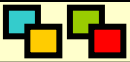


# **MEMS Automotive Collision Avoidance Radar beamformer**

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# Overview

- Introduction
- State-of-the-Art
- Design challenges
- Rotman lens beamformer
- Rotman lens parameters
- Rotman lens design equations
- Lens contour
- Target scanning overview
- Radiation patterns
- Parameters calculation without insulation
- Parameters calculation with insulation
- Target design specifications
- Fabrication
- References

# Introduction

- Rotman lens works by summing or focusing  $N$  in-phase samples of a wavefront at a focal point.
- It can be considered to be a true time-delay multiple beamformer.
- Currently used Rotman lens types
  - \_ Parallel plates (wave guide feed lines)
  - \_ Microstrip lens (stripline microwave printed circuit techniques to construct the feed section).
- For use in automotive collision avoidance systems
  - Position/Proximity sensors
  - Blind spot Measurements
  - Parking aid
  - Reverse aid
  - Pre crash
  - Stop/go sensor

# State-of-the-Art

- The most common Rotman lens used in automotive collision detection are microstrip lens.
- Microstrip Rotman lens used in the industry have the following specifications:

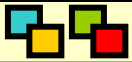
Parameter	Value
Operating Frequency	77 GHz
Operating Range	100 m
Voltage	60 V
Size	5 to 10 cm
Gain of Main lobe	> 25 db
Side lobe	< -15 db
Beam width	4°

# State-of-the-Art

- Advantages of Rotman lens:
- Monolithic construction
- Ease of manufacture
- Low cost
- Light weight
- Simultaneous availability of many beams.
- Because it is a true time-delay device, the Rotman lens produces frequency-independent beam steering and is therefore capable of extremely wide-band operation.
- These features make the Rotman lens an attractive candidate for use in multibeam satellite-based applications.

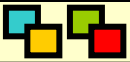
# Advantages of Rotman lens

- The Rotman lens is a true time-delay scanner that can be used either for receiving or transmitting.
- Increased range and detection performance over the ultrasonic sensors currently deployed as reverse parking aids.
- The design was improved by introducing a dielectric material in the parallel plate section (constant  $\epsilon_r$ ) reducing the dimensions of this section by  $1/\sqrt{\epsilon_r}$
- This improvement also permitted the use of microstrip and stripline microwave printed circuit techniques to construct the feed section.



# Design Challenges

- Port spacing (it have to be in the range of the fractions of the wave length)
- Suppressing and reducing the sidelobes
- Reflections at the beam and array ports
- Isolation of individual beams and cross over levels



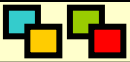
# Port spacing

- Port spacing (it have to be in the range of the fractions of the wave length)
- The distance between ports can be minimized by using sidewall layers of a relative permittivity of one forth of the lens dielectric.
- Redirecting the port can reduce the port spacing



# Reducing the sidelobes

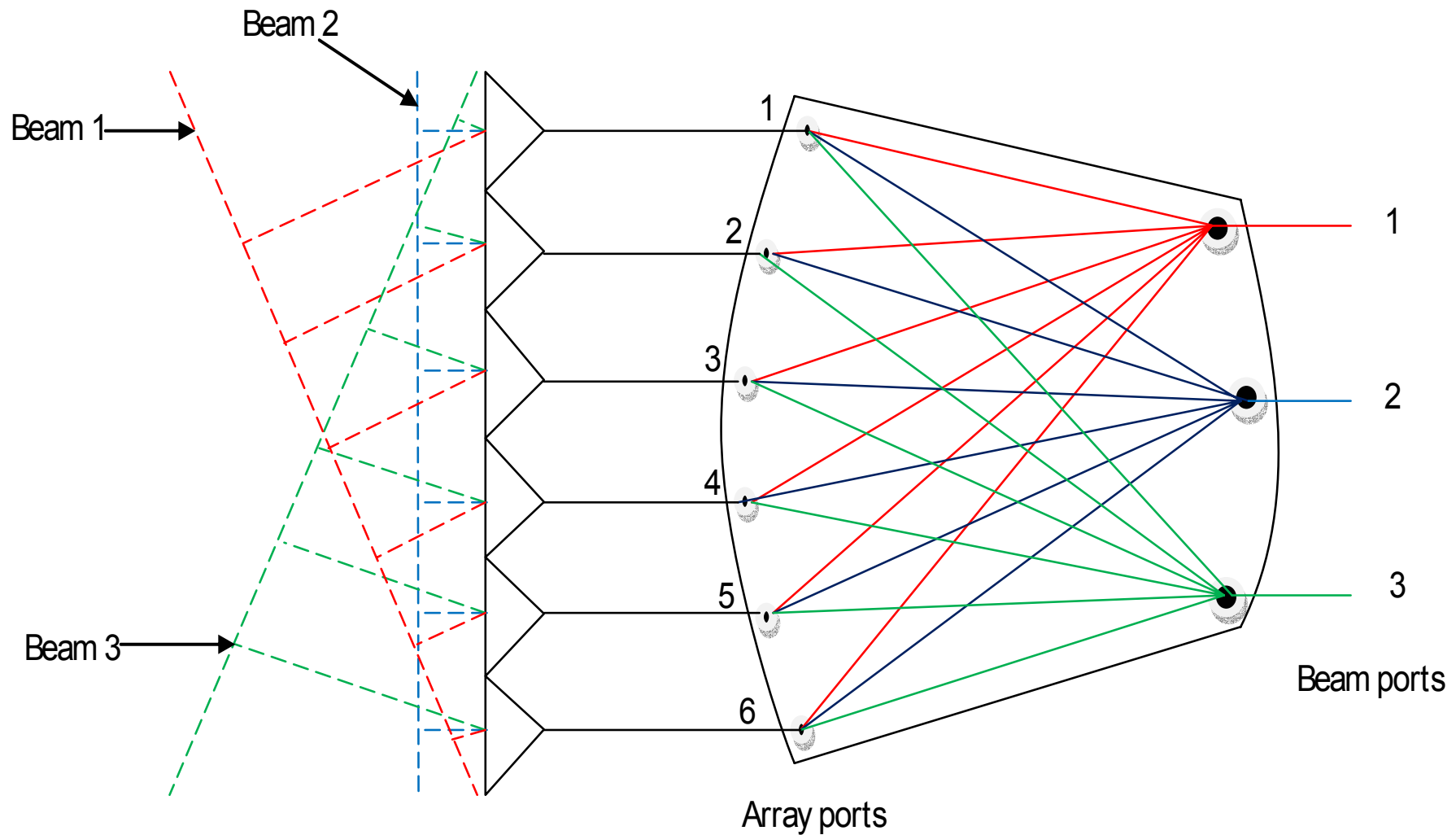
- Side lobe level improves (level of side lobe reduces with increase in element spacing).
- The sidelobe level of an array antenna fed by a Roman lens can be reduced if pairs of adjacent beam ports are combined.
- The system is fed through these summing ports.
- This procedure does not reduce the overall antenna gain.



# Reflections at the beam and array ports

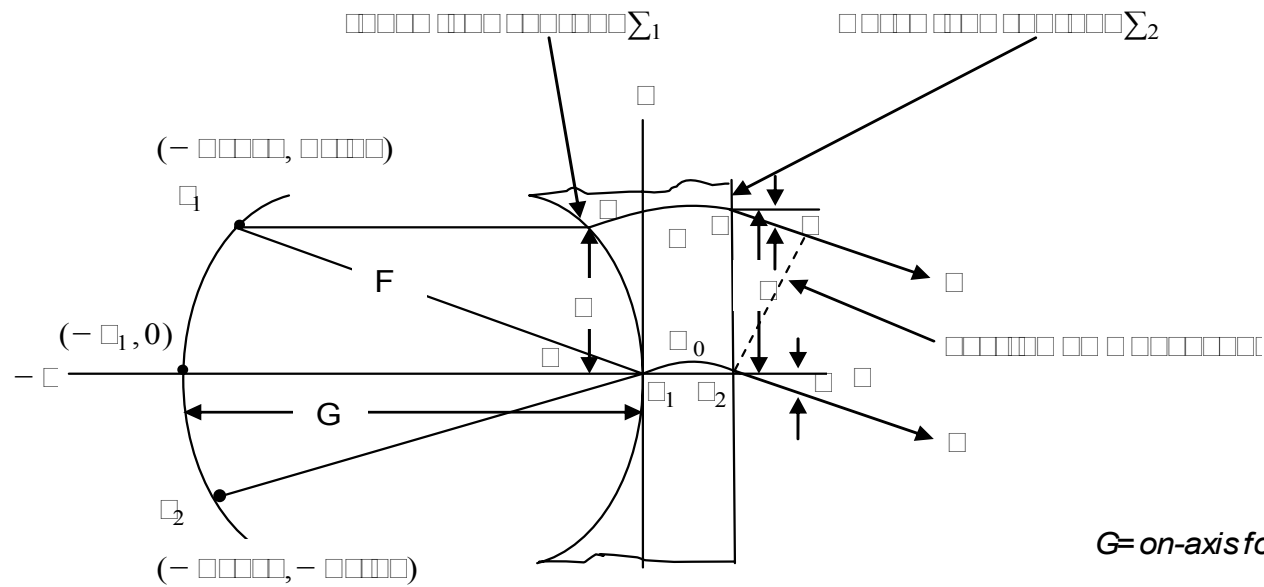
- Reflections at the beam and array ports can be reduced by introducing dummy ports that have matched resistance.
- Increasing element spacing, (but beam width decreases as element spacing increases).

# Rotman Lens beamformer



Linear array fed by a Rotman lens

# Rotman Lens parameters



$$\begin{array}{|c|c|} \hline \square & \square \\ \hline \end{array}_1 \begin{array}{|c|c|} \hline \square & \square \\ \hline \end{array} + \begin{array}{|c|} \hline \square \\ \hline \end{array} + \begin{array}{|c|c|c|c|} \hline \square & \square & \square & \square \\ \hline \end{array} = \begin{array}{|c|} \hline \square \\ \hline \end{array} + \begin{array}{|c|} \hline \square \\ \hline \end{array}_0 \quad (1)$$

$$\square_2 \square + \square - \square \square \square \square = \square + \square_0 \quad (2)$$

$$\square_1 \square + \square = \square + \square_0 \quad (3)$$

$G$  = on-axis focal length

$F$  = off-axis focal length

*W= electrical wire length*

 $\alpha = \text{scanning angle}$

# Rotman Lens Design Equations

$$\eta = N/F, \quad x = X/F, \quad y = Y/F, \quad w = \frac{W - W_0}{F}, \quad g = G/F,$$

$$a_0 = \cos \alpha, \quad b_0 = \sin \alpha$$

$$a = \left[ 1 - \eta^2 - \left( \frac{g-1}{g-a_0} \right)^2 \right]$$

$$b = \left[ 2g \left( \frac{g-1}{g-a_0} \right) - \frac{(g-1)}{(g-a_0)^2} b_0^2 \eta^2 + 2\eta^2 \right] - 2g$$

$$c = \left[ \frac{g b_0^2 \eta^2}{g-a_0} - \frac{b_0^4 \eta^4}{4(g-a_0)^2} - \eta^2 \right]$$

$$x = \frac{1-g}{g-a_0} w - \frac{b_0^2 \eta^2}{2(g-a_0)}$$

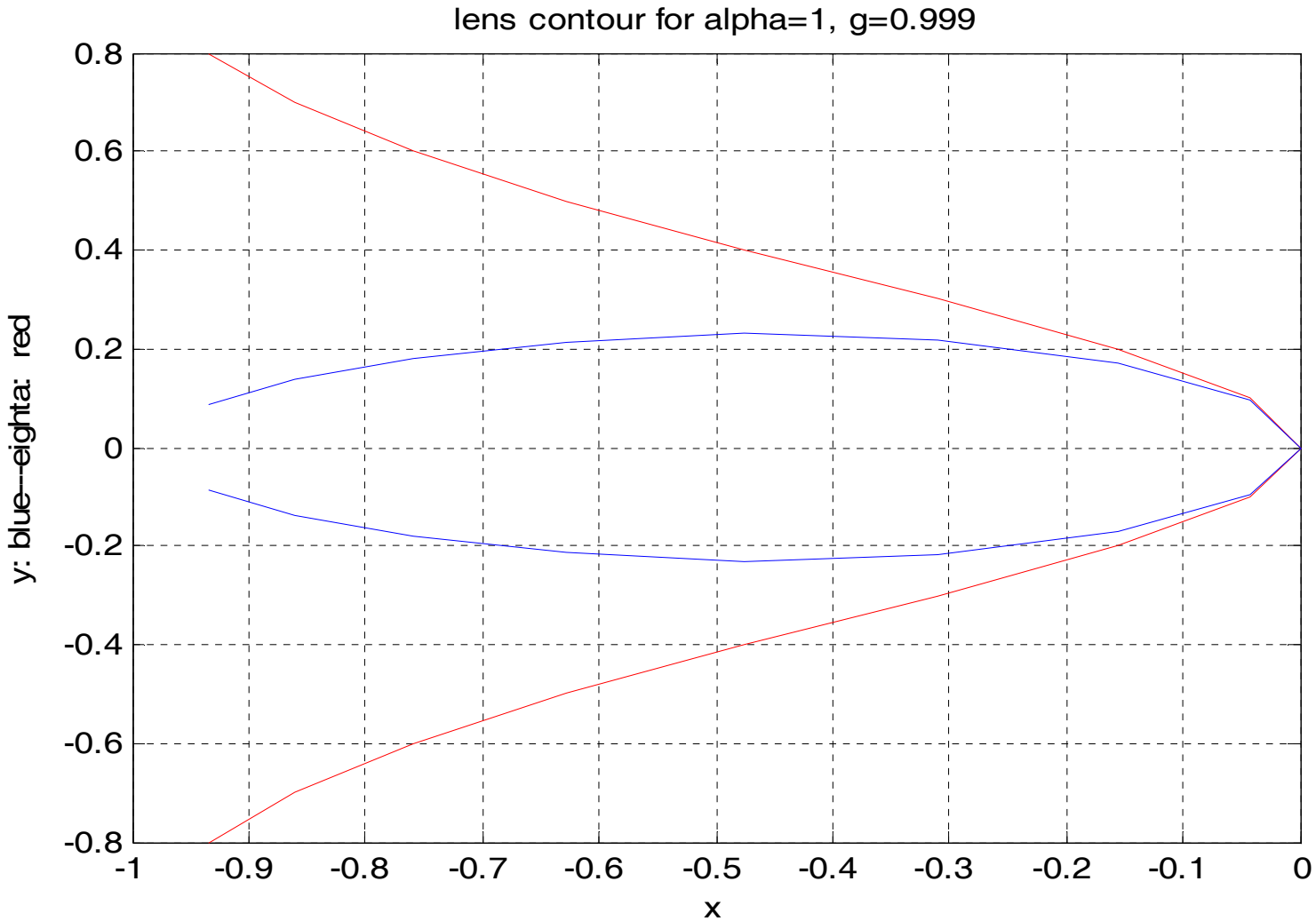
$$y = \eta(1-w)$$

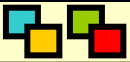
$$w = \frac{-b + \sqrt{b^2 - 4ac}}{2a}$$

We use Matlab to obtain the lens contours for different values of (G) and ( $\alpha$ ) as shown below.

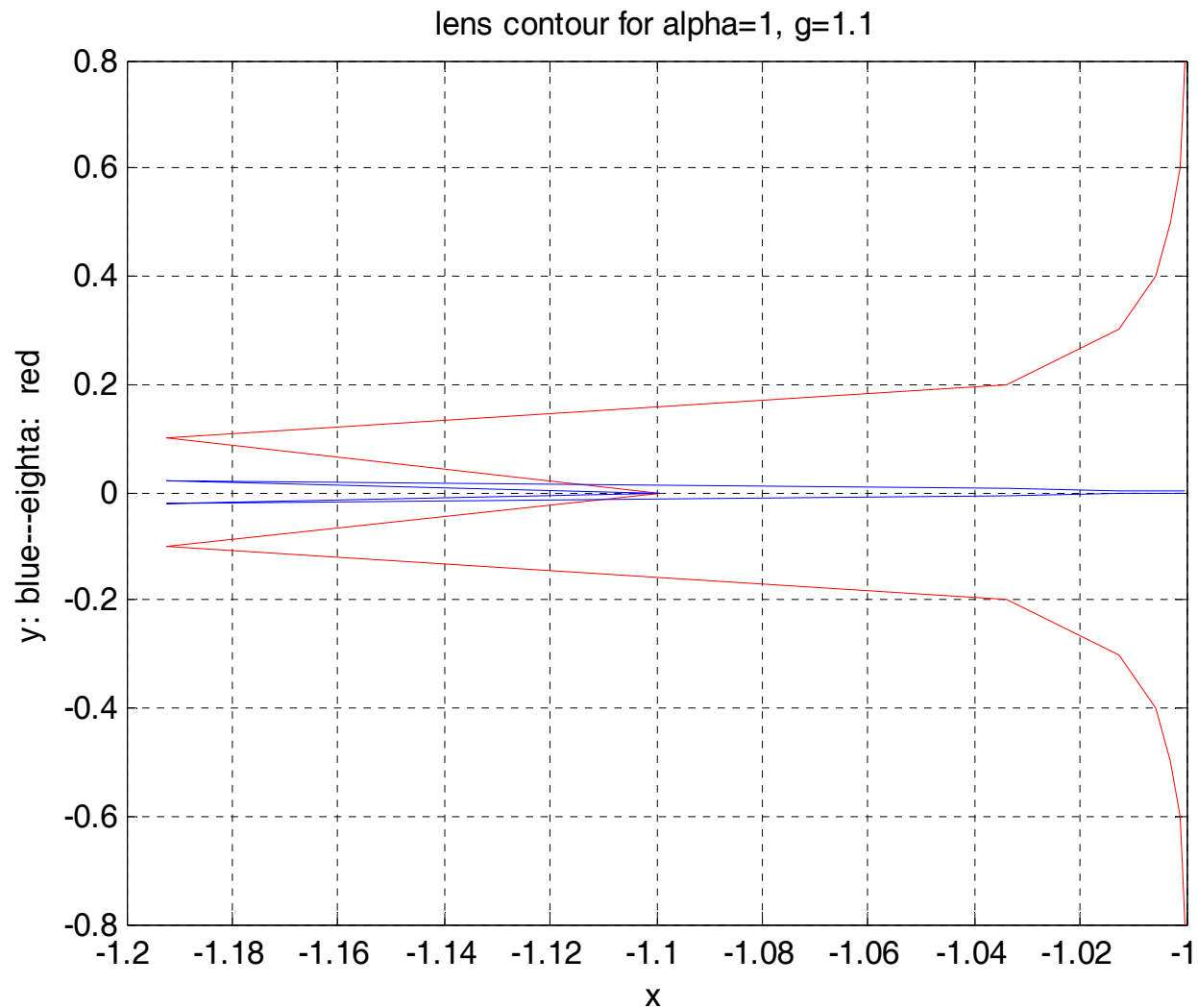


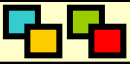
# Lens contour for $\alpha=1$ , $g=0.999$



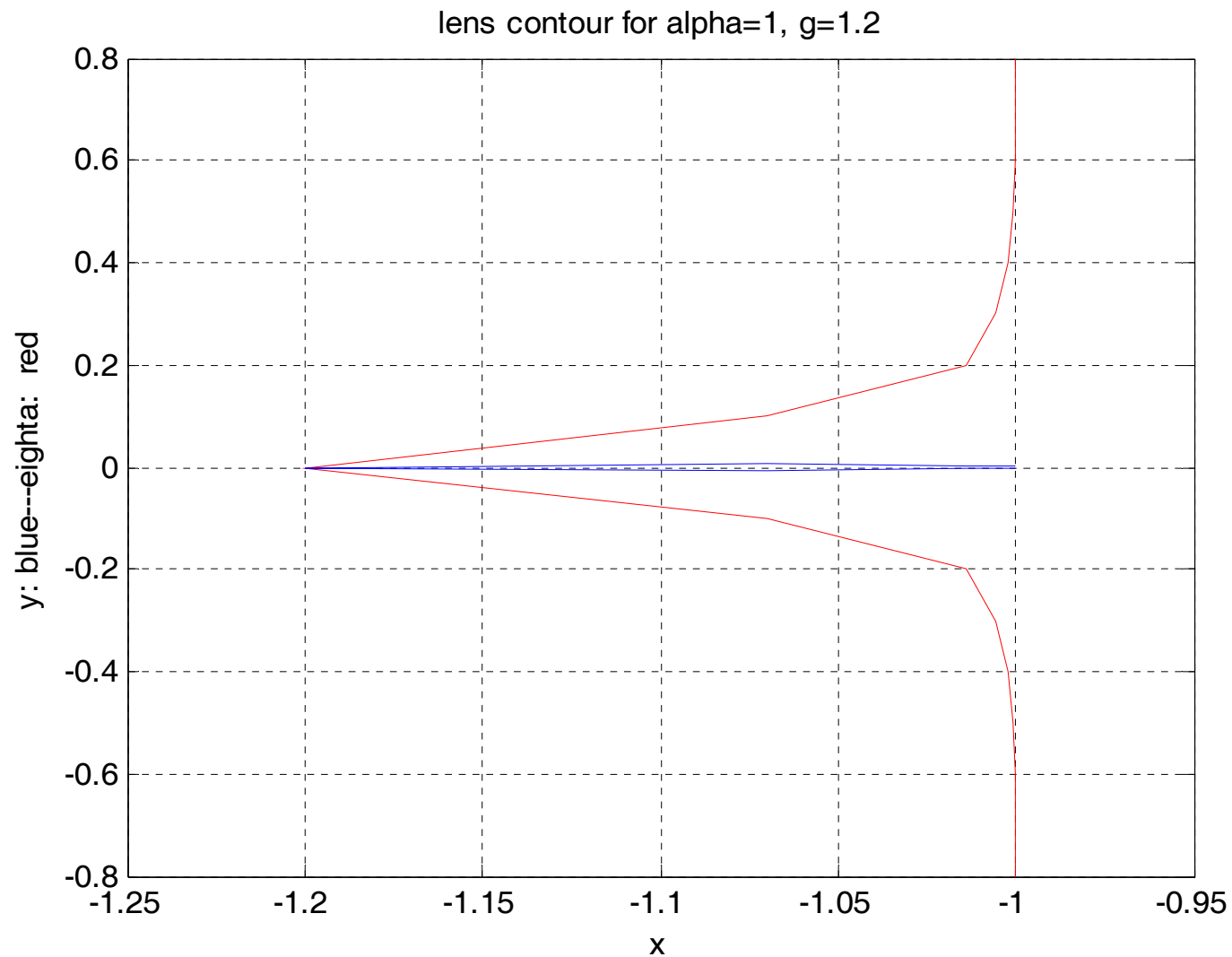


# Lens contour for $\alpha=1$ , $g=1.1$

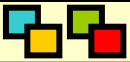




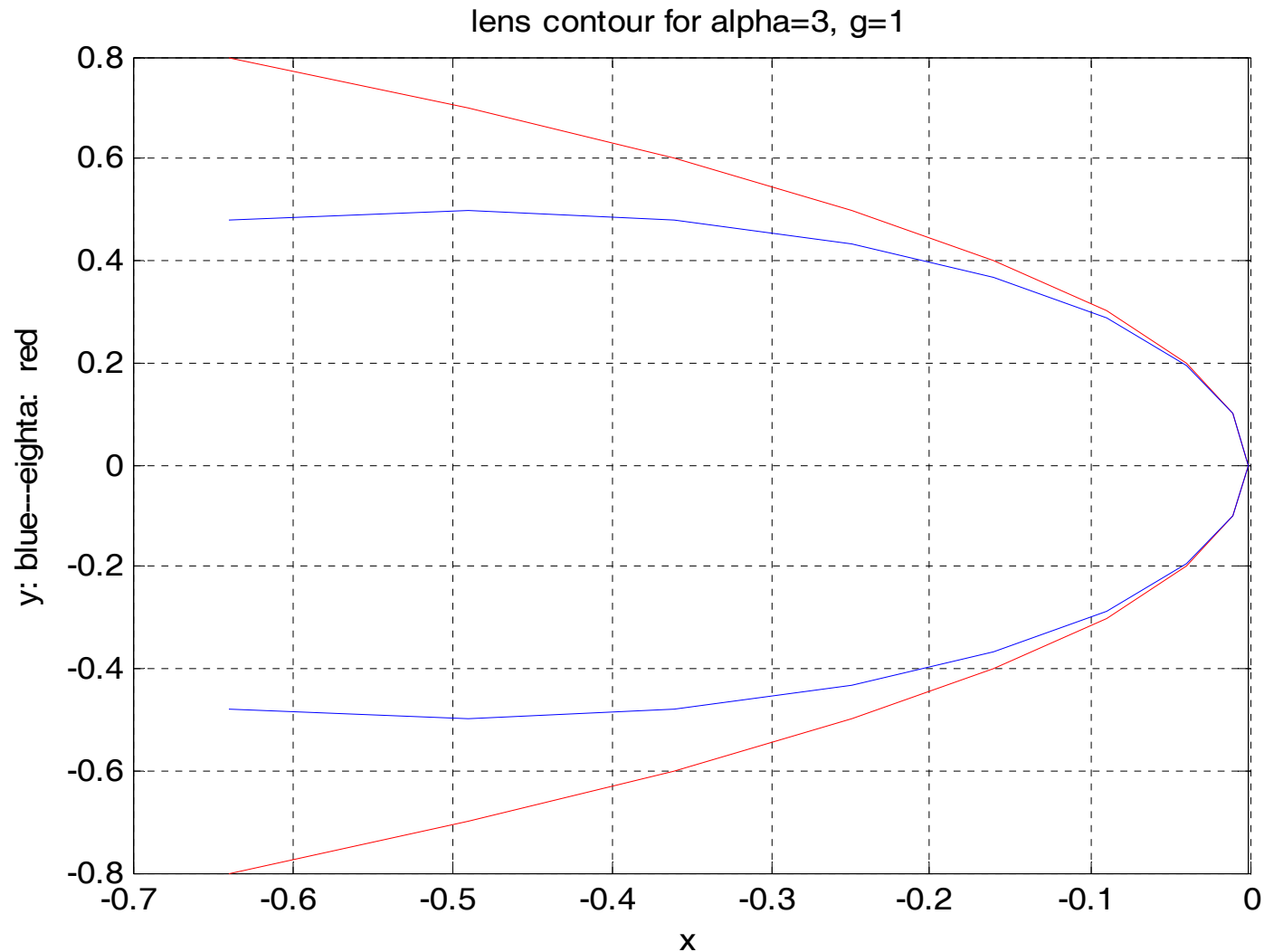
# Lens contour for $\alpha=1$ , $g=1.2$

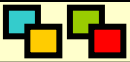




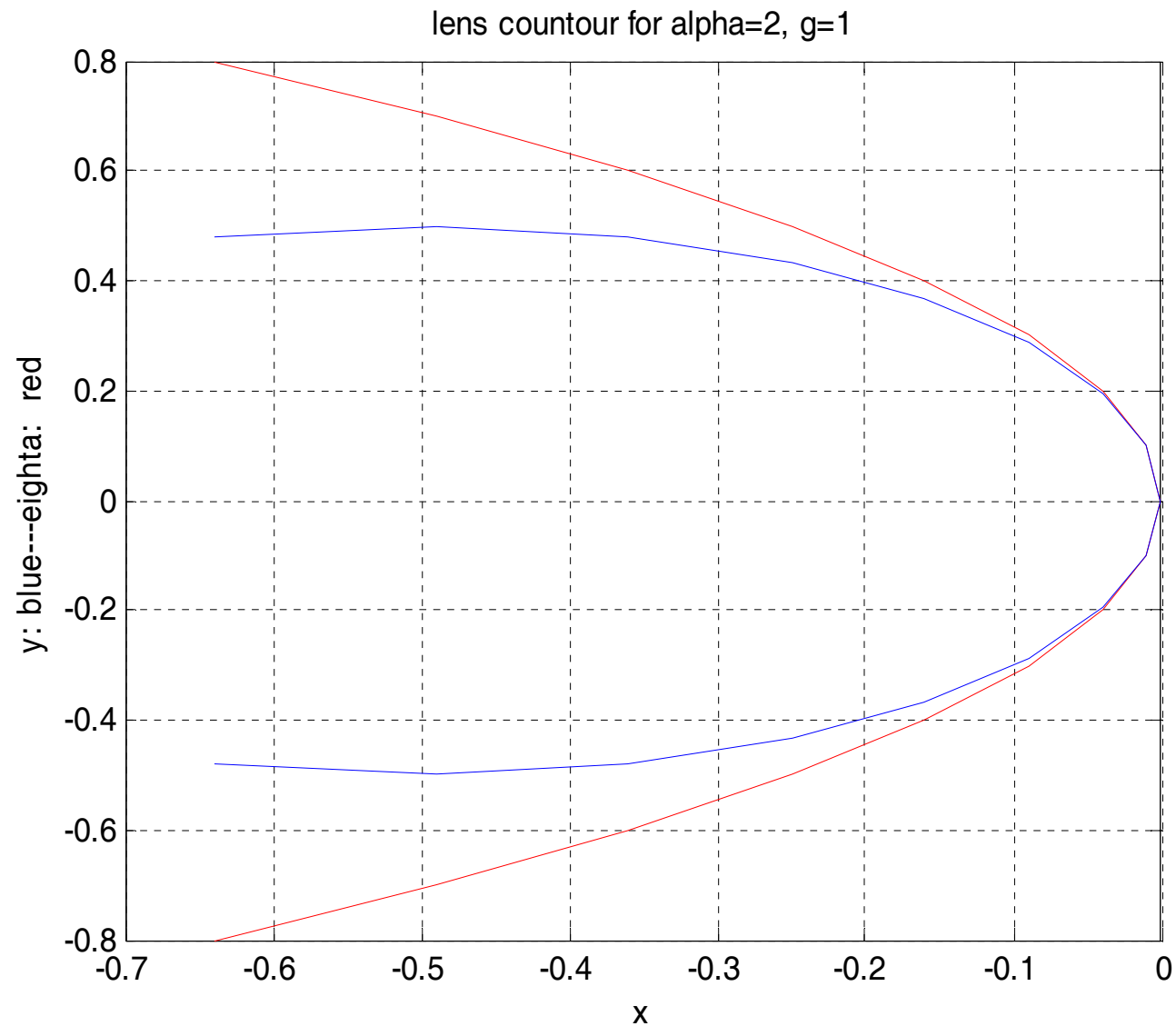


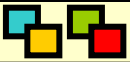
# Lens contour for $\alpha=3$ , $g=1$



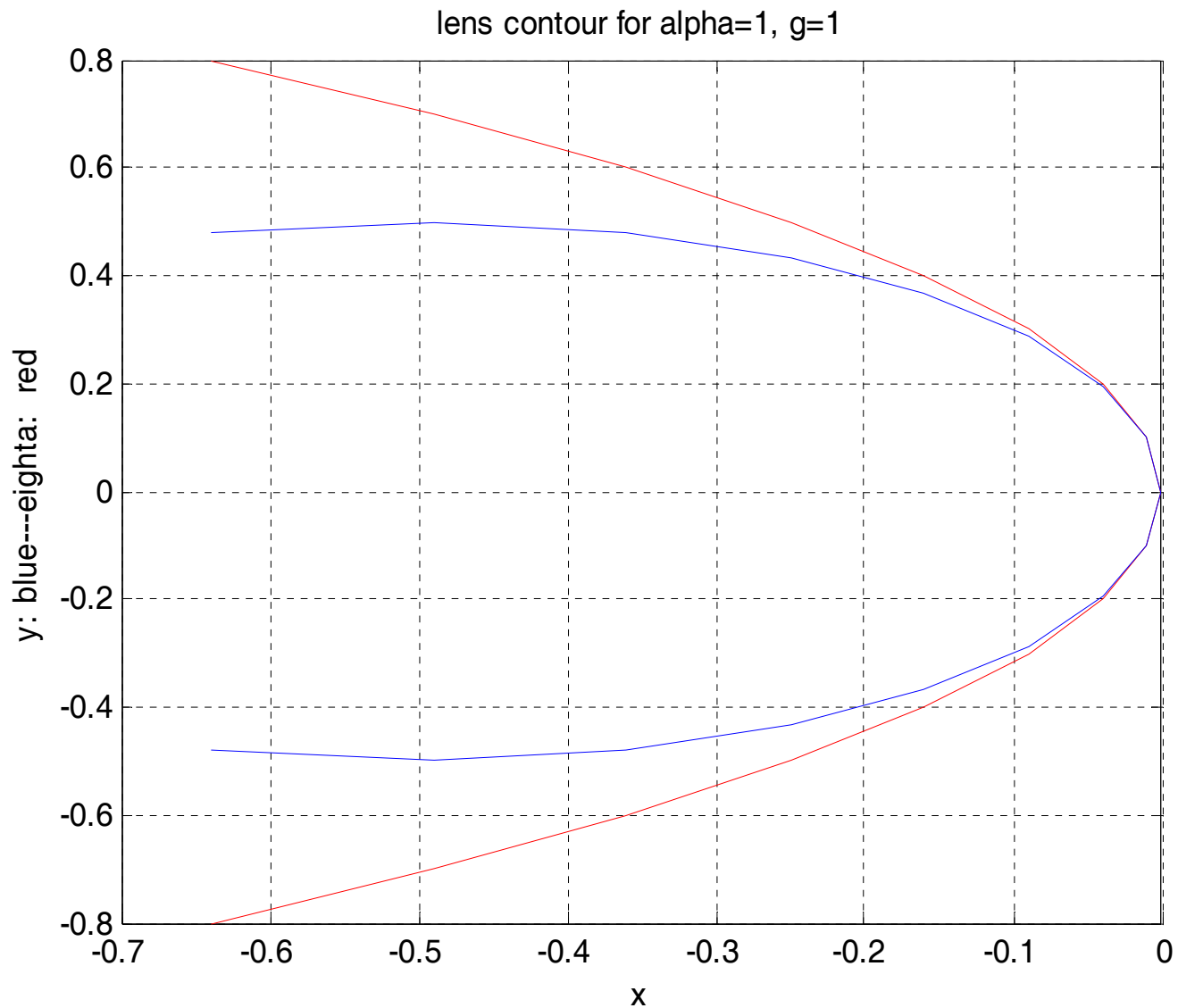


# Lens contour for $\alpha=2$ , $g=1$





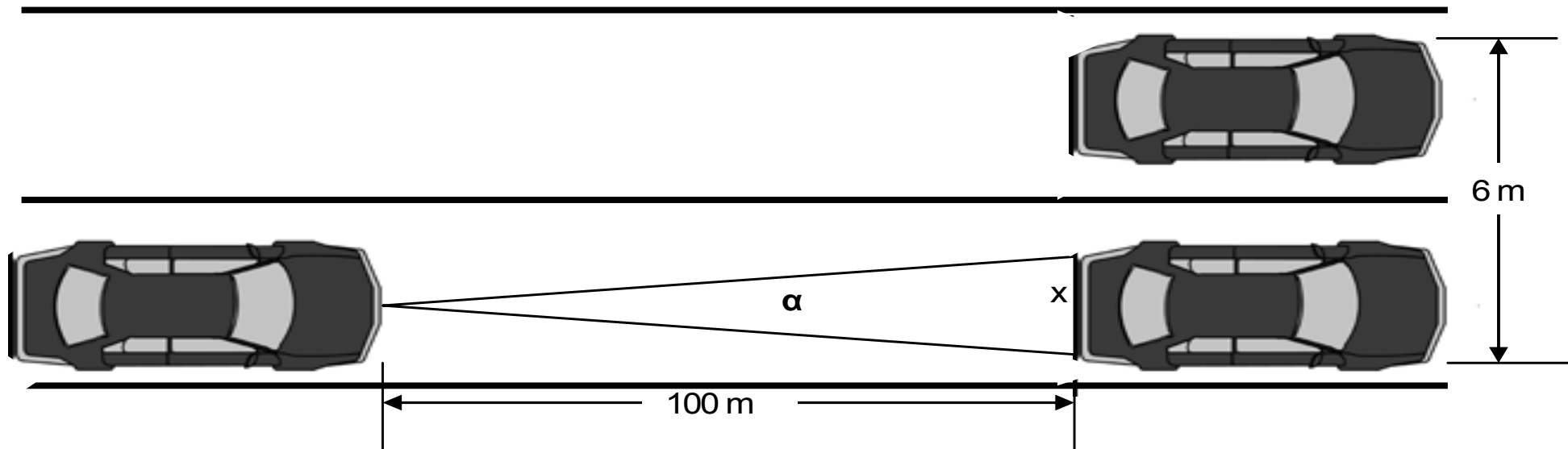
# Lens contour for $\alpha=1$ , $g=1$



# Why choosing $g=1$ and $\alpha=1$

- The focal arc is centered at the vertex  $O_1$  of the inner lens contour.
- The central ray paths from all points on the focal arc are equal in length.
- This setup of the lens is very important in monopulse applications.
- The shape of the lens will be approximated by a segment of a circle.
- The optical aberrations are are very low.

# Target scanning overview



If  $\alpha=1$        $x=1.75$  m

If  $\alpha=2$        $x=3.50$  m

If  $\alpha=3$        $x=5.24$  m

# Rotman Lens

We obtain the optimal values for the parameters from our plots as follows:

$$X = -0.63995$$

$$Y = 0.49994$$

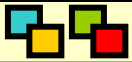
$$W = 1.6001$$

$$\theta_a = 0.7$$

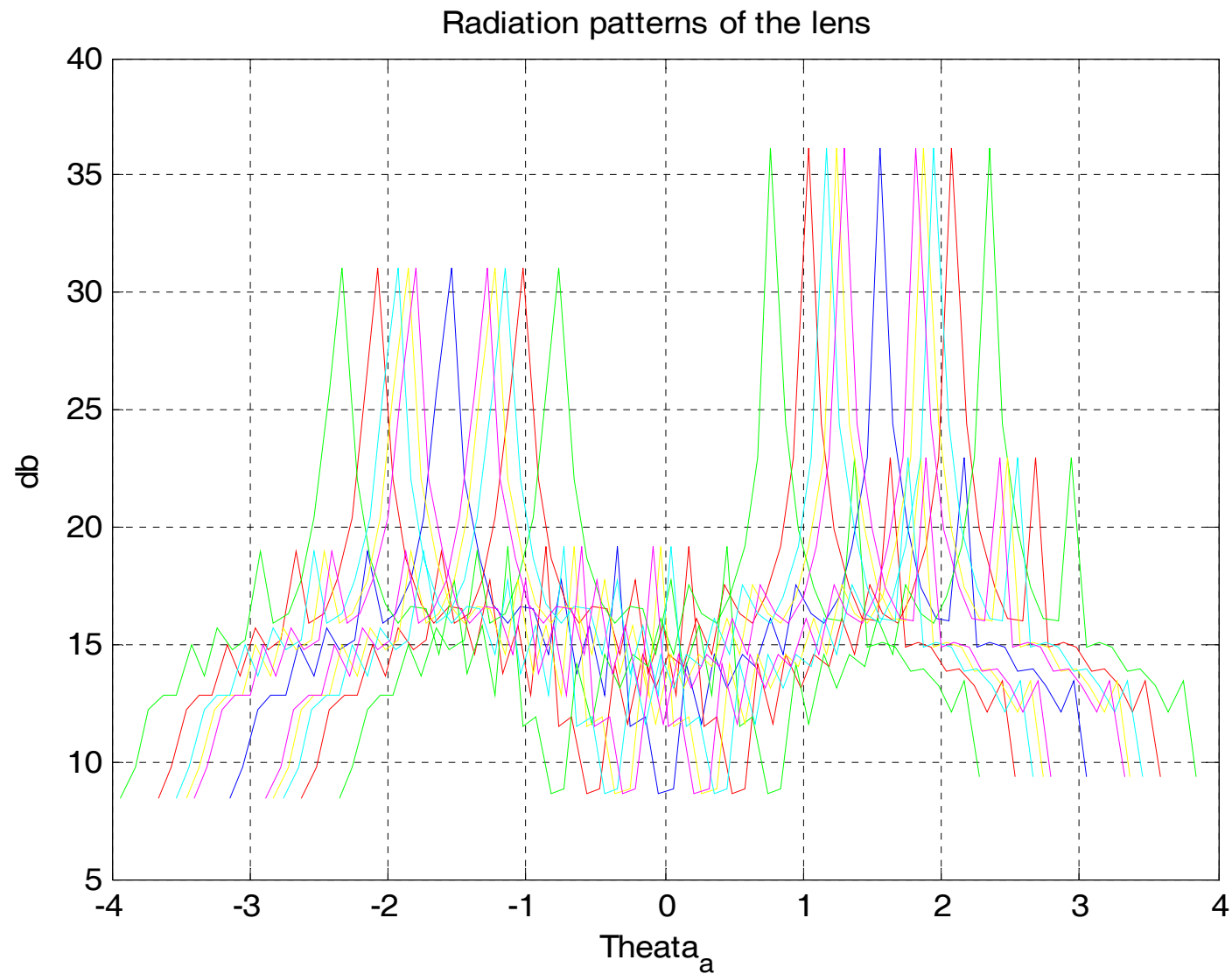
To find the radiation patterns of the lens, we have to plot gain of the lens versus the change of the array port angle.

$$\frac{P_r}{P_t} = \frac{l_a l_b}{\lambda} \cos^2 \theta_a \cos^2 \theta_b \left[ \frac{\sin (\pi l_a \sin \theta_a / \lambda)}{\pi l_a \sin \theta_a / \lambda} \right]^2 \left[ \frac{\sin (\pi l_b \sin \theta_b / \lambda)}{\pi l_b \sin \theta_b / \lambda} \right]^2$$

By taking the log of the equation above and plot it with the array angle, we got the following response.



# Radiation patterns of the lens



# Parameters calculation without insulation

we can define the length of the antenna array as

$$L_{1/2} = \frac{(N_a - 1)d}{2} = \frac{(N_a - 1)\lambda/2}{2} = \frac{(5 - 1)0.001945}{2} \\ = 3.89 \text{ mm for } N_a = 5$$

where

$L_{1/2}$  = half length of array,

$N_a$  = the number of array elements, and

$d$  = Spacing between antenna elements.

If we choose the maximum value of  $\eta$  the minimum value of scaling factor  $F$  can be obtained as

$$F_{min} = \frac{L_{1/2}}{\eta_{max}} = \frac{(N_a - 1)d}{2\eta_{max}} = \frac{0.00389}{0.7} = 0.00555 \text{ m} = 5.55 \text{ mm}$$



# Parameters calculation without insulation

$$\eta = N/F, \quad N = F\eta = 5.55 * 0.7 = 3.89 \text{ mm}$$

$$x = X/F, \quad X = Fx = 5.55 * 0.63995 = 3.55 \text{ mm}$$

$$y = Y/F, \quad Y = Fy = 5.55 * 0.49994 = 2.78 \text{ mm}$$

$$w = \frac{W - W_0}{F}, \quad W - W_0 = Fw = 6.94 * 1.6001 = 8.89 \text{ mm}$$

$$g = G/F, \quad G = Fg = 5.55 * 1 = 5.55 \text{ mm}$$

$$\text{When } g = 1 \text{ then } R = F = 5.55 \text{ mm}$$

$$\text{The arc between } F_1 \& F_2 = r\theta = 2 * \frac{\pi}{180} * 5.55 * 10^{-3} = 0.1937 \text{ mm}$$

# Parameters calculation with insulation

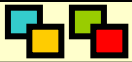
Using a high dielectric material **BaSrTiO<sub>3</sub>** (having a dielectric constant =6000), we can reduce the dimensions by a factor of  $\sqrt{\epsilon_r}$  .

Paralell plate region height

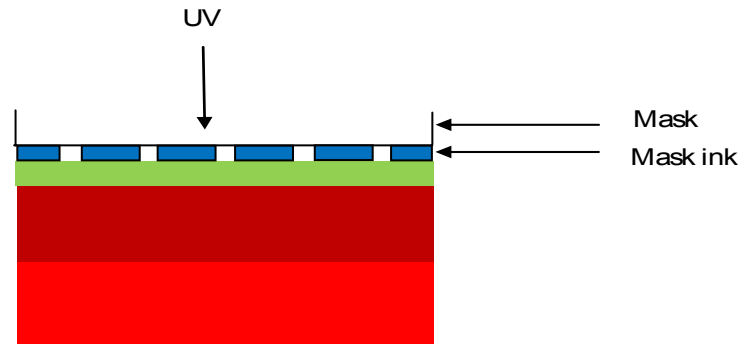
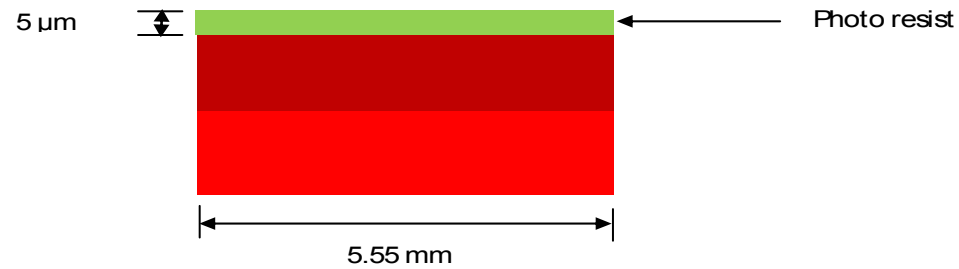
$$< \lambda_m/2 \quad (\lambda_m = \lambda_0 / \sqrt{\epsilon_r}) = \frac{c}{\sqrt{\epsilon_r} f} = \frac{3 \times 10^8}{\frac{77 \times 10^9}{\sqrt{6000}}} = 50 \mu m$$

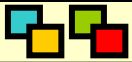
So the separation of the centres of the lens contours will be

$$\lambda_m/2 = 50/2 = 25 \mu m$$

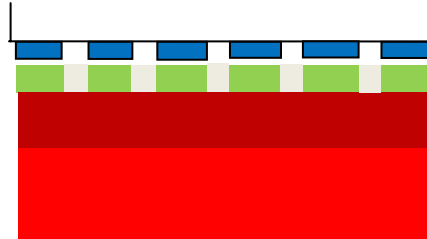


# Fabrication process

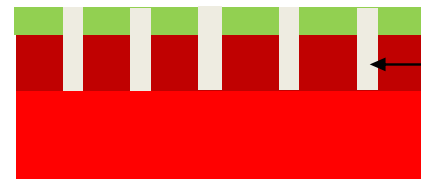




# Fabrication process



Photoresist dissolve



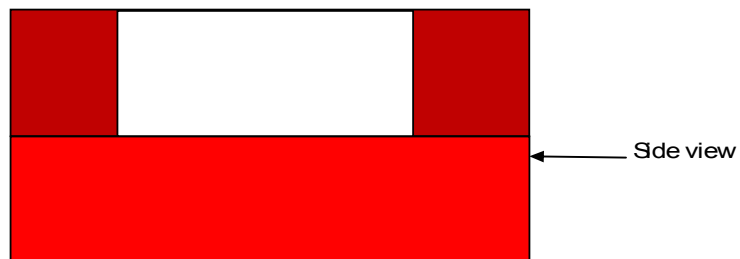
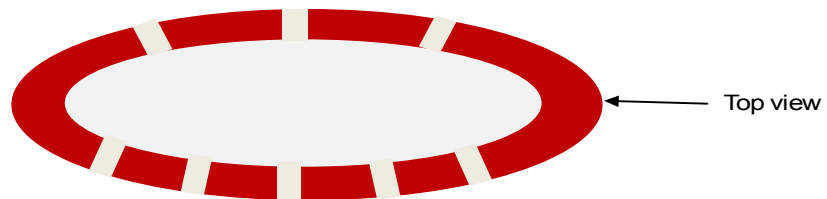
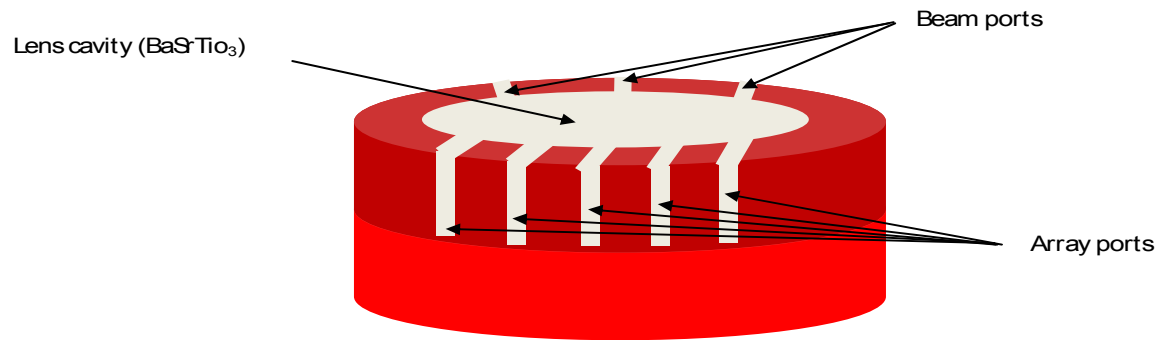
SiO<sub>2</sub> etch



Photoresist etch



# Fabrication process





# Target design specifications

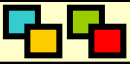
Building a safety belt around the car



Courtesy: Cambridge Consultants

# References

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2. D. Archer, "Lens-Fed Multiple Beam Arrays," *Microwave J.*, Vol. 18, 1975, pp. 37-42.
3. P. S. Simon, "Analysis and synthesis of Rotman lenses," in 22nd AIAA Internat. Comm. Satellite Sys. Conf., no. ALAA-2004-3196, 9-12 May 2004, Monterey, California.
4. Hansen, R. C., *Phased Array Antennas*, Wiley-Interscience, 1998, pp. 341-356.
5. Hansen, R. C., "Design trades for Rotman lenses," *IEEE Trans. Antennas Propagat.*, Vol. 39, No. 4, Apr. 1991, pp. 464-472.
6. Dielectric Slab Rotman Lens for Microwave/Millimeter-Wave Applications  
Jaeheung Kim, *Member, IEEE*, Choon Sik Cho, *Member, IEEE*, and Frank S. Barnes, *Life Fellow, IEEE*. IEEE TRANSACTIONS ON MICROWAVE THEORY AND TECHNIQUES, VOL. 53, NO. 8, AUGUST 2005



# THANK YOU