# Other Bias Methods

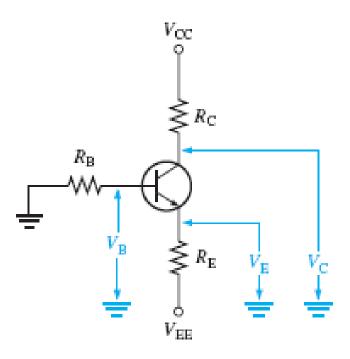
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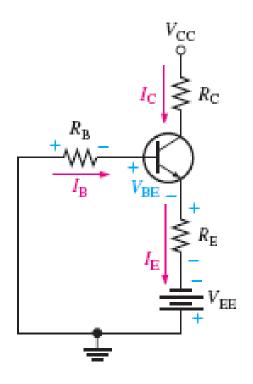
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## **Other Bias Methods**

#### **Emitter Bias**





Emitter bias provides excellent bias stability in spite of changes in  $\beta$  or temperatures.

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$$V_{\rm EE} + V_{R_{\rm B}} + V_{\rm BE} + V_{R_{\rm E}} = 0$$

Substituting, using Ohm's law,

$$V_{\rm EE} + I_{\rm B}R_{\rm B} + V_{\rm BE} + I_{\rm E}R_{\rm E} = 0$$

Substituting for  $I_B \cong I_E/\beta_{DC}$  and transposing  $V_{EE}$ ,

$$\left(\frac{I_{\rm E}}{\beta_{\rm DC}}\right)R_{\rm B} + I_{\rm E}R_{\rm E} + V_{\rm BE} = -V_{\rm EE}$$

Factoring out  $I_E$  and solving for  $I_E$ ,

$$I_{\rm E} = \frac{-V_{\rm EE} - V_{\rm BE}}{R_{\rm E} + R_{\rm B}/\beta_{\rm DC}}$$

Voltages with respect to ground are indicated by a single subscript. The emitter voltage with respect to ground is

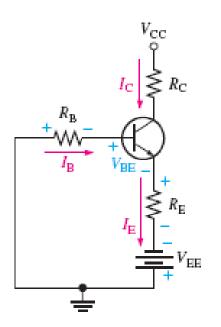
$$V_{\rm E} = V_{\rm EE} + I_{\rm E} R_{\rm E}$$

The base voltage with respect to ground is

$$V_{\rm B} = V_{\rm E} + V_{\rm BE}$$

The collector voltage with respect to ground is

$$V_{\rm C} = V_{\rm CC} - I_{\rm C}R_{\rm C}$$



### **Base Bias**

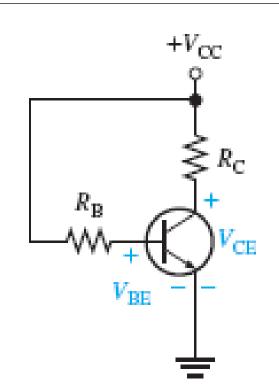
$$V_{\rm CC}-V_{R_{\rm B}}-V_{\rm BE}=0$$

Substituting  $I_BR_B$  for  $V_{R_B}$ , you get

$$V_{\rm CC} - I_{\rm B}R_{\rm B} - V_{\rm BE} = 0$$

Then solving for  $I_B$ ,

$$I_{\rm B} = \frac{V_{\rm CC} - V_{\rm BE}}{R_{\rm B}}$$



$$V_{\rm CC} - I_{\rm C}R_{\rm C} - V_{\rm CE} = 0$$

Solving for  $V_{CE}$ ,

$$V_{\rm CE} = V_{\rm CC} - I_{\rm C}R_{\rm C}$$

Substituting the expression for  $I_B$  into the formula  $I_C = \beta_{DC}I_B$  yields

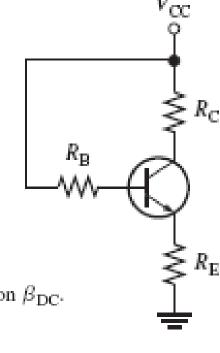
$$I_{\rm C} = \beta_{\rm DC} \left( \frac{V_{\rm CC} - V_{\rm BE}}{R_{\rm B}} \right)$$

Q-Point Stability of Base Bias Notice that Equation 5–11 shows that  $I_C$  is dependent on  $\beta_{DC}$ . The disadvantage of this is that a variation in  $\beta_{DC}$  causes  $I_C$  and, as a result,  $V_{CE}$  to change, thus changing the Q-point of the transistor. This makes the base bias circuit extremely beta-dependent and unpredictable.

Recall that  $\beta_{DC}$  varies with temperature and collector current. In addition, there is a large spread of  $\beta_{DC}$  values from one transistor to another of the same type due to manufacturing variations. For these reasons, base bias is rarely used in linear circuits but is discussed here so you will be familiar with it.

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## **Emitter Feedback Bias**



$$-V_{\rm CC} + I_{\rm B}R_{\rm B} + V_{\rm BE} + I_{\rm E}R_{\rm E} = 0$$

Substituting  $I_{\rm E}/\beta_{\rm DC}$  for  $I_{\rm B}$ , you can see that  $I_{\rm E}$  is still dependent on  $\beta_{\rm DC}$ .

$$I_{\rm E} = \frac{V_{\rm CC} - V_{\rm BE}}{R_{\rm E} + R_{\rm B}/\beta_{\rm DC}}$$

#### **Collector Feedback Bias**

Analysis of a Collector-Feedback Bias Circuit By Ohm's law, the base current can be expressed as

$$I_{\rm B} = \frac{V_{\rm C} - V_{\rm BE}}{R_{\rm B}}$$

Let's assume that  $I_C \gg I_B$ . The collector voltage is

$$V_C \cong V_{CC} - I_C R_C$$

Also,

$$I_{\rm B} = \frac{I_{\rm C}}{\beta_{\rm DC}}$$

Substituting for  $V_C$  in the equation  $I_B = (V_C - V_{BE})/R_B$ ,

$$\frac{I_{\rm C}}{\beta_{\rm DC}} = \frac{V_{\rm CC} - I_{\rm C}R_{\rm C} - V_{\rm BE}}{R_{\rm B}}$$

The terms can be arranged so that

$$\frac{I_{\rm C}R_{\rm B}}{\beta_{\rm DC}} + I_{\rm C}R_{\rm C} = V_{\rm CC} - V_{\rm BE}$$

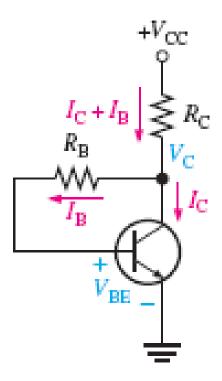
Then you can solve for  $I_C$  as follows:

$$I_{\rm C}\left(R_{\rm C} + \frac{R_B}{\beta_{\rm DC}}\right) = V_{\rm CC} - V_{\rm BE}$$

$$I_{\rm C} = \frac{V_{\rm CC} - V_{\rm BE}}{R_{\rm C} + R_{\rm B}/\beta_{\rm DC}}$$

Since the emitter is ground,  $V_{CE} = V_{C}$ .

$$V_{\rm CE} = V_{\rm CC} - I_{\rm C} R_{\rm C}$$



Q-Point Stability Over Temperature Equation 5–13 shows that the collector current is dependent to some extent on  $\beta_{DC}$  and  $V_{BE}$ . This dependency, of course, can be minimized by making  $R_C \gg R_B/\beta_{DC}$  and  $V_{CC} \gg V_{BE}$ . An important feature of collector-feedback bias is that it essentially eliminates the  $\beta_{DC}$  and  $V_{BE}$  dependency even if the stated conditions are met.

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