# Digital Communications - HW4 - MATLAB Code

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#### AWGN, DFE, OFDM systems

```
clear all; close all; clc;
format long g;
l = [PN(20); PN(20)];
sstep = 32400;
num bits = floor(length(b l) / sstep) * sstep;
b l = b l(1:num bits + 54);
sigma \ a = 2;
% ENCODE VIA LDPC
%create LDPC encoder with the dafault matrix (rate=1/2)
encoderLDPC = comm.LDPCEncoder;
enc b l = zeros(2*length(b l),1);
for i = 0: (floor(length(b_l)/sstep))-1
    %encodes block by block the input bits
    %block length is equal to 32400
    block = b l(i * sstep + 1:i * sstep + sstep);
    enc b l(2 * i * sstep + 1:2 * i * sstep + 2 * sstep) = ...
        step(encoderLDPC, block);
end
% INTERLEAVING
interl b l = interl(enc b l);
\% MAP 0 TO -1
interl b l = 2 * interl b l - 1;
a = zeros(floor(length(interl_b_l)/2),1);
for \quad i=1:2:(\,lengt\,h\,(\,int\,erl\,\_\,b\,\_\,l\,)\,\,-\,\,1)
   a((i+1)/2) = interl_b_l(i) + 1i * interl_b_l(i+1);
end
```

```
clear pn;
SNR \ vector = [0.4:0.05:1];
Pbit AWGN = zeros(length(SNR vector), 1);
for snr index = 1:length(SNR vector)
SNR = SNR \ vector(snr \ index);
% Add Gaussian Noise
\operatorname{snrlin} = 10^{(SNR/10)};
sigma_w = sigma_a * E_qc / snrlin;
w = wgn(length(a), 1, 10 * log10(sigma w), 'complex');
y k = a + w;
% Compute Log Likelihood Ratio
llr = zeros(2*length(y_k),1);
llr(1:2:end, 1) = -2*real(y_k)/(sigma_w/2);
llr(2:2:end, 1) = -2*imag(y k)/(sigma w/2);
new length = floor (length (llr) / 64800) * 64800;
llr = llr (1:new length, 1);
% Decode the bits
llr = deinterleaver(llr); % Deinterleave the loglikelihood ratio first
decoderLDPC = comm.LDPCDecoder;
dstep = 2 * sstep;
tic
for i = 0: (floor(length(llr)/dstep)) - 1
    %decodes block by block the input bits
    %block length is equal to 64800
    block = llr(i * dstep + 1:i * dstep + dstep);
    dec_b_l(i * dstep / 2 + 1:i * dstep / 2 + dstep / 2) = ...
        step(decoderLDPC, block ');
end
toc
[Pbit AWGN(snr index), errors] = BER(dec b l, b l(1:length(dec b l)));
end
clear all; close all; clc;
format long g;
\% THIS SCRIPT GENERATES THE SYMBOLS BY USING A PN SEQUENCE
% AND APPLYING LDPC, ENCODING AND INTERLEAVING
b l = [PN(20); PN(20)];
sstep = 32400;
num bits = floor(length(b l) / sstep) * sstep;
b l = b l(1:num bits + 54);
```

```
sigma \ a = 2;
% ENCODE VIA LDPC
%create LDPC encoder with the dafault matrix (rate=1/2)
encoderLDPC = comm.LDPCEncoder;
enc b l = zeros(2*length(b l),1);
for i = 0: (floor(length(b l)/sstep))-1
    %encodes block by block the input bits
    %block length is equal to 32400
     block = b_l(i * sstep + 1:i * sstep + sstep);
     enc_b_1(2 * i * sstep + 1:2 * i * sstep + 2 * sstep) = ...
         step (encoderLDPC, block);
end
% INTERLEAVING
interl b l = interl(enc b l);
\% MAP 0 TO -1
interl b l = 2 * interl b l - 1;
% Generate the channel input response
[q_c, E_qc] = channel_impulse_response();
a = zeros(floor(length(interl_b_l)/2),1);
for i=1:2:(length(interl b l) - 1)
   a((i+1)/2) = interl_b_l(i) + 1i * interl_b_l(i+1);
end
clear pn;
a prime = upsample(a, 4);
% Filter through the channel
s_c = filter(q_c, 1, a_prime);
SNR \text{ vector} = [1.4:0.025:1.75];
Pbit DFEenc = zeros(length(SNR vector),1);
for snr_index = 1:length(SNR vector)
    SNR = SNR\_vector(snr\_index);
% Add Gaussian Noise
snrlin = 10^{(SNR/10)};
sigma w = sigma a * E qc / snrlin;
\mathbf{w} = \operatorname{wgn}(\operatorname{length}(\mathbf{s}_{\underline{\phantom{a}}}\mathbf{c}), 1, 10 * \log 10(\operatorname{sigma}_{\underline{\phantom{a}}}\mathbf{w}), '\operatorname{complex}');
rcv_bits = s_c + w;
% Reciver structure
g m = conj(flipud(q c));
```

```
% compute the impulse response h
h = conv(q c, g m);
h = downsample(h, 4);
h = h(h = 0);
N1 = floor(length(h)/2);
N2 = N1;
r r = filter(g m, 1, rcv bits);
t \ 0 \ bar = length(g \ m);
x = downsample(r r(t 0 bar:end), 4);
% Filtering through C and equalization
r gm = xcorr(g m);
\% r w = N0 .* downsample(r gm, 4);
r_w_u = sigma_w / 4 * r_g;
r_w = downsample(r_w_up, 4);
M1 = 5;
D = 4;
M2 = N2 + M1 - 1 - D;
[c Jmin] = WienerC DFE(h, r w, sigma a, M1, M2, D);
psi = conv(c, h);
M1 = 5; D = 4;
M2 \; = \; N2 \; + \; M1 \; - \; 1 \; - \; D;
b = -p si (find (psi = max(psi)) + 1: end);
y_k = equalization_DFE(x, c, b, D, max(psi));
decisions = zeros(length(y k), 1);
for i = 1: length(y k)
    decisions(i) = threshold detector(y k(i));
end
Jmin lin = 10^{(Jmin/10)};
noise var = (J\min lin-sigma a*abs(1-max(psi))^2)/(abs(max(psi))^2);
% Compute Log Likelihood Ratio
llr = zeros(2*length(y_k),1);
llr(1:2:end, 1) = -2*real(y_k)/(noise_var/2);
llr(2:2:end, 1) = -2*imag(y k)/(noise var/2);
new length = floor (length (llr) / 64800) * 64800;
```

```
llr = llr (1:new_length, 1);
% Decode the bits
llr = deinterleaver(llr); % Deinterleave the loglikelihood ratio first
decoderLDPC = comm.LDPCDecoder;
dstep = 2 * sstep;
tic
for i = 0: (floor(length(llr)/dstep)) - 1
    %decodes block by block the input bits
    %block length is equal to 64800
    block = llr(i * dstep + 1:i * dstep + dstep);
    dec_b_l(i * dstep / 2 + 1:i * dstep / 2 + dstep / 2) = ...
         step(decoderLDPC, block.');
end
toc
[Pbit_DFEenc(snr_index), errors] = BER(dec_b_l, b_l(1:length(dec_b_l)));
end
clear all; close all; clc;
format long g;
sigma \ a = 2;
load('generated symbols.mat', 'b l', 'uncoded a')
% Generate the channel input response
[q c, E qc] = channel impulse response();
a prime = upsample(uncoded a, 4);
% Filter through the channel
s_c = filter(q_c, 1, a_prime);
SNR \ vector = [4:0.5:14];
Pbit DFEunc = zeros(length(SNR vector),1);
for snr index = 1:length(SNR vector)
    SNR = SNR\_vector(snr\_index);
    % Add Gaussian Noise
    snrlin = 10^{(SNR/10)};
    sigma w = sigma a * E qc / snrlin;
    \mathbf{w} = \operatorname{wgn}(\operatorname{length}(\mathbf{s}_{c}), 1, 10 * \log 10(\operatorname{sigma}_{w}), 'complex');
    rcv_bits = s_c + w;
    % Reciver structure
    g m = conj(flipud(q c));
```

```
% compute the impulse response h
    h = conv(q c, g m);
    h = downsample(h, 4);
   h = h(h = 0);
   N1 = floor(length(h)/2);
   N2 = N1;
    r r = filter(g m, 1, rcv bits);
    t = 0 \text{ bar} = length(g m);
    x = downsample(r r(t 0 bar:end), 4);
   % Filtering through C and equalization
   r gm = x corr(g_m);
   \% r w = N0 .* downsample(r gm, 4);
   r_w_u = sigma_w / 4 * r_g;
   r\ w = downsample (r_w_up, \ 4);
   M1 = 5;
   D = 4;
   M2 = N2 + M1 - 1 - D;
    [c Jmin] = WienerC DFE(h, r w, sigma a, M1, M2, D);
    psi = conv(c, h);
   M1 = 5; D = 4;
   M2 \; = \; N2 \; + \; M1 \; - \; 1 \; - \; D \, ;
   b = -p si (find (psi = max(psi)) + 1 : end);
   y_k = equalization_DFE(x, c, b, D, max(psi));
    decisions = zeros(length(y_k), 1);
    for i = 1: length(y_k)
        decisions (i) = threshold_detector(y_k(i));
    end
    detected bits = IBMAP(decisions);
    [Pbit DFEunc(snr index), errors] = \dots
        BER(detected_bits, b_l(1:length(detected_bits)));
save('DFE uncoded.mat', 'Pbit DFEunc');
```

end

```
function [Pbit dec b l] = OFDM coded(a, b l, Npx, t0 bar, SNRlin)
\% pad the last symbols with -1-1i to have an integer multiple of 512
M = 512; % number of subchannels
sigma \ a = 2;
a = [a; ones(M - mod(length(a), M), 1) * (-1-1i)];
a mat = reshape(a, M, []);
A no pr = ifft(a mat);
A = [A \text{ no } pr(M - Npx + 1:M,:); A \text{ no } pr];
s n = reshape(A, [], 1);
% only to try with ideal channel
\% sigma w OFDM = ( (sigma a/M)) / SNRlin;
\% w = wgn(length(s n), 1, 10*log10(sigma w OFDM), 'complex');
\% x = s n + w;
%channel construction
ro = 0.0625;
span = 30;
sps = 4;
g_rcos = rcosdesign(ro, span, sps, 'sqrt');
g rcos = g rcos(abs(g rcos)) = (max(g rcos)*10^-2);
s up = upsample(s n, 4);
s up rcos = filter(g_rcos, 1, s_up);
%s\_up\_rcos = s\_up\_rcos(length(g\_rcos):end);
q c = channel impulse response();
s_c = filter(q_c, 1, s_up_rcos);
%s_c = s_c (length(q_c):end);
Eimp = sum(conv(g_rcos,q_c).^2);
sigma w OFDM = (sigma a * Eimp)/ (M * SNRlin);
w = wgn(length(s c), 1, 10*log10(sigma w OFDM), 'complex');
r_c = s_c + w;
% r_c = s_c;
gq = conv(g\_rcos, q\_c);
q_rup = conv(gq, g_rcos);
q_rup = q_rup(abs(q_rup)) = (max(q_rup)*10^-3);
\% \ t0_bar = find(q_r_up == max(q_r_up));
index = find(q r up = max(q r up));
rem = mod(index, 4);
```

```
if rem == 0
    start = 4;
else
    start = rem;
end
q r = downsample(q r up(start:end), 4);
q r = q r(abs(q r) > = (max(q r) *10^{-3});
x = filter(g rcos, 1, r c);
%x = upfirdn(r_c, g_rcos, 2);
x = downsample(x(t0 bar:end), 4);
K_i = fft(q_r, 512);
K_i = K_i(:);
x = x(1: end - mod(length(x), M+Npx));
r matrix = reshape(x, M+Npx, []);
r matrix = r matrix (Npx + 1:end, :);
K \text{ i inv} = K \text{ i.}^{(-1)};
x_matrix = fft(r_matrix);
y_matrix = x_matrix .* K_i_inv;
sigma i = 0.5 * sigma w OFDM * M * abs(K i inv).^2;
LLR real = -2 * real(y matrix) .* (sigma i.^(-1));
LLR_{imag} = -2 * imag(y_{matrix}) .* (sigma_i.^(-1));
LLR_{real}_{vec} = reshape(LLR_{real}, [], 1);
LLR_imag_vec = reshape(LLR_imag, [], 1);
LLR = zeros(length(LLR real vec) + length(LLR imag vec), 1);
LLR(1:2:end) = LLR\_real\_vec;
LLR(2:2:end) = LLR_imag_vec;
LLR = LLR(1:2*length(b_l));
% to try with ideal channel
\% y = reshape(x matrix, [], 1);
\% llr = zeros (2*length(y),1);
\% sigma i = 0.5*sigma w OFDM*M;
\% llr_real = -2*times(real(y),(sigma_i.^(-1)));
\% llr_imag = -2*times(imag(y),(sigma_i.^(-1)));
% llr_real_ar = reshape(llr_real, [], 1);
\% llr imag ar = reshape(llr imag, [], 1);
\% llr (1:2:end) = llr real ar;
\% \ llr(2:2:end) = llr_imag_ar;
\% \ llr = llr (1:end - mod(length(llr), 32400));
```

```
LLR = deinterleaver(LLR);
decoderLDPC = comm.LDPCDecoder;
dstep = 64800;
tic
for i = 0: (floor (length (LLR) / dstep)) - 1
    block = LLR(i * dstep + 1:i * dstep + dstep);
    dec b l(i * dstep / 2 + 1:i * dstep / 2 + dstep / 2) = ...
        step(decoderLDPC, block.');
end
toc
Pbit = BER(b l, dec b l);
end
 function [Pbit b l hat] = OFDM uncoded(a, b l, Npx, t0 bar, SNRlin)
\% pad the last symbols with -1-1i to have an integer multiple of 512
M = 512; % number of subchannels
sigma \ a = 2;
a = [a; ones(M - mod(length(a), M), 1) * (-1-1i)];
a_mat = reshape(a, M, []);
A no prefix = ifft(a mat);
%A \text{ no pr} = A \text{ no pr} (1:512,:);
A = [A \text{ no prefix}(M - Npx + 1:M,:); A \text{ no prefix}];
s n = reshape(A, [], 1);
%channel contruction
ro = 0.0625;
span = 30;
sps = 4;
g rcos = rcosdesign (ro, span, sps, 'sqrt');
g rcos = g rcos(abs(g rcos)) = (max(g rcos)*10^-2);
s up = upsample(s n, 4);
s_{proc} = filter(g_{ros}, 1, s_{proc});
%s\_up\_rcos = s\_up\_rcos(length(g\_rcos):end);
q c = channel impulse response();
s_c = filter(q_c, 1, s_up_rcos);
%s c = s c(length(q c):end);
Eimp = sum(conv(g rcos, q c).^2);
```

```
sigma_w_OFDM = (sigma_a * Eimp)/ (M * SNRlin);
w = wgn(length(s_c), 1, 10*log10(sigma_w_OFDM), 'complex');
r_c = s_c + w;
% r_c = s_c;
gq = conv(g rcos, q c);
q_rup = conv(gq, g_rcos);
q_rup = q_rup(abs(q_rup)) = (max(q_rup)*10^-3);
\% to bar = find (q r up == max(q r up));
index = find(q_r_up = max(q_r_up));
rem = mod(index, 4);
if rem == 0
    start = 4;
else
    start = rem;
end
q_r = downsample(q_r_up(start:end), 4);
q_r = q_r(abs(q_r) > = (max(q_r) *10^-3));
x = filter(g rcos, 1, r c);
x = downsample(x(t0 bar:end), 4);
K_i = fft(q_r, M);
K_i = K_i(:);
x = x(1: end - mod(length(x), M+Npx));
r matrix = reshape(x, M+Npx, []);
r_matrix = r_matrix(Npx + 1:end, :);
K \text{ i inv} = K \text{ i.}^{(-1)};
x_matrix = fft(r_matrix);
y_matrix = x_matrix \cdot * K_i_i_inv;
y = reshape(y matrix, 1, []);
dec \ a \ k = zeros(length(y), 1);
for k=1: length(y)
    dec_a_k(k) = threshold_detector(y(k));
b_l = IBMAP(dec \ a \ k);
[Pbit ~]= BER(b_l, b_l_hat(1:end));
 end
```

#### Wiener and equalization for DFE

```
function [decisions] = equalization DFE(x, c, b, D, psiD)
%EQUALIZATION for DFE
M2 = length(b);
y = conv(x,c);
y = y(1: length(x)+D);
y = y./psiD;
detected = zeros(length(x) + D, 1);
for k=0: length (y)-1
     if (k \le M2)
        a_past = [flipud(detected(1:k)); zeros(M2 - k, 1)];
        a past = flipud (detected (k - M2 + 1: k));
    end
detected(k + 1) = y(k + 1) + b.' * a_past;
%scatterplot(y)
decisions = detected(D + 1:end);
end
 function [c opt, Jmin] = WienerC DFE(h, r w, sigma a, M1, M2, D)
    N1 = floor(length(h)/2);
    N2 = N1;
    padding = 60;
    hpad = padarray(h, padding);
    % Padding the noise correlation
    r w pad = padarray(r w, padding);
    p = z \operatorname{eros}(M1, 1);
    for i = 0 : M1-1
        p(i + 1) = sigma_a * conj(hpad(N1 + padding + 1 + D - i));
    end
    R = zeros(M1);
    for row = 0:(M1-1)
        for col = 0:(M1-1)
            fsum = (hpad((padding + 1):(N1 + N2 + padding + 1))).
                 * conj(hpad((padding + 1 - (row - col)):(N1 + N2 + ...
                 padding + 1 - (row - col)));
             if M2==0
                 ssum = 0;
             else
                 ssum = (hpad((N1+padding+1+1+D-col)): \dots
                     (N1+padding+1+M2+D-col))).**
```

### Useful scripts (interleaver, BMAP, plots, TD)

```
function [interl bits] = interl(bits)
    interl bits = zeros(1,length(bits));
    rows = 43;
    columns = 41;
    for matrix = 0:(length(bits)/(rows*columns) - 1)
        curr matrix = matrix * rows * columns;
        for col = 0:(columns-1)
             interl_bits(curr_matrix + col * rows + 1 : curr_matrix...
                 + \text{col} * \text{rows} + \text{rows}) = \text{bits}(\text{curr}_{\text{matrix}} + \text{col} + 1 : \dots)
                 columns : curr matrix + col + columns * rows);
        end
    end
end
function [deinterleaved bits] = deinterleaver (bits)
deinterleaved bits = zeros(1, length(bits));
rows = 41;
columns = 43;
  for matrix = 0:(length(bits)/(rows*columns) - 1)
    curr matrix = matrix * rows * columns;
    for col = 0:(columns-1)
      deinterleaved bits (curr matrix + col * rows + 1 : curr matrix . . .
```

```
+ \text{ col} * \text{ rows} + \text{ rows}) = \text{bits}(\text{curr}_{\text{matrix}} + \text{col} + 1 : \text{columns} : \dots)
           curr matrix + col + columns * rows);
    end
  end
end
function [Pbit, count errors] = BER(sent, detected)
% Computes the symbol-error rate, it accepts QPSK symbols
count errors = 0;
for i=1:length(sent)
    if sent(i) = detected(i)
         count errors = count errors + 1;
    end
end
Pbit = count errors/length(sent);
end
function [symbols] = BMAP(bits)
% bits are given as a row vector because of the interleaver function output
L = length(bits);
symbols = zeros(L,1);
    % gray coding of the input bits for QPSK symbols
    for k = 1:2:L-1
         if (isequal(bits(k:k+1), [0 0]))
             \operatorname{symbols}(k) = -1-1i;
         elseif (isequal(bits(k:k+1), [1 0]))
             symbols(k) = 1-1i;
         elseif (isequal(bits(k:k+1), [0 1]))
             symbols(k) = -1+1i;
         elseif (isequal(bits(k:k+1), [1 1]))
             symbols(k) = +1+1i;
         end \\
    end
    symbols = symbols (1:2:end);
end
function [b \ i] = IBMAP(a \ k)
    % Check if the input array has even length
    L = length(a k);
    b i = zeros(2*L,1);
    % Map each couple of values to the corresponding symbol
    % The real part gives the bit
    for k = 1:2: length(b i)-1
        symbol = a_k((k+1)/2);
         if (real(symbol) == 1)
            b2k = 1;
         else
```

```
b2k = 0;
         \operatorname{end}
          if (imag(symbol) == 1)
              b2k1 = 1;
          else
              b2k1 = 0;
         end
         b i(k) = b2k;
         b i(k+1) = b2k1;
     end
end
clear all; close all; clc;
set (0, 'defaultTextInterpreter', 'latex')
load ('AWGN coded.mat', 'Pbit AWGN');
load ('OFDM_uncoded_final.mat', 'Pbit_OFDM_uncoded')
{\tt load}\;(\;{\tt 'DFE\_uncoded.mat'}\;,\;{\tt 'Pbit\_DFEunc'})\;;
load ('DFE coded1.mat', 'Pbit DFEenc');
load ('OFDM_coded_final.mat', 'Pbit_OFDM_coded');
SNR_dB = [4:0.5:14];
SNR_{lin} = 10.^{(SNR_{dB.}/10)};
sigma \ a = 2;
awgn bound = qfunc(sqrt(SNR lin));
figure,
semilogy (SNR dB, Pbit DFEunc, 'g-o')
y \lim ([10^{-5} 10^{-1}])
grid on;
hold on,
semilogy (SNR_dB, Pbit_OFDM_uncoded, 'b-<')
hold on,
semilogy (SNR_dB, awgn_bound, 'k')
xlabel ('SNR [dB]')
ylabel('$P {bit}$')
title ('Uncoded')
legend({ 'DFE', 'OFDM', 'AWGN' })
SNR_dB_encDFE = [1.4:0.025:1.75];
SNR_dB_encOFDM = [0.9:0.05:1.6];
SNR dB_awgn = [0.4:0.05:1];
figure,
semilogy(SNR_dB_encDFE, Pbit_DFEenc, 'g-o')
hold on,
y \lim ([10^{-5} 10^{-1}])
x \lim (\begin{bmatrix} 0 & 3 \end{bmatrix});
```

```
semilogy (SNR dB encOFDM, Pbit OFDM coded, 'b-<')
grid on;
hold on,
semilogy (SNR_dB_awgn, Pbit_AWGN, 'k')
xlabel ('SNR [dB]')
ylabel('$P {bit}$')
title ('Coded')
legend({'DFE', 'OFDM', 'AWGN'})
clear all; close all; clc;
load('generated_symbols.mat', 'a', 'enc_b_l', 'b_l', 'uncoded_a')
%parpool;
%% CODED
SNR dB = [0.9:0.05:1.6];
parfor i = 1: length (SNR dB)
    [Pbit\_OFDM\_coded(i) b\_l\_hat] = ...
         OFDM\_coded(a, b_l, 18, 43, 10^(SNR\_dB(i)/10));
end
% save ('OFDM coded.mat', 'Pbit OFDM coded')
\% \text{ SNR } dB = 1.3;
\% [Pbit_OFDM_coded b_l_hat] = \dots
\%OFDM coded(a, b l, 18, 43, 10^{\circ}(SNR dB/10));
\% SNR dB = 1.3;
\% parfor i=1:60
       [Pbit coded(i) b l hat] = \dots
\%OFDM\_coded(a, b_l, 16, i, 10^(SNR\_dB./10));
\% end
% save ('OFDM coded.mat', 'Pbit OFDM coded')
\% \ SNR\_dB \ = \ 1.3 \quad ;
\% parfor i = 10:30
       [Pbit coded(i) b l_hat] = ...
\%OFDM\_coded(a, b_l, i, 17, 10^(SNR_dB./10));
% end
%% UNCODED
\% SNR_dB = 10;
\% for i = 10:100
\% [Pbit OFDM uncoded(i) b l hat] = ...
\%OFDM uncoded (uncoded a, b l, 20, i, 10^{(SNR dB/10)});
% end
\% SNR dB = [4:0.5:14];
```

```
\% parfor i=1: length (SNR_dB)
\% [Pbit_OFDM_uncoded(i) b_l_hat] = ...
\%OFDM\_uncoded(uncoded\_a, b\_l, 18, 43, 10^(SNR\_dB(i)/10));
%save('OFDM uncoded.mat', 'SNR dB', 'Pbit OFDM uncoded1');
\% [Pbit uncoded b l hat] = ...
\label{eq:conded} \mbox{$\langle OFDM\_uncoded(uncoded\_a, b\_l, 8, 21, 10.^(SNR\_dB/10));}
function [a_hat_kD] = threshold_detector(y_k)
if (real(y_k) > 0)
     if (imag(y_k) > 0)
         a_hat_kD = 1+1i;
     else
         a_hat_kD= 1-1i;
    end
else
     if (imag(y k) > 0)
         a_hat_kD = -1+1i;
         a_hat_kD = -1-1i;
    end
end
end
```

## Signal/channel generators

```
function [pn] = PN(r)
L = pow2(r) - 1;
pn = zeros(L, 1);
pn(1:r) = ones(1,r).;
for l=r+1:L
    switch r
        case 1
            pn(l) = pn(l-1);
        case 2
            pn(l) = xor(pn(l-1), pn(l-2));
        case 3
            pn(1) = xor(pn(1-2), pn(1-3));
        case 4
            pn(1) = xor(pn(1-3), pn(1-4));
        case 5
            pn(1) = xor(pn(1-3), pn(1-5));
        case 6
```

```
case 7
            pn(1) = xor(pn(1-6), pn(1-7));
        case 8
            pn(1) = xor(xor(pn(1-2), pn(1-3)), xor(pn(1-4), pn(1-8)));
        case 9
            pn(1) = xor(pn(1-5), pn(1-9));
        case 10
            pn(1) = xor(pn(1-7), pn(1-10));
        case 11
            pn(1) = xor(pn(1-9), pn(1-11));
        case 12
            pn(1) = xor(xor(pn(1-2), pn(1-10)), xor(pn(1-11), pn(1-12)));
        case 13
            pn(1) = xor(xor(pn(1-1), pn(1-11)), xor(pn(1-12), pn(1-13)));
        case 14
            pn(1) = xor(xor(pn(1-2), pn(1-12)), xor(pn(1-13), pn(1-14)));
        case 15
            pn(1) = xor(pn(1-14), pn(1-15));
        case 16
            pn(l) = xor(xor(pn(l-11), pn(l-13)), xor(pn(l-14), pn(l-16)));
        case 17
            pn(1) = xor(pn(1-14), pn(1-17));
        case 18
            pn(l) = xor(pn(l-11), pn(l-18));
            pn(1) = xor(xor(pn(1-14), pn(1-17)), xor(pn(1-19), pn(1-18)));
        case 20
            pn(1) = xor(pn(1-17), pn(1-20));
    end
end
end
clear all; close all; clc;
%% THIS SCRIPT GENERATES THE SYMBOLS BY USING A PN SEQUENCE
\%\% AND APPLYING LDP ENCODING AND INTERLEAVING
b l = [PN(20); PN(20)];
sstep = 32400;
num bits = floor(length(b l) / sstep) * sstep;
b l = b l(1:num bits);
sigma_a = 2;
% ENCODE VIA LDPC
%create LDPC encoder with the dafault matrix (rate=2)
encoderLDPC = comm.LDPCEncoder;
enc b l = zeros(2*length(b l),1);
```

pn(1) = xor(pn(1-5), pn(1-6));

```
for i = 0: (ceil(length(b_l)/sstep))-1
    %encodes block by block the input bits
    %block length is equal to 32400
    block = b_l(i * sstep + 1:i * sstep + sstep);
    enc_b_1(2 * i * sstep + 1:2 * i * sstep + 2 * sstep) = ...
        step(encoderLDPC, block);
end
%% INTERLEAVING
interl b l = interl(enc b l);
%% BITMAP
a = BMAP(interl b l);
uncoded a = BMAP(b l.');
save('generated symbols.mat', 'a', 'enc b l', 'b l', 'uncoded a')
function \ [q\_c\,, \ E\_qc\,] \ = \ channel\_impulse\_response\,()
q_c_num = [0 \ 0 \ 0 \ 0 \ 0 \ 0.7424];
q c denom = [1 -0.67];
q_c = impz(q_c_num, q_c_denom);
\% cut the impulse response when too small
%q c = [0; 0; 0; 0; 0; q c(q c) = max(q c)*10^(-2));
E qc = sum(q c.^2);
end
clc; clear all; close all;
%% Configuration parameters
if ~exist("Noise.mat", 'file')
    noise_seq;
end
load ('Noise', 'w');
load('generated symbols.mat', 'a');
verbose = false;
plot figure = true;
r = 20;
SNR dB = [0:14];
SNR \ lin = 10.^{SNR} \ dB./10);
sigma_a = 2;
T = 1;
q c num = [0 0 0 0 0 0.7424];
q \ c \ denom = [1 \ -0.67];
```

```
q_c = impz(q_c_num, q_c_denom);
\% cut the impulse response when too small
E_qc = sum(q_c.^2);
sigma w = (sigma a * E qc) ./ SNR lin;
N0 = sigma w./4;
%% Filtering through the channel
a_{prime} = upsample(a, 4);
s_c = filter(q_c_num, q_c_denom, a_prime);
%% Add noise
r_c = z eros(length(s_c), length(SNR_dB));
for i = 1 : length (SNR dB)
    r_c(:,i) = s_c + w(1:length(s_c))*sqrt(sigma_w(i));
end
clear w;
%% Save the workspace
save("common.mat");
```