

Teoria dei sistemi.

Z transform

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- ▶ The Z-transform is the discrete time counter-part of the Laplace transform.

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Z Transform

- ▶ The Z-transform is the discrete time counterpart of the Laplace transform.
- ▶ the Z-Transform provides analytical methods for the solution of a difference equation,
- ▶ it offers direct insight into the transient and steady state behaviour of a DT signal
- ▶ it can be used to evaluate the stability of a system.

Definition

Definition of the Z-Transform

Definition

The definition of the z-Transform is the following:

$$\mathcal{Z}(f(t)) = \sum_0^{\infty} f(t)z^{-t}.$$

- ▶ This time we associate a DT signal with a function of the complex variable z .
- ▶ We will find it convenient to use the polar representation:
 $z = \rho e^{j\phi}.$

Z-Transform of $\mathbf{1}(t)$

$$\begin{aligned}\mathcal{Z}(\mathbf{1}(t)) &= \sum_0^{\infty} \mathbf{1}(t) z^{-t} \\ &= \sum_0^{\infty} z^{-t} \\ &= \lim_{H \rightarrow \infty} \sum_0^H z^{-t} \\ &= \lim_{H \rightarrow \infty} \frac{1 - z^{-H}}{1 - z^{-1}}.\end{aligned}$$

Z-Transform of $\mathbf{1}(t)$

Setting $z = \rho e^{j\theta}$, we have $z^{-H} = \rho^{-H} e^{-jH\theta}$. We have two cases:

$$\lim_{H \rightarrow \infty} z^{-H} = \begin{cases} 0 & \text{if } \rho = |z| > 1 \\ \infty & \text{if } \rho = |z| < 1 \\ e^{-jH\theta} & \text{if } \rho = |z| = 1 \end{cases}$$

$$\mathcal{Z}(\mathbf{1}(t)) = \lim_{H \rightarrow \infty} \frac{1 - z^{-H}}{1 - z^{-1}} = \begin{cases} \frac{1}{1 - z^{-1}} & \text{if } |z| > 1 \\ \text{is not defined} & \text{otherwise} \end{cases}$$

Z-Transform of $\mathbf{1}(t)a^t$

$$\begin{aligned}\mathcal{Z}(\mathbf{1}(t)a^t) &= \sum_0^{\infty} \mathbf{1}(t)a^t z^{-t} \\ &= \sum_0^{\infty} z^{-t} \\ &= \lim_{H \rightarrow \infty} \sum_0^H \left(\frac{a}{z}\right)^t \\ &= \lim_{H \rightarrow \infty} \frac{1 - \left(\frac{a}{z}\right)^H}{1 - \frac{a}{z}}.\end{aligned}$$

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- ▶ $a > 0$: Setting $z = \rho e^{j\theta}$, we have $\left(\frac{a}{z}\right)^H = \left(\frac{a}{\rho}\right)^H e^{-jH\theta}$.
- ▶ $a < 0$: we have $\left(\frac{a}{z}\right)^H = \left(\frac{a}{\rho}\right)^H e^{j\pi - jH\theta}$.

Z-Transform of $\mathbf{1}(t)a^t$

In conclusion:

$$\mathcal{Z}(\mathbf{1}(t)a^t) = \begin{cases} \frac{z}{z-a} & \text{if } |z| > |a| \\ \text{is not defined} & \text{otherwise} \end{cases}$$

Observations

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Observations

- ▶ The z-Transform of a DT signal is a function of a complex variable z
- ▶ As for the Laplace transform, it is not typically possible to define the z-Transform for all values of z , but only for a subset that we define Region of Convergence (ROC).
- ▶ Whereas for the Laplace transform the ROC is typically an half space (**Real** (s) $> \alpha$) for the z-Transform it is an annulus $|z| \geq \rho$.

Existence of the z-Transform

The existence of the z - Transform is guaranteed by the following:

Theorem

Theorem

Consider a function $f(t)$ and assume that one of the following limits exists:

$$R_f = \lim_{t \rightarrow \infty} |f(t)|^{1/t}$$

$$R_f = \lim_{t \rightarrow \infty} \frac{f(t+1)}{f(t)}.$$

Then:

- 1. the z-Transform $\mathcal{Z}(f(t))$ exists and converges for $|z| \geq R_f$.*
- 2. the z-Transform is analytic, i.e., continuous and infinitely differentiable w.r.t. z , for $|z| \geq R_f$.*

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- ▶ Suffice it to say that the existence of a limit $R_f = \lim_{t \rightarrow \infty} |f(t)|^{1/t}$ is equivalent to that of an exponential upper bound for the function

$$f(t) \leq AR_f^t,$$

for some A .

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- ▶ There is an inverse theorem, as shown next.

Inverse theorem

Invertibility of the z- Transform

Theorem

If $F(z)$ and $G(z)$ are z-Transforms of two functions $f(t)$ and $g(t)$ and if $F(z) = G(z)$ for all $|z| > R$, for some $R > 0$ then $f(t) = g(t)$ for $t = 0, 1, 2, \dots$

Proof

- ▶ If $G(z) = F(z)$ for all $|z| > R$ then

$$\sum_{t=0}^{\infty} f(t)z^{-t} = \sum_{t=0}^{\infty} g(t)z^{-t} \leftrightarrow$$
$$\sum_{t=0}^{\infty} (f(t) - g(t)) z^{-t} = 0 \leftrightarrow \sum_{t=0}^{\infty} a_t w^t = 0$$

where we have set $w = 1/z$ and $a_t = f(t) - g(t)$.

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- ▶ It is known that if we have a power series $\sum_{t=0}^{\infty} a_t w^t$ then if $\sum_{t=0}^{\infty} a_t w^t = 0$ for all $|w| \leq W$ then $a_t = 0$.

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- ▶ It is known that if we have a power series $\sum_{t=0}^{\infty} a_t w^t$ then if $\sum_{t=0}^{\infty} a_t w^t = 0$ for all $|w| \leq W$ then $a_t = 0$.
- ▶ Using this result, we conclude $f(t) = g(t)$.

Inverse z - Transform

Inverse z- Transform

Theorem

the inverse transform of the z-Transform is defined by the following line integral.

$$f(t) = \mathcal{Z}^{-1}(F(z)) = \oint_{|z|=R} F(z)z^{t-1}dz$$

where the line integral is computed along a circle included in the ROC.

Observations

- ▶ This formula is impractical and the inverse z-Transform is best computed using a different procedure
- ▶ However, this formula reveals that the z-Transform is in some sense a decomposition of a function using a basis of functions of type z^n .

Properties of the z-Transform

- ▶ The properties of z - Transform are very similar to the ones of the Laplace transform
- ▶ We state them in the following theorem

Properties of the z - Transform

Properties of the z- Transform (1)

Theorem

Let $X(z)$ have ROC R , $X_1(z)$ have ROC R_1 and $X_2(z)$ have ROC R_2 .

Linearity $\mathcal{Z}(\alpha_1 x_1(t) + \alpha_2 x_2(t)) = \alpha_1 X_1(z) + \alpha_2 X_2(z)$
(ROC $R' = R_1 \cap R_2$)

Time-shifting if $x(t)$ is a causal signal and $k > 0$ then
 $\mathcal{Z}(x(t - k)) = z^{-k} X(z)$ (ROC
 $R' \supset R \cap \{0 < |z| < \infty\}$, and $\mathcal{Z}(x(t + k)) =$
 $z^k X(z) - z^k x(0) - z^{k-1} x(1) - \dots - z x(K - 1)$
(ROC $R' = R$)

Multiplication by exponential $\mathcal{Z}(z_0^t x(t)) = X(\frac{z}{z_0})$ (ROC
 $R' = |z_0| R$)

Multiplication by t $\mathcal{Z}(tx(t)) = -z \frac{dX(z)}{dz}$ (ROC $R' = R$)

Properties of the z - Transform

Properties of the z- Transform (2)

Theorem

Let $X(z)$ have ROC R , $X_1(z)$ have ROC R_1 and $X_2(z)$ have ROC R_2 .

Time Scaling $\mathcal{Z}(x(t/t_0)) = X(z^{t_0})$ for t_0 positive integer.

Convolution $\mathcal{Z}(x_1(t) * x_2(t)) = X_1(z)X_2(z)$. (ROC $R' \supset R_1 \cap R_2$).

Accumulation $\mathcal{Z}(\sum_{\tau=0}^t x(\tau)) = \frac{z}{z-1}X(z)$ (ROC $R' = R \cap \{|z| > 1\}$)

Initial value $x(0) = \lim_{z \rightarrow \infty} X(z)$

Final value $x(\infty) = \lim_{z \rightarrow 1} (z-1)X(z)$, if $x(\infty)$ exists.

Proof

The proof very similar to Laplace transform

We give a couple of examples.

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Time shifting (with positive shift)

$$\begin{aligned}\mathcal{Z}(x(t+k)) &= \sum_{t=0}^{\infty} x(t+k)z^{-t} = \sum_{t=0}^{\infty} x(t+k)z^{-(t+k)}z^k = \\ &= z^k \sum_{t'=k}^{\infty} x(t')z^{-t'} = \\ &= z^k \left(\sum_{t'=0}^{\infty} x(t')z^{-t'} \right) - x(0)z^k - x(1)z^{k-1} - \dots - x(k-1)z \\ &= z^k X(z) - x(0)z^k - x(1)z^{k-1} - \dots - x(k-1)z\end{aligned}$$

Proof

Convolution (case of causal signals)

$$\begin{aligned}\mathcal{Z}(x_1(t) * x_2(t)) &= \sum_{t=0}^{\infty} x_1(t) * x_2(t) z^{-t} = \sum_{t=0}^{\infty} \sum_{\tau=0}^{\infty} x_1(\tau) x_2(t - \tau) z^{-t} = \\&= \sum_{\tau=0}^{\infty} \sum_{t=0}^{\infty} x_1(\tau) x_2(t - \tau) z^{-t} = \sum_{\tau=0}^{\infty} x_1(\tau) \sum_{t=0}^{\infty} x_2(t - \tau) z^{-t} = \\&= \sum_{\tau=0}^{\infty} x_1(\tau) \sum_{t=0}^{\infty} x_2(t - \tau) z^{-(t-\tau)} z^{-\tau} \\&= \sum_{\tau=0}^{\infty} x_1(\tau) z^{-\tau} \sum_{t=-\tau}^{\infty} x_2(t') z^{-t'} = \\&= X_1(z) X_2(z)\end{aligned}$$

Example

We can use the properties to construct complex z - Transform from simpler ones.

$$s(t) = \mathbf{1}(t) \cos \Omega t$$

We can use the properties as follows:

$$\begin{aligned}\mathcal{Z}(\mathbf{1}(t) \cos \Omega t) &= \frac{1}{2} \mathcal{Z}(\mathbf{1}(t) e^{j\Omega t}) + \frac{1}{2} \mathcal{Z}(\mathbf{1}(t) e^{-j\Omega t}) \\&= \frac{1}{2} \frac{z}{z - e^{j\Omega}} + \frac{1}{2} \frac{z}{z - e^{-j\Omega}} = \\&= \frac{1}{2} \frac{z(z - e^{j\Omega} + z - e^{-j\Omega})}{z^2 - z(e^{j\Omega} + e^{-j\Omega}) + 1} \\&= \frac{z(z - \cos \Omega)}{z^2 - 2z \cos \Omega + 1}\end{aligned}$$

Transform table

We can proceed in a similar way to compute a table of transforms.

z-Transform of known Signals		
Signal	z-Tranform	ROC
$\delta(t)$	1	$z \in \mathbb{C}$
$\delta(t - t_0)$	z^{-t_0}	$z \in \mathbb{C} \setminus 0$
$\mathbf{1}(t)$	$\frac{z}{z-1}$	$ z > 1$
t	$\frac{z}{(z-1)^2}$	$ z > 1$
t^2	$\frac{z(z+1)}{(z-1)^3}$	$ z > 1$
a^t	$\frac{z}{(z-a)}$	$ z > a $
a^t	$\frac{z}{(z-a)}$	$ z > a $
ta^t	$\frac{az}{(z-a)^2}$	$ z > a $
t^2a^t	$\frac{az(z+a)}{(z-a)^3}$	$ z > a $

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z-Transform of known Signals		
Signal	z-Transform	ROC
$\cos \Omega t$	$\frac{z(z - \cos \Omega)}{z^2 - 2z \cos \Omega + 1}$	$ z > 1$
$\sin \Omega t$	$\frac{z \sin \Omega}{z^2 - 2z \cos \Omega + 1}$	$ z > 1$
$a^t \cos \Omega t$	$\frac{z(z - a \cos \Omega)}{z^2 - 2az \cos \Omega + a^2}$	$ z > a$
$a^t \sin \Omega t$	$\frac{za \sin \Omega}{z^2 - 2az \cos \Omega + a^2}$	$ z > a$

Application to difference equations

An interesting application is shown through the different example.

Difference Equations

Consider the following difference equation.

$$y(t+2) = 3y(t+1) - 2y(t) + u(t+1) - 3u(t).$$

Let us find $Y(z)$ for $u(t) = \mathbf{1}(t)$, $y(1) = 1$, $y(0) = -1$, $u(0) = 0$.

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- The z - Transform is the following

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- ▶ The z - Transform is the following

$$\mathcal{Z}(y(t+2)) = \mathcal{Z}(3y(t+1) - 2y(t) + u(t+1) - u(t)).$$

- ▶ Application of time shifting rule

$$\begin{aligned} z^2 Y(z) - z^2 y(0) - zy(1) = \\ 3zY(z) - 3zy(0) - 2Y(z) + zU(z) - zu(0) - 3U(z) \end{aligned}$$

Applications to difference equations

- The latter equation becomes

$$\begin{aligned} Y(z) &= \frac{U(z)(z-3)}{z^2-3z+2} + \frac{z^2y(0) + z(y(1) - 3y(0) - u(0))}{z^2-3z+2} \\ &= \frac{z(z-2)}{(z-1)(z^2-3z+2)} + \frac{z^2y(0) + z(y(1) - 3y(0) - u(0))}{z^2-3z+2}. \end{aligned}$$

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- ▶ We will soon see how to invert this.

Inversion of the z - Transform

- ▶ Also for LTI DT systems the evolution of the system is compounded by a free evolution and by a forced evolution.
- ▶ We have to deal with a ratio of polynomial with the numerator that is typically proportional to z .
- ▶ One of the possible ways to deal with this case is by using the same technique (partial fraction expansion) that we have used for the Laplace transform (with some care)
- ▶ We will see this through some examples

Back to example

Free evolution

Let us go to the example above and compute the free evolution for $y(1) = 1$, $y(0) = -1$, $u(0) = 0$.

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- ▶ the z - Transform

$$\begin{aligned} Y(z) &= \frac{z^2 y(0) + z(y(1) - 3y(0) - u(0))}{z^2 - 3z + 2} \\ &= \frac{-z^2 + 4z}{(z - 2)(z - 1)} \end{aligned}$$

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- ▶ the z - Transform

$$\begin{aligned} Y(z) &= \frac{z^2 y(0) + z(y(1) - 3y(0) - u(0))}{z^2 - 3z + 2} \\ &= \frac{-z^2 + 4z}{(z-2)(z-1)} \end{aligned}$$

- ▶ It is convenient to divide by z and then proceed with partial fraction expansion:

$$\frac{Y(z)}{z} = \frac{-z + 4}{(z-2)(z-1)} = \frac{2}{z-2} - \frac{3}{z-1}$$

Back to the example

- And finally....

$$Y(z) = \frac{2z}{z-2} - \frac{3z}{z-1}$$
$$y(t) = \mathbf{1}(t) (2 \cdot 2^t - 3) .$$

Back to the example

- And finally....

$$Y(z) = \frac{2z}{z-2} - \frac{3z}{z-1}$$
$$y(t) = \mathbf{1}(t) (2 \cdot 2^t - 3) .$$

- Now let us compute the response for $u(t) = \mathbf{1}(t)$ Let us compute the forced evolution for $u(t) = \mathbf{1}(t)$.
- z -Transform

$$Y(z) = \frac{z(z-3)}{(z-1)^2(z-2)} .$$

Back to the example

- It is convenient to divide by z and then proceed with partial fraction expansion:

$$\begin{aligned}\frac{Y(z)}{z} &= \frac{(z-3)}{(z-1)^2(z-2)} = \\ &= \frac{A_{1,1}}{(z-1)^2} + \frac{A_{1,2}}{z-1} + \frac{A_2}{z-2} = \\ &= \frac{A_{1,1}}{(z-1)^2} + \frac{A_{1,2}}{z-1} + \frac{A_2}{z-2} = \\ &= \frac{2}{(z-1)^2} + \frac{1}{z-1} - \frac{1}{z-2} =\end{aligned}$$

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- Which leads to

$$\begin{aligned}Y(z) &= \frac{2z}{(z-1)^2} + \frac{z}{z-1} - \frac{z}{z-2} \\ y(t) &= \mathbf{1}(t) (t+1-2^t)\end{aligned}$$