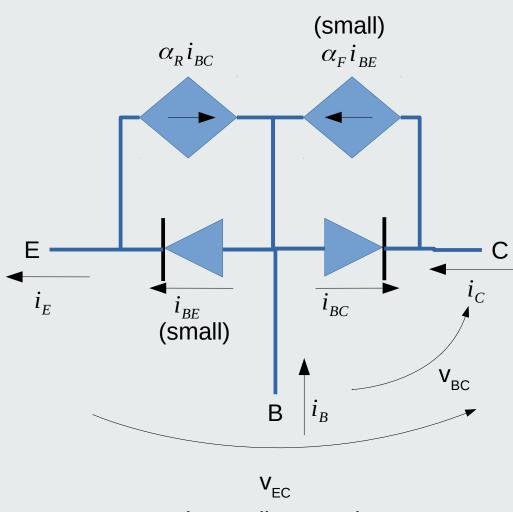


Use Ebbers-Moll equations and neglect the reverse biased terms





Reverse active region (Ebbers-Moll model)



$$i_{B} = \frac{I_{S}}{\beta_{R}} e^{\frac{V_{BC}}{V_{T}}}$$

$$-i_{E} = I_{S} e^{\frac{V_{BC}}{V_{T}}}$$

$$-i_{E} = \beta_{R} i_{B}$$

Emitter-collector voltage



Bipolar transistor large signal Ebbers-Moll model (DC)

Forward active region

$$i_C = \alpha_F i_E$$
$$i_C = \beta_F i_B$$

$$\begin{aligned} \text{KCL:} \quad & i_E \!=\! i_B \!+\! i_C \!=\! (1 \!+\! \beta_F) i_B \!=\! (1 \!+\! \frac{1}{\beta_F}) i_C \\ & \alpha_F \!=\! \frac{\beta_F}{1 \!+\! \beta_F} \! \Leftrightarrow \! \beta_F \!=\! \frac{\alpha_F}{1 \!-\! \alpha_F} \end{aligned}$$

Saturation region $i_C < \beta_E i_B$

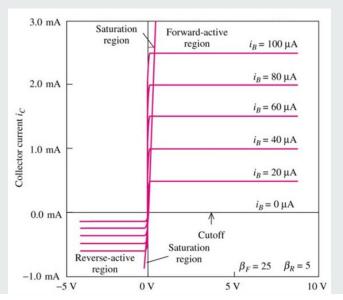
Cut-off region $i_C = i_B = i_E = 0$

Reverse active region

$$i_{E} = \alpha_{R} i_{C}$$

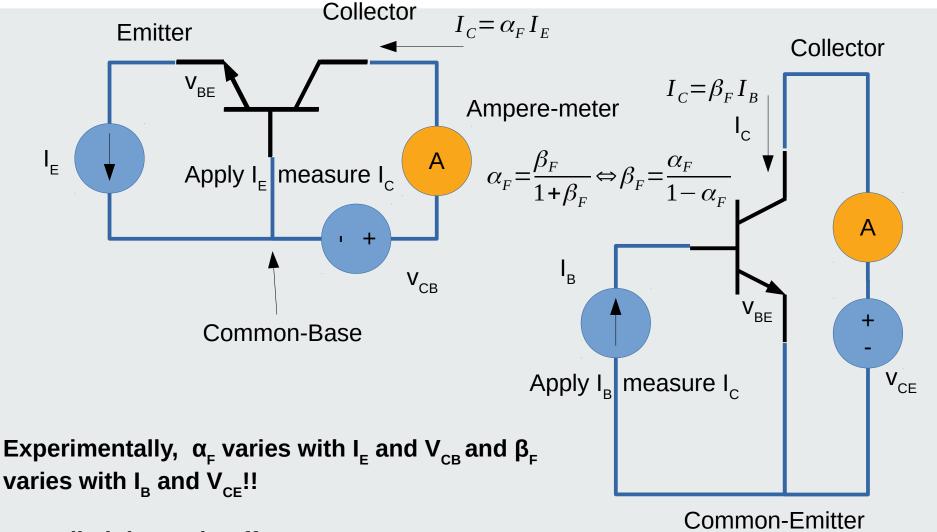
$$-i_{E} = \beta_{R} i_{B}$$
KCL:
$$-i_{C} = i_{B} - i_{E}$$

$$\alpha_{R} = \frac{\beta_{R}}{1 + \beta_{R}} \Leftrightarrow \beta_{R} = \frac{\alpha_{R}}{1 - \alpha_{R}}$$





Experimental setup to measure α_F or β_F



It's called the Early Effect!

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The Early Effect: base width modulation

$$i_C = I_S e^{\frac{V_{BE}}{V_T}} \left(1 + \frac{V_{CE}}{V_A} \right)$$

Ebbers-Moll with Early correction

$$i_C' = I_S e^{\frac{V_{BB}}{V_T}}$$

Ebbers-Moll given

$$i_C = i_C' \left(1 + \frac{V_{CE}}{V_A} \right)$$

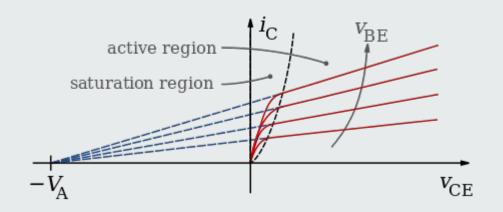
corrected

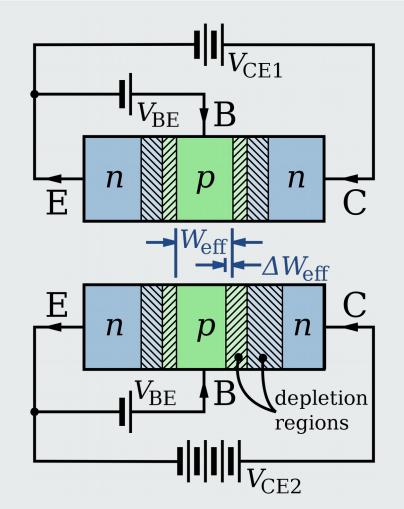
$$i_{C}' = I_{S}e^{\frac{V_{BE}}{V_{T}}}$$

$$i_{C} = i_{C}' \left(1 + \frac{V_{CE}}{V_{A}}\right)$$

$$\beta_{F} = \beta_{F0} \left(1 + \frac{V_{CB}}{V_{A}}\right)$$

corrected

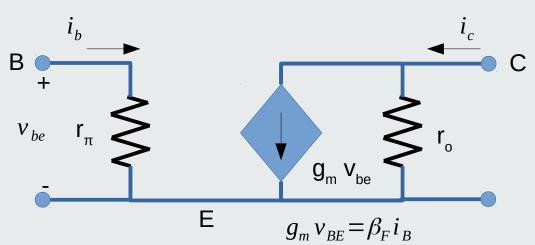






Bipolar transistor small signal model (AC)

Forward active region ONLY



$$g_{m} = \frac{\partial i_{C}}{\partial v_{BE}} = \frac{i_{C}}{V_{T}} \approx \frac{i_{C}'}{V_{T}}$$

$$r_{\pi} = \frac{\partial v_{BE}}{\partial i_{B}} = \frac{\beta_{F}}{g_{m}}$$

$$r_{o} = \frac{\partial v_{CE}}{\partial i_{C}} = \frac{V_{A} + V_{CE}}{i_{C}} \approx \frac{V_{A}}{i_{C}'}$$

Incremental parameters

 g_m Transconductance

 r_{π} Input (incremental) impedance

 r_o Output (incremental) impedance

 $I_{n} = 150 \, \mu A$ $I_{n} = 130 \, \mu A$ $I_{n} = 10 \, \mu A$ $I_{n} = 10 \, \mu A$ $I_{n} = 90 \, \mu A$ $I_{n} = 50 \, \mu A$ $I_{n} = 30 \, \mu A$ $I_{n} = 10 \, \mu A$ $V_{CE} (V)$

Computed using operating point



Bipolar pnp transistor DC model

Switch the direction of the currents and use the same equations!

Forward active region

$$\begin{split} &i_{C} = \alpha_{F} i_{E} \\ &i_{C} = \beta_{F} i_{B} \\ &i_{E} = i_{B} + i_{C} = (1 + \beta_{F}) i_{B} = (1 + \frac{1}{\beta_{F}}) i_{C} \\ &\alpha_{F} = \frac{\beta_{F}}{1 + \beta_{F}} \Leftrightarrow \beta_{F} = \frac{\alpha_{F}}{1 - \alpha_{F}} \end{split}$$

Saturation region $i_C < \beta_E i_B$

Cut-off region
$$i_C = i_B = i_E = 0$$

AC model

$$g_{m} = \frac{\partial i_{C}}{\partial v_{BE}} = \frac{i_{C}}{V_{T}}$$

$$r_{\pi} = \frac{\partial v_{BE}}{\partial i_{B}} = \frac{\beta_{F}}{g_{m}}$$

$$r_{o} = \frac{\partial v_{CE}}{\partial i_{C}} \approx \frac{V_{A}}{I_{C}}$$

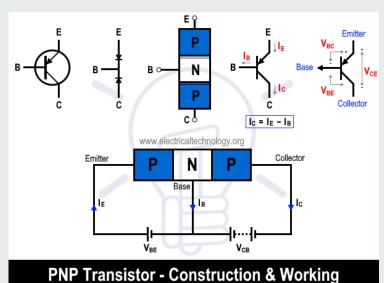
Reverse active region

$$i_{E} = \alpha_{R} i_{C}$$

$$-i_{E} = \beta_{R} i_{B}$$

$$-i_{C} = i_{B} - i_{E}$$

$$\alpha_{R} = \frac{\beta_{R}}{1 + \beta_{R}} \Leftrightarrow \beta_{R} = \frac{\alpha_{R}}{1 - \alpha_{R}}$$





Spice's BJT model

- Ngspice has very sophisticated BJT models
- Let's consider a reduced parameter set one
 - **BF**: Forward active current gain
 - VJE Base-emitter built-in potential
 - **BR**: Reverse active current gain
 - VJC Base-collector built-in potential
 - **IS**: Transport saturation current
 - VAF Forward mode Early voltage
 - CJE: Base-emitter zero-bias junction capacitance
 - VAR Reverse mode Early voltage
 - NF Forward mode ideality factor
 - NR Reverse mode ideality factor
 - CJC Base-collector zero-bias junction capacitance
- Check out l15.net: simulates $I_C(v_{CE})$ for constant v_{BE} using DC analysis (DC sweep) for a real commercial discret transistor: the BC547



Conclusion

- The transistor
- The Bipolar Junction Transistor:
- Operation regions
- Large signal model (DC)
- Small signal model (AC)
- Spice model