



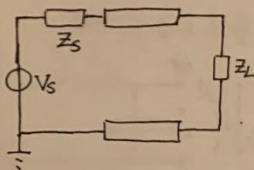
班级:

姓名: 刘开济

编号: 2

科目: 通信电路

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$$Z_s = 10\Omega$$

$$Z_L = 75\Omega$$

$$Z_0 = 50\Omega$$

$$V_{stable} = 0.8824V$$

1. 为什么是这样的波形?

由于信源内阻 Z_s 、负载电阻 Z_L 均不匹配。信源发射的信号到达负载时,由于不匹配会产生一个反射电压;反射电压回到电源后由于不匹配又会再次产生回到负载的反射信号。如此往复,导致了负载电压的振荡特性。

2. 传输线稳定后的反射电压

$$\text{先考查第一个延时有: } V_0^+ = V_0 \cdot \frac{Z_L}{Z_s + Z_L}$$

$$\text{由 } \begin{cases} T_L = \frac{Z_L - Z_0}{Z_L + Z_0} \\ T_s = \frac{Z_s - Z_0}{Z_s + Z_0} \end{cases}$$

$$\text{于是 } t_d = 1ns \text{ 时, 负载电压 } V_L(1) = (1 + T_L) V_0^+ = \frac{2Z_L}{Z_L + Z_0} \frac{Z_L}{Z_s + Z_L} V_0$$

$$\text{而 } V_L(1) = \frac{2 \cdot 75}{75 + 50} \cdot \frac{50}{10 + 50} V_0 = V_0 = 1V$$

当来自负载的反射电压 $V_{ref1} = T_L \cdot V_0^+$ 回到信源时,由于信源不匹配,立刻就有再次反射回负载的信号。

$$V_{ref}(2) = T_s \cdot V_0^+ \cdot T_L = V_0^+ (T_L T_s) = -0.1111V$$

$$\text{负载获得的信号: } V_L(3) = V_{ref}(2) (1 + T_L) = -0.133V$$

于是此时传输线负载电压

$$V = V_{L(1)} + V_{L(3)} = 0.8667V$$

$$\text{类似地: } V_{L(5)} = (1 + T_L) T_L T_s V_{ref}(2)$$

$$= (1 + T_L) T_L T_s (T_s T_L) V_0^+ = V_0^+ (1 + T_L) (T_s T_L)^2$$

$$\text{故: } V_{stable} = \left\{ V_0^+ (1 + T_L) \left[(T_s T_L)^2 + (T_s T_L)^4 + \dots \right] \right\}$$

$$= V_0^+ (1 + T_L) \cdot \frac{1}{1 - T_s T_L} = 0.8824V$$

故有传输线时的稳定电压如:

$$V_{stable} = 0.8824V$$

3. 无传输线时电压

若不存在传输线,则一个cd信号即稳定有:

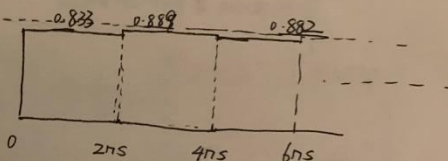
$$V = V_0 \cdot \frac{Z_L}{Z_L + Z_s} = 0.8824V$$

即两者相等。

理解:由于传输线是无损传输线,因而其两端不匹配并不影响系统的稳态解。但不匹配会放大其瞬态解,因而无传输线时1ns即到达稳态,有传输线时则需要很长时间。

4. 不匹配则数字信号不稳定,影响到晶体管的判定阈值问题。这就必须提高晶体管摆幅以维持高的电压裕度,就会影响数字电路的速度。

5. 输入端口波形:



显然,输入端口收敛速度强于输出端口。



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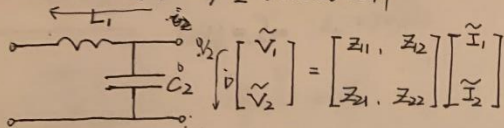
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二. 考查网络 Z 参数与 ABCD 矩阵:



立即得到:

$$\begin{cases} Z_{11} = j\omega L_1 + \frac{1}{j\omega C_2} \\ Z_{22} = \frac{1}{j\omega C_2} \\ Z_{21} = \frac{1}{j\omega C_2} \\ Z_{12} = \frac{1}{j\omega C_2} \end{cases} \quad \text{故有 } Z = \begin{bmatrix} j\omega L_1 + \frac{1}{j\omega C_2} & \frac{1}{j\omega C_2} \\ \frac{1}{j\omega C_2} & \frac{1}{j\omega C_2} \end{bmatrix}$$

$$\text{由: } \begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix}$$

$$A = \left. \frac{V_1}{V_2} \right|_{I_2=0} = \frac{\frac{1}{j\omega C_2} + j\omega L_1}{\frac{1}{j\omega C_2}} = 1 + j\omega^2 L_1 C_2$$

$$-B = \left. \frac{V_1}{I_2} \right|_{V_2=0} = -j\omega L_1 \Rightarrow B = j\omega L_1$$

$$C = \left. \frac{I_1}{V_2} \right|_{I_2=0} = j\omega C_2$$

$$-D = \left. \frac{I_1}{I_2} \right|_{V_2=0} = -1 \Rightarrow D = 1$$

$$\text{于是, LC 网络的 ABCD 参量为: } \begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} 1 + j\omega^2 L_1 C_2 & j\omega L_1 \\ j\omega C_2 & 1 \end{bmatrix}$$

$$\begin{cases} V(z) = V_0^+ \left(\exp(\gamma z) + \Gamma_0 \exp(-\gamma z) \right) \\ I(z) = V_0^+ \left(\exp(\gamma z) + \Gamma_0 \exp(-\gamma z) \right) \end{cases}$$

根据《射频电路设计》教材, 可得其 $V(z), I(z)$ 关系:

$$\begin{cases} V(z) = zj V^+ \sin \beta l \\ I(z) = \frac{V^+}{Z_0} \cos \beta l \end{cases} \quad (\text{短路时})$$

$$\begin{cases} V(z) = V^+ \cos \beta l \\ I(z) = j \frac{V^+}{Z_0} \sin \beta l \end{cases} \quad (\text{开路时})$$

由此先求其 Z 参量阵:

$$Z_{11} = \frac{V(l)}{I(l)}, \text{ 此时由于端口 2 开路, 需用同开}$$

路时的结果:

$$Z_{11} = \frac{V^+ \cos \beta l}{j \frac{V^+}{Z_0} \sin \beta l} = -j Z_0 \cot \beta l$$

$$Z_{22} = \frac{V(0)}{I(0)} = \frac{V^+}{j \frac{V^+}{Z_0} \sin \beta l} = -j Z_0 \cot \beta l$$

$$Z_{21} = \frac{V(0)}{I(l)} \bigg|_{I_2=0} = \frac{V^+}{j \frac{V^+}{Z_0} \sin \beta l} = -j Z_0 / \sin \beta l$$

$$Z_{12} = Z_{21} = -j Z_0 / \sin \beta l$$

故其 Z 参量为:

$$Z = \begin{bmatrix} -j Z_0 \cot \beta l & -j Z_0 / \sin \beta l \\ -j Z_0 / \sin \beta l & -j Z_0 \cot \beta l \end{bmatrix}$$

类似地可得其 ABCD 参量:

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} \cos \beta l & j Z_0 \sin \beta l \\ j Y_0 \sin \beta l & \cos \beta l \end{bmatrix}$$



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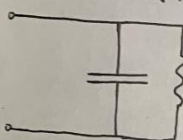
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习题三: 并联谐振及其带宽



$$C = 10 \text{ pF}$$

$$Q = 200$$

$$L = 10 \text{ } \mu\text{H}$$

对于该并联谐振系统, 立刻有: (Q 值是一个谐振概念, 要在谐振频点上查 Q 值)

自由振荡频率 $\omega_0 = \frac{1}{\sqrt{LC}} = 10^8 \text{ rad/s}$

$$f_0 = \frac{1}{2\pi} \omega_0 = 1.592 \times 10^7 \text{ Hz} = 15.92 \text{ MHz}$$

特征阻抗: $Z_0 = \sqrt{\frac{L}{C}} = 1 \text{ k}\Omega$

串联等效电阻: $r_s = \frac{Z_0}{Q} = 5 \Omega$

于是等效并联电阻: $R_p = Q r_s = 200 \text{ k}\Omega$

并联谐振频率: $f_p = \frac{1}{2\pi} \omega_0 \sqrt{1 - \frac{1}{Q^2}} \approx f_0$

于是 $BW_{3dB} = \frac{f_p}{Q} = \frac{15.92 \text{ MHz}}{200} = 79.6 \text{ kHz}$

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习题四: 抽头部分接入:

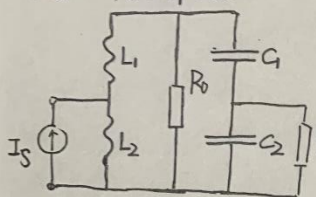
$$L_1 = L_2 = 5 \mu\text{H}$$

$$Q_0 = 100$$

$$C_1 = C_2 = 8 \text{ pF}$$

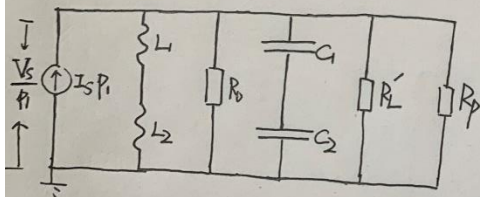
$$R_0 = 40 \text{ k}\Omega$$

$$R_L = 10 \text{ k}\Omega$$



$$P_1 = \frac{L_2}{L_1 + L_2} = \frac{1}{2}, \quad P_2 = \frac{C_1}{C_1 + C_2} = \frac{1}{2}$$

故需做两次部分接入等效, 有等效电路



$$\text{无阻尼谐振频率有: } f_m = \frac{1}{2\pi} \sqrt{\frac{1}{(L_1 + L_2) \parallel (C_1 \parallel C_2)}} \sqrt{1 - \frac{1}{Q^2}}$$

$$\text{得: } f_m = 25.163 \text{ MHz}$$

$$\text{并联谐振回路特征阻抗: } Z_0 = \sqrt{\frac{L_1 + L_2}{C_1 \parallel C_2}} = \frac{1.58 \text{ k}}{1.58 \text{ k}} = 1 \text{ k}\Omega$$

$$\text{故并联电阻: } R_p = Z_0 \cdot Q = 158.11 \text{ k}\Omega$$

$$R_L' = \frac{R_L}{P_2} = 40 \text{ k}\Omega$$

$$\text{而: } \frac{V_s/P_1}{I_s/P_1} = \frac{V_s}{I_s} \cdot \frac{1}{P_1} = R_0 \parallel R_p \parallel R_L'$$

$$\Rightarrow Z_{in} \Big|_{\omega=\omega_p} = P_1^2 (R_0 \parallel R_p \parallel R_L') = 4.44 \text{ k}\Omega$$

考查其 3dB 带宽

考查电路整体 Q 值, 若拿掉负载电阻, 其有

$$Q_{RL} = \frac{Y_0}{G} = \frac{\sqrt{\frac{C}{L}}}{G_0 + G_p + G_L'} = 11.2288$$

若拿掉负载电阻, 则:

$$Q_{no, RL} = \sqrt{\frac{C_1 \parallel C_2}{L_1 + L_2}} \Big/ \frac{1}{G_0 + G_p} = 20.1904$$

故无 RL 时带宽有:

$$BW_{3dB, no, RL} = \frac{f_m}{Q_{no, RL}} = 1.2463 \text{ MHz}$$

有 RL 时带宽有:

$$BW_{3dB, RL} = \frac{f_m}{Q_{RL}} = 2.2410 \text{ MHz}$$



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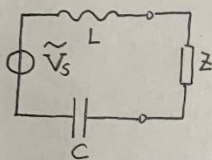
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习题五: 阻抗测量: $f_s = 1 \text{ MHz}$

$$V_{ps} = 0.1 \text{ V}$$



一: 短接电路.

$$\text{短接电路时有谐振频率: } f_m = \frac{1}{2\pi} \sqrt{\frac{1}{LC}}$$

$$\text{短接时, 有: } 1 \text{ MHz} = \frac{1}{2\pi} \sqrt{\frac{1}{L \cdot 100 \text{ pF}}}$$

$$\Rightarrow L = \frac{1}{(2\pi f_m)^2 C} = 2.533 \times 10^{-4} \text{ H} = 253.30 \mu\text{H}$$

谐振时, 电路整体呈阻性, 电压电源全部加在 L 的串联等效电阻 r_s 上, 故有

$$Q = \frac{V_{pp} C}{2 V_{ps}} = 100 = \frac{1}{\omega_0 C r_s}$$

$$\text{于是有: } r_s = \frac{1}{Q \omega_0 C} = 15.916 \Omega$$

$$\text{于是有 } \begin{cases} \text{电感无载品质因数: } Q = 100 \\ \text{电感数值: } L = 253.30 \mu\text{H} \end{cases}$$

二: 接某阻抗元件, 记其有 $Z_L = R_L + j\omega Q$

$$\text{立刻有: } C_2 \parallel C_L = C \Rightarrow C_L = 200 \text{ pF}$$

$$\frac{V_{pp} C}{2 \cdot V_{ps}} = Q = \frac{\sqrt{\frac{L}{C_2 \parallel C_L}}}{r_s + R_L}$$

$$\Rightarrow R_L = 47.75 \Omega$$

$$\text{故 } \begin{cases} \text{阻抗电阻 } R_L = 47.75 \Omega \\ \text{阻抗电容: } C_L = 200 \text{ pF} \end{cases}$$