# 06\_03\_Passive FTC: Recent Advances in Adaptive Methods

1. Structure of Passive Fault Tolerant Control

Motivation

Example Methods

2. An Example of PFTC Method: Subspace Predictive FTC

Introduction

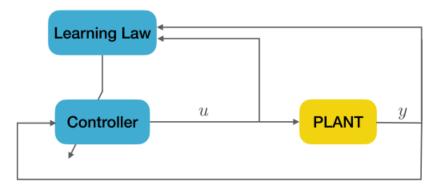
Process

Properties

Summary

## 1. Structure of Passive Fault Tolerant Control

## A Passive, Adaptive FTC block diagram



#### **Motivation**

- Active Approaches need an explicit FDI decision and fault estimation (and of course model)
  - Fault Classes for FDI needs to be designed by experts
  - FDI adds computational complexity'
- Passive adaptive approaches do not need FDI (do not need model or automatically learn a model)
  - o less computational complexity
  - o work in healthy and faulty conditions, even for unknown faults
- distinction between Active and Passive is blurry

# **Example Methods**

> Examples:

> Neural Networks + Back Propagation

> Direct/Indirect Adaptive Control

> Polynomial/Wavelets etc. controllers

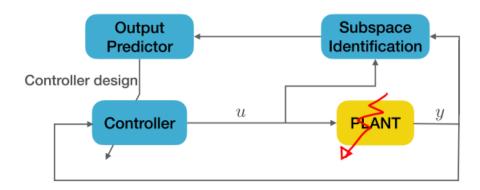
> Reinforcement Learning

> Active Inference

> Other Machine Learning approaches

> ..

# 2. An Example of PFTC Method: Subspace Predictive FTC



#### Introduction

Subspace Predictive Fault-Tolerant Control = Subspace Identification + Output Predictor + Controller design law

- subspace-identification: learning model of the plan
  - ∘ Fault Happen → learn faulty model
- model-free, data-driven approach

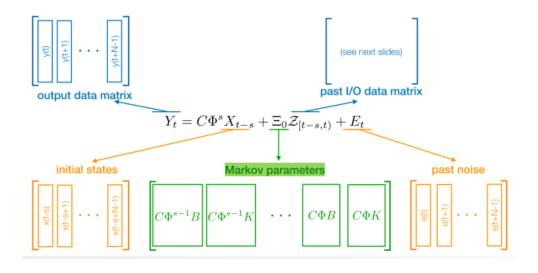
#### **Process**

The following discrete-time state space model in innovation form is assumed (but not known!)

$$\begin{cases} x(k+1) = Ax(k) + Bu(k) + Ke(k) \\ y(k) = Cx(k) + \underline{e(k)} \end{cases}$$
 zero-mean white noise (innovation)

**Assumption:**  $\Phi = A - KC$  is stable and the system is minimal

Step 1: online recursive subspace identification

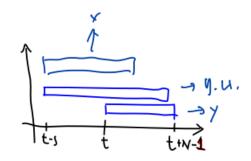


$$\mathcal{Z}_{[t-s,t)} = \begin{bmatrix} u(t-s) & u(t-s+1) & \cdots & u(t-s+N-1) \\ y(t-s) & y(t-s+1) & \cdots & y(t-s+N-1) \\ u(t-s+1) & u(t-s+2) & \cdots & u(t-s+N) \\ y(t-s+1) & y(t-s+2) & \cdots & y(t-s+N) \\ \vdots & \vdots & & \vdots \\ u(t-1) & u(t) & \cdots & u(t+N-2) \\ y(t-1) & y(t) & \cdots & y(t+N-2) \end{bmatrix}$$

- Initial states is unknown
- · Past noise is unknown

Then we can estimate the system parameter by estimate the <u>Markov Parameters</u>(by solving a least square question)

$$\hat{\Xi}_0 = Y_t \cdot \mathcal{Z}^+_{[t-s,t)}$$



#### Illustration

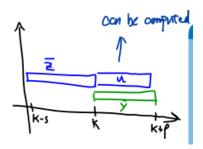
$$\left\{egin{array}{l} x(k+1) = Ax(k) + Bu(k) \ y(k) = Cx(k) \end{array}
ight. \ y(k) = C\left[A^{k-k_0}x\left(k_0
ight) + \sum_{h=k_0}^{k-1}A^{k-1-h}Bu(h)
ight] 
ight.$$

#### **Step 2: Output Predictor**

• Future Outputs can be approximated by:

$$\hat{\mathbf{y}}_{[k,k+f)} \approx \begin{bmatrix} \Xi_0 \\ \Xi_1 \\ \vdots \\ \Xi_{f-1} \end{bmatrix} \bar{Z}_{[k-s,k)} + \begin{bmatrix} 0 \\ \Psi_1 & 0 \\ \vdots & \vdots & \ddots \\ \Psi_{f-1} & \Psi_{f-2} & \cdots & \Psi_1 & 0 \end{bmatrix} \cdot \begin{bmatrix} u(k) \\ y(k) \\ \vdots \\ u(k+f-1) \\ y(k+f-1) \end{bmatrix} + \underbrace{\begin{bmatrix} C\Phi^s x(k-s) \\ C\Phi^{s+1} x(k-s) \\ \vdots \\ C\Phi^{s+f-1} x(k-s) \end{bmatrix}}_{\mathbf{b}_x}$$

- However, actually we do not know the real u and y after timestamp k, but can be computed
- The process can be illustrated in the following graph:



#### **Step 3: Fault Tolerant Control Law**

- Assume a fault occurs at time  $T_0$
- The faulty dynamics get learned by the Subspace Identification (Step 1) and the predictor is updated to match them
- Then we can use a lot of method to design a controller
  - For example use MPC

Future reference 
$$\mathbf{r}_{[k+1,k+f)} = [r^T(k+1) \ r^T(k+2) \ \cdots \ r^T(k+f-1)]^T,$$
 Cost function 
$$J(k+1) = \|\mathbf{r}_{[k+1,k+f)} - \hat{\mathbf{y}}_{[k+1,k+f)}\|_Q^2 + \|\Delta\mathbf{u}_{[k,k+f-1)}\|_{R^2}^2$$
 Input change 
$$\Delta u(k+i) = u(k+i) - u(k+i-1)$$
 Weighting matrices 
$$Q, R \succ 0$$

$$S_{\Delta} = \begin{bmatrix} I_m & & & & \\ -I_m & I_m & & & & \\ & \ddots & \ddots & & \\ & & -I_m & I_m \end{bmatrix}, \quad S_{k-1} = \begin{bmatrix} 0_{m \times (s-1)(m+l)} & I_m & 0_{m \times l} \\ 0_{(f-2)m \times (s-1)(m+l)} & 0_{(f-2)m \times m} & 0_{(f-2)m \times l} \end{bmatrix}$$

$$\mathbf{u}_{[k,k+f-1)} = \arg\min_{\mathbf{u}_{[k,k+f-1)}} J(k+1),$$

$$\mathbf{u}_{(k,k+f-1)}^* = [\Lambda^T Q \Lambda + S_{\Delta}^T R S_{\Delta}]^{-1} [\Lambda^T Q (\mathbf{r}_{(k+1,k+f)} - \tilde{\Gamma} \mathbf{Z}_{(k-s,k)}) + S_{\Delta}^T R S_{k-1} \mathbf{Z}_{(k-s,k)}]$$

## **Properties**

#### **Benefits**

- no requirement for FDI
- · works both during healthy and faulty conditions
- conventional control design laws such as LQR, MPC, LPV-based, etc. can be used

#### Limitations

- there is **no guarantee** of closed-loop stability
- adapting to a fault can be slower than in active approaches
  - o abrupt faults are more difficult to accommodate quick enough

# **Summary**

- Passive Fault Tolerant Control
  - No offline model needed
  - May have delay
- Subspace Predictive FTC
  - Learn model by **subspace identification**
  - **Predict** future output and future state
  - Use control law design control input based on learnt system model and prediction (such as MPC controller)