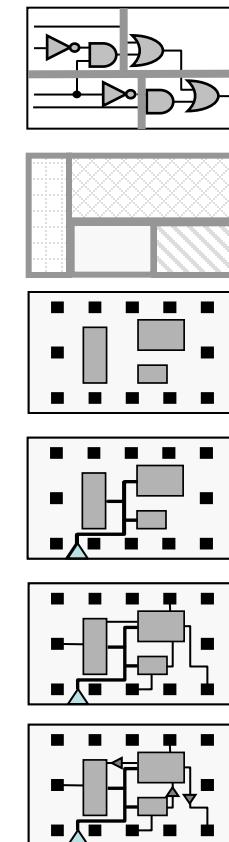
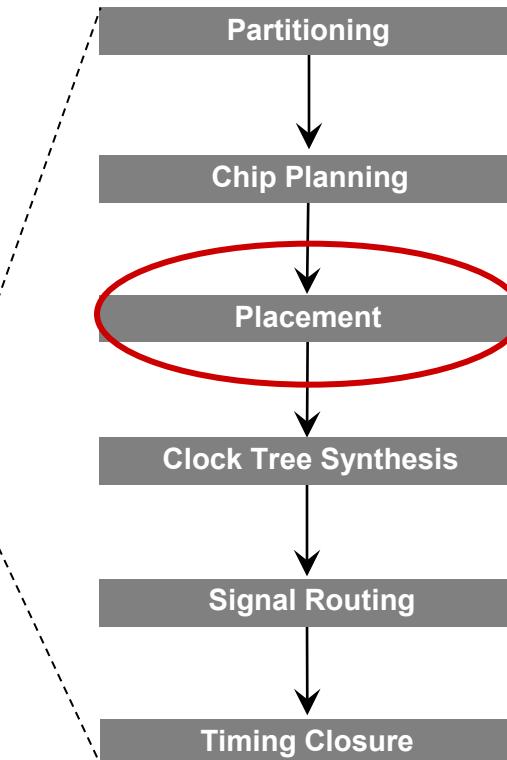
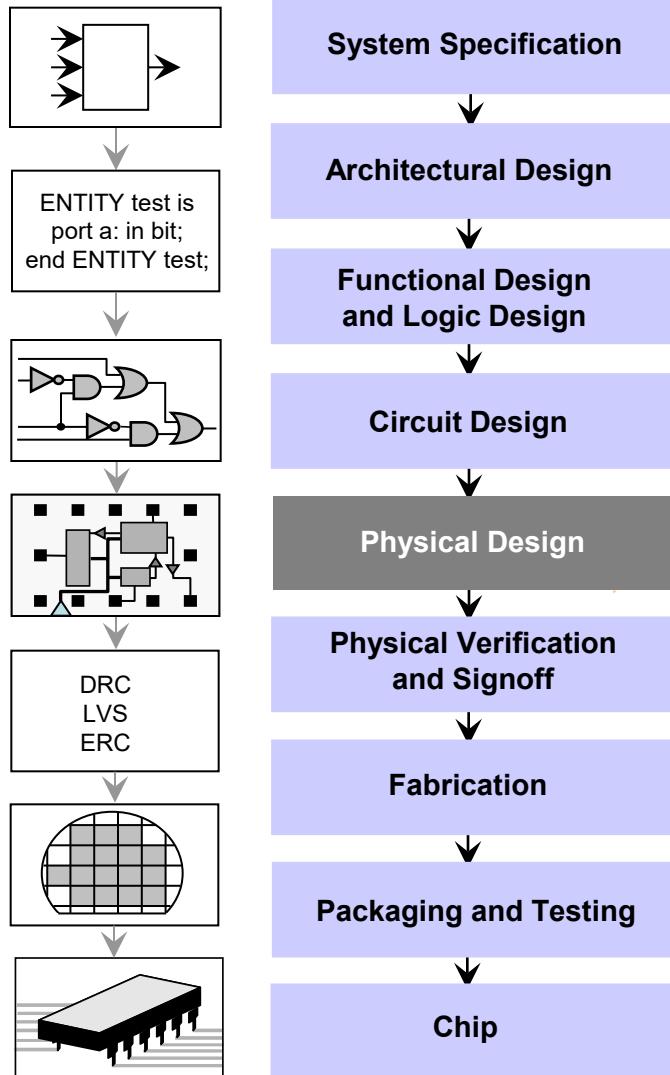


# Lecture 9: Placement

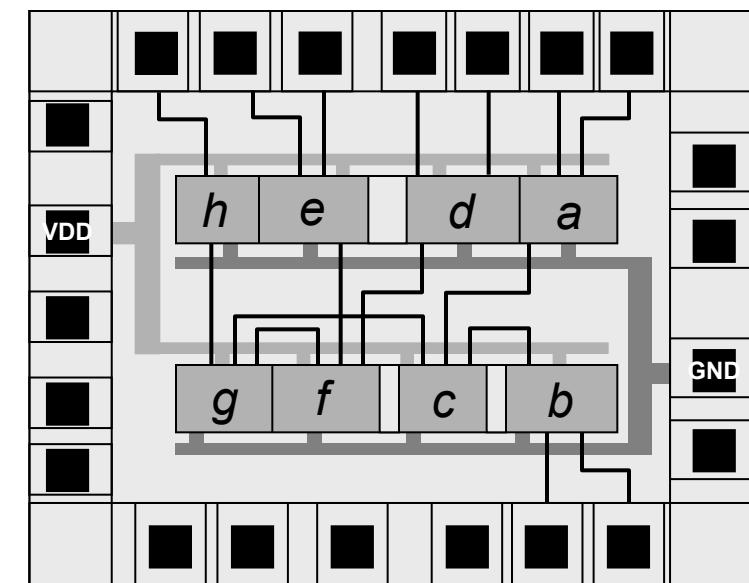
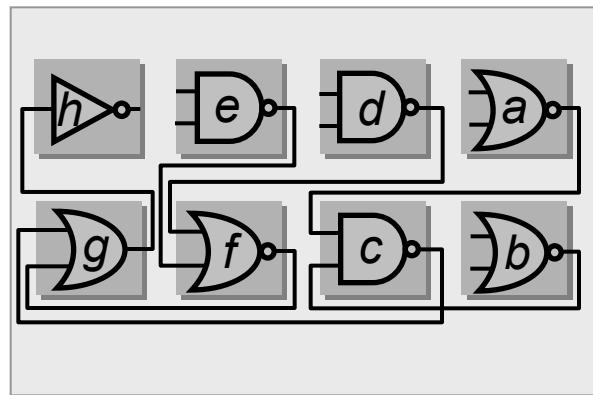
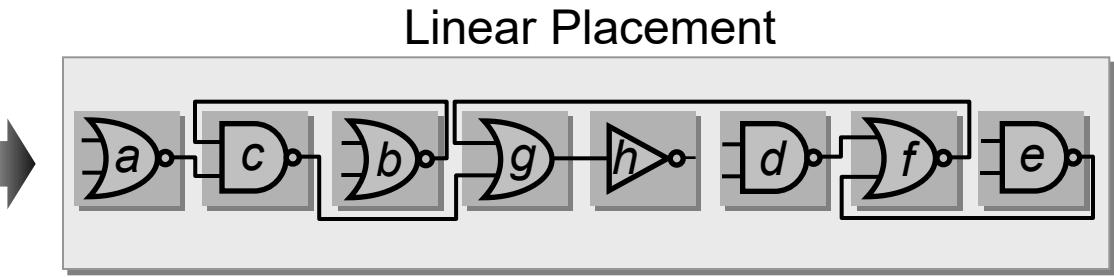
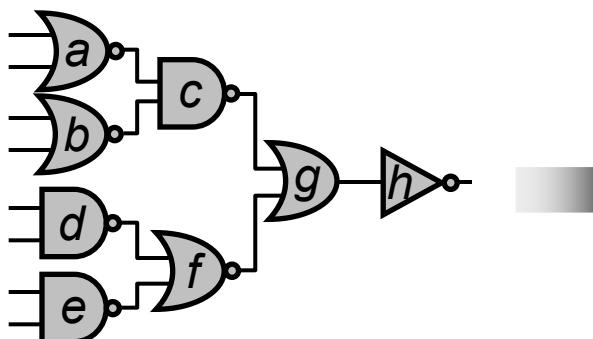
ECE201A

Some notes adopted from  
Andrew B. Kahng  
Lei He  
Igor Markov  
Mani Srivastava  
Mohammad Tehranipoor

# Introduction

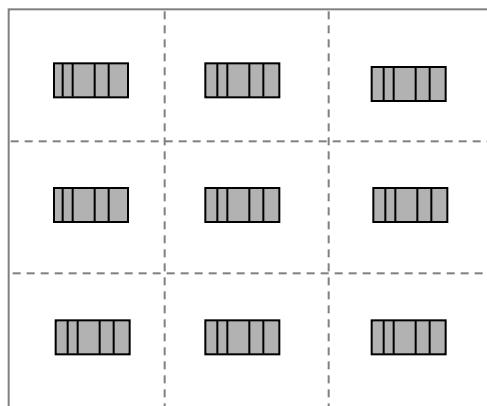


# Placement

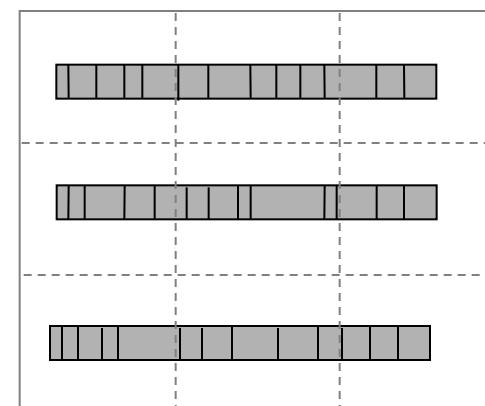


# Introduction

Global  
Placement

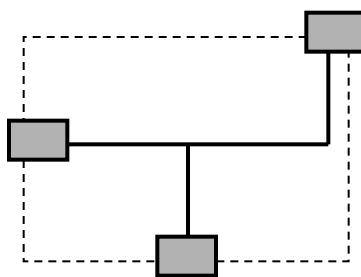


Detailed  
Placement

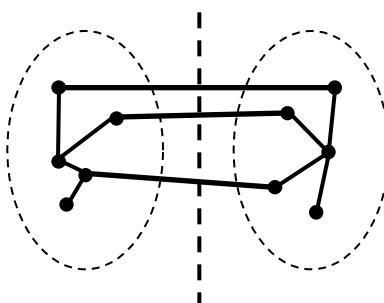


# Optimization Objectives

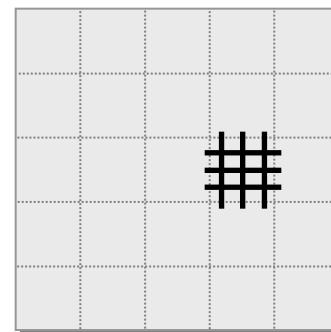
Total  
Wirelength



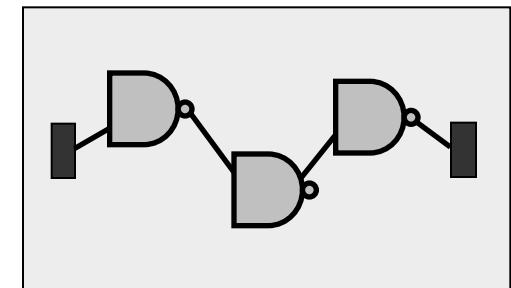
Number of  
Cut Nets



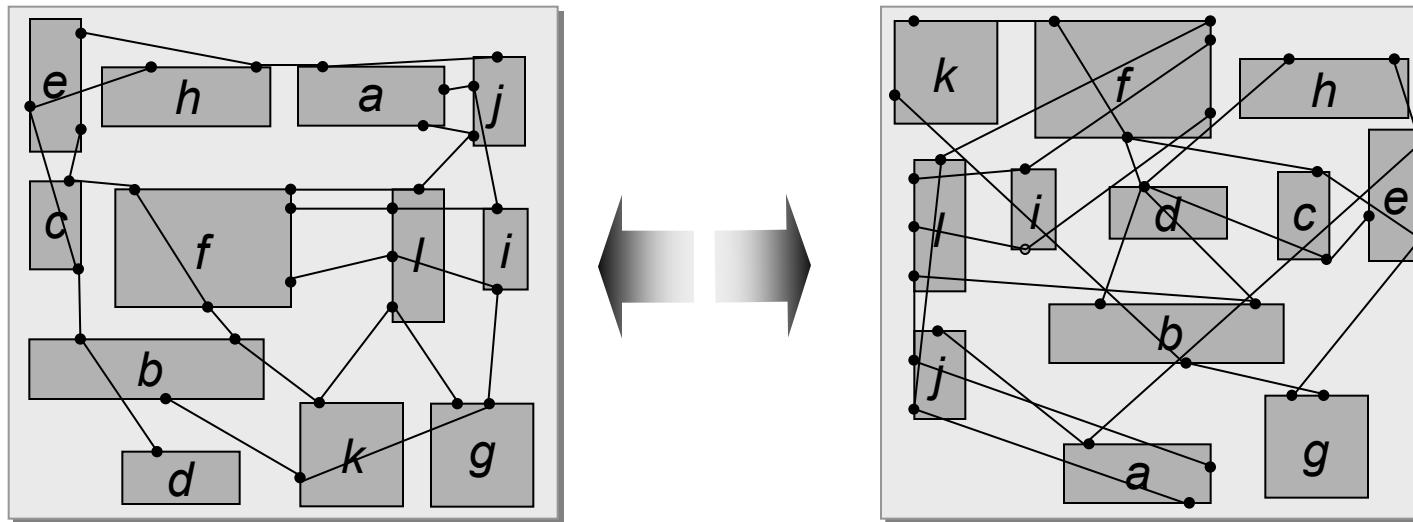
Wire  
Congestion



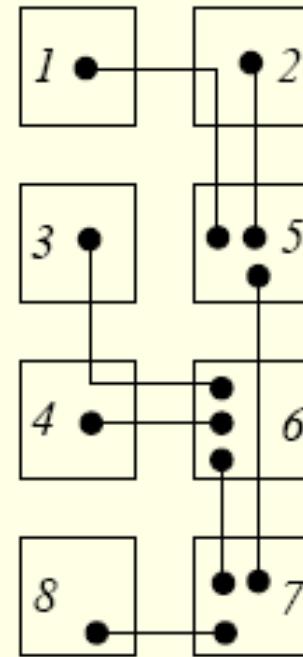
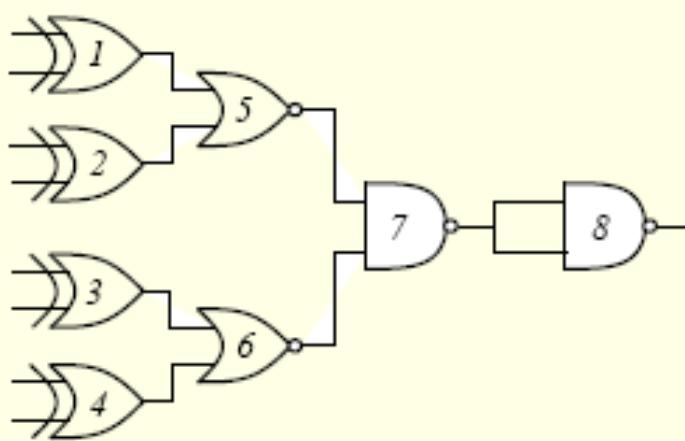
Signal  
Delay



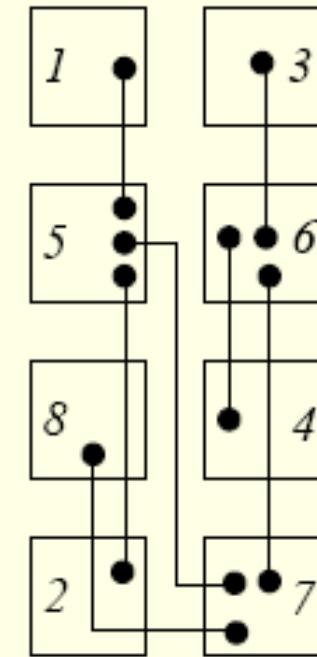
# Optimization Objectives – Total Wirelength



# Wirelength = Manhattan ( $L_1$ ) Metric



*wirelength = 10*



*wirelength = 12*

$L_p$  norm:  $((\Delta x)^p + (\Delta y)^p)^{1/p}$

$p = 1$  (Manhattan);  $p = 2$  (Euclidean);  
 $p = \infty$  (max, Chebyshev)

# How to measure WL ?

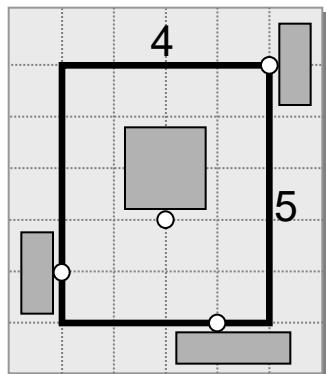
Wirelength estimation for a given placement

Half-perimeter  
wirelength  
(HPWL)

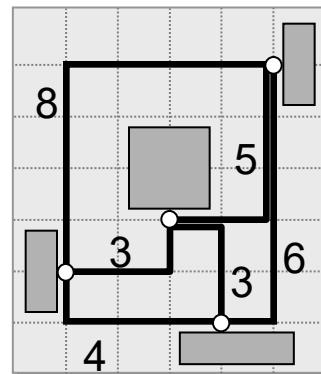
Complete  
graph  
(clique)

Monotone  
chain

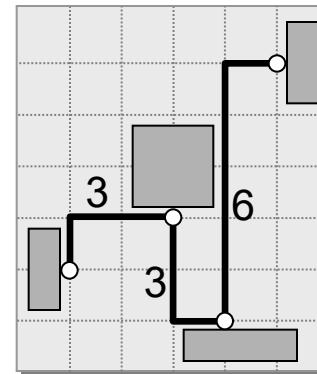
Star model



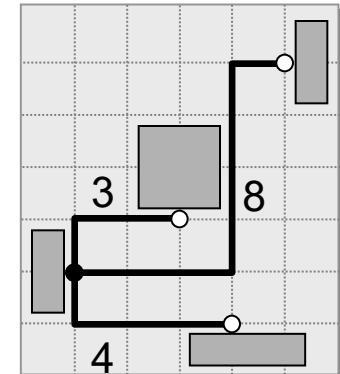
HPWL = 9  
Exact for #pins = 2, 3



Clique Length =  
 $(2/p)\sum_{e \in \text{clique}} d_M(e) = 14.5$



Chain Length = 12

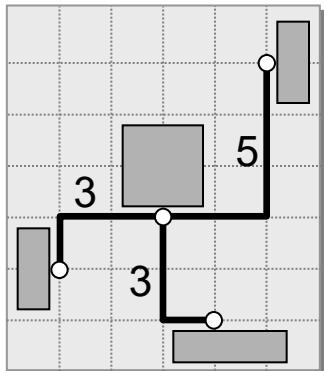


Star Length = 15

# How to measure WL ?

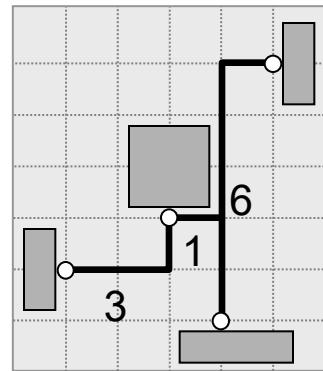
Wirelength estimation for a given placement (cont'd.)

Rectilinear  
minimum  
spanning  
tree (RMST)



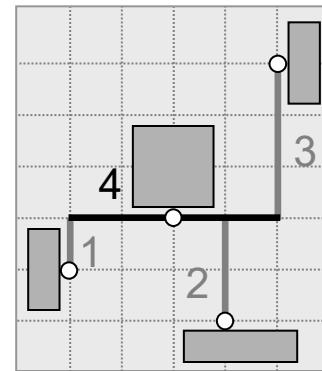
RMST Length = 11  
 $O(n \log n)$

Rectilinear  
Steiner  
minimum  
tree (RSMT)



RSMT Length = 10  
NP-hard optimization  
True WL

Single-trunk  
Steiner  
tree (STST)



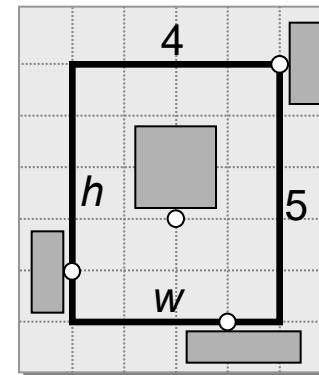
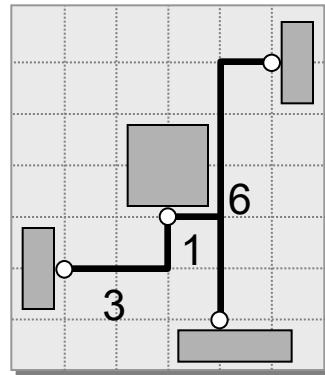
STST Length = 10

# Optimization Objectives – Total Wirelength

## Wirelength estimation for a given placement (cont'd.)

Preferred method: Half-perimeter wirelength (HPWL)

- Fast (order of magnitude faster than RSMT)
- Equal to length of RSMT for 2- and 3-pin nets
- Margin of error for real circuits approx. 8%



$$L_{\text{HPWL}} = w + h$$

# Optimization Objectives – Total Wirelength

Total wirelength with net weights (weighted wirelength)

- For a placement  $P$ , an estimate of total weighted wirelength is

$$L(P) = \sum_{net \in P} w(net) \cdot L(net)$$

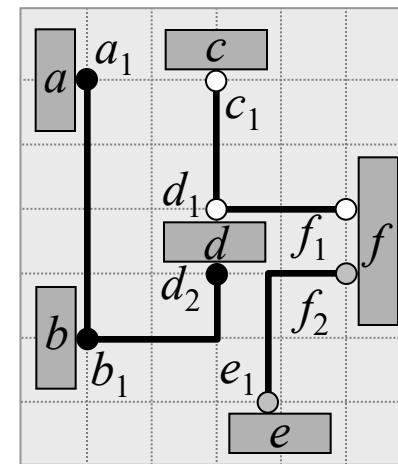
where  $w(net)$  is the weight of  $net$ , and  $L(net)$  is the estimated wirelength of  $net$ .

Weight could be timing criticality

- Example:

Nets	Weights
$N_1 = (a_1, b_1, d_2)$	$w(N_1) = 2$
$N_2 = (c_1, d_1, f_1)$	$w(N_2) = 4$
$N_3 = (e_1, f_2)$	$w(N_3) = 1$

$$L(P) = \sum_{net \in P} w(net) \cdot L(net) = 2 \cdot 7 + 4 \cdot 4 + 1 \cdot 3 = 33$$



# Optimization Objectives – Number of Cut Nets



Cut sizes of a placement

- To improve total wirelength of a placement  $P$ , separately calculate the number of crossings of global vertical and horizontal cutlines, and minimize

$$L(P) = \sum_{v \in V_P} \psi_P(v) + \sum_{h \in H_P} \psi_P(h)$$

where  $\Psi_P(cut)$  be the set of nets cut by a cutline  $cut$

- Net-cut cost = number of external nets between different global bins
  - Asymptotically, net-cut cost = wire length
  - Minimizing net-cut in global placement tends to put highly connected cells close to each other.

# Optimization Objectives – Number of Cut Nets

Cut sizes of a placement

- Example:  
Nets  
 $N_1 = (a_1, b_1, d_2)$   
 $N_2 = (c_1, d_1, f_1)$   
 $N_3 = (e_1, f_2)$
- Cut values for each global cutline

$$\psi_P(v_1) = 1 \quad \psi_P(v_2) = 2$$

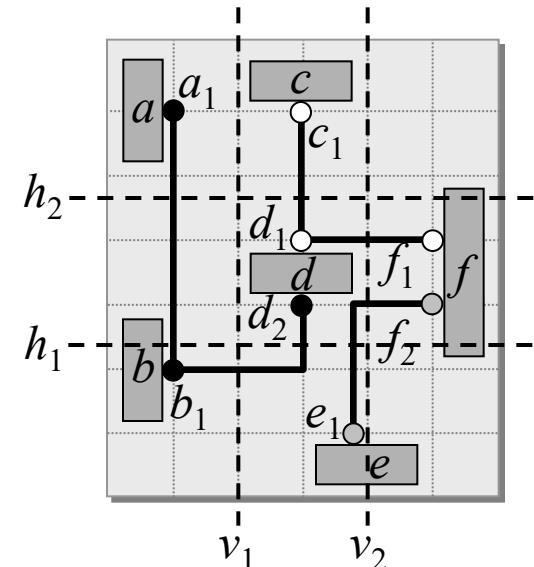
$$\psi_P(h_1) = 3 \quad \psi_P(h_2) = 2$$

- Total number of crossings in  $P$
- $$\psi_P(v_1) + \psi_P(v_2) + \psi_P(h_1) + \psi_P(h_2) = 1 + 2 + 3 + 2 = 8$$

- Cut sizes

$$X(P) = \max(\psi_P(v_1), \psi_P(v_2)) = \max(1, 2) = 2$$

$$Y(P) = \max(\psi_P(h_1), \psi_P(h_2)) = \max(3, 2) = 3$$



# Optimization Objectives – Wire Congestion



Routing congestion of a placement

- Formally, the local wire density  $\varphi_P(e)$  of an edge  $e$  between two neighboring grid cells is

$$\varphi_P(e) = \frac{\eta_P(e)}{\sigma_P(e)}$$

where  $\eta_P(e)$  is the estimated number of nets that cross  $e$  and  
 $\sigma_P(e)$  is the maximum number of nets that can cross  $e$

- If  $\varphi_P(e) > 1$ , then too many nets are estimated to cross  $e$ , making  $P$  more likely to be unroutable.
- The wire density of  $P$  is  $\Phi(P) = \max_{e \in E}(\varphi_P(e))$

where  $E$  is the set of all edges

- If  $\Phi(P) < 1$ , then the design is estimated to be fully routable, otherwise routing will need to detour some nets through less-congested edges

# Optimization Objectives – Wire Congestion

Wire Density of a placement

$$\eta_P(h_1) = 1$$

$$\eta_P(h_2) = 2$$

$$\eta_P(h_3) = 0$$

$$\eta_P(h_4) = 1$$

$$\eta_P(h_5) = 1$$

$$\eta_P(h_6) = 0$$

$$\eta_P(v_1) = 1$$

$$\eta_P(v_2) = 0$$

$$\eta_P(v_3) = 0$$

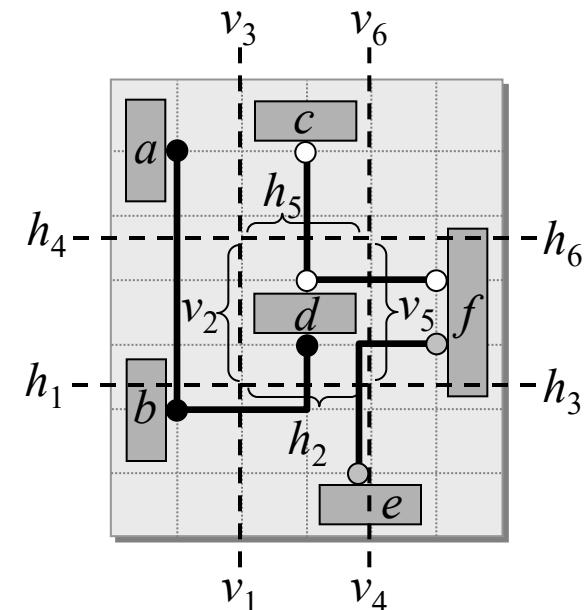
$$\eta_P(v_4) = 0$$

$$\eta_P(v_5) = 2$$

$$\eta_P(v_6) = 0$$

Maximum:

$$\eta_P(e) = 2$$



$$\Phi(P) = \frac{\eta_P(e)}{\sigma_P(e)} = \frac{2}{3}$$

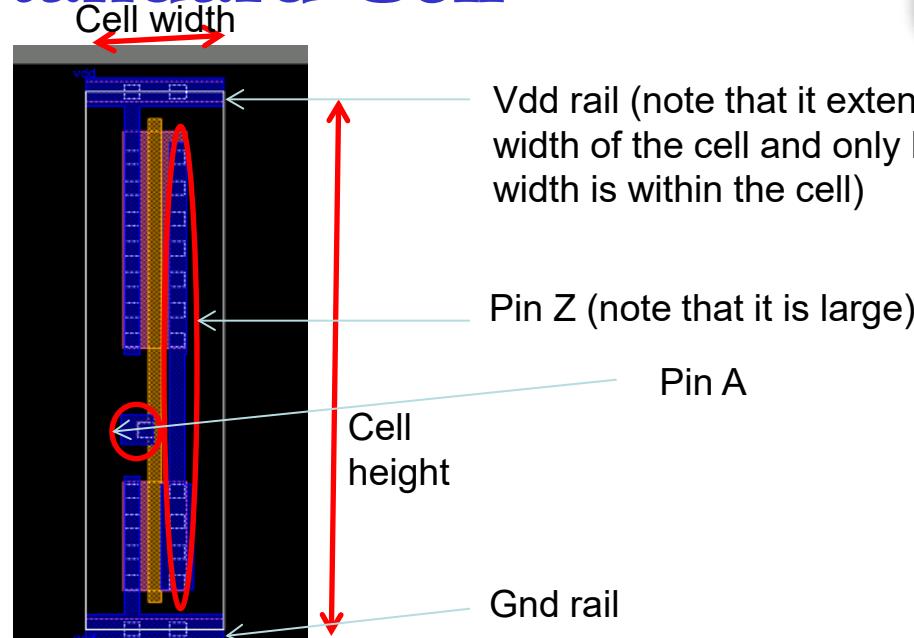


Routable

# The Standard Cell



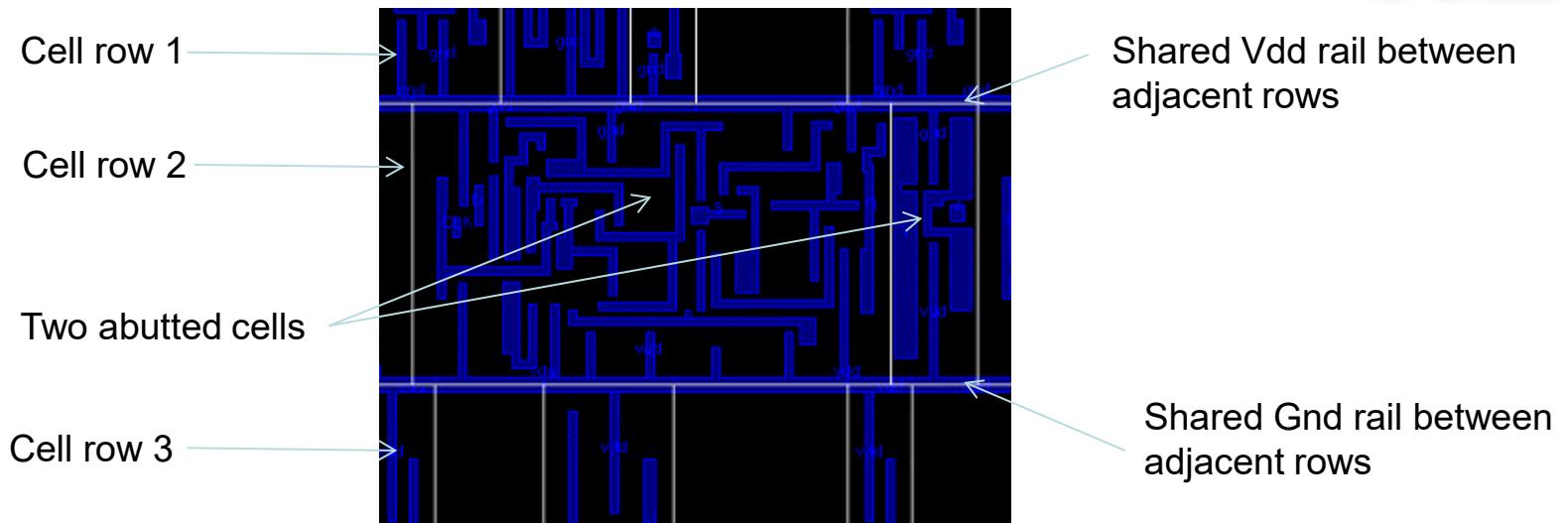
Cell layouts are front-end only:  
typically use only up to M1  
(sometimes M2)



An INV Cell

- Almost all modern digital designs are cell-based.
  - Cells are typically small gates (NAND, FF, etc) containing 2-100 transistors.
- There are  $\sim$ 1000 cells in a library
  - Typically one design uses one library only.
- Cell width and height are quantized (measured in #metal tracks)
  - Cell height usually fixed for a library (8-track, 10-track, etc)
  - Cell width a multiple of minimum sized inverter (again multiple of min. metal pitch)
  - Constraint ensures that when cells are placed abutted, all pins fall on metal track intersections  
→ easier for router to connect to them.

# Cell-Based Placement

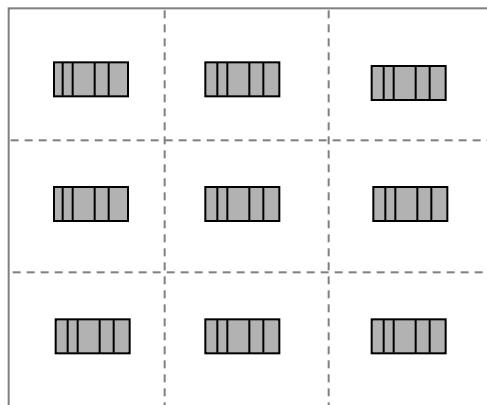


- Cells are placed in rows
- Almost all logic is automatically synthesized, placed using cell libraries
- Modern designs easily have 10M+ cells
- Goal of placement: timing correct, routable placement of all cells

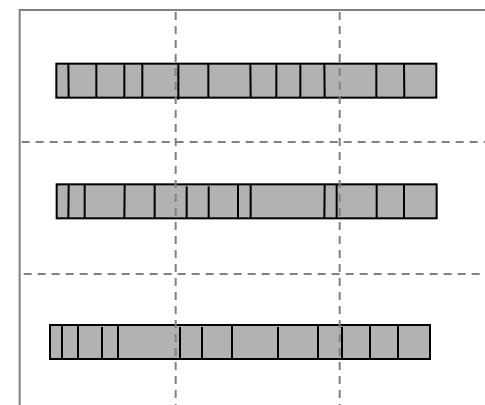
# 5 min break

# Introduction

Global  
Placement



Detailed  
Placement



# Global Placement

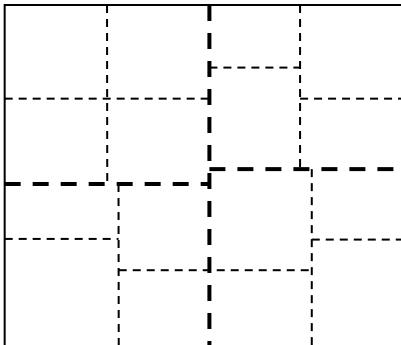
- Partitioning-based algorithms:
  - The netlist and the layout are divided into smaller sub-netlists and sub-regions, respectively
  - Process is repeated until each sub-netlist and sub-region is small enough to be handled optimally
  - Detailed placement often performed by optimal solvers, facilitating a natural transition from global placement to detailed placement
  - Example: min-cut placement
- Analytic techniques:
  - Model the placement problem using an objective (cost) function, which can be optimized via numerical analysis
  - Examples: quadratic placement and force-directed placement
- Stochastic algorithms:
  - Randomized moves that allow hill-climbing are used to optimize the cost function
  - Example: simulated annealing

# Global Placement

Partitioning-based



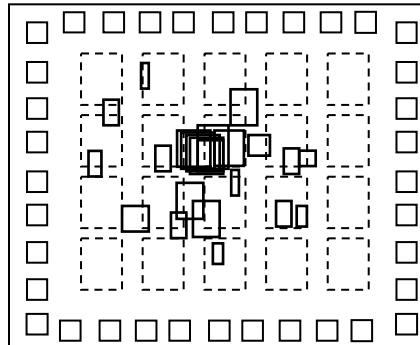
Min-cut  
placement



Analytic



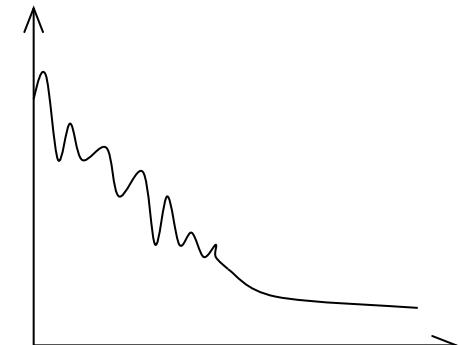
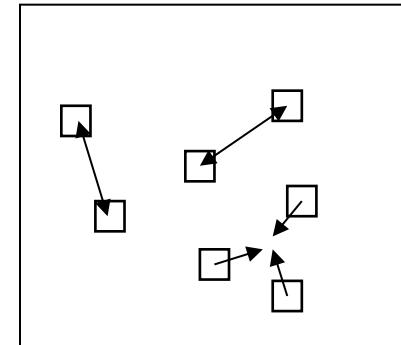
Quadratic  
placement



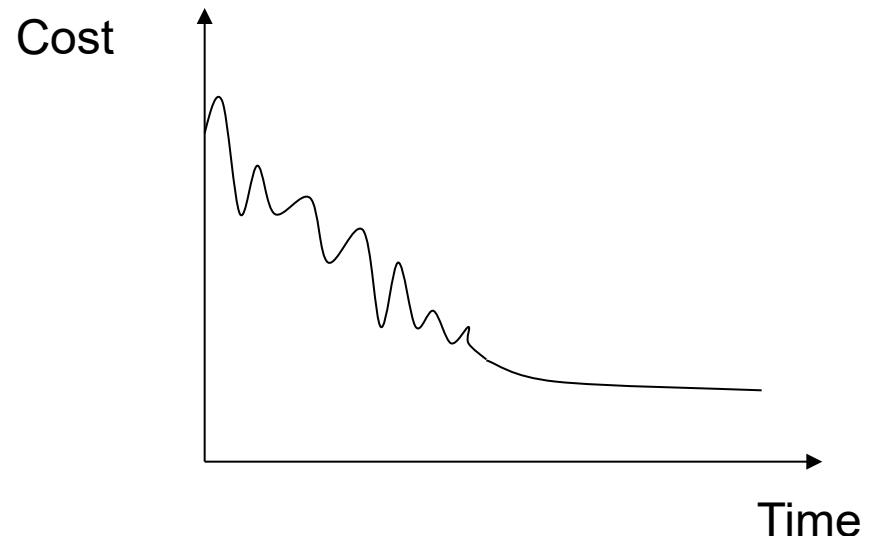
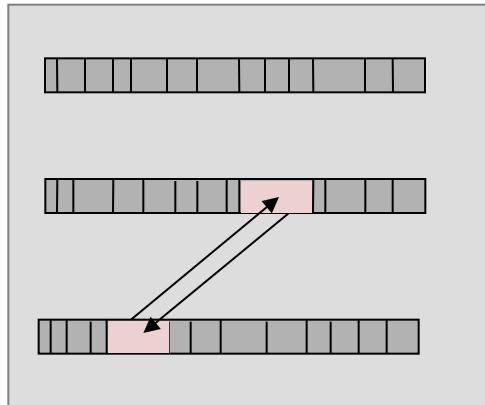
Stochastic



Simulated  
annealing



# Simulated Annealing



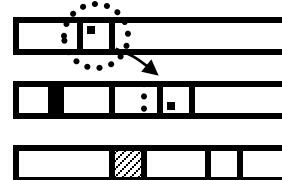
- Analogous to the physical **annealing process**
  - Melt metal and then slowly cool it
  - Result: energy-minimal crystal structure
- Modification of an initial configuration (placement) by moving/exchanging of randomly selected cells
  - Accept the new placement if it improves the objective function
  - If no improvement: Move/exchange is accepted with temperature-dependent (i.e., decreasing) probability

# Neighborhood Structure for SA

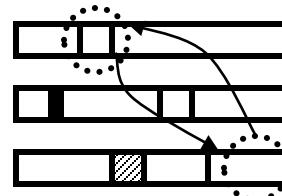


**Solution space:** All arrangements of cells into rows, possibly with overlaps

Three types of moves:

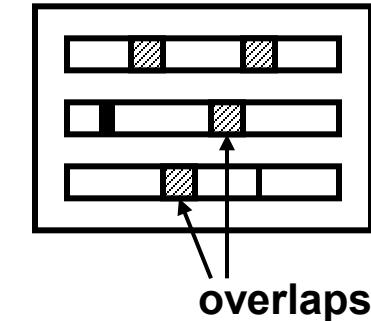
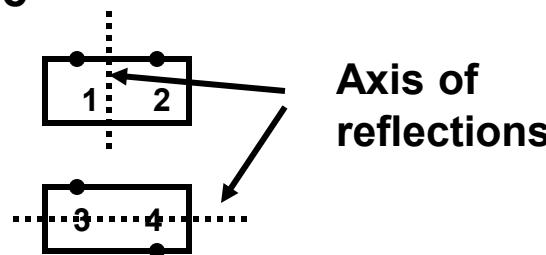
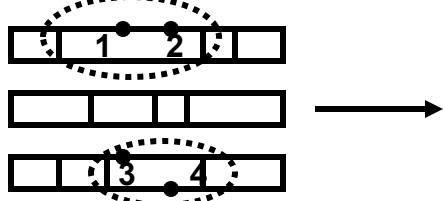


**M1:** Displace a module to a new location



**M2:** Interchange two modules

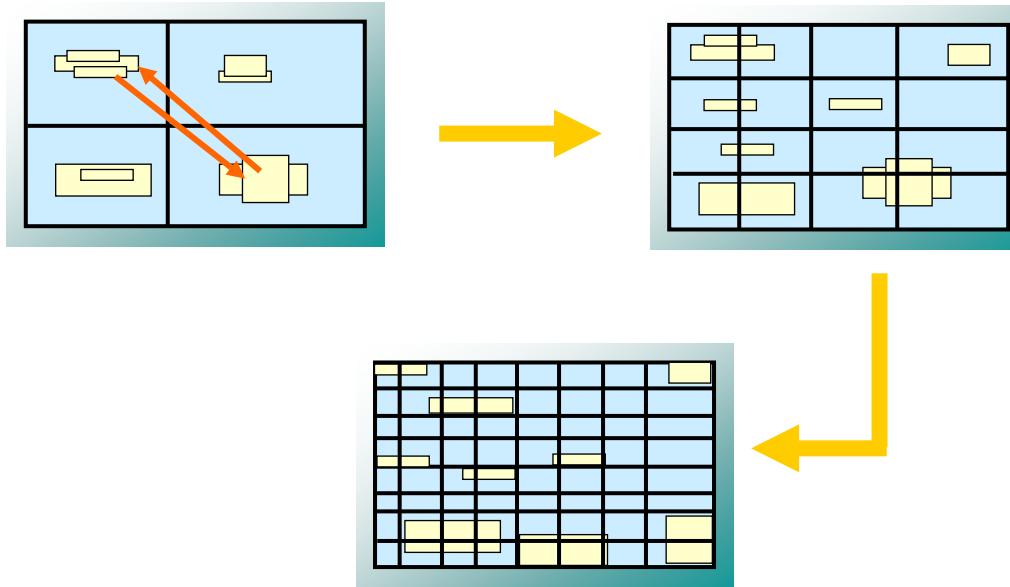
**M3:** Change orientation of a module



# Simulated Annealing

- Advantages:
  - Can find global optimum (given sufficient time)
  - Well-suited for detailed placement
- Disadvantages:
  - Very slow
  - To achieve high-quality implementation, laborious parameter tuning is necessary
  - Randomized, chaotic algorithms - small changes in the input lead to large changes in the output
- Practical applications of SA:
  - Very small placement instances with complicated constraints
  - Detailed placement, where SA can be applied in small windows (not common anymore)
  - FPGA layout, where complicated constraints are becoming a norm

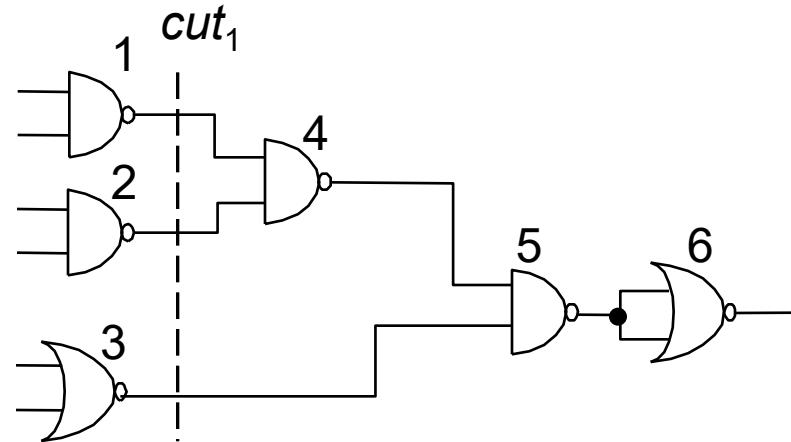
# Min-Cut Placement



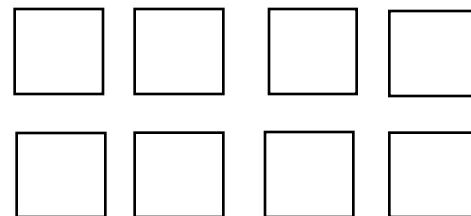
- Uses partitioning algorithms to divide (1) the netlist and (2) the layout region into smaller sub-netlists and sub-regions
- Conceptually, each sub-region is assigned a portion of the original netlist
- Each cut heuristically minimizes the number of cut nets using,
  - Fiduccia-Mattheyses (FM) algorithm

# Min-Cut Placement – Example

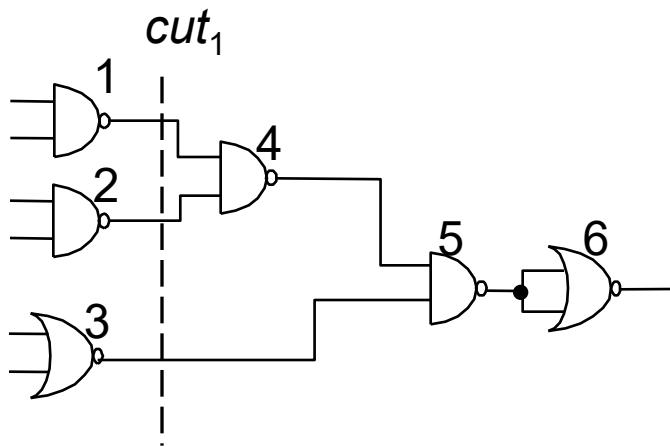
Given:



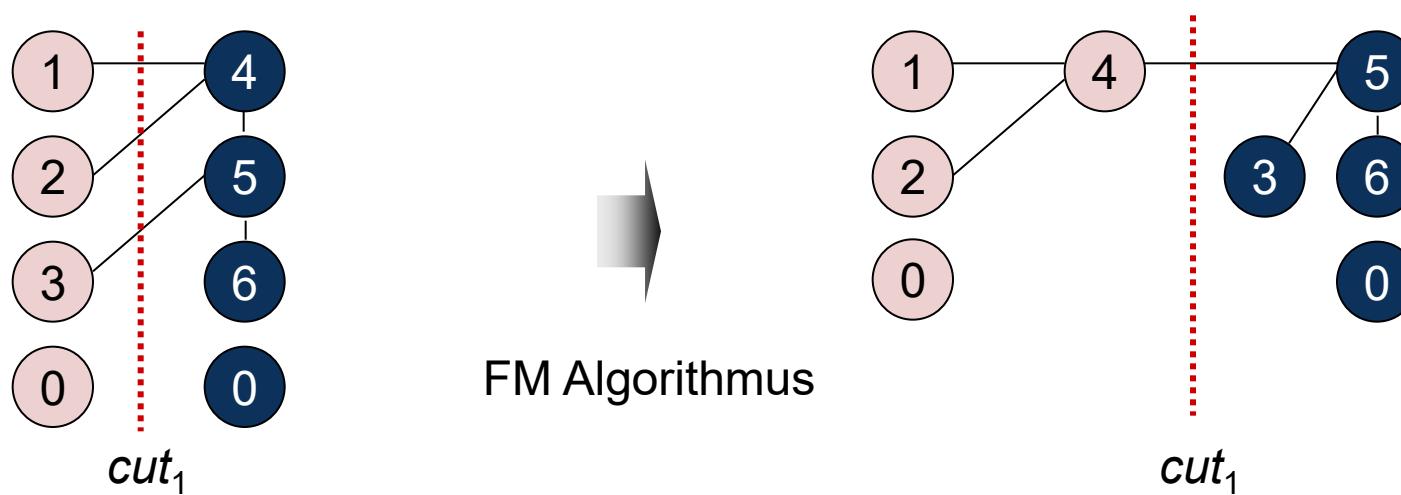
Task:  $4 \times 2$  placement with minimum wirelength using alternative cutline directions and the FM algorithm

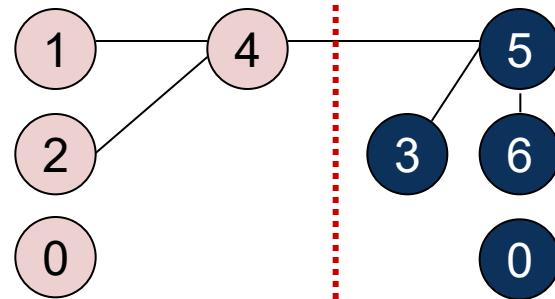


# Min-Cut-Placement



Vertical cut  $cut_1$ :  $L=\{1,2,3\}$ ,  $R=\{4,5,6\}$

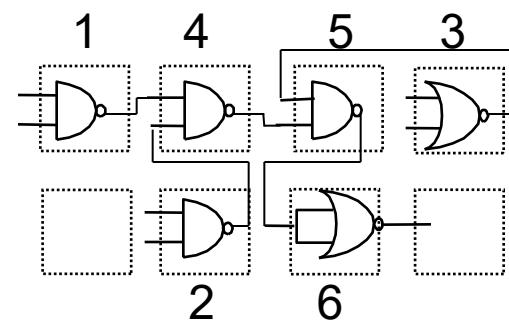
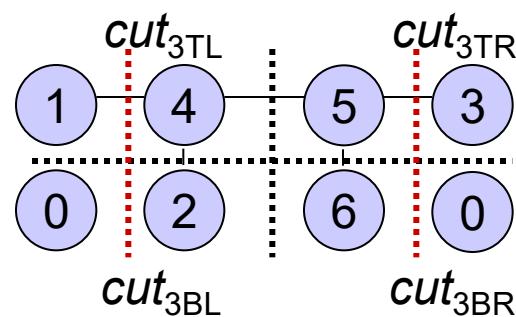
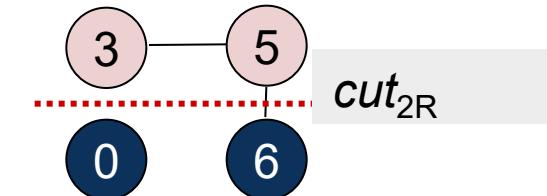
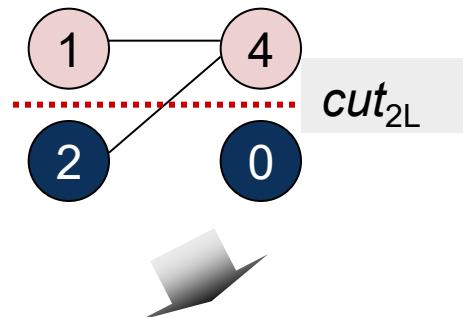




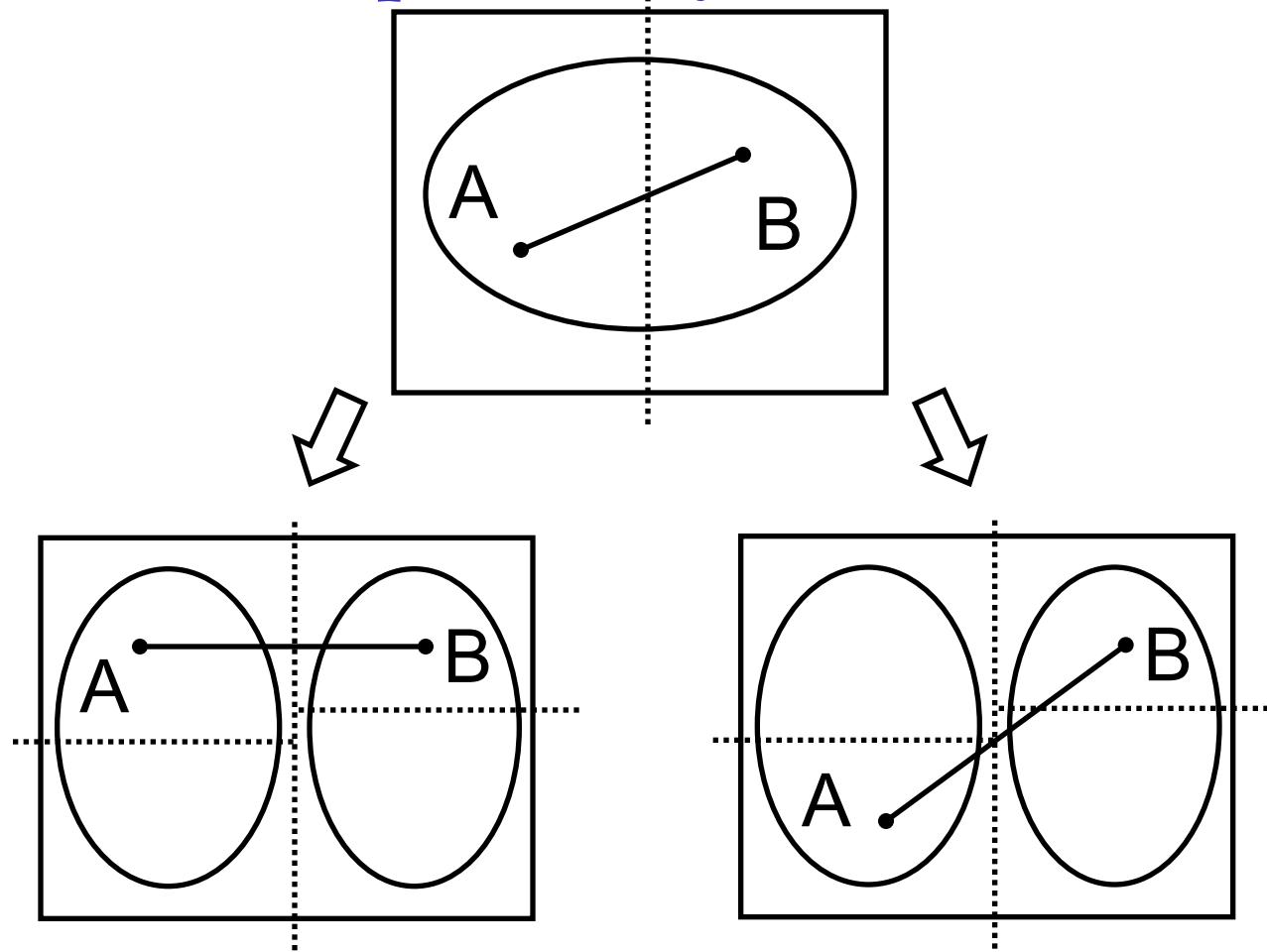
Horizontal cut  $cut_{2L}$ :  $T=\{1,4\}$ ,  $B=\{2,0\}$

$cut_1$

Horizontal cut  $cut_{2R}$ :  $T=\{3,5\}$ ,  $B=\{6,0\}$

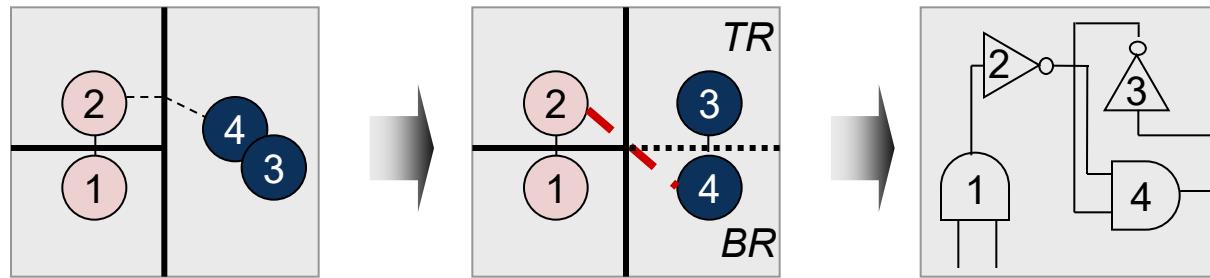


# How to Partition Subcircuits “Independently”?

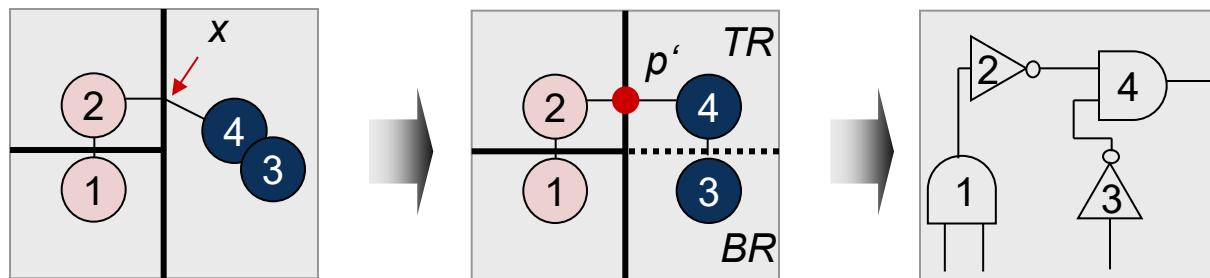


The costs of these two solutions are not the same

# Min-Cut Placement – Terminal Propagation



- Terminal Propagation
  - External connections are represented by artificial connection points on the cutline
  - Dummy nodes in hypergraphs



# Min-Cut Placement Summary

- Advantages:
  - Reasonable fast
  - Objective function can be adjusted, e.g., to perform timing-driven placement
  - Hierarchical strategy applicable to large circuits
- Disadvantages:
  - Randomized, chaotic algorithms – small changes in input lead to large changes in output
  - Optimizing one cutline at a time may result in routing congestion elsewhere

# Analytic Placement – Quadratic Placement



- Objective function is quadratic; sum of (weighted) squared Euclidean distance represents placement objective function

$$L(P) = \frac{1}{2} \sum_{i,j=1}^n c_{ij} \left( (x_i - x_j)^2 + (y_i - y_j)^2 \right)$$

where  $n$  is the total number of cells, and  $c_{ij}$  is the connection cost between cells  $i$  and  $j$ .

- Only two-point-connections
- Minimize objective function by equating its derivative to zero which reduces to solving a system of linear equations

# Analytic Placement – Quadratic Placement



$$L(P) = \frac{1}{2} \sum_{i,j=1}^n c_{ij} \left( (x_i - x_j)^2 + (y_i - y_j)^2 \right)$$

where  $n$  is the total number of cells, and  $c(i,j)$  is the connection cost between cells  $i$  and  $j$ .

- Each dimension can be considered independently:

$$L_x(P) = \sum_{i=1, j=1}^n c(i, j)(x_i - x_j)^2 \quad L_y(P) = \sum_{i=1, j=1}^n c(i, j)(y_i - y_j)^2$$

- Convex quadratic optimization problem: any local minimum solution is also a global minimum
- Optimal  $x$ - and  $y$ -coordinates can be found by setting the partial derivatives of  $L_x(P)$  and  $L_y(P)$  to zero

# Analytic Placement – Quadratic Placement



$$L(P) = \frac{1}{2} \sum_{i,j=1}^n c_{ij} \left( (x_i - x_j)^2 + (y_i - y_j)^2 \right)$$

where  $n$  is the total number of cells, and  $c(i,j)$  is the connection cost between cells  $i$  and  $j$ .

$$L_x(P) = \sum_{i=1, j=1}^n c(i, j)(x_i - x_j)^2 \quad L_y(P) = \sum_{i=1, j=1}^n c(i, j)(y_i - y_j)^2$$



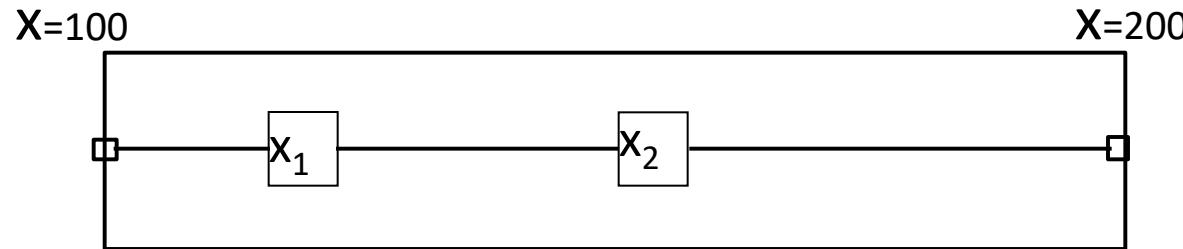
$$\frac{\partial L_x(P)}{\partial X} = AX - b_x = 0$$



$$\frac{\partial L_y(P)}{\partial Y} = AY - b_y = 0$$

- Each dimension can be considered independently

# Analytical Placement Example



$$Cost = (x_1 - 100)^2 + (x_1 - x_2)^2 + (x_2 - 200)^2$$

$$\frac{\partial}{\partial x_1} Cost = 2(x_1 - 100) + 2(x_1 - x_2)$$

$$\frac{\partial}{\partial x_2} Cost = -2(x_1 - x_2) + 2(x_2 - 200)$$

Setting the partial derivatives = 0, we solve for the minimum Cost :

$$Ax = b$$

$$\begin{bmatrix} 4 & -2 \\ -2 & 4 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} 200 \\ 400 \end{bmatrix}$$

$$\begin{bmatrix} 2 & -1 \\ -1 & 2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} 100 \\ 200 \end{bmatrix}$$

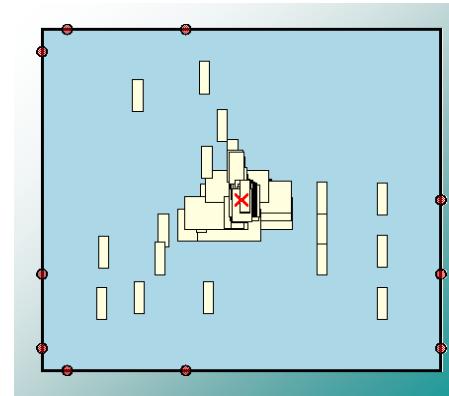
$$x_1 = \frac{400}{3} \quad x_2 = \frac{500}{3}$$

**A<sub>ii</sub> = degree of a node**  
**A<sub>ij</sub> = -(i-j) connectivity**  
**b<sub>i</sub> = sum of locations connected to cell i**

# Why “Squared Wirelength” ?

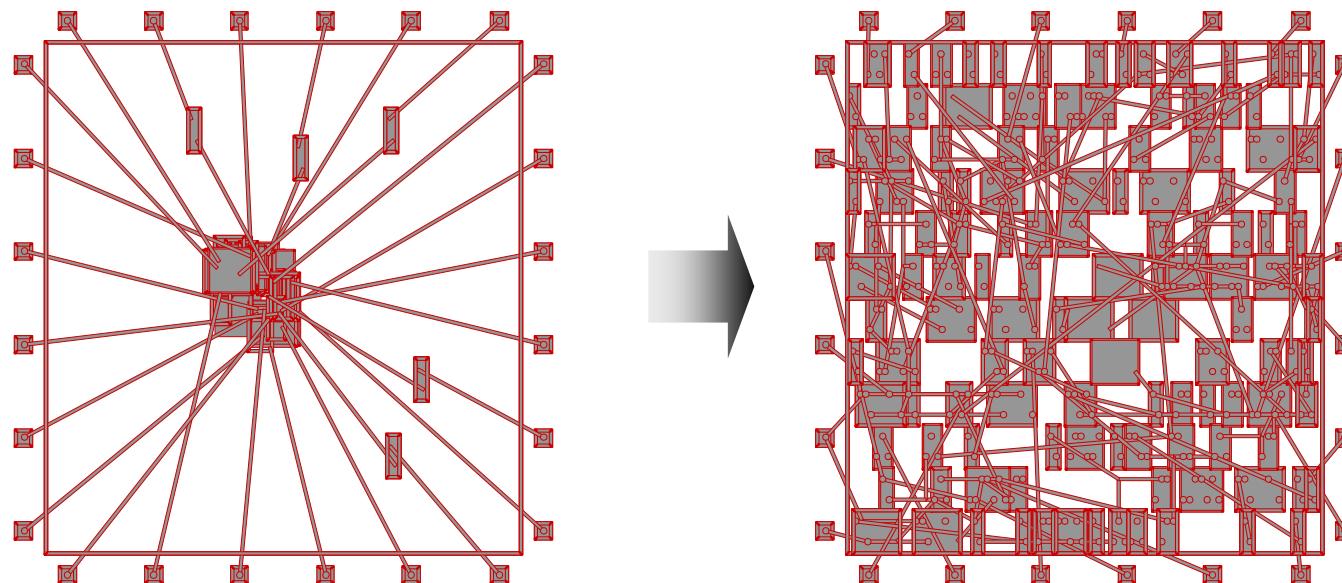


- Because we can
- Because it is trivial to solve
- Because there is only one solution
- Because the solution is a global optimum
- Because the solution conveys “relative order” information
- (Because the solution conveys “global position” information)
- **Key issue: “spreading”**
  - What is the optimal Solution in previous case if No pin locations are there ?



# Analytic Placement – Quadratic Placement

- Second stage of quadratic placers: cells are spread out to remove overlaps
- Methods:
  - Adding fake nets that pull cells away from dense regions toward anchors
  - Geometric sorting and scaling
  - Repulsion forces, etc.



# Analytic Placement – Quadratic Placement



- Advantages:
  - Captures the placement problem concisely in mathematical terms
  - Leverages efficient algorithms from numerical analysis and available software
  - Can be applied to large circuits without netlist clustering (flat)
  - Stability: small changes in the input do not lead to large changes in the output
- Disadvantages:
  - Connections to fixed objects are necessary: I/O pads, pins of fixed macros, etc.