

# Chapter 2. Continuous-Time Systems

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# Introduction

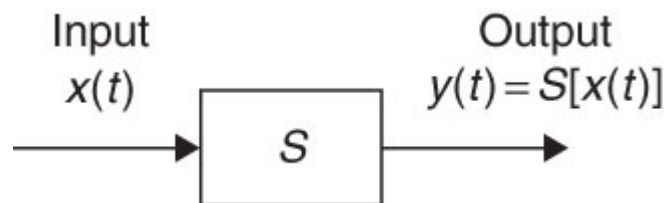
- **Systems and their classification**
  - The concept of **system** is useful in dealing with actual devices or processes for purposes of analysis and synthesis.
- **Linear time-invariant (LTI) systems**
  - We propose the **LTI model** as a mathematical idealization of the behavior of systems.
- **Convolution integral, causality, and stability**
  - **Causality**, or **non-anticipatory behavior of the system**, relates to the **cause-effect relationship** between the input and the output.

# System Concept

- We view a **system** as a mathematical transformation of an input signal into an output signal.
  - A system can be considered as a connection of subsystems.
- System classification
  - A **dynamic system** has the capability of storing energy or remembering its state, while a **static system does not**.
  - In the **lumped-parameter systems**, the behavior of spatially distributed systems is modeled as a **lumped element**.
  - A system is **passive** if it is not able to deliver energy to the outside world.
- We consider only **dynamic systems** with lumped parameters possibly changing with time, with a single input and a single output.

# LTI Continuous-Time Systems

A **continuous-time system** is a **system in which the signals at its input and output are continuous-time signals**. Mathematically, we represent it as a transformation  $S$  that converts an input signal  $x(t)$  into an output signal  **$y(t) = S[x(t)]$** .



- Characteristics of continuous-time systems
  - **Linearity, time invariance, causality, and stability**

# Linearity

A system represented by  $S$  is said to be **linear** if for inputs  $x(t)$  and  $v(t)$ , and any constants  $\alpha$  and  $\beta$ , superposition holds, i.e.,

$$\begin{aligned} S[\alpha x(t) + \beta v(t)] &= S[\alpha x(t)] + S[\beta v(t)] \\ &= \alpha S[x(t)] + \beta S[v(t)] \end{aligned}$$

– **Linearity = Additivity + Scaling**

[Ex 2.1 – **Biased Averager**] For an input  $x(t)$ , the output  $y(t)$  of such a system is given by

$$y(t) = \frac{1}{T} \int_{t-T}^t x(\tau) d\tau + B.$$

Is this system linear? If not, is there a way to make it linear? Explain.

# Example: Op-Amp

[Ex 2.2] Consider the following input-output relations that show the corresponding systems are nonlinear:

$$y(t) = |x(t)|$$

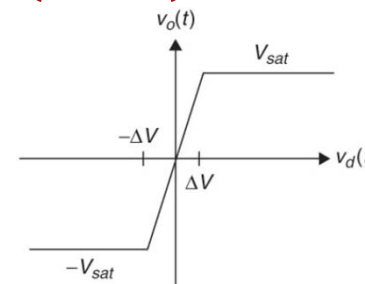
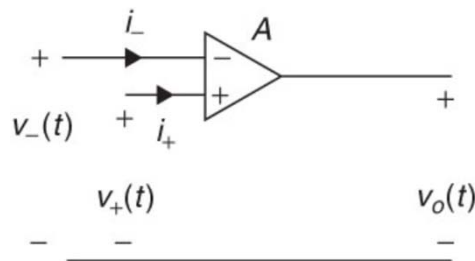
$$z(t) = \cos(x(t)) \text{ assuming } |x(t)| \leq 1$$

$$v(t) = x^2(t)$$

[Ex 2.3] Consider each of the components of an RLC circuit and determine under what conditions they are linear.

- Operational Amplifier (Op-Amp)**

- Input:  $v_-(t)$  inverting terminal and  $v_+(t)$ : noninverting terminal
- Output:  $v_o(t) = f[v_+(t) - v_-(t)] = f(v_d(t)) = Av_d(t)$



# Time Invariance

- Ideal Op-Amp
  - Assuming that  $A \rightarrow \infty$  and  $R \rightarrow \infty$

$$\begin{aligned}i_- &= i_+ = 0 \\v_d(t) &= v_+(t) - v_-(t) = 0 \\-V_{sat} &\leq v_o(t) \leq V_{sat}\end{aligned}$$

A continuous-time system  $S$  is **time invariant** if whenever for an input  $x(t)$  with a corresponding output  $S[x(t)]$ , the output corresponding to **a shifted input  $x(t \pm \tau)$**  is the **original output shifted in time  $S[x(t \pm \tau)]$** , i.e.,

$$\begin{aligned}x(t) &\Rightarrow y(t) = S[x(t)] \\x(t \pm \tau) &\Rightarrow y(t \pm \tau) = S[x(t \pm \tau)]\end{aligned}$$

The system **does not age** – its parameters are **constant!**

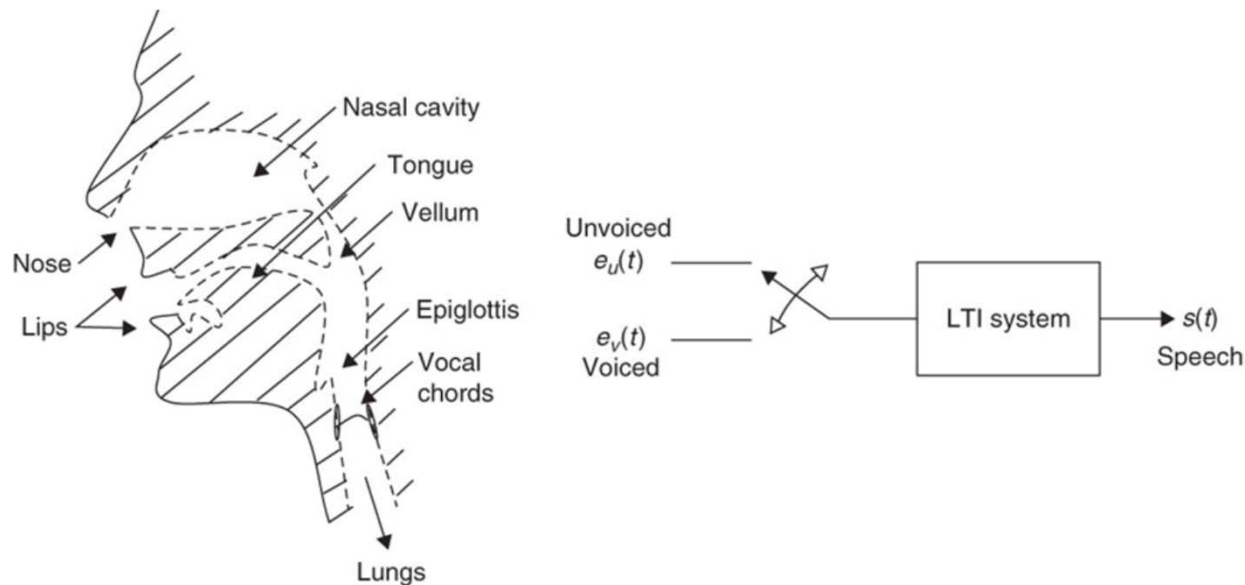


# Example: AM Communication Systems

- Remarks
  - Linearity and time invariance are **independent of each other**.
  - Most actual systems are **nonlinear** and **time varying**.
    - **Linear models** are used to approximate **around an operating point of the nonlinear behavior**, while **time-invariant models** are used to approximate in **short segments of the system's time-varying behavior**.
- A direct transmission of our voice  $m(t)$  requires **huge antenna**.
  - **AM communication systems** modulates this signal into
$$y(t) = m(t) \cos \Omega_0 t : \text{linear and time varying}$$
  - **FM communication systems** modulates this signal into
$$z(t) = \cos(\Omega_c t + \int_{-\infty}^t m(\tau) d\tau) : \text{nonlinear}$$

# Example: Vocal System

- A **typical vocal system** is modeled as a distributed system and represented by **partial differential equations**.
  - A typical LTI model for speech production considers segments of speech of about 20 msec, and **for each develops a low-order LTI system**.



# Examples

[Ex 2.4] Characterize time-varying resistors, capacitors and inductors. Assume zero initial conditions in the capacitors and inductors.

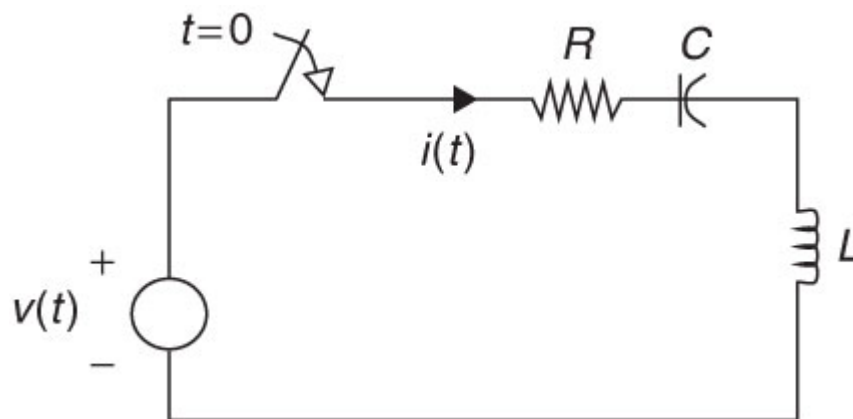
[Ex 2.5] Consider constant linear capacitors and inductors, represented by differential equations

$$\frac{dv_c(t)}{dt} = \frac{1}{C} i(t), \quad \frac{di_L(t)}{dt} = \frac{1}{L} v(t)$$

with initial conditions  $v_c(0) = 0$  and  $i_L(0) = 0$ . Under what conditions are these time-invariant systems?

## Example: RLC Circuits

- If the **initial conditions of the RLC circuit are zero**, and the **input is zero for  $t < 0$** , then the system represented by the linear differential equation with constant coefficients is **LTI**.



$$v(t) = Ri(t) + L \frac{di(t)}{dt} + \frac{1}{C} \int_0^t i(\tau) d\tau$$



$$\frac{dv(t)}{dt} = R \frac{di(t)}{dt} + L \frac{d^2 i(t)}{dt^2} + \frac{1}{C} i(t)$$

# Systems in forms of Differential Equations

Given a **dynamic system** represented by a **linear differential equations** with constant coefficients,

$$a_0 y(t) + a_1 \frac{dy(t)}{dt} + \dots + \frac{d^N y(t)}{dt^N} = b_0 x(t) + b_1 \frac{dx(t)}{dt} + \dots + b_M \frac{d^M x(t)}{dt^M}, \quad t \geq 0$$

with  **$N$  conditions**  $y(0), d^k y(t)/dt^k|_{t=0}$  for  $k = 0, \dots, N - 1$  and **input**  $x(t) = 0$  for  $t < 0$ , its **complete response**  $y(t)$  for  $t \geq 0$  has two components:

- The **zero-state response**  $y_{zs}(t)$  due exclusively to the input, because the initial conditions are zero.
- The **zero-input response**  $y_{zi}(t)$  due exclusively to the initial conditions, because the input is zero.

# Systems in forms of Differential Equations

- Most **continuous-time dynamic systems** with lumped parameters are represented by **linear ordinary differential equations with constant coefficients**.
  - Defining derivative operator as  $D^n[y(t)] = \frac{d^n y(t)}{dt^n}$   $n > 0$ , integer, the differential equation of the system is written by
 
$$(a_0 + a_1 D + \cdots + D^N)[y(t)] = (b_0 + b_1 D + \cdots + b_M D^M)[x(t)], \quad t \geq 0$$
  - This system is **LTI** if **the initial conditions as well as the input are zero for  $t < 0$**  – that is, the system is not energized for  $t < 0$ .
 
$$(a_0 + a_1 D + \cdots + D^N)[y_{zi}(t)] = 0, \quad D^k[y_{zi}(t)]_{t=0} \quad k = 0, \dots, N-1$$

$$(a_0 + a_1 D + \cdots + D^N)[y_{zs}(t)] = (b_0 + b_1 D + \cdots + b_M D^M)[x(t)]$$
  - The **characteristic polynomial** is

$$a_0 + a_1 s + \cdots + s^N = \prod_k (s - p_k)$$

eigenvalues

# Analog Mechanical Systems

[Ex 2.7] Consider a circuit that is a series of a resistor  $R = 1 \Omega$  and an inductor  $L = 1 H$ , with a voltage source  $v(t) = Bu(t)$ , and  $I_o$  amps is the initial current in the inductor. Find and solve the differential equation for  $B = 1$  and  $B = 2$  for initial condition  $I_o = 1$  and  $I_o = 0$ , respectively. Determine the zero-input and the zero-output responses. Under what conditions is the system linear and time invariant?

- Analog mechanical systems
  - The applied force  $f(t)$  equals the sum of the forces generated by the mass  $M$  and the damper  $D$

$$f(t) = M \frac{dw(t)}{dt} + Dw(t)$$

- If **the initial velocity and the external force are zero for  $t < 0$** , the above differential equation represents a **LTI mechanical system**.

# Superposition and Time Invariance

If  $S$  is the transformation corresponding to an LTI system

$$\mathbf{y}(t) = \mathbf{S}[\mathbf{x}(t)] \text{ for an input } \mathbf{x}(t)$$

Then we have

$$\begin{aligned} \mathbf{S}\left[\sum_k A_k \mathbf{x}(t - \tau_k)\right] &= \sum_k A_k \mathbf{S}[\mathbf{x}(t - \tau_k)] = \sum_k A_k \mathbf{y}(t - \tau_k) \\ \mathbf{S}\left[\int g(\tau) \mathbf{x}(t - \tau) d\tau\right] &= \int g(\tau) \mathbf{S}[\mathbf{x}(t - \tau)] d\tau = \int g(\tau) \mathbf{y}(t - \tau) d\tau \end{aligned}$$

This property allows us to **find the response of an LTI system due to any signal**, if we know the response of the system to an **impulse signal**.



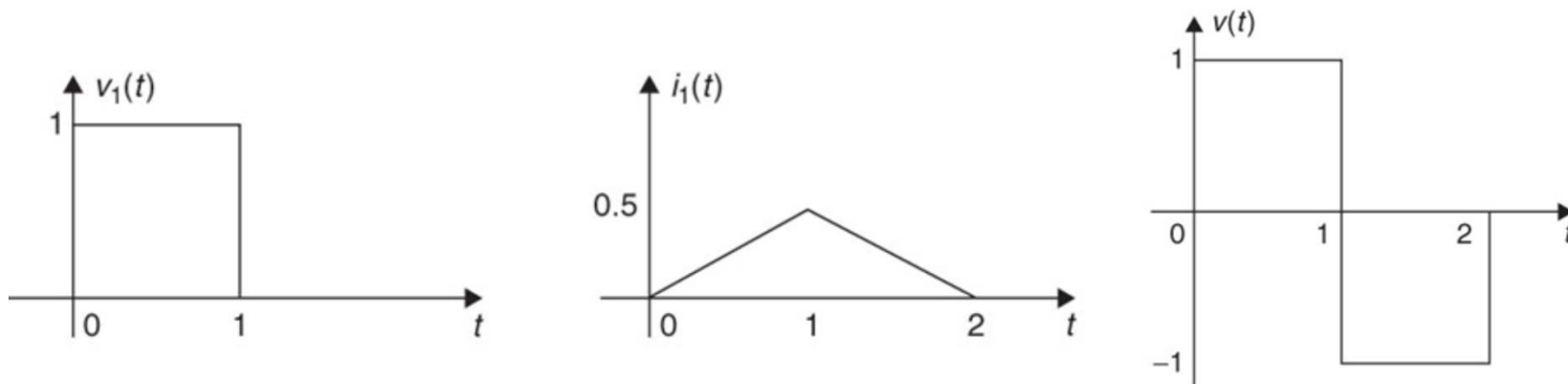
# Examples

[Ex 2.8] The response of RL circuit to a unit-step function  $u(t)$  is

$$i(t) = (1 - e^{-t})u(t)$$

Find the response to a source  $v(t) = u(t) - u(t - 2)$ .

[Ex 2.9] Suppose we know that the response to a rectangular pulse  $v_1(t)$  is the current  $i_1(t)$  shown in the figure below. If the input voltage is a train of two pulses,  $v(t)$ , find the corresponding current  $i(t)$ .



# Convolution Integral

The **impulse response** of an analog LTI system,  $h(t)$ , is the **output of the system** corresponding to **an impulse  $\delta(t)$  as input** and **initial conditions equal to zero**.

The **response of an LTI system  $S$**  with  $h(t) = S[\delta(t)]$  to any signal  $x(t)$  is the **convolution integral**

$$\begin{aligned} y(t) &= \int_{-\infty}^{\infty} x(\tau)h(t - \tau) d\tau = \int_{-\infty}^{\infty} x(t - \tau)h(\tau) d\tau \\ &= [x * h](t) = [h * x](t) \end{aligned}$$

assuming that no energy is initially stored in the system.

- The impulse response is fundamental in the characterization of LTI systems.
- A system characterized by the convolution integral is **linear and time invariant** by the above construction.

# Examples

[Ex 2.10] Obtain the impulse response of a capacitor and use it to find its unit-step response by means of the convolution integral. Let  $C = 1\text{ F}$ .

[Ex 2.11] The output of an analog average is given by

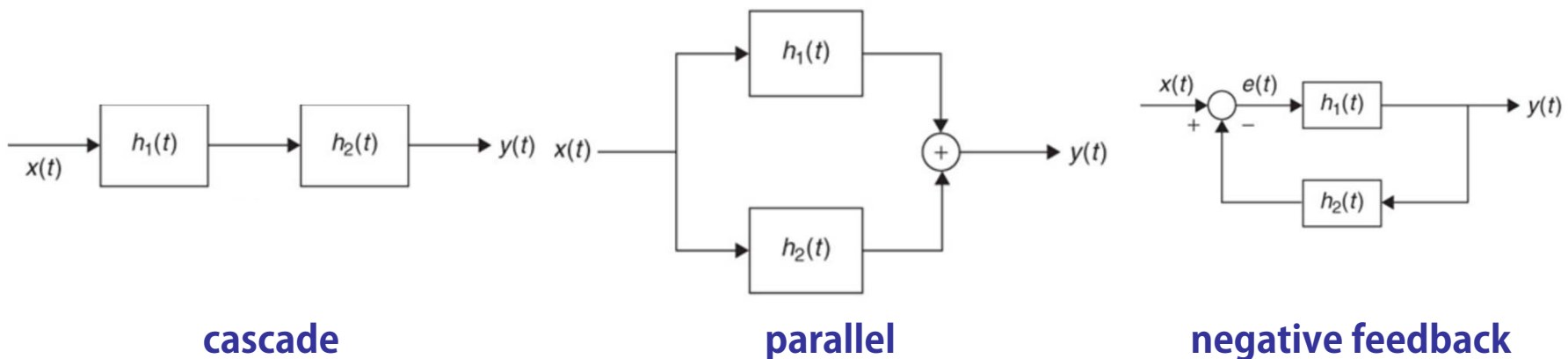
$$y(t) = \frac{1}{T} \int_{t-T}^t x(\tau) d\tau$$

which corresponds to the accumulation of values of  $x(t)$  in a segment  $[t - T, t]$  divided by its length  $T$ , or the average of  $x(t)$  in  $[t - T, t]$ . Use the convolution integral to find the response of the average to a ramp.

[Ex 2.12] The impulse response of a LTI system is  $h(t) = u(t) - u(t - 1)$ . Consider inputs  $x_2(t) = 0.5[\delta(t) + \delta(t - 0.5)]$ ,  $x_4(t) = 0.25[\delta(t) + \delta(t - 0.25) + \delta(t - 0.5) + \delta(t - 0.75)]$  find the corresponding outputs of the system.

# Interconnection of Systems - Block Diagrams

- The **flow of signals** is indicated by **arrows**, and the **addition of signals or multiplication of a signal by a constant** is indicated by means of **circles**.
  - The **cascade** and the **parallel** connections.
  - The **feedback** connection is found in control systems.



# Cascade and Parallel Connection

Two LTI systems with **impulse responses**  $h_1(t)$  and  $h_2(t)$  connected in **cascade** have an overall impulse response

$$h(t) = [h_1 * h_2](t) = [h_2 * h_1](t)$$

where  $h_1(t)$  and  $h_2(t)$  **commute**.

- When dealing with **time-varying systems**, the **order in which we connect the systems in cascade is important**.

If we **connect in parallel** two LTI systems with **impulse responses**  $h_1(t)$  and  $h_2(t)$ , the impulse response of the overall system is

$$h(t) = h_1(t) + h_2(t)$$

# Feedback Connection

- The feedback output is either **added to the input** giving a **positive feedback** system or **subtracted from the input** giving a **negative feedback** system.

Given two LTI systems with **impulse responses**  $h_1(t)$  and  $h_2(t)$ , a **negative feedback** connection is such that the output is

$$y(t) = [h_1 * e](t)$$

where the **error signal**

$$e(t) = x(t) - [y * h_2](t)$$

The **impulse response**  $h(t)$  is given by

$$h(t) = [h_1 - h * h_1 * h_2](t)$$

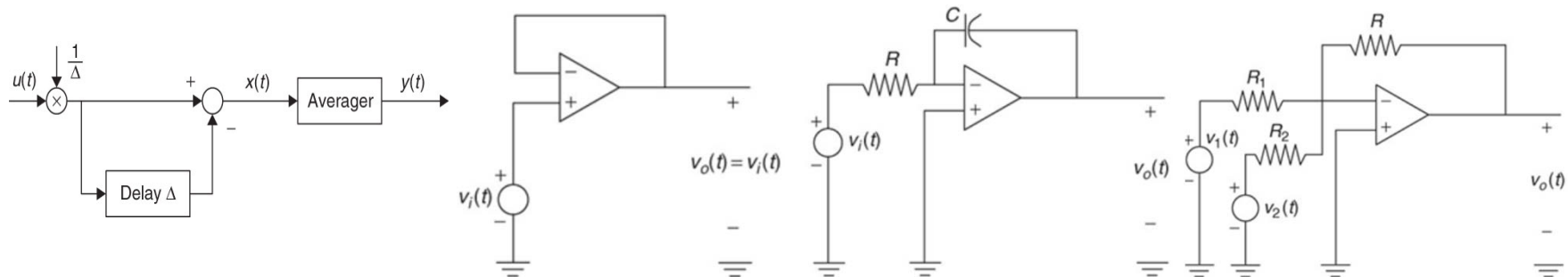
# Examples

[Ex 2.13] Consider a block diagram with input a the unit-step signal  $u(t)$ . The average is such that for an input  $x(t)$  its output is

$$y(t) = \frac{1}{T} \int_{t-T}^t x(\tau) d\tau$$

Determine what the system is doing as we let the delay  $\Delta \rightarrow 0$ . Consider that the average and the system with input  $u(t)$  and output  $y(t)$  are LTI.

[Ex 2.14] Consider the circuits obtained with an operational amplifier when we feed back its output with a wire, a resistor, and a capacitor. Assume the linear model for the op-amp. The circuits are called a voltage follower, an integrator, and an adder.



# Causality

A **continuous-time system**  $S$  is called **causal** if

- Whenever the **input**  $x(t) = 0$  and there are **no initial conditions**, the **output**  $y(t) = 0$ .
- The output  $y(t)$  **does not depend on future inputs**.

An **LTI system** represented by its **impulse response**  $h(t)$  is **causal** if

$$h(t) = 0, \quad \text{for } t < 0$$

The output of a causal LTI system with a **causal input**  $x(t)$  (i.e.  $x(t) = 0$  for  $t < 0$ ) is

$$y(t) = \int_0^t x(\tau) h(t - \tau) d\tau$$



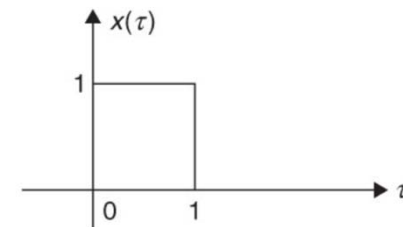
# Graphical Computation of Convolution

[Ex 2.15] Graphically find the unit-step  $y(t)$  response of an average, with  $T = 1$  sec, which has an impulse response

$$h(t) = u(t) - u(t - 1)$$

[Ex 2.16] Consider the graphical computation of the convolution integral of two pulses of the same duration.

The length of the support  $y(t) = [x * h](t)$  is equal to the **sum of the lengths of the supports of  $x(t)$  and  $h(t)$** .



[Ex 2.17<sup>MATLAB</sup>] Compute the corresponding output using convolution integral of the following inputs and impulse responses of LTI systems:

(a)  $x_1(t) = u(t) - u(t - 1)$ ,  $h_1(t) = u(t) - u(t - 2)$ ,

(b)  $x_2(t) = h_2(t) = r(t) - 2r(t - 1) + r(t - 2)$ ,

(c)  $x_3(t) = e^{-t}u(t)$ ,  $h_3(t) = e^{-10t}u(t)$ .

# Bounded-Input Bounded-Output Stability

- A **stable system** is such that **well-behaved outputs are obtained for well-behaved inputs**.

**Bounded-input bounded-output (BIBO) stability** establishes that for a bounded input  $x(t)$  the output of a BIBO stable system  $y(t)$  is also bounded.

An **LTI system with an absolutely integrable impulse response** – that is

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

is **BIBO stable**.

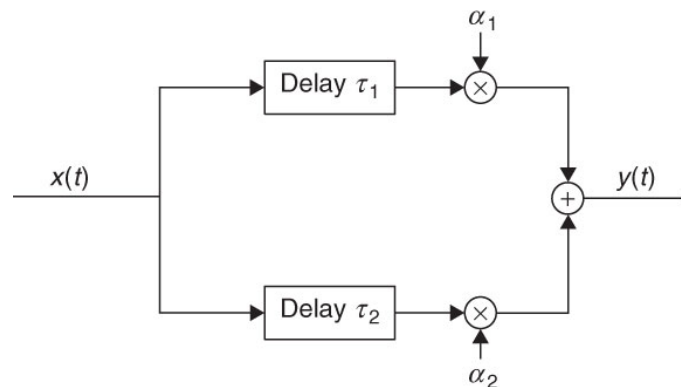
# Examples

[Ex 2.18] Consider the BIBO stability and causality of RLC circuits. Consider, for instance, a series RL circuit where  $R = 1 \Omega$  and  $L = 1 H$ , and a voltage source  $v_s(t)$ , which is bounded. Discuss why such a system would be causal and stable.

[Ex 2.19] Consider the causality and BIBO stability of an echo system (or a multipath system). Let the output  $y(t)$  be given by

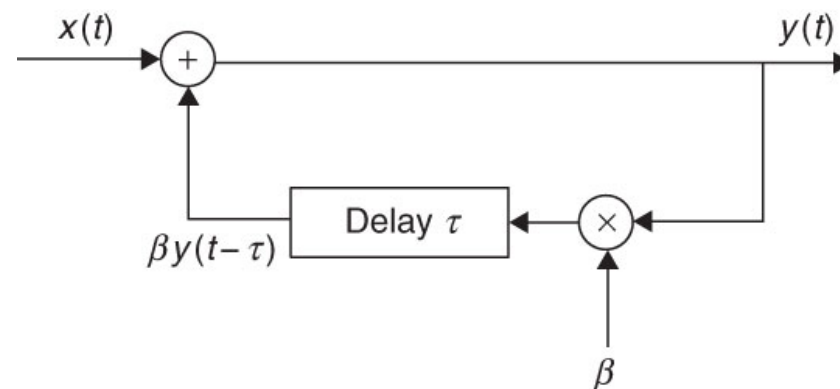
$$y(t) = \alpha_1 x(t - \tau_1) + \alpha_2 x(t - \tau_2)$$

where  $x(t)$  is the input, and  $|\alpha_i| < 1$ ,  $\tau_i > 0$ , for  $i = 1$  and  $2$ , are attenuation factors and delays. Is this system causal and BIBO stable?



# Example

[Ex 2.20] Consider a positive feedback system created by a microphone close to a set of speakers that are putting out an amplified acoustic signal. The microphone picks up the input signal  $x(t)$  as well as the amplified and delayed signal  $\beta y(t - \tau)$ ,  $|\beta| \geq 1$ . Find the equation that connects the input  $x(t)$  and output  $y(t)$  and recursively from it obtain an expression for  $y(t)$  in terms of past values of the input. Determine if the system is BIBO stable or not – use  $x(t) = u(t)$ ,  $\beta = 2$ , and  $\tau = 1$  in doing so.





*Thank You*