# **EECS151: Introduction to Digital Design and ICs**

# **Lecture 6 – Combinational Logic**

## **Bora Nikolić**



September 14, 2021. Apple announces A15 Bionic processor

- 6 cores
- New GPU
- New neural engine
- •

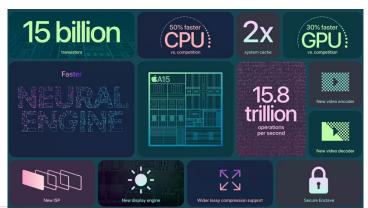




Image source: Apple



#### Review

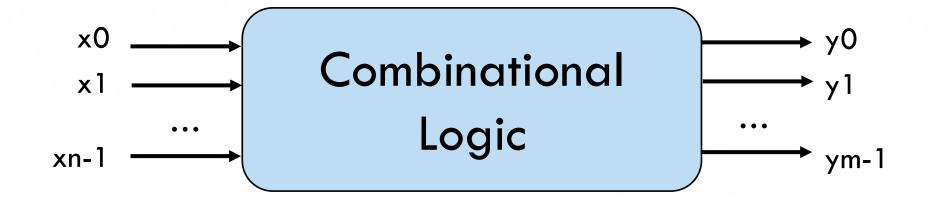
- Sequential logic uses flip-flops and (sometimes) latches
- Flip-flops and latches are inferred in Verilog
  - Always blocks
- Practice is the best way to learn a new language...
- Blocking and non-blocking assignments



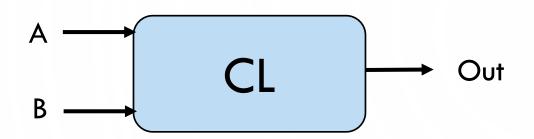
# **Combinational Logic**

#### **Combinational Logic**

- The outputs depend \*only\* on the current values of the inputs.
  - Memoryless: compute the output values using the current inputs.



## Combinational Logic Example



#### **Boolean Equations:**

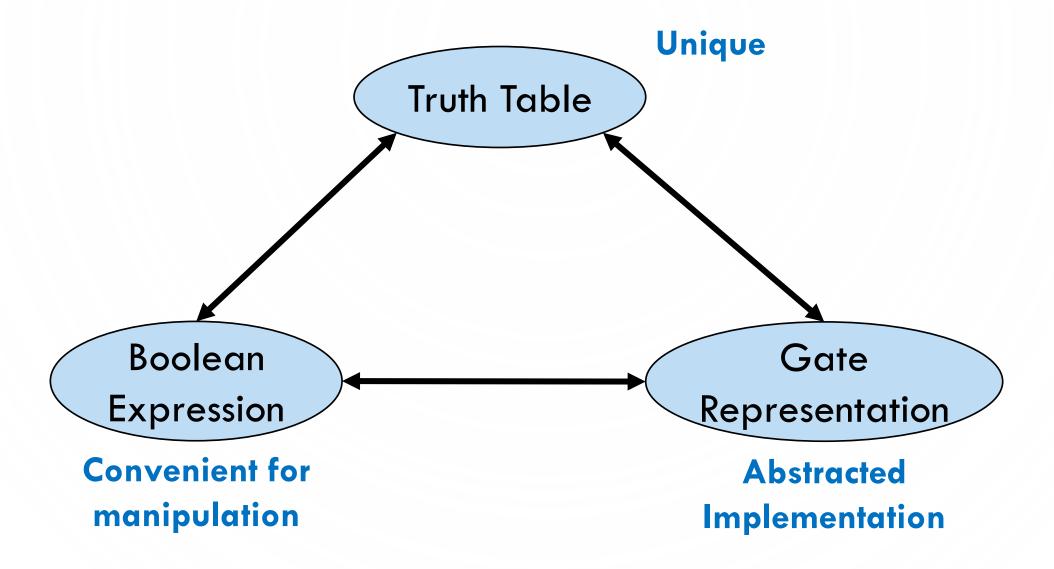
#### Truth Table Description:

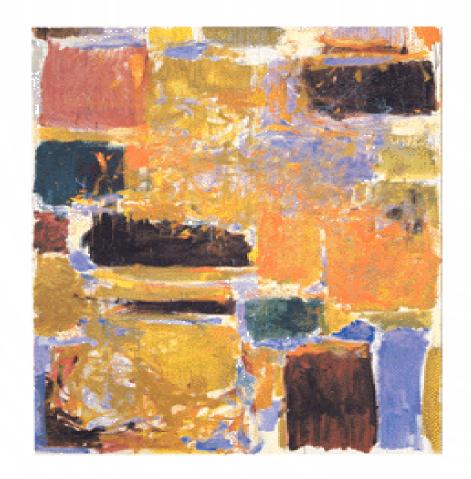
A	В	Out
0	0	0
0	1	1
1	0	1
1	1	1

#### Gate Representations:



#### Relationship Among Representations



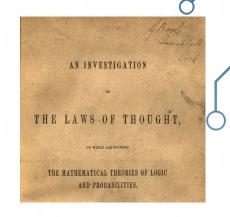


Boolean Algebra

## Boolean Algebra Background

- Logic: The study of the principles of reasoning.
- The 19th Century Mathematician, George Boole, developed a math. system (algebra) involving logic, Boolean Algebra.
  - His variables took on TRUE, FALSE.
- Later Claude Shannon (father of information theory) showed (in his Master's thesis!) how to map Boolean Algebra to digital circuits.





georgeboole.com Digitized by Google





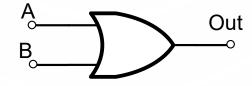
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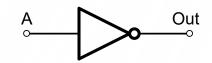


## Boolean Algebra Fundamentals

- Two elements {0, 1}
- Two binary operators: AND (·) OR (+)
- One unary operator: NOT ( , ')







Α	В	Out
0	0	0
0	1	0
1	0	0
1	1	1

A	В	Out
0	0	0
0	1	1
1	0	1
1	1	1

A	Out
0	1
1	0

## Axioms of Boolean Algebra

Axiom	Dual	Name
$B = 0 \text{ if } B \neq 1$	$B = 1 \text{ if } B \neq 0$	Binary field
$\overline{O} = 1$	<u>1</u> = 0	NOT
0 • 0 = 0	1 + 1 = 1	AND/OR
1 • 1 = 1	0 + 0 = 0	AND/OR
$0 \cdot 1 = 0 \cdot 1 = 0$	1 + 0 = 0 + 1 = 1	AND/OR

In mathematical logic, axioms are given
Anything else can be derived from these axioms
Each axiom has a dual

## **Boolean Operations**

• Given two variables (A, B), 16 logic functions

A	В	$\boldsymbol{F_0}$	$F_1$	$F_2$	$F_3$	$F_4$	$F_5$	$F_6$	$F_7$	<i>F</i> <sub>8</sub>	<b>F</b> <sub>9</sub>	$F_A$	$F_B$	Fc	$F_D$	$F_E$	$F_F$
0	0	0	0	0	0	0	0	0	0	1	1	1	1	1	1	1	1
0	1	0	0	0	0	1	1	1	1	0	0	0	0	1	1	1	1
1	0	0	0	1	1	0	0	1	1	0	0	1	1	0	0	1	1
1	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1	0	1

#### **Boolean Algebra Theorems**

#### Null elements, identities:

#### • Idempotency:

#### Complements:

• Involution:

Commutativity:

#### Associativity:

• 
$$(A + B) + C = A + (B + C) = A + B + C$$

• 
$$(A \cdot B) \cdot C = A \cdot (B \cdot C) = A \cdot B \cdot C$$

#### • Distributivity:

• 
$$A \cdot (B+C) = (A \cdot B) + (A \cdot C)$$

• 
$$A + (B \cdot C) = (A + B) \cdot (A + C)$$

#### • Covering:

• 
$$A \cdot (A+B) = A, A + (A \cdot B) = B$$

#### Consensus

• 
$$(A \cdot B) + (A' \cdot C) + (B \cdot C) = (A \cdot B) + (A' \cdot C)$$

## Proving Distributive Law

• A • 
$$(B+C) = (A \cdot B) + (A \cdot C)$$

A	В	C	(B+C)	A • (B+C)	(A•B)	(A•C)	(A•B) + (A•C)
0	0	0					
0	0	1					
0	1	0					
0	1	1					
1	0	0					
1	0	1					
1	1	0					
1	1	1					

## Proving Distributive Law

• A • 
$$(B+C) = (A \cdot B) + (A \cdot C)$$

A	В	С	(B+C)	A • (B+C)	(A•B)	(A•C)	(A•B) + (A•C)
0	0	0	0	0	0	0	0
0	0	1	1	0	0	0	0
0	1	0	1	0	0	0	0
0	1	1	1	0	0	0	0
1	0	0	0	0	0	0	0
1	0	1	1	1	0	1	1
1	1	0	1	1	1	0	1
1	1	1	1	1	1	1	1

## DeMorgan's Law

• Theorem for complementing a complex function.

$$(A + B)' = A' B'$$

$$(A B)' = A' + B'$$

A	В	A'	В'	(A + B)'	A' B'
0	0				
0	1				
1	0				
1	1				

A	В	Α'	Α'	(A B)"	A' + B'
0	0				
0	1				
1	0				
1	1				

#### DeMorgan's Law

• Procedure for complementing a complex function.

$$(A + B)' = A' B'$$

$$(A B)' = A' + B'$$

A	В	Α'	В'	(A + B)"	A' B'
0	0	1	1	1	1
0	1	1	0	0	0
1	0	0	1	0	0
1	1	0	0	0	0

A	В	Α'	В'	(A B)"	A' + B'
0	0	1	1	1	1
0	1	1	0	1	1
1	0	0	1	1	1
1	1	0	0	0	0

#### **Canonical Forms**

- Two types:
  - Sum of Products (SOP)
  - Product of Sums (POS)
- Sum of Products
  - a.k.a Disjunctive normal form, minterm expansion
  - Minterm: a product (AND) involving all inputs
  - SOP: Summing minterms for which the output is True

Minterms	a	b	С	f	f′
a'b'c'	0	0	0	0	1
a'b'c'	0	0	1	0	1
a'b c'	0	1	0	0	1
a'b c	0	1	1	1	0
a b'c'	1	0	0	1	0
a b'c	1	0	1	1	0
a b c'	1	1	0	1	0
a b c	1	1	1	1	0

One product (and) term for each 1 in f:

#### Sum of Products (cont.)

- Canonical Forms are usually not minimal:
- Example:

#### **Canonical Forms**

- Two types:
  - Sum of Products (SOP)
  - Product of Sums (POS)
- Product of Sums:
  - a.k.a. conjunctive normal form, maxterm expansion
  - Maxterm: a sum (OR) involving all inputs
  - POS: Product (AND) maxterms for which the output is FALSE
  - Can obtain POSs from applying DeMorgan's law to the SOPs of F (and vice versa)

Maxterms	a	b	С	f	f′
a+b+c	0	0	0	0	1
a+b+c'	0	0	1	0	1
a+b′+c	0	1	0	0	1
a+b'+c'	0	1	1	1	0
a'+b+c	1	0	0	1	0
a'+b+c'	1	0	1	1	0
a'+b'+c	1	1	0	1	0
a'+b'+c'	1	1	1	1	0

One sum (or) term for each 0 in f:



#### Quiz

ullet Derive the product-of-sums form of  $\overline{Y}$  based on the truth table.

a) 
$$\overline{Y} = (A + B)(A + \overline{B})$$

b) 
$$\overline{Y} = A\overline{B} + AB$$

c) 
$$\bar{Y} = \bar{A}\bar{B} + \bar{A}B$$

Α	В	Y	Ÿ
0	0	0	1
0	1	0	1
1	0	1	0
1	1	1	0



# **Boolean Simplification**

#### Example: Full Adder (FA) Carry out

```
co = a'bc + ab'c + abc' + abc
   = a'bc + ab'c + abc' + abc + abc
   = a'bc + abc + ab'c + abc' + abc
   = (a' + a)bc + ab'c + abc' + abc
   = (1)bc + ab'c + abc' + abc
   = bc + ab'c + abc' + abc + abc
   = bc + ab'c + abc + abc' + abc
   = bc + a(b' +b)c + abc' +abc
   = bc + a(1)c + abc' + abc
   = bc + ac + ab(c' + c)
   = bc + ac + ab(1)
   = bc + ac + ab
```

С	b	С	S	со
0	0	0	0	0
0	0	1	1	0
0	1	0	1	0
0	1	1	0	1
1	0	0	1	0
1	0	1	0	1
1	1	0	0	1
1	1	1	1	1

## Why do Boolean simplification?

- Minimize number of gates in circuit
  - Gates take area
- Minimize amount of wiring in circuit
  - Wiring takes space and is difficult to route
  - Physical gates have limited number of inputs
- Minimize number of gate levels
  - Faster is better
- How to systematically simplify Boolean logics?
  - Use tools!

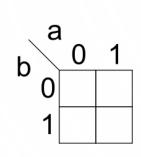
#### Practical methods for Boolean simplification

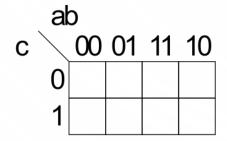
- Still based on Boolean algebra, but more systematic
- 2-level simplification -> multilevel
- Key tool: The Uniting Theorem

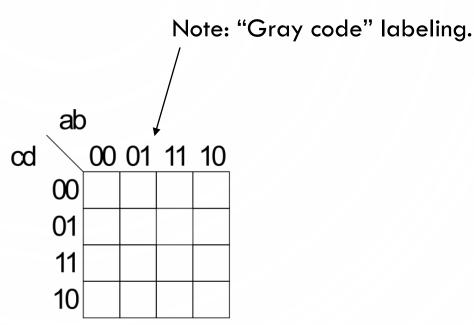
$$ab' + ab = a (b' + b) = a (1) = a$$

ab f	f = ab' + ab = a(b'+b) = a
00 0 01 0 10 1 11 1	<b>b</b> values change within rows
01 0	a values don't change
10 1	<b>b</b> is eliminated, a remains
11 1	

• K-map is an alternative method of representing the truth table and to help visual the adjacencies.



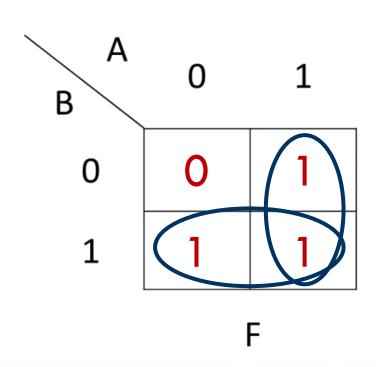




Adjacent groups of 1's represent product terms

Α	В	F
0	0	0
0	1	1
1	0	1
1	1	1

# Karnaugh Map

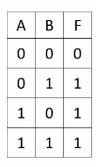


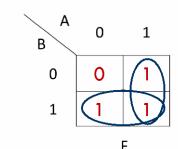
$$F = A + B$$

- 1. Draw K-map of the appropriate number of variables.
- 2. Fill in map with function values from truth table.
- 3. Form groups of 1's.
  - $\sqrt{}$  Dimensions of groups must be even powers of two (1x1, 1x2, 1x4, ..., 2x2, 2x4, ...)
  - √ Form as large as possible groups and as few groups as possible.
  - ✓ Groups can overlap (this helps make larger groups)
  - √ Remember K-map is periodical in all dimensions (groups can cross over edges of map and continue on other side)
- 4. For each group write a product term.
  - The term includes the "constant" variables (use the uncomplemented variable for a constant 1 and complemented variable for constant 0)
- 5. Form Boolean expression as sum-of-products.

OR

Karnaugh Map

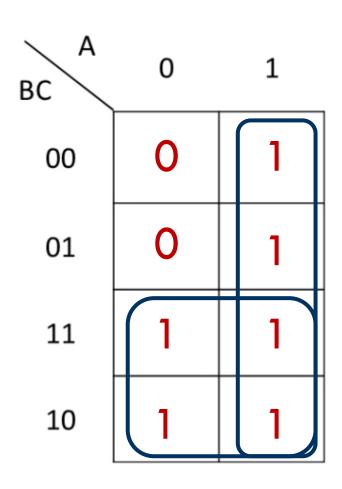




F = A + B

Α	В	С	F
0	0	0	0
0	0	1	0
0	1	0	1
0	1	1	1
1	0	0	1
1	0	1	1
1	1	0	1
1	1	1	1





#### Product-of-Sums Version

- 1. Form groups of 0's instead of 1's.
- 2. For each group write a sum term.
  - the term includes the "constant" variables (use the uncomplemented variable for a constant 0 and complemented variable for constant 1)
- 3. Form Boolean expression as product-of-sums.

	<sub>(</sub> ab	)				
$\infty$		00	01	11	10	
	00	1(	$(\bigcirc)$	6	1	
	01	6	1	0	0	
	11	1	1	1	1	
	10	1	1	1	1	

$$f = (b' + c + d)(a' + c + d')(b + c + d')$$

#### Karnaugh Maps with Don't Cares

• Don't cares (x's) in the truth table can be either 0's or 1's

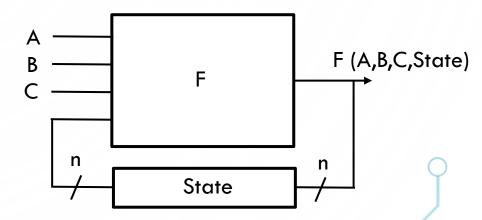
A	AB			
CD	00	01	11	10
00	1	1	X	1
01	1	0	Х	1
11	1	1	Х	Х
10	1	0	Х	Х



Finite-State Machines

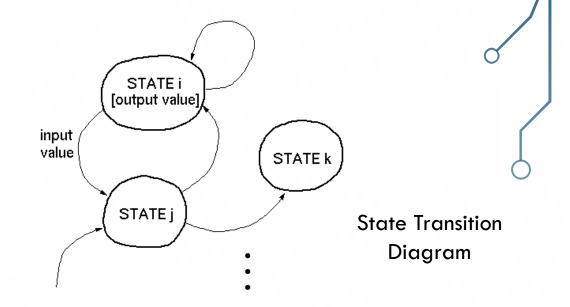
#### Sequential logic

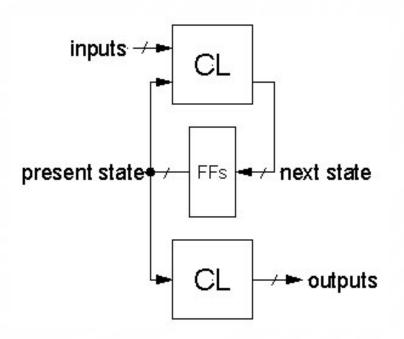
- Combinational logic:
  - Memoryless: the outputs only dependent on the current inputs.
- Sequential logic:
  - Memory: the outputs depend on both current and previous values of the inputs.
    - Distill the prior inputs into a smaller amount of information, i.e., states.
  - State: the information about a circuit
    - Influences the circuit's future behavior
    - Stored in Flip-flops and Latches
  - Finite State Machines:
    - Useful representation for designing sequential circuits
    - As with all sequential circuits: output depends on present and past inputs
    - We will first learn how to design by hand then how to implement in Verilog.



#### Finite State Machines

- A sequential circuit which has
  - External inputs
  - Externally visible outputs
  - Internal states
- Consists of:
  - State register
    - Stores current state
    - Loads previously calculated next state
    - # of states <= 2^(# of FFs)
  - Combinational logic
    - Computes the next state
    - Computes the outputs

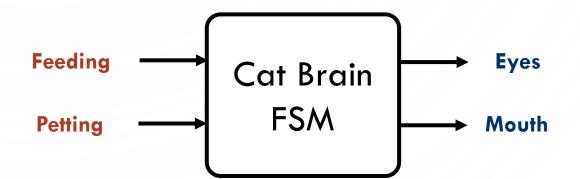




#### FSM Example

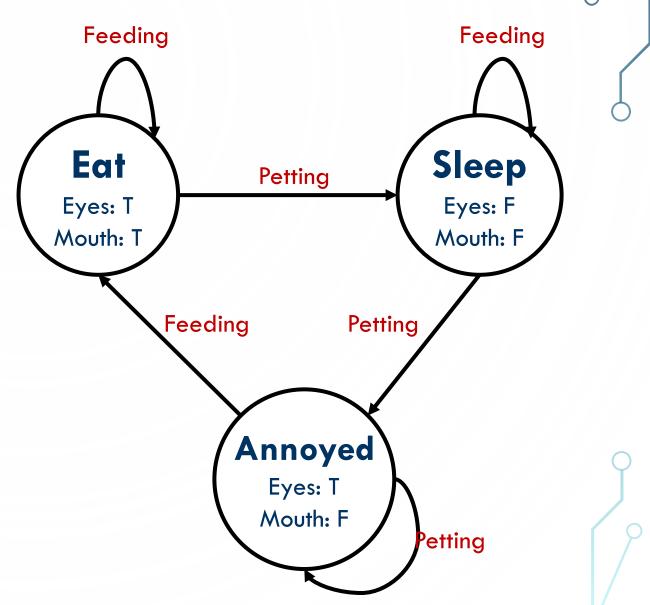
- Cat Brain (Simplified...)
  - Inputs:
    - Feeding
    - Petting
  - Outputs:
    - Eyes: open or close
    - Mouth: open or close
  - States:
    - Eating
    - Sleeping
    - Annoyed...





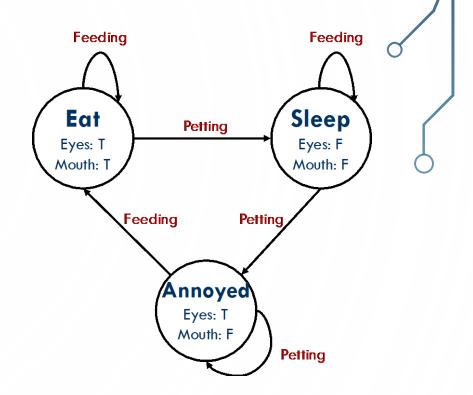
#### FSM State Transition Diagram

- States:
  - Circles
- Outputs:
  - Labeled in each state
  - Arcs
- Inputs:
  - Arcs



## FSM Symbolic State Transition Table

Current State	Inputs	Next State
Eat	Feeding	Eat
Eat	Petting	Sleep
Sleep	Feeding	Sleep
Sleep	Petting	Annoyed
Annoyed	Feeding	Eat
Annoyed	Petting	Annoyed

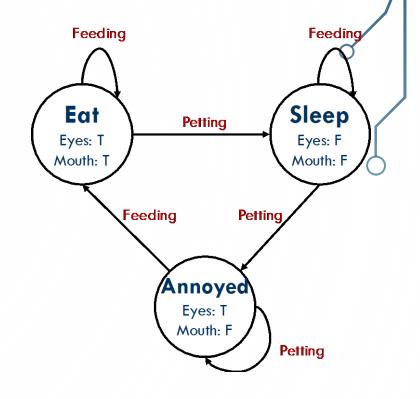


#### FSM Encoded State Transition Table

State	Encoding
Eat	00
Sleep	01
Annoyed	10

Current State		Next	State
SO	X	<b>S1</b> '	<b>SO'</b>
0	0	0	0
0	1	0	1
1	0	0	1
1	1	1	0
0	0	0	0
0	1	1	0
	\$0 0 0 1 1 0	SO     X       0     0       0     1       1     0       1     1       0     0	SO     X     S1'       0     0     0       0     1     0       1     0     0       1     1     1       0     0     0

$$S0' = \overline{S1S0}X + \overline{S1}S0\overline{X} = \overline{S1}(\overline{S0}X + S0\overline{X}) = \overline{S1}(S0 \oplus X)$$
  
$$S1' = \overline{S1}S0X + S1\overline{S0}X = (S1 \oplus S0)X$$



Current State	Inputs	Next State
Eat	Feeding	Eat
Eat	Petting	Sleep
Sleep	Feeding	Sleep
Sleep	Petting	Annoyed
Annoyed	Feeding	Eat
Annoyed	Petting	Annoyed



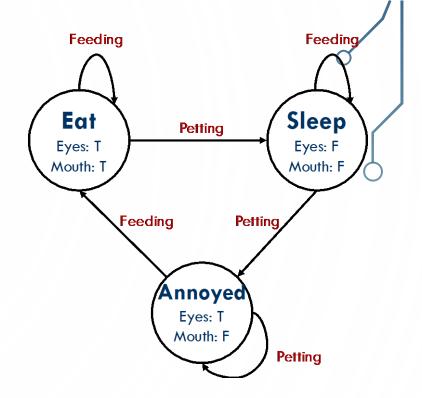
## FSM Output Table

State	Encoding
Eat	00
Sleep	01
Annoyed	10

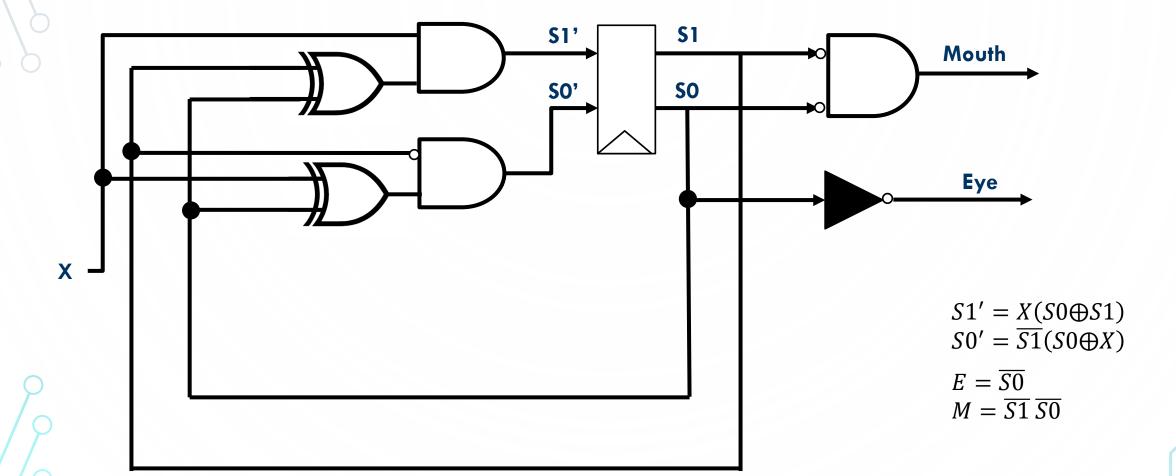
Current State		Outputs	
<b>S</b> 1	SO	E	M
0	0	1	1
0	1	0	0
1	0	1	0

Outputs		Encoding
Eyes	Mouth	
Open	Open	11
Close	Close	00
Open	Close	10

$$E = \overline{S1S0} + S1\overline{S0} = \overline{S0}$$
$$M = \overline{S1}\overline{S0}$$



## FSM Gate Representation



#### Summary

- Combinational logic:
  - The outputs only depend on the current values of the inputs (memoryless)
  - The functional specification of a combinational circuit can be expressed as:
    - A truth table
    - A Boolean equation
- Boolean algebra
  - Deal with variables that are either True or False
  - Map naturally to hardware logic gates
  - Use theorems of Boolean algebra and Karnaugh maps to simplify equations
- Finite state machines: Common example of sequential logic
- Common job interview questions

