

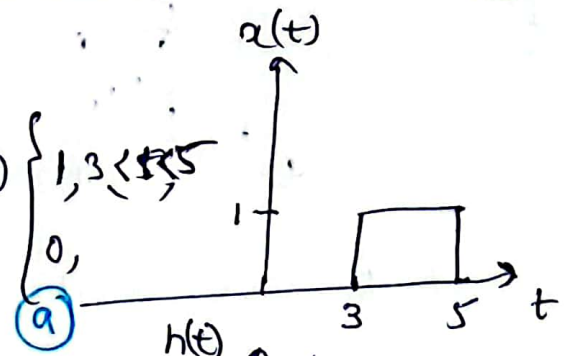
Practice sheet 3 solutions

(1)

Q1. Given,

$$x(t) = U(t-3) - U(t-5) \quad \begin{cases} 1, 3 < t < 5 \\ 0, \text{ otherwise} \end{cases}$$

$$h(t) = e^{-3t} \cdot U(t)$$



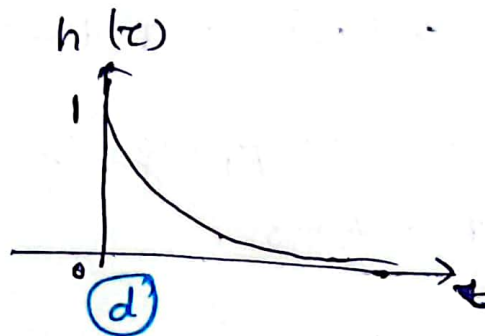
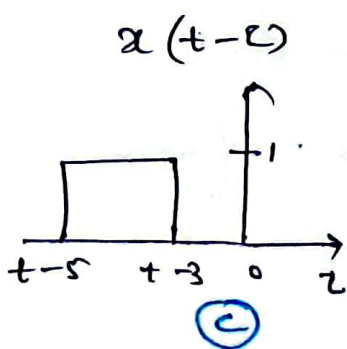
(a) $y(t) = x(t) * h(t)$

$$y(t) = \int_{-\infty}^{\infty} x(\tau) \cdot h(t-\tau) \cdot d\tau$$

$$y(t) = \int_{-\infty}^{\infty} h(\tau) \cdot x(t-\tau) \cdot d\tau$$

The signal $x(t)$ & $h(t)$ is represented above.

Now.



$$y(t) = x(t) * h(t) \rightarrow$$

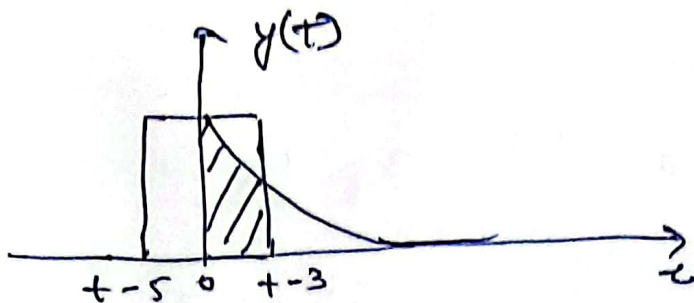
No overlapping
of
signal

$$y(t) = 0, \text{ if } t < 3$$

$$\text{if } t-3 < 0 \Rightarrow t < 3$$

(e)

(f)

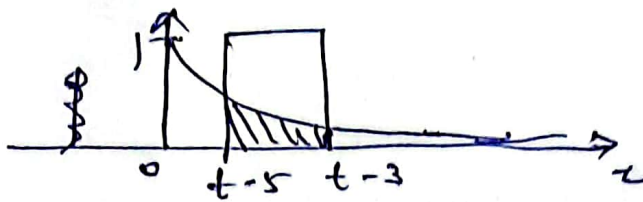


if $t-3 \geq 0$ and (2)

$$t-5 < 0$$

$$3 \leq t < 5$$

(g)



if $t-5 \geq 0$

$$t \geq 5$$

Now if $t-3 \geq 0 \Rightarrow t \geq 3$
 $t-5 < 0 \Rightarrow t < 5$

there will be overlapping of signal. from
 $z = 0$ to $z = t-3$. for $3 \leq t < 5$

$$y(t) = \int_0^{t-3} x(t-z) \cdot h(z) \cdot dz$$

$$= \int_0^{t-3} (1) \cdot e^{-3z} \cdot dz = \frac{1 - e^{-3(t-3)}}{3}$$

Now if $t-5 \geq 0 \Rightarrow t \geq 5$

$$t-3 \leq \infty \Rightarrow t < \infty$$

$z = t-5$ to $z = t-3$. for $5 \leq t \leq \infty$

$$y(t) = \int_{t-5}^{t-3} x(t-z) \cdot h(z) \cdot dz$$

$$y(t) = \int_{t-5}^{t-3} (1) \cdot e^{-3z} dz = \frac{(1 - e^{-6}) e^{-3(t-5)}}{3}$$

$$y(t) = \begin{cases} 0 & -\infty < t < 3 \\ \frac{1 - e^{-3(t-3)}}{3} & 3 < t \leq 5 \\ \frac{(1 - e^{-6})e^{-3(t-5)}}{3} & 5 < t \leq \infty \end{cases}$$

(b) By differentiating $x(t)$ w.r.t time we get

$$\frac{dx(t)}{dt} = \delta(t-3) - \delta(t-5)$$

$$\begin{aligned} g(t) &= \frac{dx(t)}{dt} * h(t) \\ &= e^{-3(t-3)} \cdot u(t-3) - e^{-3(t-5)} \cdot u(t-5) \end{aligned}$$

Explanation

$$g(t) = (\delta(t-3) * h(t) - \delta(t-5) * h(t))$$

$$g(t) = h(t-3) - h(t-5) \quad \begin{array}{l} \text{distributive} \\ \text{property of convolution} \end{array}$$

$$= e^{-3(t-3)} \cdot u(t-3) - e^{-3(t-5)} \cdot u(t-5)$$

$$g(t) = \begin{cases} 0 & -\infty < t < 3 \\ e^{-3(t-3)} & 3 < t \leq 5 \\ \begin{array}{l} e^{-3(t-3)} - e^{-3(t-5)}, 5 < t < \infty \\ \text{or} \\ (e^{-6} - 1)e^{-3(t-5)}, \end{array} & 5 < t < \infty \end{cases}$$

(3)

∴ (c) From the result of part (a) we can compute the derivative of $y(t)$.

$$\frac{dy(t)}{dt} = \begin{cases} 0 & -\infty < t < 3 \\ e^{-3(t-3)} & 3 < t \leq 5 \\ (e^{-6} - 1)e^{-3(t-5)} & 5 < t \leq \infty \end{cases}$$

We can say that

$$g(t) = \frac{dy(t)}{dt}$$

~~Q1.99~~

~~Q2. Consider~~

Q

Q20 (a) $y(t) = t^2 x(t-1)$
①

$$x_1(t) \longrightarrow y_1(t) = t^2 x_1(t-1)$$

$$x_2(t) \longrightarrow y_2(t) = t^2 x_2(t-1)$$

$$x_3 = a x_1(t) + b x_2(t)$$

$$y_3(t) = t^2 x_3(t-1)$$

$$= t^2 (a x_1(t-1) + b x_2(t-1))$$

$$y_3(t) = a y_1(t) + b y_2(t)$$

Therefore the system is linear.

② $y_1(t) = t^2 x_1(t-1)$

Now $x_2(t) = x_1(t-t_0)$

$$y_2(t) = t^2 x_2(t-1) = t^2 x_1(t-1-t_0)$$

$$y_1(t-t_0) = (t-t_0)^2 x_1(t-1-t_0) \neq y_2(t)$$

System is not time invariant.

Q2. (b) $y(t) = x(t^2)$

(4)

(i) $y_1(t) = x_1(t^2)$ — (1)

(ii) $y_2(t) = x_2(t^2)$ — (2)

$$y_1(t) + y_2(t) = x_1(t^2) + x_2(t^2)$$

$$y_3(t) = a x_1(t^2) + b x_2(t^2)$$

$$y_3(t) = a y_1(t) + b y_2(t)$$

The system is linear

(i) $y_1(t) = x_1(t^2)$

$$x_2(t) = x_1(t^2 - t_0) \rightarrow \text{delayed signal}$$

$$y_2(t) = x_2(t) \\ = x_1(t^2 - t_0)$$

$$y_1(t - t_0) = x_1((t - t_0)^2) \rightarrow \text{delayed output signal}$$

$$\therefore y_1(t) \neq y_2(t)$$

Hence system is time variant.

$$(c) \quad y[n] = x^2[n-2]$$

$$(i) \quad y_1[n] = x_1^2[n-2]$$

$$y_2[n] = x_2^2[n-2]$$

$$~~y_3[n] = a x_1[n] + b x_2[n]~~ \quad x_3[n] = a x_1[n] + b x_2[n]$$

$$y_3[n] = x_3^2[n-2]$$

$$= (a x_1^2[n-2] + b x_2^2[n-2])^2$$

$$y_3[n] \neq a y_1[n] + b y_2[n]$$

Hence system is nonlinear

$$(ii) \quad y_1[n] = x_1^2[n-2]$$

$$x_2[n] = x_1[n] = x_1^2[n-2-n_0]$$

$$y_2[n] = x_2^2[n-2] = x_1^2[n-2-n_0]$$

$$y_1[n-n_0] = x_1^2[n-2-n_0]$$

$$y_2[n] = y_1[n-n_0]$$

This implies the system is time invariant.

$$(d) \quad y[n] = x[n+1] - x[n-1]$$

$$(i) \quad y_1[n] = x_1[n+1] - x_1[n-1]$$

$$y_2[n] = x_2[n+1] - x_2[n-1]$$

$$x_3[n] = a y_1[n] + b y_2[n] \Rightarrow x_3[n] = a x_1[n] + b x_2[n]$$

$$y_3[n] = x_3[n+1] - x_3[n-1]$$

$$= a \{x_1[n+1] - x_1[n-1]\} + b \{x_2[n+1] - x_2[n-1]\}$$

$$\boxed{y_3[n] = a y_1[n] + b y_2[n]}$$

Hence, system is linear.

$$(ii) \quad y_1[n] = x_2[n+1] - x_1[n-1]$$

$$x_2[n] = x_1[n-n_0]$$

$$y_2[n] = x_2[n+1] - x_2[n-1]$$

$$= \cancel{x_1[n-n_0+1]} - \cancel{x_1[n-n_0-1]}$$

$$= x_1[n+1-n_0] - x_1[n-1-n_0]$$

$$y[n-n_0] = x_1[n+1-n_0] - x_1[n-1-n_0]$$

$$y_2[n] = y[n-n_0]$$

Hence the system is time invariant.

3.

$$\begin{aligned} \text{(a)} \quad y_2[n] &= x_2[n-2] + \frac{1}{2} x_2[n-3] \\ &= y_1[n-2] + \frac{1}{2} y_1[n-3] \\ &= 2x_1[n-2] + 4x_1[n-3] + \\ &\quad \frac{1}{2} (2x_1[n-3] + 4x_1[n-4]) \\ &= 2x_1[n-2] + 4x_1[n-3] + x_1[n-3] + 4x_1[n-4] \\ &= 2x_1[n-2] + 5x_1[n-3] + 4x_1[n-4] \end{aligned}$$

The input output relationship of S —

$$y[n] = 2x[n-2] + 5x[n-3] + 2x[n-4]$$

(b) Now let's reverse the S_1 & S_2



$$\begin{aligned} S_2: \quad y_1[n] &= 2x_1[n] + 4x_1[n-1] \\ &= 2y_2[n] + 4y_2[n-1] \\ &= 2(x_2[n-2] + \frac{1}{2}x_2[n-3]) + \\ &\quad 4(x_2[n-3] + \frac{1}{2}x_2[n-4]) \\ &= 2x_2[n-2] + x_2[n-3] + 4x_2[n-3] + 2x_2[n-4] \\ &= 2x_2[n-2] + 5x_2[n-3] + 2x_2[n-4] \end{aligned}$$

we observe there is no change. The output remain same.

(4)

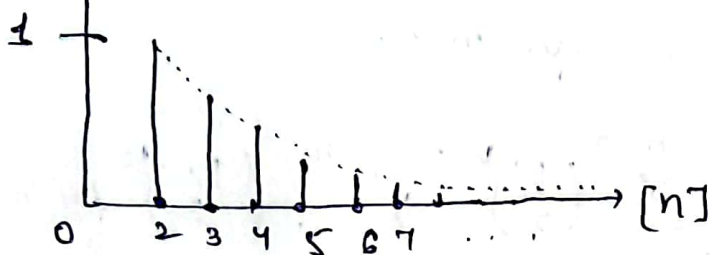
Given,

$$x[n] = \left(\frac{1}{2}\right)^{n-2} u[n-2]$$

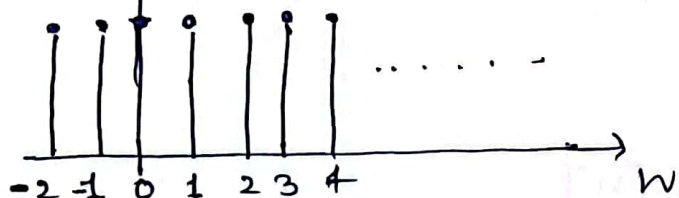
$$h[n] = u[n+2]$$

$$y[n] = x[n] * h[n] \quad ?$$

$$x[n] = \left(\frac{1}{2}\right)^{n-2} u[n-2]$$



$$h[n] = u[n+2]$$



We note that,

$$y[n] = x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

$$= \sum_{k=-\infty}^{\infty} \left(\frac{1}{2}\right)^{k-2} u[k-2] u[n+2-k]$$

Now by looking into $x[n]$ & $h[n]$ plot -

Conditions

$$u[k-2] = 1 \quad \text{when } k-2 \geq 0 \text{ or } k \geq 2 \\ = 0 \quad \text{otherwise}$$

$$u[n+2-k] = 1 \quad n+2-k \geq 0 \text{ or } k \leq n+2$$

There are two conditions

① $n+2 < 2 \rightarrow y[n] = 0 \quad \text{for } n < 0$

② $n+2 \geq 2 \rightarrow y[n] = \sum_{k=2}^{n+2} \left(\frac{1}{2}\right)^{k-2} \Rightarrow \sum_{k=0}^n \left(\frac{1}{2}\right)^k \quad \text{for } n \geq 0$

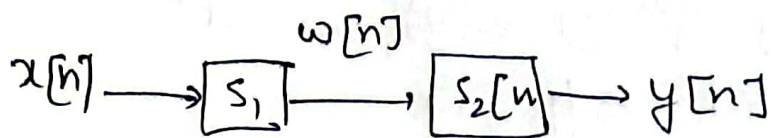
On applying finite sum formula.

$$\sum_{k=0}^n a^k = \frac{1-a^{n+1}}{1-a} \quad \text{for } n \geq 0 \text{ and } |a| < 1$$

$$y[n] = \frac{1 - \left(\frac{1}{2}\right)^{n+1}}{1 - \frac{1}{2}} u[n]$$

$$y[n] = 2 \left[1 - \left(\frac{1}{2}\right)^{n+1} \right] \cdot u[n]$$

5.



7

$$w[n] = \frac{1}{2} w[n-1] + x[n] \quad \text{--- (1)}$$

$$y[n] = \alpha y[n-1] + \beta w[n] \quad \text{--- (2)}$$

$$w[n] = \frac{1}{\beta} \{y[n] - \alpha y[n-1]\} \quad \text{from eqn (2)} \quad \text{--- (a)}$$

$$\frac{1}{2} w[n-1] = \frac{1}{2\beta} \{y[n-1] - \alpha y[n-2]\} \quad \text{--- (b)}$$

from eqn (1)

$$w[n-1] =$$

$$w[n] - \frac{1}{2} w[n-1] = x[n]$$

from eqn (a) & (b),

$$w[n] - \frac{1}{2} w[n-1] = \frac{1}{\beta} y[n] - \frac{\alpha}{\beta} y[n-1] - \frac{1}{2\beta} y[n-1] + \frac{\alpha}{2\beta} y[n-2]$$

$$w[n] - \frac{1}{2} w[n-1] = \frac{1}{\beta} y[n] - \left(\frac{\alpha}{\beta} + \frac{1}{2\beta}\right) y[n-1] + \frac{\alpha}{2\beta} y[n-2]$$

$$x[n] = \frac{1}{\beta} y[n] - \left(\frac{\alpha}{\beta} + \frac{1}{2\beta}\right) y[n-1] + \frac{\alpha}{2\beta} y[n-2] \quad \text{--- (3)}$$

On Comparing $y[n] = -\frac{1}{8} y[n-2] + \frac{3}{4} y[n-1] + x[n]$

$$x[n] = +y[n] + \frac{1}{8} y[n-2] - \frac{3}{4} y[n-1] \quad \text{--- (4)}$$

On comparing eqⁿ (3) & (4), we get

$$\frac{1}{\beta} = 1 \Rightarrow \boxed{\beta = 1}$$

$$\frac{\alpha}{2\beta} = \frac{1}{8} \Rightarrow \frac{\alpha}{2} = \frac{1}{8}$$

$$\boxed{\alpha = 1/4}$$

⑥

$$x(t) = \delta(t) + \delta(t-1) + \delta(t-2)$$

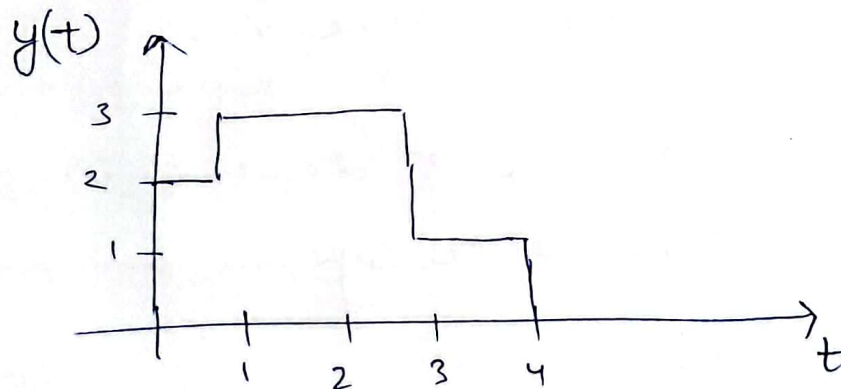
$$h(t) = 2u(t) - u(t-1) - u(t-2)$$

$$y(t) = x(t) * h(t)$$

$$= \{2u(t) - u(t-1) - u(t-2)\} * \{\delta(t) + \delta(t-1) + \delta(t-2)\}$$

$$= 2u(t) - u(t-1) - u(t-2) + 2u(t-1) - u(t-2) - u(t-3) + 2u(t-2) - u(t-3) - u(t-4)$$

$$y(t) = 2u(t) + u(t-1) - 2u(t-3) - u(t-4)$$



$$y(2) = 3$$

7.

$$y[n] = x^2[n] + \frac{1}{x^2[n-1]}$$

d)

a) Let input = $x_1[n]$

$$y_1[n] = x_1^2[n] + \frac{1}{x_1^2[n-1]}$$

$$\text{input} = x_2[n]$$

$$y_2[n] = x_2^2[n] + \frac{1}{x_2^2[n-1]}$$

$$y[n] = [x_1[n] + x_2[n]]^2 + \frac{1}{(x_1^2[n-1] + x_2^2[n-1])^2}$$

$$y[n] \neq y_1[n] + y_2[n]$$

Hence system is not linear ..

b) The present output depends on past & present values hence system is causal.

$$y[1] = x^2[1] + \frac{1}{x^2[0]}$$

$$y[0] = x^2[0] + \frac{1}{x^2[-1]}$$

c) For delayed input $x(n-t)$

$$y_1(n) = x_1^2(n-t) + \frac{1}{x_1^2(n-t-1)}$$

for delayed output $y(n-t)$

$$y(n-t) = x_1^2(n-t) + \frac{1}{x_1^2(n-t-1)}$$

$$y(n-t) = y_1(n)$$

Hence system is time invariant system

8. Given,

(a) ~~$h[n] = a^n u[n]$~~ $h[n] = a^n u[n+2]$

Stability is determined by checking whether the impulse response is absolutely summable —

$$\sum_{k=-\infty}^{\infty} |h[k]| = \sum_{k=-2}^{\infty} |a^k| \quad \text{--- (1)}$$
$$= a^{-2} + a^{-1} + \sum_{k=0}^{\infty} |a|^k$$

The infinite geometric sum in the equation (1) converges only if $|a| < 1$. Hence the system is stable and provide $0 < |a| < 1$. The system is not causal, since the impulse response $h[n]$ is non zero for $n = -1, -2$, The system is not memoryless because $h[n]$ is non zero for $n = -1, -2$.

The system is not memoryless because $h[n]$ is non zero for some values $n \neq 0$.

$$(b) \quad h[n] = n \cos\left(\frac{\pi}{4}n\right) \cdot u[n]$$

we know that

$$\begin{aligned} \sum_{k=-\infty}^{\infty} |h[k]| &= \sum_{k=-\infty}^{\infty} \left| k \cos\left(\frac{\pi}{4}k\right) u[k] \right| \\ &= \sum_{k=0}^{\infty} \left| k \cos\left(\frac{\pi}{4}k\right) \right| \end{aligned}$$

This sum does not have a finite value. because function $\left| k \cos\left(\frac{\pi}{4}k\right) \right|$ increases as the value of k increases. Therefore, $h_1[n]$ cannot be impulse response of a stable LTI system.

(19.1) Given,

$$h[n] = \left(\frac{1}{5}\right)^n u[n]$$

$$(a) \quad h[n] - A h[n-1] = \delta[n]$$

$$\left(\frac{1}{5}\right)^n u[n] - A \left(\frac{1}{5}\right)^{n-1} u[n-1] = \delta[n]$$

Putting $n=1$ and solving for A gives

$$A = 1/5$$

(b) from part (a)

$$h[n] - \frac{1}{5} h[n-1] = \delta[n]$$

$$h[n] * \delta[n] - \frac{1}{5} \delta[n-1] = \delta[n]$$

From the definition of inverse system,

we can say —

$$g[n] = \delta[n] - \frac{1}{5} \delta[n-1]$$