

PCM Input Class D Audio Power Amplifiers

General Description

Features

The MAX98357A/MAX98357B are digital pulse-code modulation (PCM) input Class D power amplifiers that provide Class AB audio performance with Class D efficiency. These ICs offer five selectable gain settings (3dB, 6dB, 9dB, 12dB, and 15dB) in I²S/left-justified mode set by a single gain select input and a fixed 12dB gain in TDM mode.

The digital audio interface is highly flexible with the MAX98357A supporting I²S data and the MAX98357B supporting left-justified data. Both ICs support 8 channel time division multiplexed (TDM) data. The digital audio interface accepts specified sample rates between 8kHz and 96kHz for all supported data formats. The ICs can be configured to produce a left channel, right channel, or (left/2 + right/2) output from the stereo input data. The ICs operate using 16/24/32-bit data for I²S and left-justified modes as well as 16-bit or 32-bit data using TDM mode. The ICs eliminate the need for the external MCLK signal that is typically used for PCM communication. This reduces EMI and possible board coupling issues in addition to reducing the size and pin count of the ICs.

The ICs also feature a very high wideband jitter tolerance (12ns typ) on BCLK and LRCLK to provide robust operation.

Active emissions-limiting, edge-rate limiting, and overshoot control circuitry greatly reduce EMI. A filterless spread-spectrum modulation scheme eliminates the need for output filtering found in traditional Class D devices and reduces the component count of the solution.

The ICs are available in 9-pin WLP (1.345mm \times 1.435mm \times 0.64mm) and 16-pin TQFN (3mm \times 3mm \times 0.75mm) packages and are specified over the -40°C to +85°C temperature range.

Applications

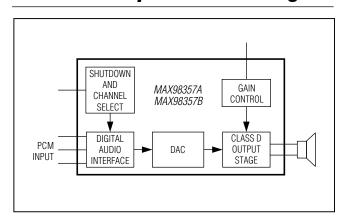
Notebook and Netbook Computers Cellular Phones Tablets

<u>Ordering Information</u> appears at end of data sheet. <u>Functional Diagram</u> appears at end of data sheet.

For related parts and recommended products to use with this part, refer to www.maximintegrated.com/MAX98357A.related.

- ♦ Single-Supply Operation (2.5V to 5.5V)
- ♦ 3.2W Output Power into 4Ω at 5V
- ♦ 2.4mA Quiescent Current
- ♦ 92% Efficiency (R_I = 8Ω , $P_{OUT} = 1W$)
- ♦ 25µV_{RMS} Output Noise (A_V = 15dB)
- ♦ Low 0.015% THD+N at 1kHz
- **♦ No MCLK Required**
- ♦ Sample Rates of 8kHz to 96kHz
- ♦ Supports Left, Right, or (Left/2 + Right/2) Output
- ♦ Sophisticated Edge Rate Control Enables Filterless Class D Outputs
- ♦ 77dB PSRR at 1kHz
- ♦ Low RF Susceptibility Rejects TDMA Noise from GSM Radios
- **♦ Extensive Click-and-Pop Reduction Circuitry**
- ♦ Robust Short-Circuit and Thermal Protection
- ♦ Available in Space-Saving Packages: 1.345mm x 1.435mm WLP (0.4mm Pitch) and 3mm x 3mm TQFN

Simplified Block Diagram



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ABSOLUTE MAXIMUM RATINGS

V _{DD} , LRCLK, BCLK, and DIN to GND0.3V to +6V	Continuous Power Dissipation ($T_A = +70^{\circ}C$)
All Other Pins to GND0.3V to (V _{DD} + 0.3V)	WLP (derate 13.7mW/°C above +70°C)1096mW
Continuous Current In/Out of V _{DD} /GND/OUT±1.6A	TQFN (derate 20.8mW/°C above +70°C)1666mW
Continuous Input Current (all other pins)±20mA	Junction Temperature+150°C
Duration of OUT_ Short Circuit to GND or V _{DD} Continuous	Operating Temperature Range40°C to +85°C
Duration of OUTP Short to OUTNContinuous	Storage Temperature Range65°C to +150°C
	Soldering Temperature (reflow)+230°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

PACKAGE THERMAL CHARACTERISTICS (Note 1)

WLP	TQFN
Junction-to-Ambient Thermal Resistance (θ _{JA})73°C/W	Junction-to-Ambient Thermal Resistance (θ _{JA})48°C/W
Junction-to-Case Thermal Resistance (θ _{JC})50°C/W	Junction-to-Case Thermal Resistance (θ _{JC})7°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

ELECTRICAL CHARACTERISTICS

 $(V_{DD}=5V,\ V_{GND}=0V,\ GAIN_SLOT=V_{DD}.\ BCLK=3.072MHz,\ LRCLK=48kHz,\ speaker\ loads\ (Z_{SPK})\ connected\ between\ OUTP\ and\ OUTN,\ Z_{SPK}=\infty,\ T_A=T_{MIN}\ to\ T_{MAX},\ unless\ otherwise\ noted.$ Typical values are at $T_A=+25^{\circ}C.)$ (Note 2)

PARAMETER	SYMBOL	CONE	DITIONS	MIN	TYP	MAX	UNITS
Supply Voltage Range	V _{DD}	Guaranteed by PSS	R test	2.5		5.5	V
Undervoltage Lockout	UVLO			1.4	1.8	2.3	V
Quiescent Current	1	$T_A = +25^{\circ}C$			2.75	3.35	- mA
Quiescent Current	I _{DD}	$T_A = +25$ °C, $V_{DD} =$	3.7V		2.4	2.85	IIIA
Shutdown Current	ISHDN	SD_MODE = 0V, TA	= +25°C		0.6	2	μΑ
Standby Current	ISTNDBY	<u>SD_MODE</u> = 1.8V, 1	no BCLK, T _A = +25°C		340	400	μA
Turn-On Time	t _{ON}				7	7.5	ms
Output Offset Voltage	Vos	T _A = +25°C, gain =	15dB		±0.3	±2.5	mV
Clink and Day Lavel	l/	Peak voltage, T _A = +25°C, A-weighted,	Into shutdown		-72		4DV
Click-and-Pop Level	K _{CP}	132 Samples per	Out of shutdown		-66		- dBV
		$V_{DD} = 2.5V \text{ to } 5.5V,$	T _A = +25°C	60	75		
Power-Supply Rejection Ratio	PSRR	T _A = +25°C	f = 217Hz, $200mV_{P-P}$ ripple		77		dB
		(Notes 3, 4)	f = 10kHz, 200mV _{P-P} ripple		60		

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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD}=5V,\,V_{GND}=0V,\,GAIN_SLOT=V_{DD}.\,BCLK=3.072MHz,\,LRCLK=48kHz,\,speaker\,loads\,(Z_{SPK})$ connected between OUTP and OUTN, $Z_{SPK}=\infty$, $T_A=T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A=+25^{\circ}C.$) (Note 2)

PARAMETER	SYMBOL	CON	DITIONS	MIN	TYP	MAX	UNITS
			$Z_{SPK} = 4\Omega + 33\mu H$		3.2		
		THD+N 10%, gain = 12dB	$Z_{SPK} = 8\Omega + 68\mu H$		1.8		1
			$Z_{SPK} = 8\Omega + 68\mu H,$ $V_{DD} = 3.7V$		0.93]
Output Power (Note 3)	Pout		$Z_{SPK} = 4\Omega + 33\mu H$		2.5		W
		THD+N = 1%,	$Z_{SPK} = 8\Omega + 68\mu H$		1.4		1
		gain = 12dB	$Z_{SPK} = 8\Omega + 68\mu H,$ $V_{DD} = 3.7V$		0.77		
		$f = 1 \text{kHz}, P_{OUT} = 1$ $Z_{SPK} = 4\Omega + 33 \mu \text{H}$, WLP		0.02	0.06	
Total Harmonic Distortion + Noise	THD+N	$f = 1 \text{kHz}, P_{OUT} = 1$ $Z_{SPK} = 4\Omega + 33 \mu \text{H}$			0.02		%
		$f = 1 \text{kHz}, P_{OUT} = 0$ $Z_{SPK} = 8\Omega + 68 \mu \text{H}$			0.013		
Dynamic Range	DR	A-weighted, V _{RMS} = 2.54V, 24-	or 32-bit data		105		dB
Output Noise	V _N	A-weighted, 24- or	32-bit data (Note 4)		25		μV _{RMS}
	A _V	GAIN_SLOT = GND through $100k\Omega$		14.4	15	15.6	dB
0 : (0 1 : 1 0 10)		GAIN_SLOT = GND		11.4	12	12.6	
Gain (Relative to a 2.1dBV Reference Level)		GAIN_SLOT = unconnected		8.4	9	9.6	
		GAIN_SLOT = V _{DD}		5.4	6	6.6	
		GAIN_SLOT = V _{DD}	through 100kΩ	2.4	3	3.6	
Current Limit	I _{LIM}				2.8		А
Efficiency	3	$Z_{SPK} = 8\Omega + 68\mu H$ f = 1kHz, gain = 12			92		%
DAC Gain Error					1		%
Frequency Response				-0.2		+0.2	dB
Class D Switching Frequency	fosc				330		kHz
Spread-Spectrum Bandwidth					±12.5		kHz
DAC DIGITAL FILTERS							
VOICE MODE IIR LOWPASS F	ILTER (LRC	LK < 30kHz)					
Passband Cutoff	t.	Ripple limit cutoff		0.443 x f _S			
	f _{PLP}	-3dB cutoff		0.446 x f _S			- Hz
Stopband Cutoff	f _{SLP}					0.464 x f _S	Hz
Stopband Attenuation		f > f _{SLP}		75			dB

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ELECTRICAL CHARACTERISTICS (continued)

(V_{DD} = 5V, V_{GND} = 0V, GAIN = V_{DD} . BCLK = 3.072MHz, LRCLK = 48kHz, speaker loads (Z_{SPK}) connected between OUTP and OUTN, Z_{SPK} = ∞ , T_A = T_{MIN} to T_{MAX} , unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
AUDIO MODE FIR LOWPASS	FILTER (30k	Hz < LRCLK < 50kHz)				
		Ripple limit cutoff	0.43 x f _S			
Passband Cutoff	f _{PLP}	-3dB cutoff	0.47 x f _S			Hz
		-6.02dB cutoff	0.5 x f _S			
Stopband Cutoff	f _{SLP}				0.58 x f _S	Hz
Stopband Attenuation		f > f _{SLP}	60			dB
AUDIO MODE FIR LOWPASS	FILTER (LRC	CLK > 50kHz)				
5 10 10	,	Ripple limit cutoff	0.24 x f _S			
Passband Cutoff	f _{PLP}	-3dB cutoff	0.31 x f _S			Hz
Stopband Cutoff	f _{SLP}				0.477 x f _S	Hz
Stopband Attenuation		f < f _{SLP}	60			dB
DIGITAL AUDIO INTERFACE						
LRCLK Range 1	f _{S1}		7.6	8	8.4	
LRCLK Range 2	f _{S2}		15.2	16	16.8	kHz
LRCLK Range 3	f _{S3}		30.4	48	50.4	KHZ
LRCLK Range 4	f _{S4}		83.8	96	100.8	
Resolution		I ² S/left justified mode TDM mode		16/24/32 16/32		Bits
BCLK Frequency Range	fBCLKH	BCLK must be 32, 48, or 64X of LRCLK	0.2432	<u> </u>	25.804	MHz
BCLK High Time	†BCLKH		15			ns
BCLK Low Time	†BCLKL		15			ns
Maximum Low Frequency BCLK and LRCLK Jitter		RMS jitter below 40kHz		0.5		
Maximum High Frequency BCLK and LRCLK Jitter		RMS jitter above 40kHz		12		ns
Input High Voltage	V _{IH}	Digital audio inputs	1.3			V

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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD}=5V,\ V_{GND}=0V,\ GAIN=V_{DD}.\ BCLK=3.072MHz,\ LRCLK=48kHz,\ speaker\ loads\ (Z_{SPK})\ connected\ between\ OUTP\ and\ Seaker\ Seak$ OUTN, $Z_{SPK} = \infty$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25$ °C.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Input Low Voltage	V _{IL}	Digital audio inputs			0.6	V
Input Leakage Current	I _{IH} , I _{IL}	$V_{IN} = 0V, V_{DD} = 5.5V, T_A = +25^{\circ}C$	-1		+1	μΑ
Input Capacitance	C _{IN}			3		pF
DIN to BCLK Setup Time	tSETUP		10			ns
LRCLK to BCLK Setup Time	tsyncset		10			ns
DIN to BCLK Hold Time	t _{HOLD}		10			ns
LRCLK to BCLK Hold Time	tsynchold		10			ns
SD_MODE COMPARATOR TR	IP POINTS					
В0			0.08	0.16	0.355	
B1		See SD_MODE and shutdown operation for details	0.65	0.77	0.825	V
B2]	ioi detaiis	1.245	1.4	1.5	1
SD_MODE Pulldown Resistor	R _{PD}		92	100	108	kΩ
GAIN_SLOT COMPARATOR T	RIP POINTS					
		A _V = 3dB gain	0.65 x		0.85 x	
		Ay = Sub gairi	V_{DD}		V_{DD}	
		A _V = 6dB gain	0.9 x		V_{DD}	
		/\varphi = 0db gaiii	V_{DD}		• DD	
	V_GAIN_	A _V = 9dB gain	0.4 x		0.6 x	V
SLOT	SLOT	Ny = 332 gain	V_{DD}		V_{DD}	
		A _V = 12dB gain	0		0.1 x	
		v 9			V_{DD}	
		A _V = 15dB gain	0.15 x		0.35 x	
		V _{DD}		V_{DD}		

Note 2: 100% production tested at $T_A = +25$ °C. Specifications over temperature limits are guaranteed by design.

Note 3: Class D amplifier testing performed with a resistive load in series with an inductor to simulate an actual speaker load. For $R_L=8\Omega,\ L_L=68\mu H.\ For\ R_L=4\Omega,\ L_L=33\mu H.$ Note 4: Digital silence used for input signal.

Note 5: Dynamic range measured using the EIAJ method. -60dBFS 1kHz output signal, A-weighted, and normalized to 0dBFS. f = 20Hz to 20kHz.

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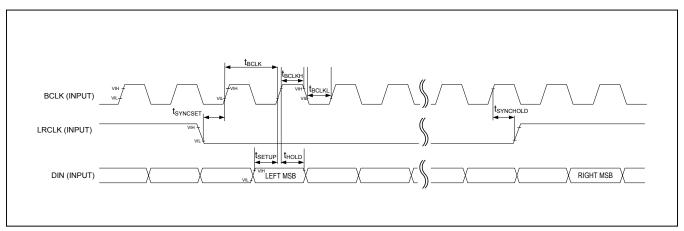


Figure 1. I²S Audio Interface Timing Diagram (MAX98357A)

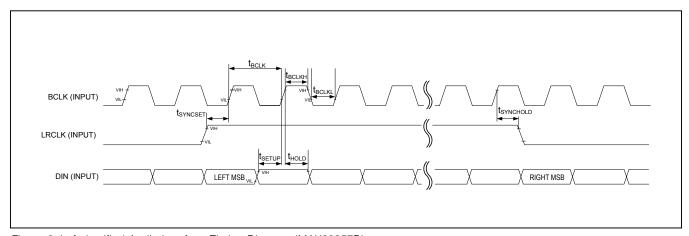


Figure 2. Left-Justified Audio Interface Timing Diagram (MAX98357B)

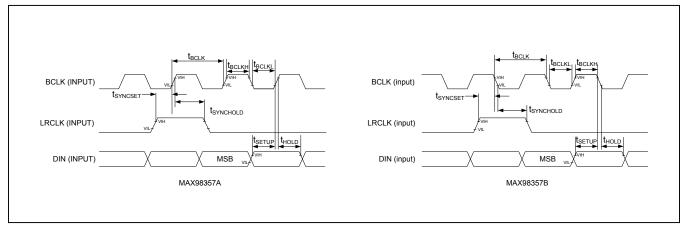
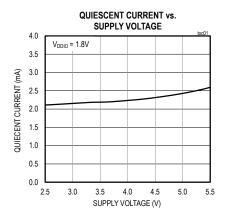


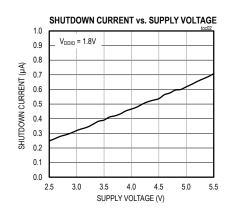
Figure 3. TDM Audio Interface Timing Diagram

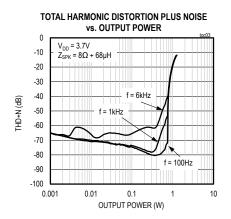
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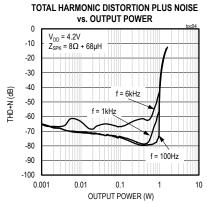
Typical Operating Characteristics

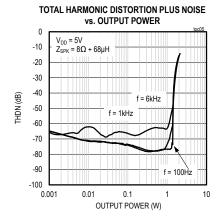
 $(V_{DD} = 5V, V_{GND} = 0V, GAIN_SLOT = GND (+12dB). BCLK = 3.072MHz, LRCLK = 48kHz, speaker loads (Z_{SPK}) connected between OUTP and OUTN, <math>T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25$ °C.)

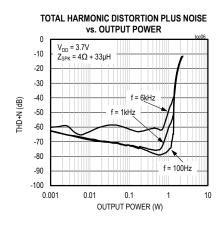


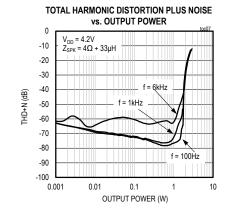


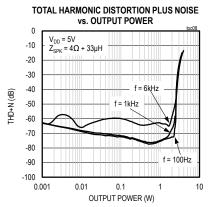








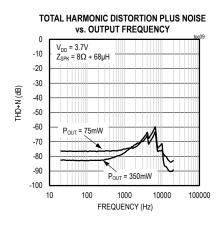


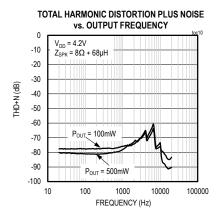


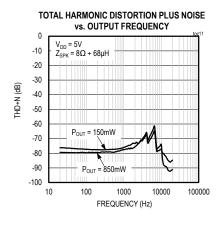
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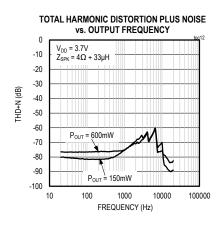
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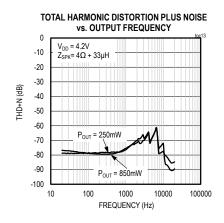
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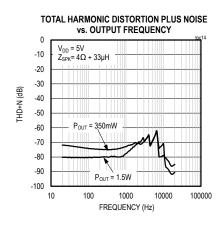


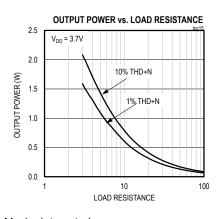


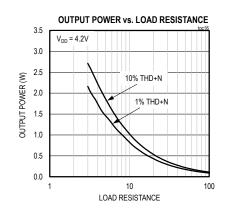


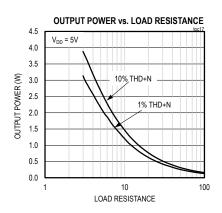








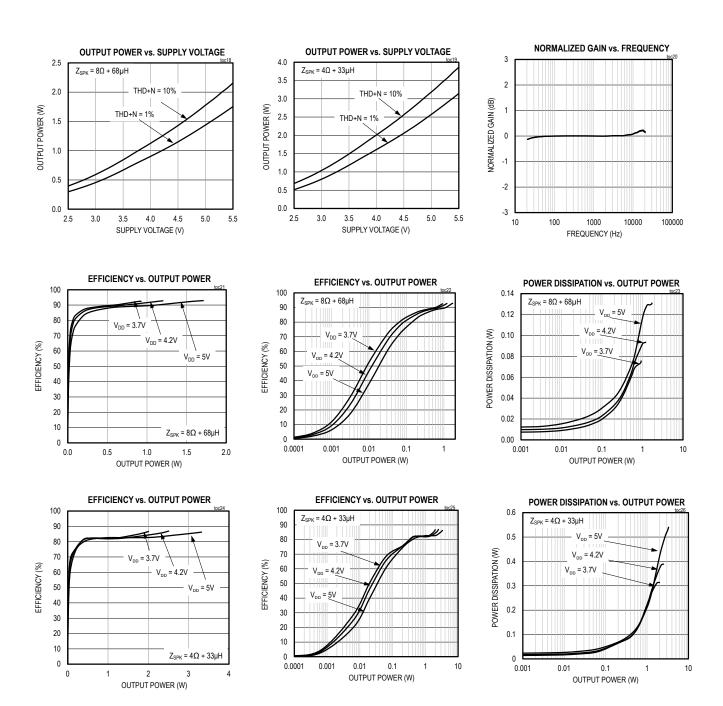




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Typical Operating Characteristics (continued)

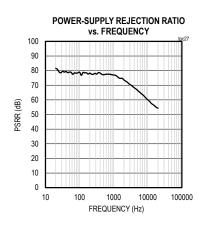
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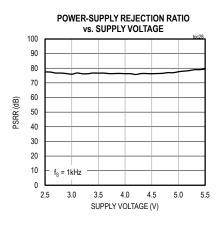


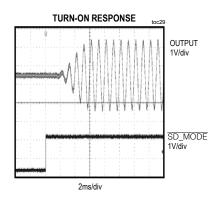
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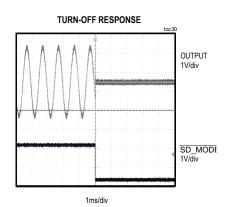
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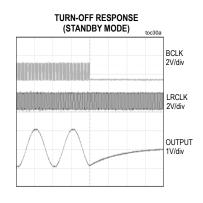
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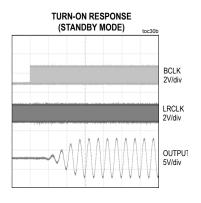


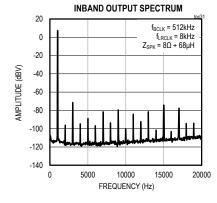


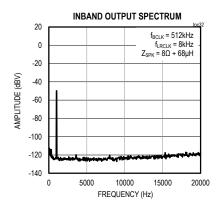


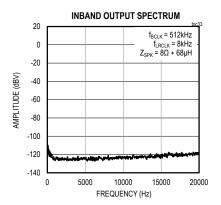








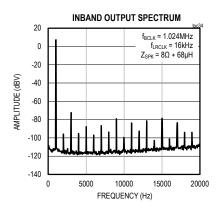


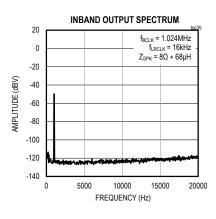


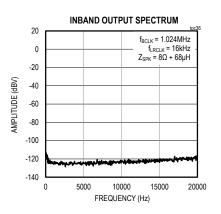
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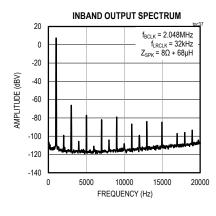
Typical Operating Characteristics (continued)

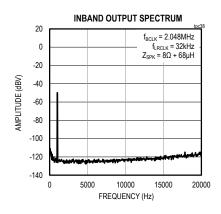
 $(V_{DD} = 5V, V_{GND} = 0V, GAIN_SLOT = GND (+12dB). BCLK = 3.072MHz, LRCLK = 48kHz, speaker loads (Z_{SPK}) connected between OUTP and OUTN, <math>T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25$ °C.)

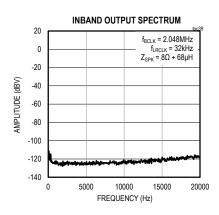


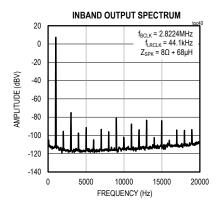


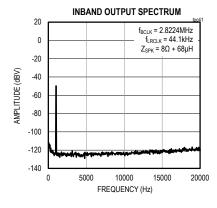


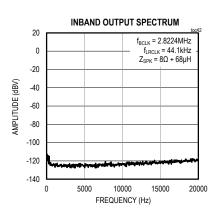








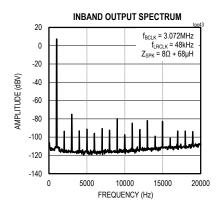


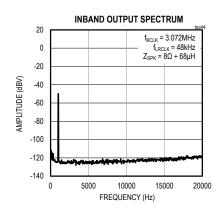


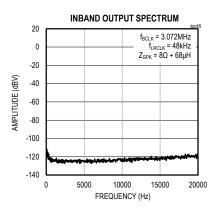
PCM Input Class D Audio Power Amplifiers

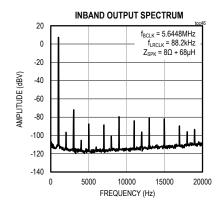
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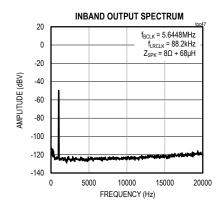
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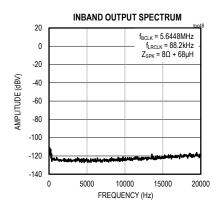


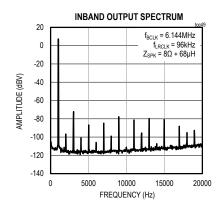


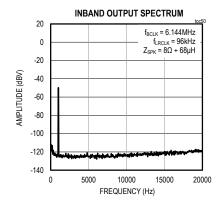


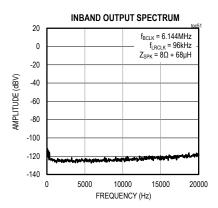






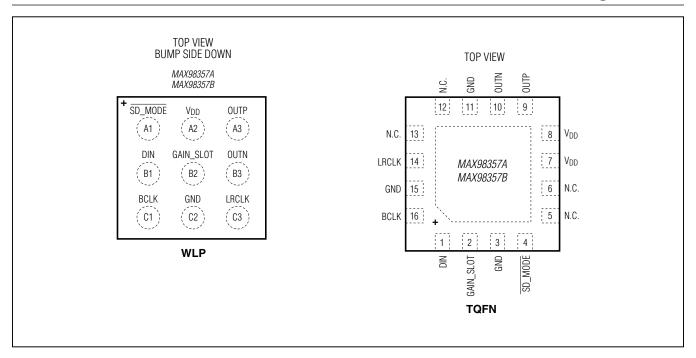






PCM Input Class D Audio Power Amplifiers

Pin Configurations



Pin Description

Р	IN	NAME	FUNCTION		
WLP	TQFN	INAIVIE	FUNCTION		
A1	4	SD_MODE	Shutdown and Channel Select. Pull SD_MODE low to place the device in shutdown. In I2S or LJ mode, SD_MODE selects the data channel (Table 5). In TDM mode, SD_MODE and GAIN_SLOT are both used for channel selection (Table 7).		
A2	7, 8	V_{DD}	Power-Supply Input		
АЗ	9	OUTP	Positive Speaker Amplifier Output		
B1	1	DIN	Digital Input Signal		
B2	2	GAIN_ SLOT	Gain and Channel Selection. In I ² S and LJ mode determines amplifier output gain (Table 8) In TDM mode, used for channel selection with SD_MODE (Table 7). In TDM mode, gain is fixed at 12dB.		
В3	10	OUTN	Negative Speaker Amplifier Output		
C1	16	BCLK	Bit Clock Input		
C2	3, 11, 15	GND	Ground		
СЗ	14	LRCLK	Frame Clock. Left/right clock for I ² S and LJ mode. Sync clock for TDM mode.		
_	5, 6, 12, 13	N.C.	No Connection		
	_	EP	Exposed Pad. The exposed pad is not internally connected. Connect the exposed page to a solid ground plane for thermal dissipation.		

PCM Input Class D Audio Power Amplifiers

Detailed Description

The MAX98357A/MAX98357B are digital PCM input Class D power amplifiers. The MAX98357A accepts standard I²S data through DIN, BCLK, and LRCLK while the MAX98357B accepts left-justified data through the same inputs. Both versions also accept 16-bit or 32-bit TDM data with up to eight slots. The digital audio interface eliminates the need for an external MCLK signal that is typically required for I²S data transmission.

SD_MODE selects which data word is output by the amplifier and is used to put the ICs into shutdown. These devices offer five gain settings in I2S/left-justified mode and a fixed 12dB gain in TDM mode. Channel selection in TDM mode is set with the combination of SD_MODE and GAIN_SLOT (Table 7).

The MAX98357A/MAX98357B feature low-quiescent current, comprehensive click-and-pop suppression, and excellent RF immunity. The ICs offer Class AB audio performance with Class D efficiency in a minimal board-space solution. The Class D amplifier features spread-spectrum modulation with edge-rate and overshoot control circuitry that offers significant improvements in switch-mode amplifier radiated emissions. The amplifier features click-and-pop suppression that reduces audible transients on startup and shutdown. The amplifier includes thermal-overload and short-circuit protection.

Digital Audio Interface Modes

The input stage of the digital audio interface is highly flexible, supporting 8kHz–96kHz sampling rates with 16/24/32-bit resolution for I²S/left justified data as well as up to a 8-slot, 16-bit or 32-bit time division multiplexed (TDM) format. When LRCLK has a 50% duty cycle the data format is determined by the part number selection (MAX98357A/MAX98357B). When a frame sync pulse is used for the LRCLK the data format is automatically configured in TDM mode. The frame sync pulse indicates the beginning of the first time slot.

MCLK Elimination

The ICs eliminate the need for the external MCLK signal that is typically used for PCM communication. This reduces EMI and possible board coupling issues in addition to reducing the size and pin-count of the ICs.

BCLK Jitter Tolerance

The ICs feature a BCLK jitter tolerance of 0.5ns for RMS jitter below 40kHz and 12ns for wideband RMS jitter while maintaining a dynamic range greater than 98dB (Table 1).

BCLK Polarity

When operating in I²S/left justified mode, incoming serial data is always clocked-in on the rising edge of BCLK. In TDM mode, the MAX98357A clocks-in serial data on the rising edge of BCLK while the MAX98357B clocks in serial data on the falling edge of BCLK (Table 2).

LRCLK Polarity

LRCLK specifies whether left-channel data or right-channel data is currently being read by the digital audio interface. The MAX98357A indicates the left channel word when LRCLK is low, and the MAX98357B indicates the left channel word when LRCLK is high (Table 3).

LRCLK ONLY supports 8kHz, 16kHz, 32kHz, 44.1kHz, 48kHz, 88.2kHz and 96kHz frequencies. LRCLK clocks at 11.025kHz, 12kHz, 22.05kHz and 24kHz are **NOT** supported. Do not remove LRCLK while BCLK is present. Removing LRCLK while BCLK is present can cause unexpected output behavior including a large DC output voltage.

Standby Mode

The ICs automatically enter standby mode when BCLK is removed. If BCLK stops toggling, the ICs automati-

Table 1. RMS Jitter Tolerance

FREQUENCY	RMS JITTER TOLERANCE (ns)
< 40kHz	0.5
40kHz-BCLK	12

Table 2. BCLK Polarity

MODE	PART NUMBER	BCLK POLARITY	
I2S	MAX98357A	Rising edge	
Left-justified	MAX98357B	Rising edge	
TDM	MAX98357A	Rising edge	
I DIVI	MAX98357B	Falling edge	

Table 3. LRCLK Polarity

PART NUMBER	LRCLK POLARITY (LEFT CHANNEL)
MAX98357A	Low
MAX98357B	High

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cally enter standby mode. In standby mode, the Class D speaker is turned off and the outputs go into a high-impedance state, ensuring that unwanted current is not transferred to the load during this condition. Standby mode has reduced power consumption from normal operation (340 μ A), but does not reach as low as full shutdown (0.6 μ A). Standby mode can be used to reduce power consumption when no GPIO us available to pull SD MODE low.

DAC Digital Filters

The DAC features a digital lowpass filter that is automatically configured for voice playback or music playback based on the sample rate that is used. This filter eliminates the effect of aliasing and any other high-frequency noise that might otherwise be present. Table 4 shows the digital filter settings that are automatically selected.

SD_MODE and Shutdown Operation

The ICs feature a low-power shutdown mode, drawing less than $0.6\mu A$ (typ) of supply current. During shutdown, all internal blocks are turned off, including setting the output stage to a high-impedance state. Drive $\overline{\text{SD_MODE}}$ low to put the ICs into shutdown.

The state of SD_MODE determines the audio channel that is sent to the amplifier output (Table 5).

Drive SD_MODE high to select the left word of the stereo input data. Drive SD_MODE high through a sufficiently small resistor to select the right word of the stereo input data. Drive SD_MODE high through a sufficiently large resistor to select both the left and right words of the stereo input data (left/2 + right/2). R_{LARGE} and R_{SMALL} are determined by the V_{DDIO} voltage (logic voltage from control interface) that is driving SD_MODE according to the following two equations:

$$R_{SMALL}$$
 (k Ω) = 94.0 x V_{DDIO} - 100
 R_{LARGE} (k Ω) = 222.2 x V_{DDIO} - 100

When the devices are configured in left-channel mode $(\overline{SD_MODE}$ is directly driven to logic-high by the control interface), take care to avoid violating the Absolute Maximum Ratings limits for $\overline{SD_MODE}$. Ensuring that $\underline{V_{DD}}$ is always greater than V_{DDIO} is one way to prevent $\overline{SD_MODE}$ from violating the Absolute Maximum Ratings limits. If this is not possible in the application (e.g., if V_{DD} < 3.0V and V_{DDIO} = 3.3V), then it is necessary to add a small resistance (~2k Ω) in series with $\overline{SD_MODE}$ to limit the current into the $\overline{SD_MODE}$ pin. This is not a concern when using the right channel or (left/2 + right/2) modes.

Figure 4 and Figure 5 show how to connect an external resistor to SD_MODE when using an open-drain driver or a push-pull driver.

Table 4. Digital Filter Settings

LRCLK FREQUENCY	-3dB CUTOFF FREQUENCY	RIPPLE LIMIT CUTOFF FREQUENCY	STOPBAND CUTOFF FREQUENCY	STOPBAND ATTENUATION (dB)
f _{LRCLK} < 30kHz	0.446 x f _{LRCLK}	0.443 x f _{LRCLK}	0.464 x f _{LRCLK}	75
30kHz < f _{LRCLK} < 50kHz	0.47 x f _{LRCLK}	0.43 x f _{LRCLK}	0.58 x f _{LRCLK}	60
f _{LRCLK} > 50kHz	0.31 x f _{LRCLK}	0.24 x f _{LRCLK}	0.477 x f _{LRCLK}	60

Table 5. SD_MODE Control

SD_MODI	SELECTED CHANNEL	
High	V _{SD_MODE} > B2 trip point	Left
Pullup through R _{SMALL}	B2 trip point > V _{SD_MODE} > B1 trip point	Right
Pullup through R _{LARGE}	B1 trip point > V _{SD_MODE} > B0 trip point	(Left/2 + right/2)
Low	B0 trip point > VSD_MODE	Shutdown

Table 6. Examples of SD_MODE Pullup Resistor Values

LOGIC VOLTAGE LEVEL (V _{DDIO}) (V)	R _{SMALL} (kΩ)	R _{LARGE} (kΩ)
1.8	69.8	300
3.3	210.2	634

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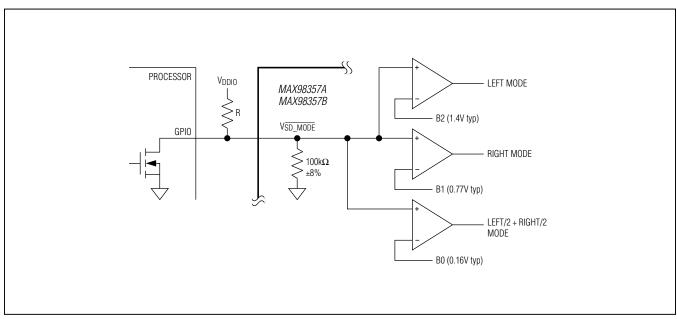


Figure 4. SD_MODE Resistor Connected Using Open-Drain Driver

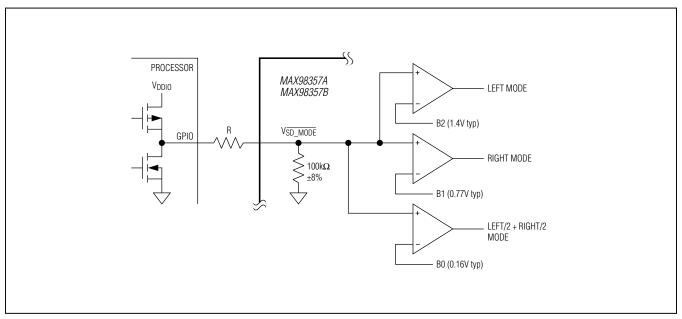


Figure 5. SD_MODE Resistor Connected Using Push-Pull Driver

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I²S and Left Justified Mode

The MAX98357A follows standard I2S timing by allowing a delay of one BCLK cycle after the LRCLK transition before the beginning of a new data word (Figure 6 and Figure 7). The MAX98357B follows the left justified timing specification by aligning the LRCLK transitions with the beginning of a new data word (Figure 8 and Figure 9). LRCLK ONLY supports 8kHz, 16kHz, 32kHz, 44.1kHz, 48kHz, 88.2kHz, and 96kHz frequencies. LRCLK clocks at 11.025kHz, 12kHz, 22.05kHz and 24kHz are **NOT** supported. Do not remove LRCLK while BLCK is present. Removing LRCLK while BCLK is present. Removing LRCLK while BCLK is present can cause unexpected output behavior, including a large DC output voltage.

The digital <u>audio interface</u> output mode is chosen by the voltage at <u>SD_MODE</u>. <u>Table 5</u> shows how the available modes are selected. Trip point <u>B0-B2</u> are shown the <u>Electrical Characteristics</u> in the <u>SD_MODE</u> Comparator Trip Points section. Values for <u>SD_MODE</u> pullup resistors

 R_{SMALL} and R_{LARGE} are dependent on the voltage level of V_{DDIO} . See Table 6 for pullup resistor values.

TDM Mode

TDM mode is automatically detected by monitoring the short channel sync pulse on LRCLK. The frequency detector circuit detects the bit depth. In TDM mode, the MAX98357A/MAX98357B has a fixed gain of 12dB. GAIN_SLOT and SD_MODE are used to select to which of 8 channels of TDM data the parts respond. Table 7 shows the connections for GAIN_SLOT and SD_MODE for channel selection. The MAX98357A data is valid on the BCLK rising edge. The MAX98357B data is valid on the BCLK falling edge.

Figure 10, Figure 11, Figure 12, and Figure 13 show TDM operation, in which a frame-sync pulse is used for LRCLK. In TDM mode, there must be 128 (16-bit mode) or 256 (32-bit mode) BCLK cycles per frame. In TDM mode, the ICs only accept 16-bit or 32-bit formatted data and any of the 8 TDM slots can be selected.

Table 7. TDM Mode Channel Selection

SD_MODE	GAIN_SLOT	CHANNEL	BITS
Low	X	Off	N/A
V_{DD}	GND	0	16/32
V_{DD}	V_{DD} with 0Ω	1	16/32
V_{DD}	Float	2	16/32
V _{DD}	V_{DD} with 100k Ω	3	16/32
V_{DD}	GND with 100kΩ	4	16/32
V _{DD} through R _{LARGE}	GND	5	16/32
V _{DD} through R _{LARGE}	Float	6	16/32
V _{DD} through R _{LARGE}	V _{DD}	7	16/32

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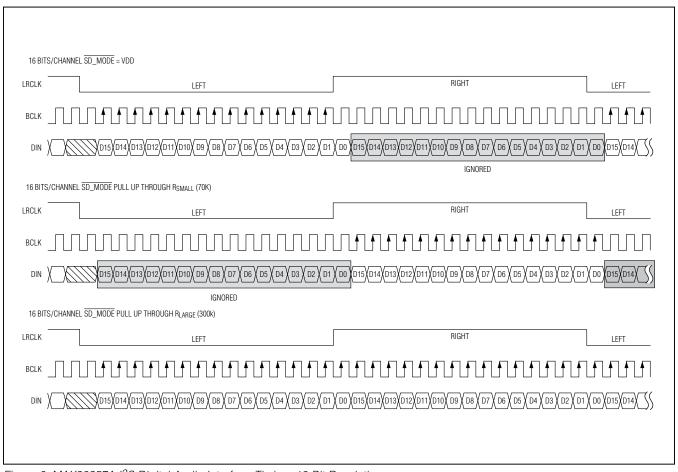


Figure 6. MAX98357A I²S Digital Audio Interface Timing, 16-Bit Resolution

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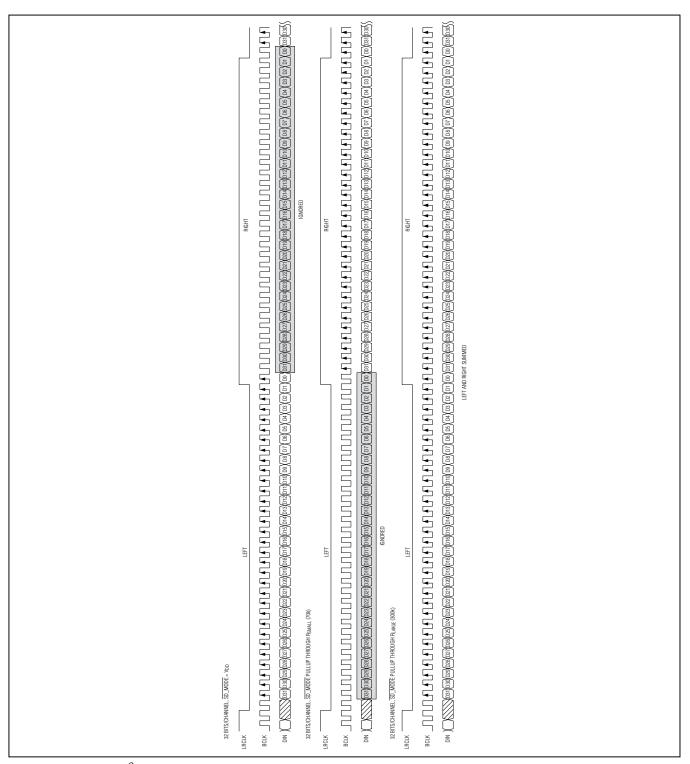


Figure 7. MAX98357A I²S Digital Audio Interface Timing, 32-Bit Resolution

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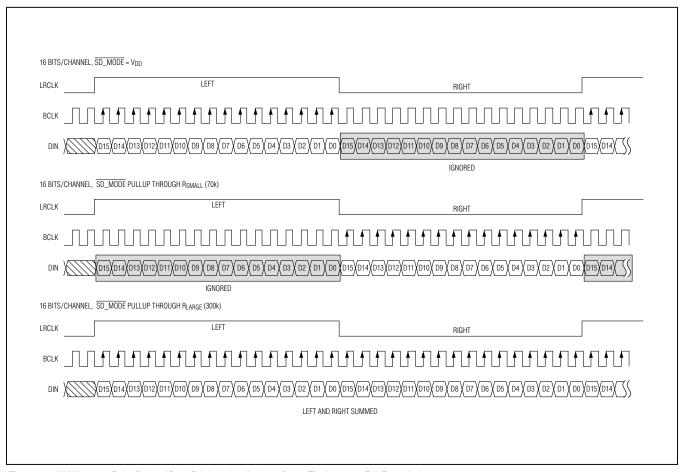


Figure 8. MAX98357B Left-Justified Digital Audio Interface Timing, 16-Bit Resolution

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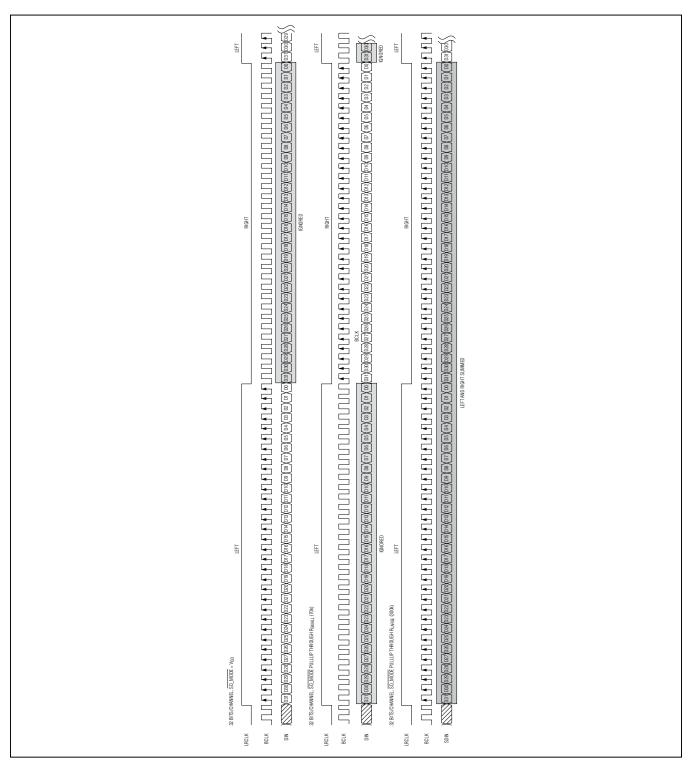


Figure 9. MAX98357B Left-Justified Digital Audio Interface Timing, 32-Bit Resolution

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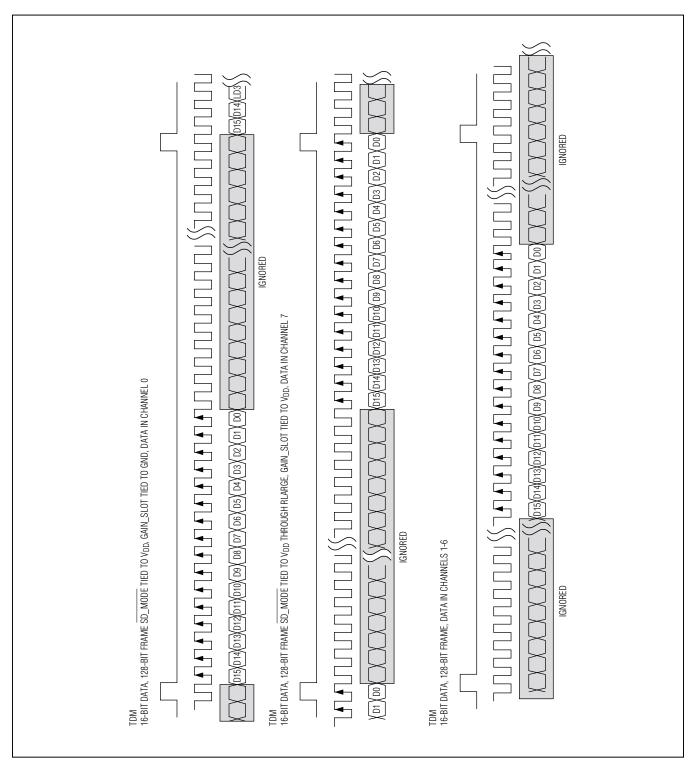


Figure 10. MAX98357A TDM 16-Bit DAI Timing

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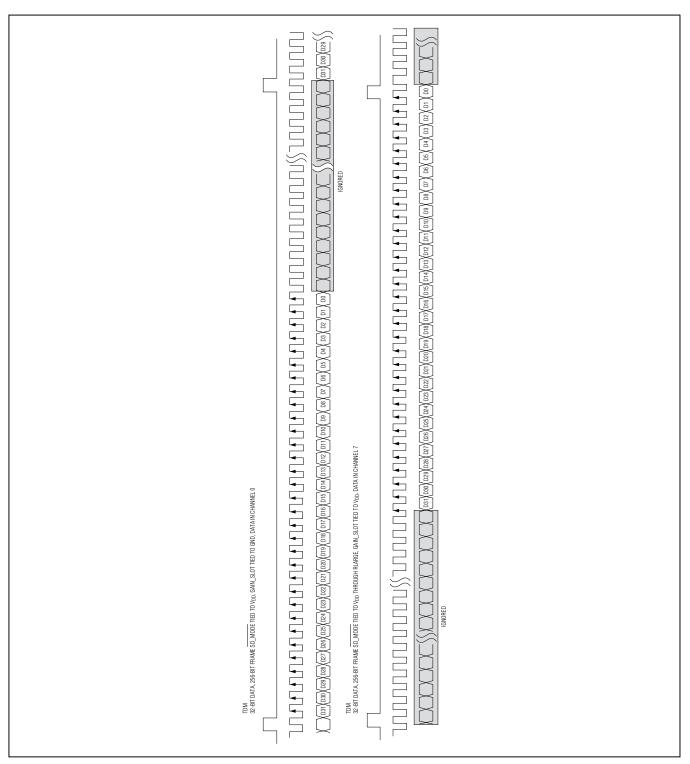


Figure 11. MAX98357A TDM 32-Bit DAI Timing

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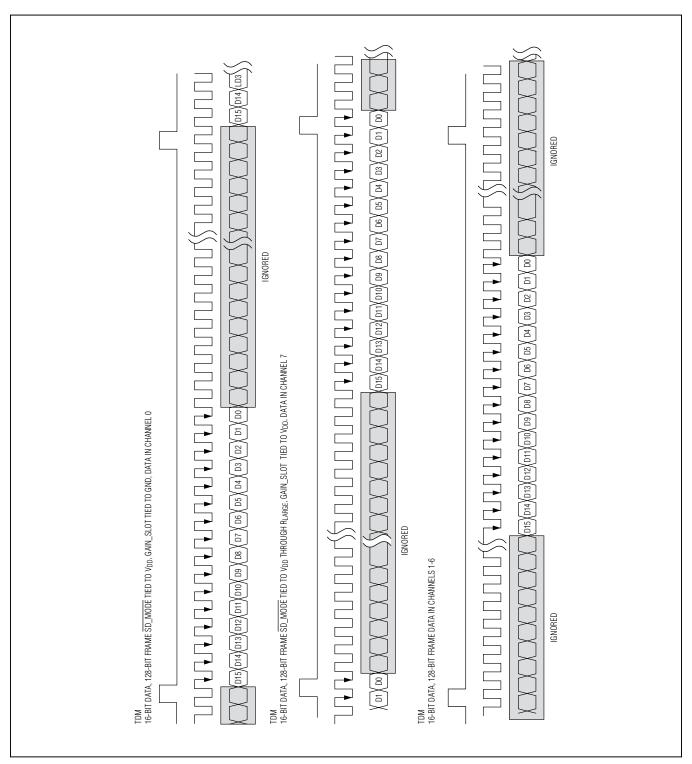


Figure 12. MAX98357B TDM 16-Bit DAI Timing

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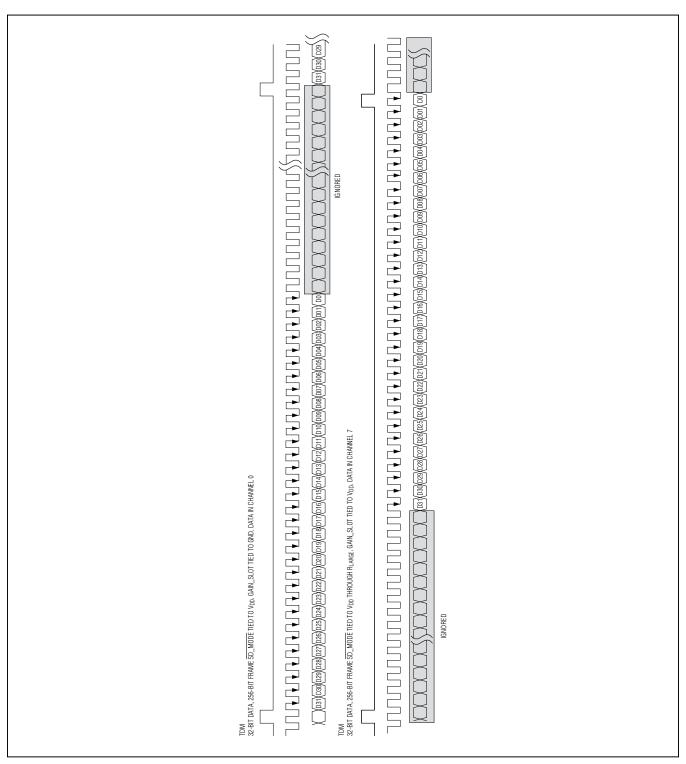


Figure 13. MAX98357B TDM 32-Bit DAI Timing

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Class D Speaker Amplifier

The filterless Class D amplifier offers much higher efficiency than Class AB amplifiers. The high efficiency of a Class D amplifier is due to the switching operation of the output stage transistors. Any power loss associated with the Class D output stage is mostly due to the I²R loss of the MOSFET on-resistance and quiescent current overhead.

Ultra-Low EMI Filterless Output Stage

Traditional Class D amplifiers require the use of external LC filters, or shielding, to meet EN55022B electromagnetic-interference (EMI) regulation standards. Maxim's active emissions-limiting edge-rate control circuitry and spread-spectrum modulation reduces EMI emissions while maintaining up to 92% efficiency.

Maxim's spread-spectrum modulation mode flattens wideband spectral components while proprietary techniques ensure that the cycle-to-cycle variation of the switching period does not degrade audio reproduction or efficiency. The ICs' spread-spectrum modulator randomly varies the switching frequency by ±12.5kHz around the center frequency (330kHz). Above 10MHz, the wideband spectrum looks like noise for EMI purposes (Figure 14).

Speaker Current Limit

If the output current of the speaker amplifier exceeds the current limit (2.8A typ), the IC disables the outputs for approximately 100µs. At the end of the 100µs, the outputs are re-enabled. If the fault condition still exists, the IC continues to disable and reenable the outputs until the fault condition is removed.

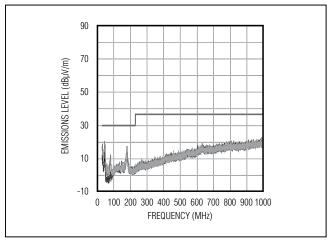


Figure 14. EMI with 12in of Speaker Cable and No Output Filtering

Gain Selection

The ICs offer five programmable gain selections through a single gain input (GAIN_SLOT) in I²S/left justified mode. Gain is referenced to the full-scale output of the DAC, which is 2.1dBV (Table 8). In TDM mode, the gain is automatically set at a fixed 12dB. Assuming that the desired output swing is not limited by the supply voltage rail, the IC's output level can be calculated based on the digital input signal level and selected amplifier gain according to the following equation:

Output signal level (dBV) = input signal level (dBFS) + 2.1dB + selected amplifier gain (dB)

where OdBFS is referenced to OdBV.

Click-and-Pop Suppression

The IC speaker amplifier features Maxim's comprehensive click-and-pop suppression. During startup, the click-and-pop suppression circuitry reduces audible transient sources internal to the device by ramping the input signal from mute to 0dB. When entering shutdown, the differential speaker outputs simultaneously drop to GND.

The comprehensive click-and-pop suppression of the MAX98357 is unaffected by power-up or power-down sequencing. Applying the DAI clocks before or after the transition of SD_MODE yields the same click-and-pop performance. The MAX98357 does not have a volume ramp-down response when entering shutdown. For optimal click-and-pop performance, ramp down the digital data on SDIN before powering down the MAX98357.

Table 8. Gain Selection

GAIN_SLOT	I ² S/LJ GAIN (dB)
Connect to GND through 100kΩ ±5% resistor	15
Connect to GND	12
Unconnected	9
Connect to V _{DD}	6
Connect to V _{DD} through 100kΩ ±5% resistor	3

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Applications Information

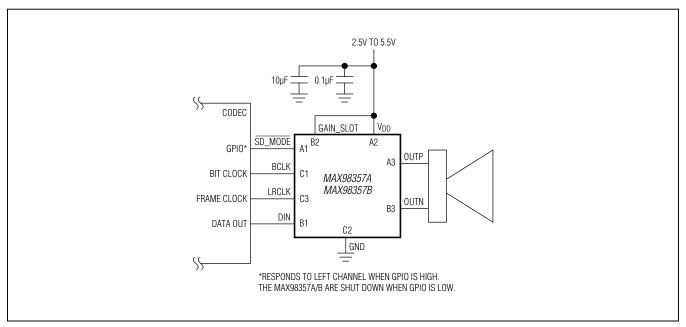


Figure 15. Left-Channel PCM Operation with 6dB Gain

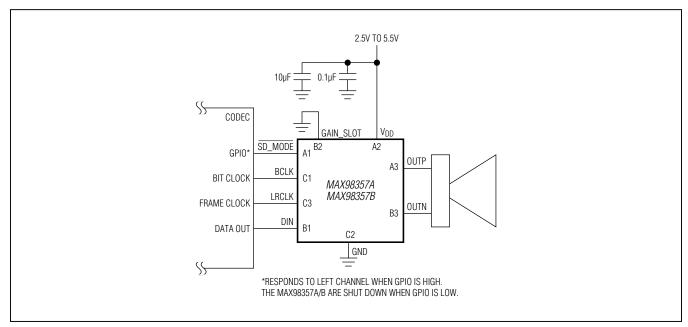


Figure 16. Left-Channel PCM Operation with 12dB Gain

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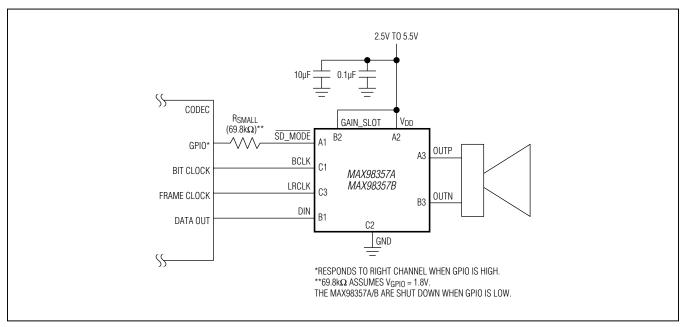


Figure 17. Right-Channel PCM Operation with 6dB Gain

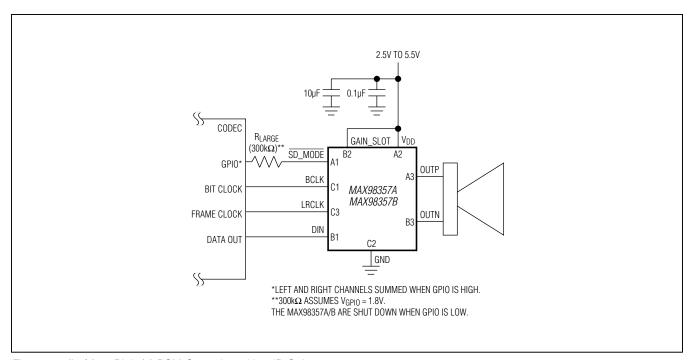


Figure 18. (Left/2 + Right/2) PCM Operation with 6dB Gain

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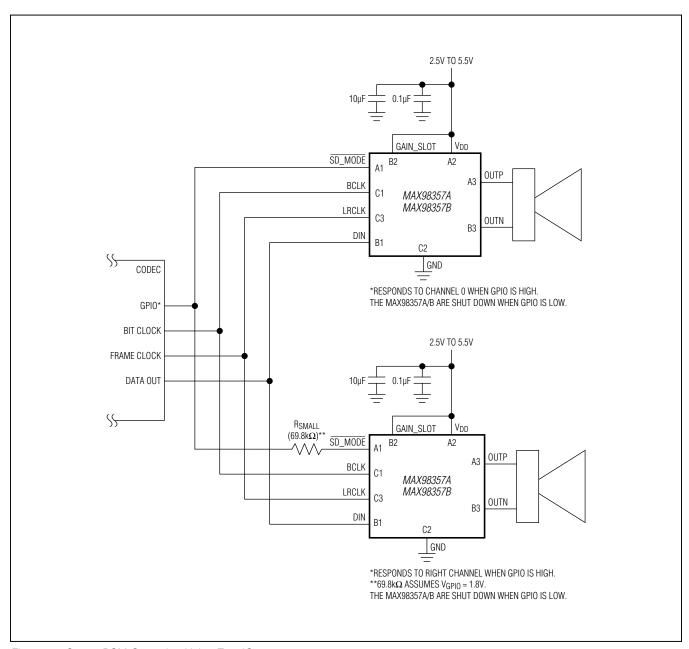


Figure 19. Stereo PCM Operation Using Two ICs

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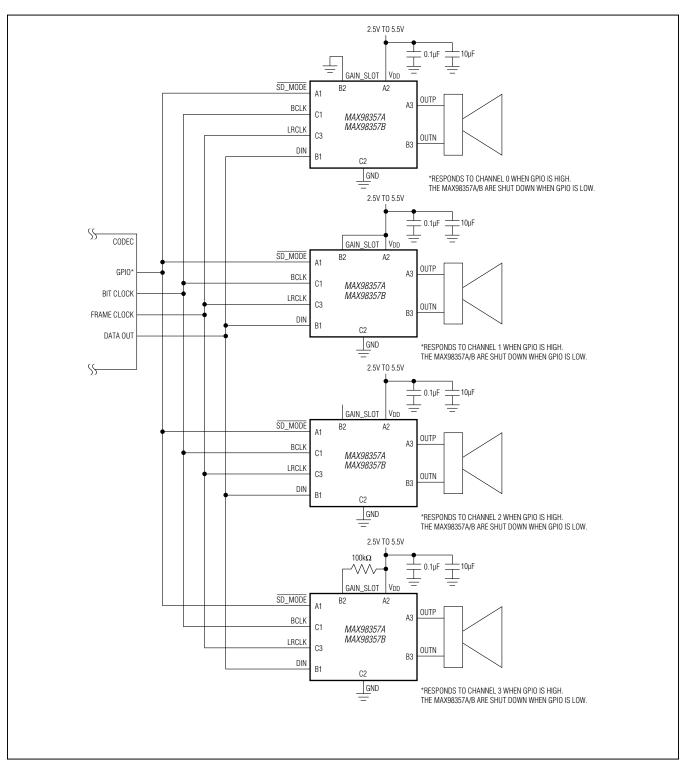
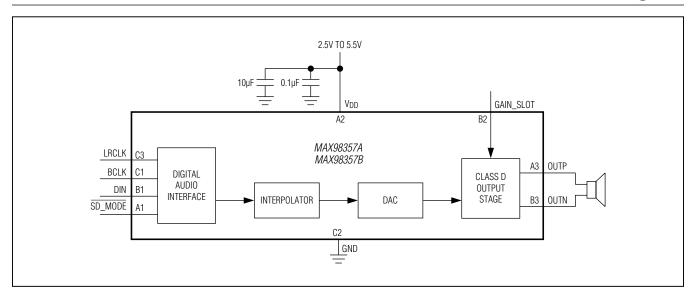


Figure 20. Channel TDM Operation (Gain Fixed at 12dB)

PCM Input Class D Audio Power Amplifiers

Functional Diagram



Filterless Class D Operation

Traditional Class D amplifiers require an output filter to recover the audio signal from the amplifier's output. The filter adds cost, size, and decreases efficiency and THD+N performance. The ICs' filterless modulation scheme does not require an output filter. The device relies on the inherent inductance of the speaker coil and the natural filtering of both the speaker and the human ear to recover the audio component of the square-wave output.

Because the switching frequency of the ICs is well beyond the bandwidth of most speakers, voice coil movement due to the switching frequency is very small. Use a speaker with a series inductance > $10\mu H$. Typical 8Ω speakers exhibit series inductances in the $20\mu H$ to $100\mu H$ range.

Power-Supply Input

 V_{DD} , which ranges from 2.5V to 5.5V, powers the IC, including the speaker amplifier. Bypass V_{DD} with a 0.1 μ F and 10 μ F capacitor to GND. Some applications might require only the 10 μ F bypass capacitor, making it possible to operate with a single external component. Apply additional bulk capacitance at the ICs if long input traces between V_{DD} and the power source are used.

Layout and Grounding

Proper layout and grounding are essential for optimum performance. Good grounding improves audio performance and prevents switching noise from coupling into the audio signal.

Use wide, low-resistance output traces. As load impedance decreases, the current drawn from the device outputs increases. At higher current, the resistance of the output traces decreases the power delivered to the load. For example, if 2W is delivered from the speaker output to a 4Ω load through $100m\Omega$ of total speaker trace, 1.904W is being delivered to the speaker. If power is delivered through $10m\Omega$ of total speaker trace, 1.951W is being delivered to the speaker. Wide output, supply, and ground traces also improve the power dissipation of the ICs. Parasitic capacitance on the output traces cause higher quiescent current by V_{DD} x 330kHz x $C_{PARASITIC}$.

For example, at V_{DD} = 5V and a total parasitic capacitance of 100pF (50pF on each output trace), the increase in quiescent current is 5 x 330kHz x 100pF = 165 μ A.

The ICs are inherently designed for excellent RF immunity. For best performance, add ground fills around all signal traces on top or bottom PCB planes.

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WLP Applications Information

For the latest application details on WLP construction, dimensions, tape carrier information, PCB techniques, bump-pad layout, and recommended reflow temperature profile, as well as the latest information on reliability testing results, refer to the Application Note 1891: Wafer-Level Packaging (WLP) and Its Applications. Figure 21 shows the dimensions of the WLP balls used on the ICs.

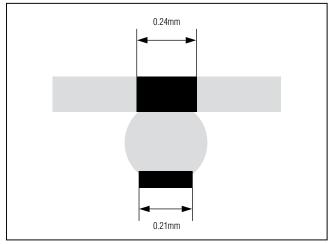


Figure 21. MAX98357A/MAX98357B WLP Ball Dimensions

Ordering Information

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX98357AETE+*	-40°C to +85°C	16 TQFN	_
MAX98357AETE+T*	-40°C to +85°C	16 TQFN	_
MAX98357AEWL+T	-40°C to +85°C	9 WLP	+AKM
MAX98357BETE+*	-40°C to +85°C	16 TQFN	_
MAX98357BETE+T*	-40°C to +85°C	16 TQFN	<u> </u>
MAX98357BEWL+T	-40°C to +85°C	9 WLP	+AKN

⁺Denotes a lead(Pb)-free/RoHS-compliant package.

T = Tape and reel.

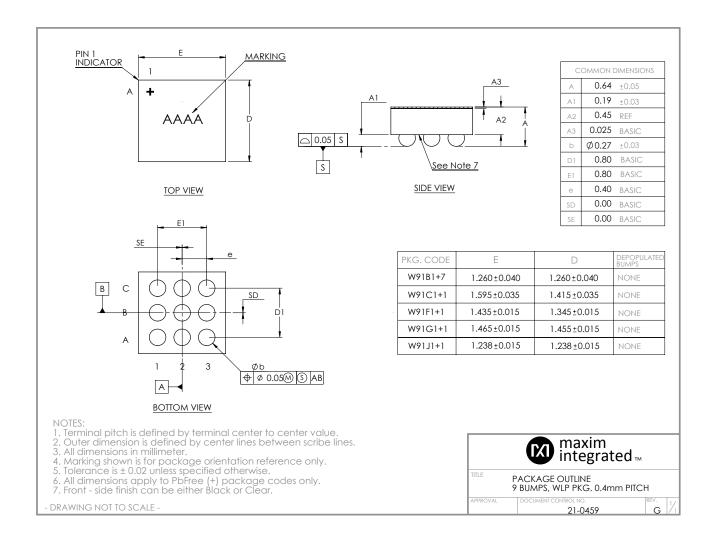
^{*}Future product—contact factory for availability.

PCM Input Class D Audio Power Amplifiers

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

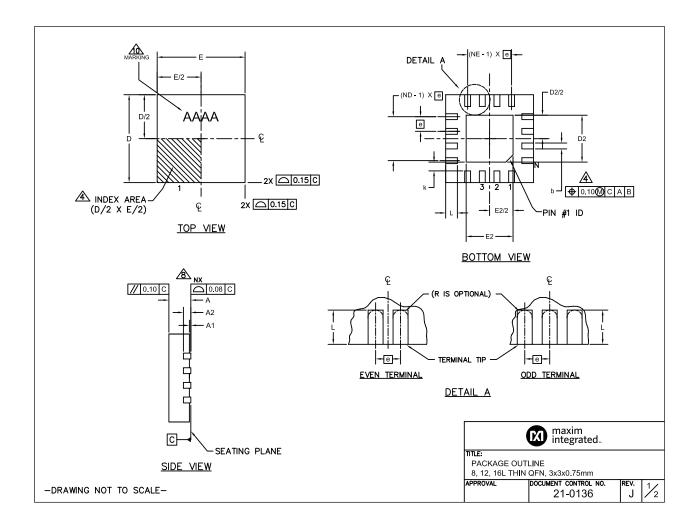
PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
9 WLP	W91F1+1	<u>21-0459</u>	Refer to Application Note 1891
16 TQFN	T1633+4	21-0136	90-0031



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Package Information (continued)

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PKG		8L 3x3		12L 3x3			16L 3x3			
REF.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	
Α	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	
b	0.25	0.30	0.35	0.20	0.25	0.30	0.20	0.25	0.30	
О	2.90	3.00	3.10	2.90	3.00	3.10	2.90	3.00	3.10	
П	2.90	3.00	3.10	2.90	3.00	3.10	2.90	3.00	3.10	
е	0	.65 BS	C.	0	0.50 BSC.		0	0.50 BSC.		
L	0.35	0.55	0.75	0.45	0.55	0.65	0.30	0.40	0.50	
Z		8			12 16					
ND		2			3			4		
NE		2		3 4						
A1	0	0.02	0.05	0	0.02	0.05	0	0.02	0.05	
A2	0	.20 RE	F	0.20 REF 0.20 REF			F			
k	0.25	-	-	0.25	-	-	0.25	-	-	

EXPOSED PAD VARIATIONS									
PKG.	D2			E2			DIN ID	IEDEO	
CODES	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	PIN ID	JEDEC	
TQ833-1	0.25	0.70	1.25	0.25	0.70	1.25	0.35 x 45°	WEEC	
T1233-1	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-1	
T1233-3	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-1	
T1233-4	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-1	
T1633-2	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-2	
T1633F-3	0.65	0.80	0.95	0.65	0.80	0.95	0.225 x 45°	WEED-2	
T1633FH-3	0.65	0.80	0.95	0.65	0.80	0.95	0.225 x 45°	WEED-2	
T1633-4	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-2	
T1633-5	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-2	

- 1. DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994.
- 2. ALL DIMENSIONS ARE IN MILLIMETERS. ANGLES ARE IN DEGREES.
- N IS THE TOTAL NUMBER OF TERMINALS.
- 1 THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE OPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR
- ⚠ DIMENSION b APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.20 mm AND 0.25 mm FROM TERMINAL TIP.
- 6 ND AND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY.
- DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION.
- DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION.
 COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS.
- DRAWING CONFORMS TO JEDEC MO220 REVISION C.
- DRAWING CONFORMS TO JEDEC M0220 REVISION C.
 MARKING SHOWN IS FOR PACKAGE ORIENTATION REFERENCE ONLY.
- 11. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.
- 12. WARPAGE NOT TO EXCEED 0.10mm.
- 13. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND Pb FREE (+) PARTS.

-DRAWING NOT TO SCALE-



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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	9/13	Initial release	_
1	11/13	Added two new TOCs, replaced TOC 29, updated Figures 1–3, and made various corrections	1, 4–20, 29–32, 34
2	8/14	Added THD+N for TQFN package with typical spec	5



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