

## Practice sheet 7

①

$$x[n] = \begin{cases} 1 & 0 \leq n \leq 5 \\ 0 & \text{ow} \end{cases}$$

$$g[n] = x[n] - x[n-1]$$

(a)

$$g[n] = \delta[n] - \delta[n-6]$$

$$\therefore G(z) = 1 - z^{-6} \quad |z| > 0$$

$$(b) \quad x[n] = \sum_{k=-\infty}^{\infty} g[k] \xleftrightarrow{z} X(z) = \frac{1}{1-z^{-1}} G(z)$$

For  $|z| < 1$

$$\therefore X(z) = \frac{1 - z^{-6}}{1 - z^{-1}}, \quad |z| > 0$$

The ROC is  $|z| > 0$  because  $x[n]$  is a finite length signal.

②

(a) NO, finite length signal ROC is the entire  $z$ -plane  
 $\therefore$  no poles in the finite  $z$ -plane for a finite length signal.

So this is not the case for given problem.

(b) No. If a signal is absolutely summable, the ROC must include the unit circle.

Signal has a pole at  $z = \frac{1}{2}$ , ROC can never be of the form  $0 \leq |z| < \infty$ . So signal cannot be left sided.

(c) YES. Since the signal is absolutely summable, the ROC must include the unit circle.

Since it is given that the signal has a pole at  $z = \frac{1}{2}$ , a valid ROC for this signal would be  $|z| > \frac{1}{2}$ . So this would correspond to a right sided signal.

(d) Yes, since the signal is absolutely summable, the ROC must include the unit circle. Clearly, we can define an ROC which is a ring in the  $z$ -plane and includes the unit circle. The signal is two sided.

(3)  $x(n) = \left(\frac{1}{3}\right)^n \cos\left(\frac{\pi}{4}n\right) \quad n \leq 0$

$$\therefore X(z) = \sum_{n=-\infty}^0 \left(\frac{1}{3}\right)^n \cos\left(\frac{\pi}{4}n\right) z^{-n}$$

$$= \frac{1}{2} \sum_{n=-\infty}^0 \left(\frac{1}{3}\right)^n e^{j\pi n/4} z^{-n} + \left(\frac{1}{2}\right) \sum_{n=-\infty}^0 \left(\frac{1}{3}\right)^n e^{j\pi n/4} z^{-n}$$

$$= \frac{1}{2} \frac{1}{1 - 3e^{j\pi/4} z} + \frac{1}{2} \frac{1}{1 - 3e^{j\pi/4} z}, \quad |z| < \frac{1}{3}$$

$$= \frac{1}{2} \frac{1}{1 - 3e^{-j\pi/4} z} + \frac{1}{2} \frac{1}{1 - 3e^{j\pi/4} z} \text{ for}$$

poles at  $z = \frac{1}{3}e^{j\pi/4}$  and  $z = \frac{1}{3}e^{-j\pi/4}$ .

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④ Let  $y[n] = x[n] \sum_{k=-\infty}^{\infty} \delta[n-3k]$

Then  $Y(e^{j\omega}) = \frac{1}{3} \sum_{k=0}^3 X(e^{j(\omega - 2\pi k/3)})$

Here kindly note  $\frac{\sin(\pi n/3)}{\pi n/3}$  is the impulse response of an ideal LPF with cut off frequency  $\pi/3$  and pass band freq. is 3.

$\therefore$  It is required when  $y[n]$  passed through this filter should yield  $x[n]$ .

Not to overlap,  $X(e^{j\omega}) = 0$  for  $\pi/3 \leq |\omega| \leq \pi$

(b) ⑤ If  $x(j\omega) = 0$  for  $|\omega| > \omega_1$ ,

then  $x(j\omega) * x(j\omega) = 0$  for  $|\omega| > 2\omega_1$ ,

$\therefore x(j\omega) = 0$  for  $|\omega| > 7000\pi$ .

Nyquist rate is  $N_R = 2 \times 7000\pi = 14000\pi$

In order to be able to retrieve  $x(t)$  from  $x_s(t)$

Maximum sampling period time  $= \frac{2\pi}{14000\pi}$   
 $= 1.33 \times 10^{-4} \text{ sec}$

Sampling period  $T = 10^{-4} \text{ sec} < T_{\max}$ .

⑤ (a) Nyquist rate in this case is  $2 \times 5000\pi = 10000\pi$ .

So for proper recovery of  $x(t)$ ,

sampling period must be at most

$T_{\max} = \frac{2\pi}{10000\pi} = 2 \times 10^{-4} \text{ sec}$

$\therefore$  Sampling period  $10^{-4} \text{ sec} < T_{\max}$ .

$\therefore x(t)$  is recovered.

(b)

(a) Yes, Aliasing will occur in this case,

Consider the sinusoidal signal  $x(t)$  for  $k=5$ .

$$y(t) = \left(\frac{1}{2}\right)^5 \sin(5\pi t)$$

If  $x(t)$  sampled at  $T=0.2$ , then we will always be sampling  $y(t)$  at exactly zero crossings.

$\therefore$  Sinusoidal  $y(t)$  of freq.  $5\pi$  is aliased into sinusoidal  
freq zero of sampled signal.

(b)

Since aliasing resulted in the loss of signal  $\left(\frac{1}{2}\right)^5 \sin(5\pi t)$ , the output is

$$x_0(t) = \sum_{k=0}^4 \left(\frac{1}{2}\right)^k \sin(k\pi t)$$

The Fourier series representation

$$x_0(t) = \sum_{k=-4}^4 a_k e^{j(k\pi t)}$$

$$\text{where } a_k = \begin{cases} 0 & k=0 \\ -j\left(\frac{1}{2}\right)^{k+1} & 1 \leq k \leq 4 \\ +j\left(\frac{1}{2}\right)^{-k+1} & -4 \leq k \leq -1 \end{cases}$$