

电力电子公式一览

Fourier Transform

$$f(x) = \frac{a_0}{2} + \sum_{n=1}^{\infty} [a_n \cos(nx) + b_n \sin(nx)]$$

$$a_n = \frac{2}{T} \int f(t) \cos \omega t dt$$

$$b_n = \frac{2}{T} \int f(t) \sin \omega t dt$$

DCDC

BUCK

电压变比

$$M = D \quad (1)$$

临界电流

$$I_{OB} = \frac{V_o}{2fL} (1 - D) \quad (2)$$

DCM电压变比

$$M = \frac{2}{1 + \sqrt{1 + \frac{4I_o^*}{D^2}}} \quad (3)$$

$$I_o^* = \frac{I_o}{V_o/2fL}$$

BOOST

电压变比

$$M = \frac{1}{1 - D} \quad (4)$$

临界电流

$$I_{OB} = \frac{V_o}{2fL} (1 - D)^2 D \quad (5)$$

DCM电压变比

$$M = \frac{1 + \sqrt{1 + \frac{4D^2}{I_o^*}}}{2} \quad (6)$$

$$I_O^* = \frac{I_O}{V_o/2fL}$$

Cuk/Buck-Boost

电压变比

$$M = \frac{D}{1-D} \quad (7)$$

临界电流

$$I_{OB} = \frac{V_o}{2fL} (1-D)^2 \quad (8)$$

DCM电压变比

$$M = \frac{D}{\sqrt{I_O^*}} \quad (9)$$

$$I_O^* = \frac{I_O}{V_o/2fL}$$

DCAC

上下方波的傅里叶变换

$$v(t) = \sum_{n=1,3,5}^{\infty} \frac{4V_D}{n\pi} \sin(n\omega t)$$

有效值 $V_1 = 0.9V_D$

接Z阻抗 $(R + j\omega L)$

$$(10)$$

$$i_{a1} = \frac{4V_D}{\pi \sqrt{R^2 + (\omega L)^2}} \sin\left(\omega t - \arctan\left(\frac{\omega L}{R}\right)\right)$$

单脉波脉冲宽度调制

$$v_{ab}(\omega t) = \sum_{n=1,3,4}^{\infty} -(-1)^{(n+2)/2} \frac{4V_D}{n\pi} \sin \frac{n\theta}{2} \sin n\omega t \quad (11)$$

逆变器的性能指标

$$\text{谐波系数 } HF_n = \frac{V_n}{V_1}$$

$$\text{总谐波系数 } THD = \sqrt{\sum_{n=1,2,3}^{\infty} \left(\frac{V_n}{V_1}\right)^2 - 1}$$

畸变系数

$$\text{一阶滤波 } DF_1 = \sqrt{\sum_{n=2,3,4}^{\infty} \left(\frac{V_n}{nV_1}\right)^2}$$

$$\text{二阶滤波 } DF_2 = \sqrt{\sum_{n=2,3,4}^{\infty} \left(\frac{V_n}{n^2V_1}\right)^2}$$

$$(12)$$

$$\sqrt[n=2,3,4]{u'v'1}$$

数据

- BSPWM：0.707VD
- 方波：0.9VD
- 三相SPWM相电压幅值：0.5
- 三相SPWM线电压幅值：0.866
- 三相方波相电压幅值：1.1

ACDC公式

不控整流

- 1. 输出电压：
 - 1. 双半波： $\frac{2\sqrt{2}}{\pi}$
 - 2. 单相桥： $\frac{2\sqrt{2}}{\pi}$
 - 3. 三相半桥： $\frac{3\sqrt{6}}{2\pi}$
 - 4. 三相桥式： $\frac{3\sqrt{6}}{\pi}$

单相相控整流电压

平均值

$$\frac{2\sqrt{2}}{\pi}V_s\frac{1+\cos\alpha}{2}\tag{13}$$

功率因数

$$PF=\sqrt{\frac{\sin2\alpha}{2\pi}+\frac{\pi-\alpha}{\pi}}\tag{14}$$

电阻负载的电压有效值

$$V_s*PF\tag{15}$$

电阻负载的电流有效值

$$\frac{V_s}{R}PF\tag{16}$$

晶闸管的电流有效值

$$\frac{V_s}{2R}PF\tag{17}$$

三相相控整流电压

平均值

$$\frac{3\sqrt{6}}{\pi}V_s\cos\alpha\tag{18}$$

单相整流器导通角求取（有阻感性负载） $(\alpha > \phi)$

$$\tan(\alpha - \phi) = \frac{\sin \theta}{e^{-\frac{\theta}{\tan \theta}} - \cos \theta} \quad (19)$$

在有反电动势存在时单相相控整流器停止导电角

$$\sin \delta = \frac{E}{\sqrt{2}V_s} \quad (20)$$

单相桥相控整流时电流连续条件

$$L \geq \frac{\frac{2\sqrt{2}}{\pi} V_s}{\omega I_D} \quad (21)$$

换相整流区

m脉波整流电路

单相桥式m=2，核心在Vs=2Vs

两相半波m=2

三相半波m=3

三相全桥m=6

换相电阻

$$R = \frac{m\omega L}{2\pi} \quad (22)$$

换相重叠区角度

$$\cos \alpha - \cos(\alpha + \gamma) = \frac{\omega L_s I_D}{\sqrt{2}V_s \sin \frac{\pi}{m}} \quad (23)$$

有源逆变输出电压

$$\frac{3\sqrt{6}}{\pi} V_s \cos \beta - \frac{3\omega L_s}{2\pi} I_D * 2 \quad (24)$$

电力电子公式推导**DCAC****上下方波的傅里叶变换**

$$f(t) = \sum_{n=1,3,5}^{\infty} a_n \sin \omega t$$

$$a_n = \frac{2}{\pi} \int_0^{\pi} V_D \sin n\omega t d(\omega t) \quad (25)$$

$$a_n = \frac{4}{n\pi} V_D$$

三相方波的傅里叶变换

$$v_{ab}(t) = \sum_{n=1,3,5}^{\infty} a_n \sin wt$$

$$a_n = \frac{2}{\pi} \int_{\pi/6}^{5\pi/6} V_D \sin nwt d(wt) \quad (26)$$

$$a_n = \frac{2\sqrt{3}}{n\pi}$$

ACDC

不控整流平均输出电压计算

单相

$$V_o = \frac{1}{\pi} \int_0^{\pi} \sqrt{2} V_s \sin wt d(wt) = \frac{2\sqrt{2}}{\pi} V_s \quad (27)$$

三相

$$V_o = \frac{1}{\pi/3} \int_{\pi/3}^{2\pi/3} \sqrt{3} * \sqrt{2} V_s \sin wt d(wt) = \frac{3\sqrt{6}}{\pi} V_s \quad (28)$$

相控整流平均电压计算

单相

$$V_o = \frac{1}{\pi} \int_{\alpha}^{\pi} \sqrt{2} V_s \sin wt d(wt) = \frac{2\sqrt{2}}{\pi} V_s \frac{1 + \cos \alpha}{2} \quad (29)$$

三相

$$V_o = \frac{1}{\pi/3} \int_{\pi/3+\alpha}^{2\pi/3+\alpha} \sqrt{3} * \sqrt{2} V_s \sin wt d(wt) = \frac{3\sqrt{6}}{\pi} V_s \cos \alpha (\alpha < 60^\circ) \quad (30)$$

单相相控整流电压电阻负载的电压有效值计算

$$V_{rms} = \sqrt{\frac{1}{\pi} \int_{\alpha}^{\pi} [\sqrt{2} V_s \sin wt]^2 d(wt)} = V_s \sqrt{\frac{\sin 2\alpha}{2\pi} + \frac{\pi - \alpha}{\pi}} \quad (31)$$

单相相控整流电压电阻负载的功率因数计算

$$PF = \frac{P_{in}}{S_{in}} = \frac{P_{out}}{S_{in}} = \frac{V_{orms}^2/R}{V_{irms} I_{irms}} = \frac{V_{orms}^2/R}{V_s V_{orms}/R} = \sqrt{\frac{\sin 2\alpha}{2\pi} + \frac{\pi - \alpha}{\pi}} \quad (32)$$

单相整流器导通角求取（有阻感性负载）（ $\alpha > \phi$ ）

原理

$$\sqrt{2} V_s \sin wt = L \frac{di_D}{dt} + Ri_D$$

$$\text{when } wt = \alpha, i_D = 0$$

$$\text{when } wt = \alpha + \theta, i_D = 0 \quad (33)$$

求解此方程组可得形式（齐次解+非齐次解）

$$i_D = \frac{\sqrt{2} V_s}{\omega L} \sin (wt - \phi) - \frac{R}{\omega L} e^{-\frac{R}{L} t} \quad (34)$$

$$i_D = \frac{E}{Z} + Ae^{-\frac{R}{L}t} \quad (34)$$

进而可以解得 A, θ

$$\tan(\alpha - \phi) = \frac{\sin \theta}{e^{-\frac{\theta}{\tan \theta}} - \cos \theta} \quad (35)$$

在有反电动势存在时单相相控整流器停止导电角

易于理解，画一条线

$$\sin \delta = \frac{E}{\sqrt{2}V_s} \quad (36)$$

单相桥相控整流时电流连续条件

书上官方

取 α 处为时间坐标原点，在临界连续时有

$$i_D(\omega t = \pi) = 0 \quad (37)$$

而 i_D 的计算为

$$\begin{aligned} L \frac{di_D}{dt} + Ri_D &= \sqrt{2}V_s \sin(\omega t + \alpha) - E \\ i_D(\omega t = \pi) &= 0 \\ i_D(\omega t = 0) &= 0 \end{aligned} \quad (38)$$

$$Ri_D + E = \frac{2\sqrt{2}}{\pi} V_s \cos \alpha$$

故可解得临界连续时的电流为

$$i_D(t) = \frac{\sqrt{2}V_s}{\omega L} (\cos \alpha - \cos(\omega t + \alpha)) - \frac{2}{\pi} \cos \alpha * \omega t \quad (39)$$

电流平均值为

$$I_{Dmin} = \int_0^\pi i_D(\omega t) d\omega t = \frac{2\sqrt{2}}{\pi} V_s \frac{1}{\omega L} \sin \alpha \quad (40)$$

为使得什么时候都连续，取临界值 $\alpha = \pi/2$

则

$$I_{Dmin} = \frac{2\sqrt{2}}{\pi} V_s \frac{1}{\omega L} \quad (41)$$

直接看 $\alpha = \pi/2$

此时输出电压为0

$$L \frac{di_D}{dt} = \sqrt{2}V_s \cos(\omega t) \quad (42)$$

则

$$i_D = \frac{\sqrt{2}V_s \sin \omega t}{\omega L} \quad (43)$$

平均临界电流为

$$I_D = \frac{2\sqrt{2}}{\pi} \frac{V_s}{wL} \quad (44)$$

换相整流区

注意m为脉波数

原理

$$\begin{aligned} v_1 &= \sqrt{2}V_s \cos(wt - \pi/m) \\ v_2 &= \sqrt{2}V_s \cos(wt + \pi/m) \\ v_D &= \frac{1}{2}(v_1 + v_2) \\ \Delta V_S &= \frac{1}{2\pi/m} \int_{\alpha}^{\alpha+\gamma} \frac{1}{2}(v_b - v_a)d(wt) = \frac{m}{\pi} \sqrt{2}V_s \sin \frac{\pi}{m} \frac{\cos \alpha - \cos(\alpha + \gamma)}{2} \\ I_D &= \int_{\alpha}^{\alpha+\gamma} \frac{1}{wL_s} \frac{1}{2}(v_b - v_a)dwt = \frac{1}{wL_s} 2\sqrt{2}V_s \sin \frac{\pi}{m} \frac{\cos \alpha - \cos(\alpha + \gamma)}{2} \end{aligned} \quad (45)$$

则

$$\begin{aligned} R_s &= \frac{\Delta V_S}{I_D} = \frac{mwL_s}{2\pi} \\ \cos \alpha - \cos(\alpha + \gamma) &= \frac{wL_s I_D}{\sqrt{2}V_s \sin \frac{\pi}{m}} \end{aligned} \quad (46)$$