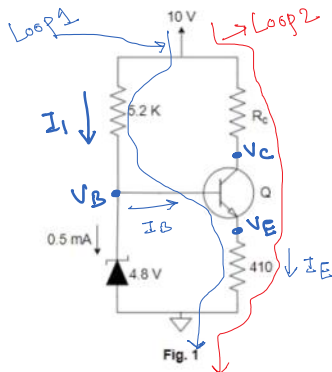


IE Quiz 1 Solution

22 February 2022 12:08

Q1. In the given circuit in Fig. 1, for the transistor Q, base-emitter drop ($V_{BE(ON)}$) is 0.7 V, $V_{CE(sat)} = 0.2$ V and zener breakdown voltage is given to be 4.8 V.

- a.) Calculate the value for the dc current gain β for Q? Take $R_C = 330 \text{ Ohm}$. [10 Marks]
b.) What is the range of R_C for which Q1 will remain in active region. [15 Marks]



Solution (a.) Since 0.5mA current is flowing in the Zener, hence it is in breakdown.

$$V_B = 4.8 \text{ V}$$

Applying KVL in loop 1

$$-10 + (10 - V_B) + V_{BE(ON)} + I_E \times 410 = 0$$

$$\Rightarrow -10 + (10 - 4.8) + 0.7 + 410 I_E = 0$$

$$\Rightarrow I_E = \frac{4.8 - 0.7}{410} = \frac{4.1}{410} = \frac{41}{410} \times 10^{-1} = 10^{-2}$$

$$\therefore I_E = 10 \text{ mA} \quad \text{--- (1)}$$

$$I_1 \text{ in the above fig} = \frac{10 - 4.8}{5.2 \text{ K}} = \frac{5.2}{5.2} \times 10^{-3} = 1 \text{ mA}$$

$$\therefore I_B = I_1 - 0.5 = 0.5 \text{ mA}$$

$$\text{Now, } I_E = (\beta + 1) I_B \therefore \beta + 1 = \frac{I_E}{I_B} = \frac{10 \text{ mA}}{0.5 \text{ mA}} = 20$$

$$\therefore \beta = 19$$

Since, KVL in loop 1 was applied by assuming Q to be in active region.
 \therefore checking the terminal voltages of Q:

$$V_E = 410 \times I_E = 4.1 \text{ V}$$

$$I_C = \beta \times I_B = 19 \times 0.5 \text{ mA} = 9.5 \text{ mA}$$

$$V_C = 10 - I_C \times R_C = 10 - 9.5 \text{ mA} \times 330 \Omega = 10 - 3.135 = 6.865$$

$$\therefore V_{CB} = 6.865 - 4.8 = 2.065 \rightarrow \text{Reverse Biasing of CB jn}$$

$$V_{BE} = 0.7 \rightarrow \text{EB jn is in forward bias}$$

Hence, our assumption was correct and Q is operating in active region with $\beta = 19$

Solution (b)

Applying KVL in loop 2 (as indicated in the figure)

$$-10 + I_C \times R_C + V_{CE} + I_E \times 410 = 0 \quad \text{--- (1)}$$

Since, current through the Zener is maintained at 0.5mA (as per question)

$$I_B = 0.5 \text{ mA} \quad (\text{for the transistor to work in active region})$$

$$\therefore I_C = 9.5 \text{ mA}$$

$$I_E = 10 \text{ mA}$$

∴ Putting all these values in ①

$$-10 + 9.5 \times 10^{-3} R_C + V_{CE} + 9.5 \times 10^{-3} \times 410 = 0$$

$$+ 9.5 \times 10^{-3} R_C + V_{CE} = 10 - 3.895 = 6.105$$

$$V_{CE} = 6.105 - 9.5 \times 10^{-3} R_C \quad \text{--- ②}$$

Now for the Q to remain in active region

$$V_{CE} > 0.2 \text{ V}$$

$$6.105 - 9.5 \times 10^{-3} R_C > 0.2 \text{ V}$$

$$9.5 \times 10^{-3} R_C < 6.105 - 0.2 = 5.905$$

$$\therefore R_C < \frac{5.905}{9.5} \times 10^3 \approx 0.62158 \times 10^3 \Omega$$

$$\therefore R_C < 621.58 \Omega$$

∴ Range of R_C will be $[0, 621.58 \Omega)$.

Q2. In the circuit shown in Fig. 2, diode is ideal with a cut-in voltage of 0.7 V.

a.) Find the magnitude of current I_2 . [10 Marks]

b.) If 2 K ohm arm is removed from the circuit, find the bias point for D1. [15 Marks]

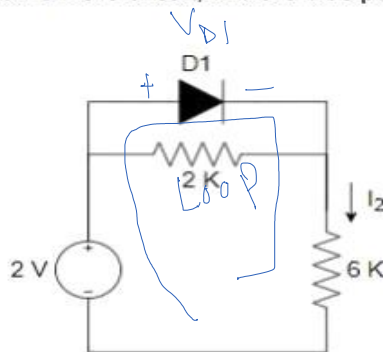
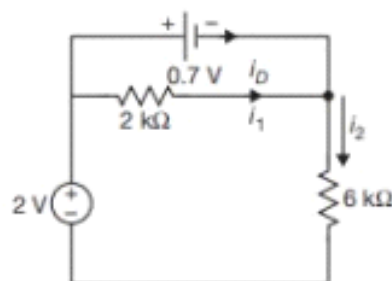


Fig. 2

Solution(a.)

We begin by converting the ideal diode to a voltage drop.



Let $i_2 = i_1 + i_D$. Applying KVL,

$$i_2 = \frac{2 - 0.7}{6} = 0.216 \text{ mA}$$

$$i_1 = \frac{0.7}{2} = 0.35 \text{ mA}$$

$$i_2 < i_1$$

$$\therefore i_D = -0.134$$

The diode is forward biased and it requires positive current to flow through it for conduction, but i_D is negative, so the diode is in OFF state.

$$\therefore i_2 = \frac{2}{2+6} = 0.25mA$$

Solution (b): If $2k$ ohm is removed then applying KVL in the loop

$$-2V + V_{D1} + 6k \cdot I_{D1} = 0 \quad (\text{Since, } I_{D1} = \text{diode current} = I_2)$$

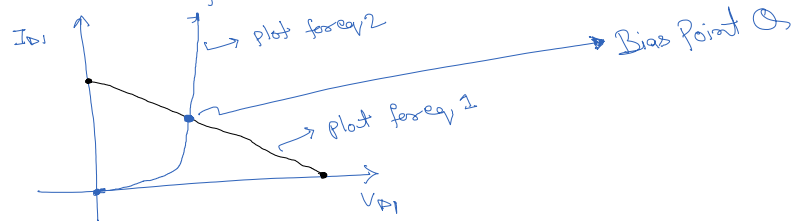
$$\therefore 6kI_{D1} + V_{D1} = 2 \quad \text{--- (1)}$$

Now the relation between V_{D1} and I_{D1} is known to us as

$$I_{D1} = I_0 \left(e^{\frac{qV_{D1}}{nV_T}} - 1 \right) \quad \text{--- (2)} \quad \{ \text{Take } n=1 \}$$

where, I_0 = Reverse saturation current

Plotting I_{D1} vs V_{D1} for both (1) and (2)



From the above figure by solving eq^{ns} (1) & (2) we can ^{get} the exact bias point (indicated in the plot by Q) for diode $D1$.

However, if we assume that diode drop will be approximately $0.7V$

$$\text{then } I_{D1} = \frac{2-0.7}{6k} = \frac{1.3}{6k} \approx 0.217mA$$

$\therefore (0.217mA, 0.7V)$ will be an approximate estimation of the bias point for $D1$.

Q3. In the circuit shown in Fig. 3, breakdown voltage of the zener is $5V$, β of the NPN transistor is 100 . Assume that $V_{BE} (ON)$ is $0.7V$

a.) Find the current and voltage at each terminal of the transistor. [15 marks]

b.) Can you make an NPN-BJT by connecting two PN junctions together from their P-terminal and using the N-terminals as Collector and Emitter (Refer Fig. 4). Give a detailed explanation.

[10 Marks]

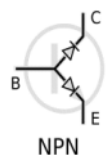
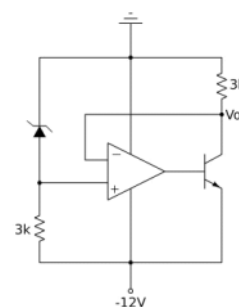
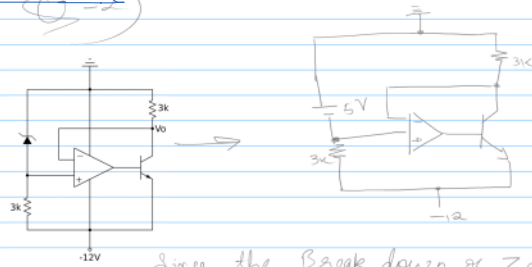


Fig. 4

Solution(a)



Since the Break down of Zener Diode is 5V, it will act like an Ideal Voltage Source in this case &

$$V_+ = -5V$$

$$\Rightarrow V_- = -5V \text{ (Virtual short)}$$

$$\Rightarrow I_c = \frac{5}{3k} = 1.66 \text{ mA} \quad V_c = -5V$$

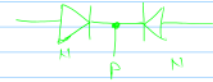
$$\Rightarrow I_B = \frac{I_c}{\beta} = \frac{1.66 \text{ mA}}{100} = 16.66 \mu A$$

$$\Rightarrow I_E = I_B + I_c = 1.6766 \text{ mA}$$

$$V_B = -11.3V \quad V_E = -12V$$

Solution(b)

B) A BJT can not be formed by



The N-Side of both the Diodes is at same Doping Concentration

The base Width of this design can not be ensured to be ultra-thin. The carrier Concentration profile is not guaranteed to be linearly decreasing in p-region. Thus there will not be a proper diffusion of n-type carriers from N to N. Thus the design will NOT work

Q4. For the circuit shown in Fig. 5, diode is ideal with a negligible cut-in voltage.

- Find the relation between V_o and V_i . Note that V_i can be both +ve and -ve. [15 Marks]
- Which circuit you have studied so far mimics the above circuit's behavior. Is there any advantage of the below circuit over the other circuit? [10 Marks]

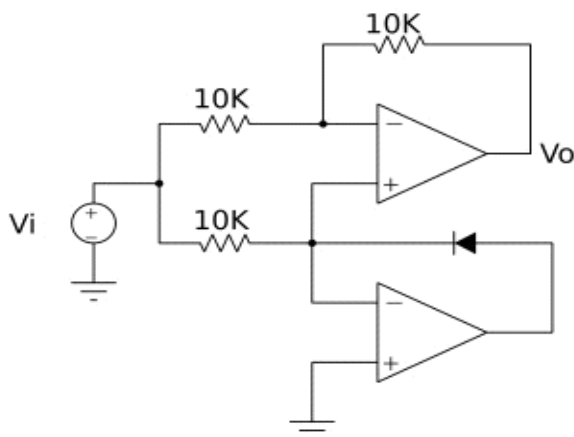
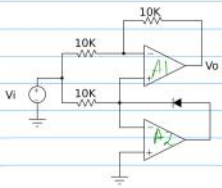


Fig. 5

Solution (a):

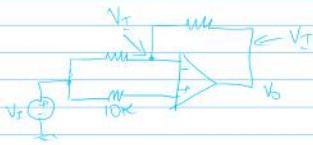


a) This question has to be solved in two steps:
 i) For the case when $V_i > 0$
 ii) $V_i < 0$

i) When $V_i > 0$:

Virtual short concept can not be applied in the op-amp for this case as the diode will be in Reverse Bias. Thus there will be no -ve feedback from diode to the op-amp.

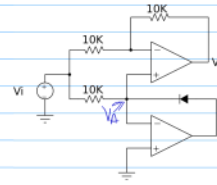
So the voltage at non-inverting terminal will not be 0 and since the diode is Reverse Biased, we can open it.



$$\Rightarrow V_o = V_i$$

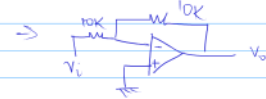
$$\Rightarrow \text{for } V_i > 0 : V_o = V_i$$

Case 2: $V_i < 0$:



When $V_i < 0$, the Diode can be forward Biased. So the diode will act as a wire. So the op-amp's Virtual Short will be valid in this case.

$$\Rightarrow V_A = 0V$$



$$\Rightarrow V_o = -V_i$$

Solution b \rightarrow for $V_i < 0 \rightarrow V_o = -V_i$

\rightarrow Combining: $V_o = |V_i|$

B) This acts as a Rectifier.

The potential drop across the diode does not appear in the output of this circuit. Thus this circuit can be used for rectifying low voltage signals!