# Control Systems Design Lab 2

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Section: 8

### A. Evaluate Closed Loop Transfer Function.

$$\int \ddot{\theta} + \beta \dot{\theta} = |\chi(\theta y - \theta)|$$

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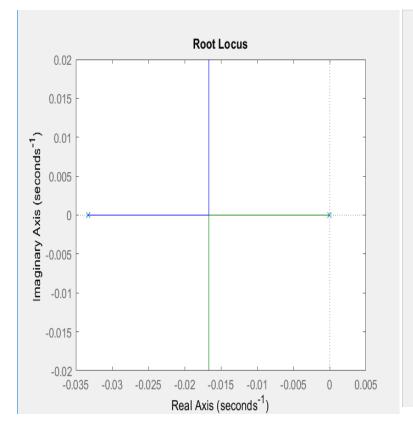
$$\int \partial \theta + \beta \dot{\theta} = |\chi(\theta y - \theta)|$$

$$\int \partial$$

## B. Use Matlab to generate the state space.

#### • Code:

Output of the matlab



# C. Maximum value of k that makes the Closed loop system stable :

• Using Routh Herwitz:

$$Js^2 + Bs + k = 0$$

$s^2$	J	k
s <sup>1</sup>	В	0
s <sup>0</sup>	k > 0	0

 So for any positive value of k the closed loop system will be positive.

#### D. Maximum value of k to have $M_P < 10\%$ :

K = 477

```
20000
      j = 600e3; B = 20e3; k = 1;
                                                 info
                                                            1x1 struct
     trans = tf(k, [j B k]);
10 -
                                                            600000
      info = stepinfo(ss(trans));
11 -
12 - ∃while info.Overshoot < 10
                                                            477
13 -
         k = k + 1;
                                                max_k_oversh... 477
14 -
        trans = tf(k, [j B k]);
                                                🗾 state
                                                            1x1 ss
         info = stepinfo(ss(trans));
15 -
                                                trans
                                                            1x1 tf
16 -
    ∟end
      max k overshot = k;
17 -
18
```

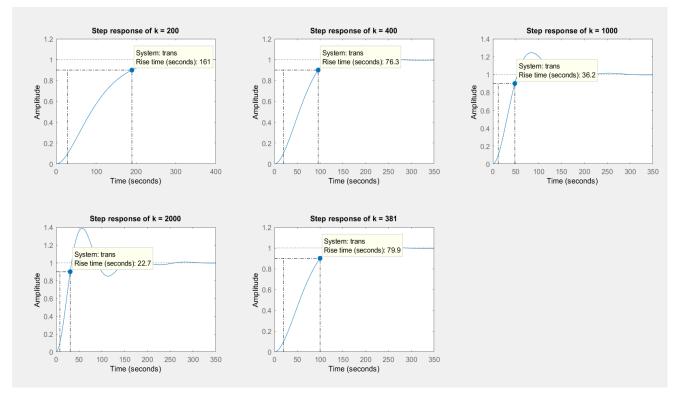
## E. Max value of k to have risetime less than 80 msec:

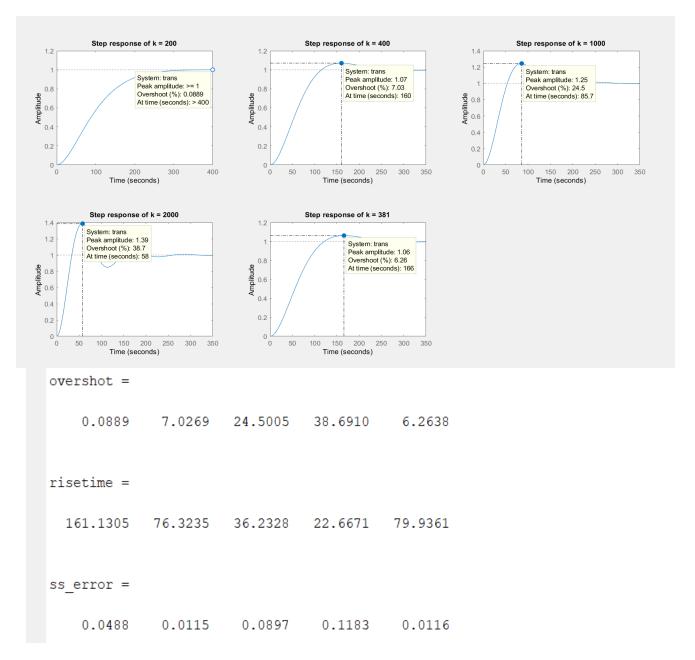
K = 381

```
%% *********** part e ************** Name ▲
19
                                                                        Value
20 -
       j = 600e3; B = 20e3; k = 1;
                                                         <del>Н</del> В
                                                                        20000
21 -
      trans = tf(k, [j B k]);
                                                          info
                                                                        1x1 struct
      info = stepinfo(ss(trans));
22 -
                                                                        600000
23 - ☐ while info.RiseTime > 80
                                                                        381
                                                         🕇 max_k_oversh... 477
24 -
           k = k + 1;
                                                           max_k_risetime 381
25 -
           trans = tf(k, [j B k]);
                                                         state
                                                                        1x1 ss
           info = stepinfo(ss(trans));
26 -
                                                         🗐 trans
                                                                        1x1 tf
27 -
      ∟end
28 -
       max k risetime = k
29
```

# F. Step response of k = 200, 400, 1000, 2000, 381 rise time and over shot:

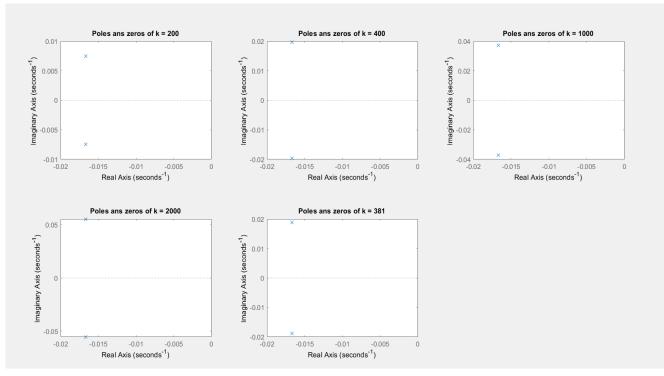
```
%% ************ part f, g and h *************
       j = 600e3; B = 20e3; k = [200, 400, 1000, 2000, 381];
31 -
32 -
       overshot = []; risetime = []; ss error = [];
33 -
     \Box for i=1:1:5
           trans = tf(k(i), [j B k(i)]);
34 -
35 -
           info = stepinfo(ss(trans));
           overshot(i) = info.Overshoot;
36 -
37 -
           risetime(i) = info.RiseTime;
           ss error(i) = abs(1-(info.SettlingMax + info.SettlingMin)/2);
38 -
39 -
           figure(1);
40 -
           subplot(2,3,i);
           stepplot(trans); title(sprintf('Step response of k = %d',k(i)));
41 -
42 -
           figure(2);
43 -
           subplot(2,3,i);
           pzplot(trans); title(sprintf('Poles ans zeros of k = d', k(i));
44 -
45 -
       end
```





As shown in the photo above the rise time and overshot output of the plots is the same as the code.

#### G. Poles and zeros



As show when the value of k increases the poles become farther. However ther is no zeros in the transfer fuction so the increasing of k won't affect zeros due to absence of s in the numerator.

Same code as part F

#### H. Steady state error

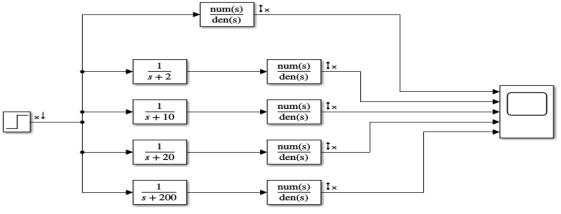
To get the steady-state error I used the following equation

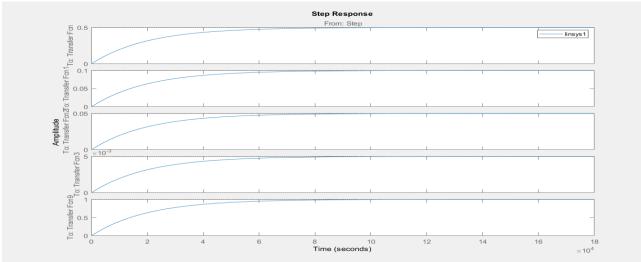
$$e_{SS} = \left| 1 - \frac{SettlingTime_{Min} + SettlingTime_{Max}}{2} \right|$$

$$= \frac{1}{2}$$

Same code as part F and G

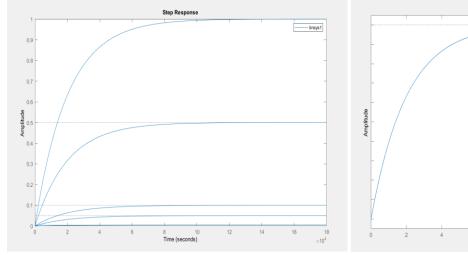
### I. Poles at -2, -10, -20, -200 at k=1

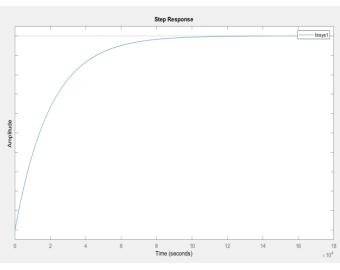


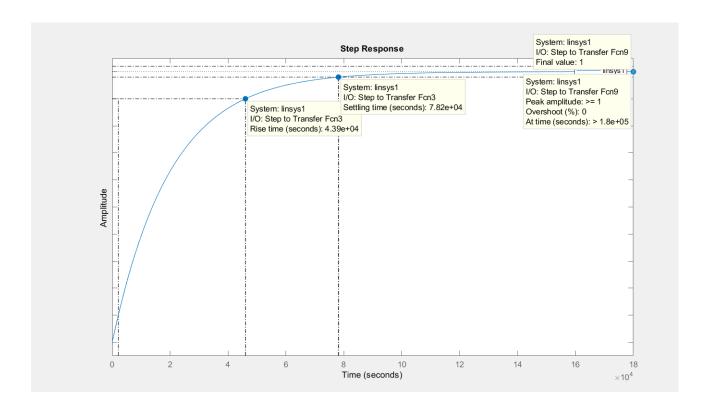


#### After groping all

#### normalized



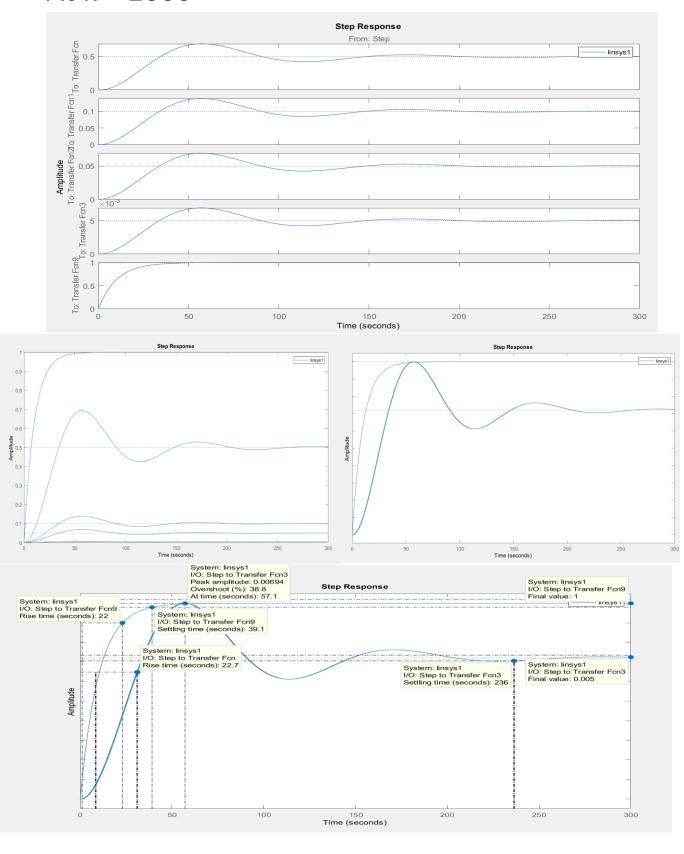




Rise time =  $4.39 \times 10^4$  seconds Settling time =  $7.82 \times 10^4$  seconds Peak time =  $1.8 \times 10^5$  seconds

Max overshot = 0%

#### At k = 2000



Rise time = 22.7 seconds

Settling time = 236 seconds

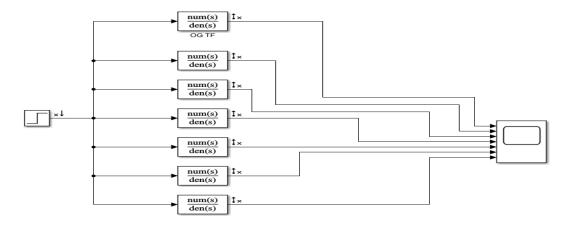
Peak time = 57.1 seconds

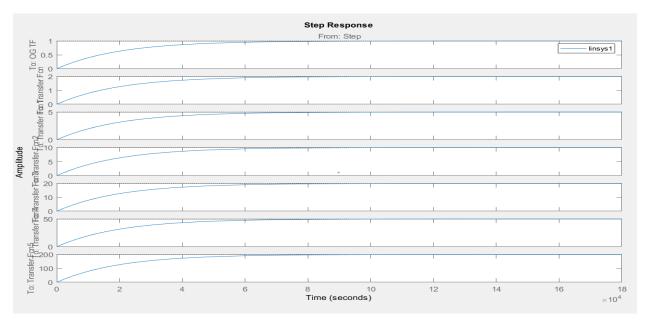
Max overshot = 38.8%

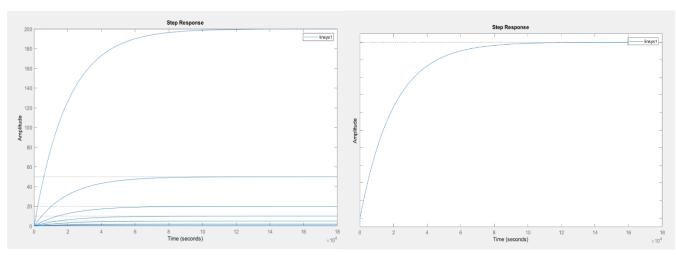
#### J.

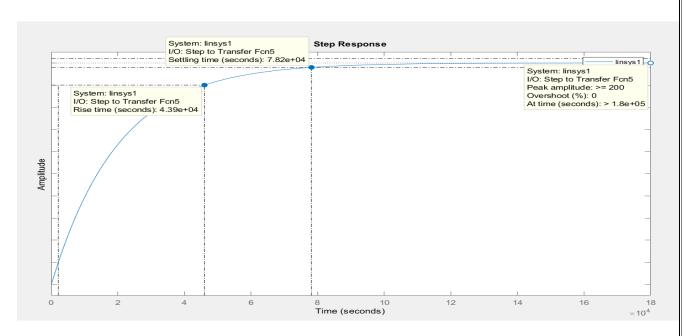
As the number of poles in a system increase, the transient response becomes slower. However, if they are quite far from the imaginary axis compared to the pole pair of the original second order system, the effect of slowing the transient response decreases, and as the new poles become nearer to the proximity of the pole pair, their effect increases and the pole pair dominance decreases. As the pole value rises, the final value decreases, while everything else remains largely the same.

### K. Zeros at -2, -5, -10, -20, -50, -200 at k=1









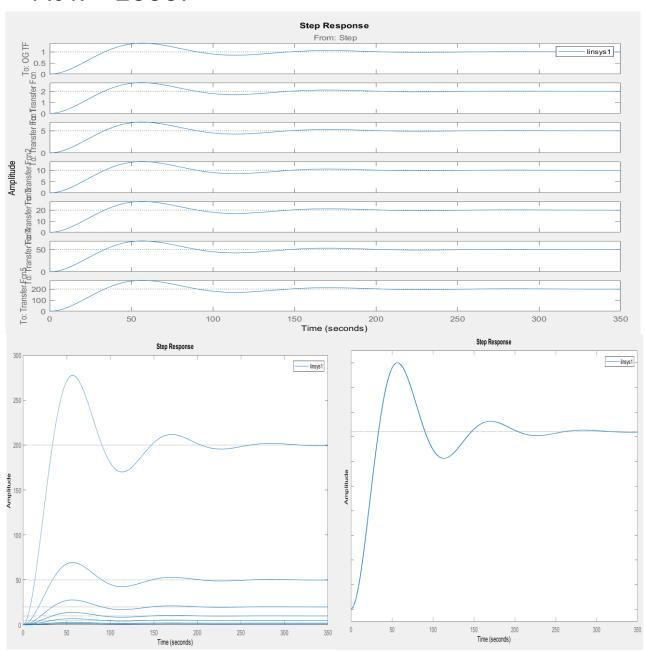
Rise time =  $4.39 \times 10^4$  seconds

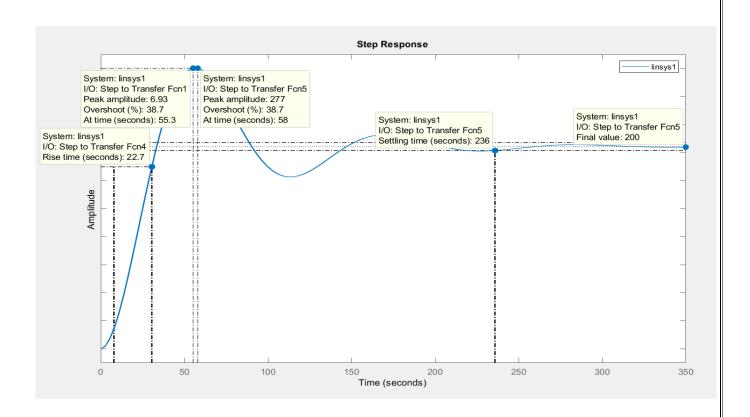
Settling time =  $7.82 \times 10^4$  seconds

Peak time = 1.8 x 10<sup>5</sup> seconds

Max overshot = 0%

### At k = 2000:





Rise time = 22.7 seconds

Settling time = 236 seconds

Peak time = 58 seconds

Max overshot = 38.7%

#### L.

The dominant pole pair is the pair closest to the imaginary axis of the S-plane plot, far closer than the rest of system roots, making the pole pair have a greater effect on the system, while the rest have little to none, and thus negligible.

As number of zeros increases, stability and overshoot increases, but rise time and peak time decrease. However, as they remain

farther from the imaginary axis on the S-plane plot, their effect decreases. As such, even for a higher-order system, the zero can be ignored if it is too far from the rest of the roots.

The dominant second-order pair exist when they are far closer to the imaginary axis than the rest of the poles and as a result let us assume the system to be of the second order since the rest of the roots are too far to affect the system.

As the zero value rises, the final value rises, while everything else remains the same.