

# **PID AND PREDICTIVE CONTROL OF ELECTRICAL DRIVES AND POWER CONVERTERS USING MATLAB<sup>®</sup>/SIMULINK<sup>®</sup>**

# **PID AND PREDICTIVE CONTROL OF ELECTRICAL DRIVES AND POWER CONVERTERS USING MATLAB<sup>®</sup>/SIMULINK<sup>®</sup>**

**Liuping Wang, Shan Chai, Dae Yoo, Lu Gan and Ki Ng**

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# Preface

## About This Book

Electrical drives play a critical role in electromechanical energy conversions. They are seen everywhere in our daily life from the cooling fans, washing machines to computers. They are the fundamental building blocks in manufacturing, transportation, mineral processing, wind energy and many other industries. For the last several decades, the advances of electronically switched semiconductors in the form of power electronics have made *AC* motor drives gain more prominence over the *DC* machines in industries since they allow a direct connection to power grids via grid connected power converters and have a more reliable physical structure. The grid connected three phase power converter has wide applications in renewable energy generation.

This book gives an introduction to the automatic control of electrical drives and grid connected three phase power converters, and to recent developments in design and implementation. When they are combined together as one unit, it will provide a direct connection for the electrical drives to the power grid for electromechanical energy conversions and renewable wind energy applications. In the context of control system design, electrical drives and grid connected three phase power converters share similar characteristics in their dynamic models and use the same type of semiconductors as actuators in the implementation of control systems. Therefore, in this book, electrical drives and power converters will be studied as individual components of the larger system and examined in the same framework.

As electrical drives and power converters have restricted operations imposed by electronically switched semiconductors, their operational constraints are paramount in the design and implementation of the control systems. In this regard, model predictive control has an established reputation in successfully handling the operational constraints in an optimal manner. Two chapters of this book will be devoted to seeking new predictive control technologies that address the specific needs of controlling electrical drives and power converters, and an additional two chapters will apply the existing predictive control technologies to these systems. Since PID control systems are used in the majority of industrial electrical drives and power converters, understanding these control systems and having the capability to design and implement them are important to a control engineer. There are three chapters in the book that will systematically cover PID control system design, PID control system implementation with anti-windup mechanisms and tuning of PID control systems. All control systems presented in this book have been experimentally validated using self-built test-beds with industrial sized motors. To assist the reader, tutorials about the real-time control system implementation and the physical model based simulators are presented in this book.

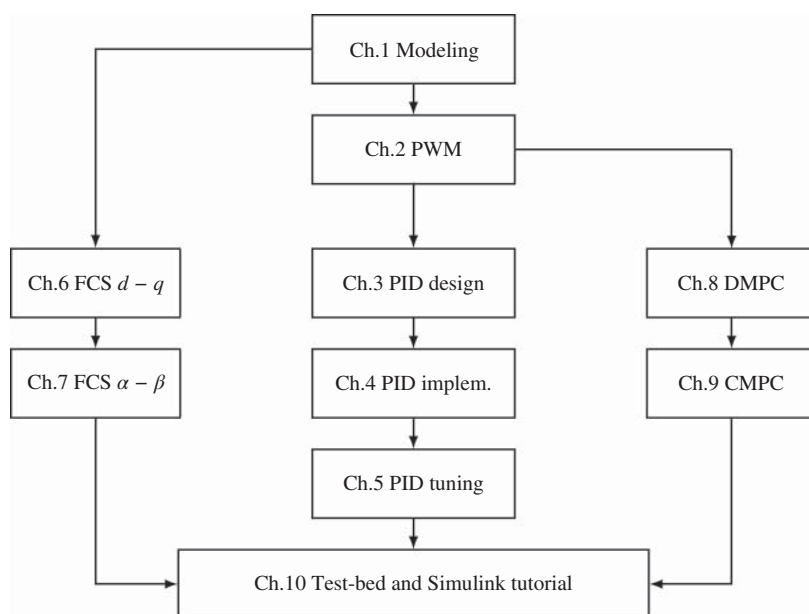
This book is intended for readers who have completed or are about to complete four years engineering studies with some basic knowledge in electrical and control systems. The targeted readers are students, practitioners, instructors and researchers who wish to learn electrical motor control and power converter control. The book is self-contained with MATLAB/Simulink tutorials and supported with simulation and experimental results. It is worth mentioning that the material contained in the first five chapters is aimed at readers who are working or are going to work in the relevant engineering field.

## Outline of This Book

The structure of the book is illustrated by the block diagram as shown in Figure 0.1. There are ten chapters in this book, covering the topics of mathematical modeling, control of semiconductor switches, PID control system design, implementation and tuning, Finite Control Set (FCS)-predictive control in both  $d-q$  and  $\alpha-\beta$  reference frames, traditional predictive control in both continuous-time and discrete-time. PID controllers (see Chapters 3–5) are implemented using Pulse-Width-Modulation (PWM) technologies introduced in Chapter 2. The traditional model predictive controllers (see Chapters 8–9) use this technology too. However, FCS-predictive controllers (see Chapters 6–7) are implemented without PWM mechanisms by directly optimizing the switching patterns of semiconductors. Hence, this has significantly simplified the implementation procedure of control systems.

This book begins by discussing the physical models of electrical drives and grid connected three phase power converter since mathematical modeling is the first step toward the design and implementation of control systems. In Chapter 1, the mathematical models of machine drives and power converter are derived in a unified way that firstly uses space vector description of physical variables such as voltage, current and flux, and secondly converts the space vector based model to various reference frames. By adopting this unified framework, it is hoped that through the derivations in a similar process, the dynamic models of drives and power converters can be easily understood by a reader who does not have extensive background in AC machines and power converters. It must be emphasized, due to the efforts of generations of electrical engineers (see for example Park (1929), Duesterhoeft *et al.* (1951), Vas (1992), Leonhard (2001), Drury (2009), Hughes and Drury (2013), Quang and Dittrich (2008)), that the dynamics models are highly structured and have incredibly high fidelity, which forms the solid basis for control system designs introduced in the book.

From a control engineer's perspective, the next natural question following from mathematical modeling is how to realize manipulated control variables in applications. It has been well established that control of semiconductor switches is the most efficient and convenient means to achieve control of AC



**Figure 0.1** Book structure diagram

machine drives and power converters. It is shown in Chapter 2 that they act as actuators in the implementation of control systems where the manipulated control inputs in the form of three phase voltage signals are realized by turning on and off semiconductor switches. Also, the PWM implementation of control systems dictates the operational limits termed linear modulation range, which, in later chapters, will be translated into constraints imposed in the PID and predictive controllers using PWM mechanisms for implementation.

The next three chapters of this book will see the developments of PID control systems for electrical drives and power converter (see Chapters 3–5). In Chapter 3, for AC motor control systems, electromechanical torque control is achieved using PI control of currents in the  $d-q$  reference frame, followed by achieving further requirements of controlling angular velocity and position via a cascade control system architecture. Identical control strategies are deployed to control the currents of a three phase power converter and its DC voltage in a grid connected environment. In all PID controller designs presented in the book, the pole-assignment control method is used. The reasons for this choice of design method is that it is perhaps among the simplest control system design methods and yet offers an effective means of selecting desired closed-loop performance in terms of response to reference signals and to disturbance rejection. In Chapter 4, PI controller implementation is discussed for both current controllers as inner-loop controllers, and velocity and DC voltage controllers as outer-loop controllers. In particular, continuous-time controllers are discretized for digital implementation, and operational constraints imposed by PWM operations are taken into consideration in the implementation of PI controllers. In order to avoid integrator wind-up in the presence of control signal reaching saturation limits, anti-windup mechanisms are proposed together with digital implementation, which leads to the so-called velocity form that has naturally embedded anti-windup mechanisms and is convenient for implementation. A MATLAB tutorial is introduced in this chapter to show how an embedded function can be created for the PI controller with its anti-windup mechanism, which has been directly used in the experimental validation. In Chapter 5, sensitivity functions in feedback control systems are introduced to measure the closed-loop control system performance against set-point following, disturbance rejection and noise attenuation in the frequency domain. Current control systems are analyzed for the effects of current sensor errors and harmonics caused by the voltage source inverter used in implementation of the control system. When velocity control, position control or DC voltage control is required in a cascade control structure, performance robustness in the outer-loop control system is considered where a weighting function is introduced to quantify the difference between the desired closed-loop performance and the actual closed-loop performance. Parameter variations are also studied using Nyquist plots. A large number of experiments are conducted in this chapter to demonstrate tuning procedures of the PI cascade control systems.

There are two approaches used in this book to generate the gate signal for the semiconductor switches. The first approach uses Pulse Width Modulation (PWM) based on which PID controllers (see Chapters 3–5) and traditional model predictive controllers are implemented (see Chapters 8–9). In control applications, the control signals calculated are the three phase voltage signals that are obtained from one of the controller designs using the model either in the  $d-q$  reference frame or  $\alpha-\beta$  reference frame. The role of the voltage source inverter with power electronics devices is to realize three phase voltage control signals as closely as possible. Namely, the sinusoidal phase voltage signals created by turning on-off each power switch with PWM technologies are aimed to be closely matched with three phase voltage control signals. The second methodology features a much simpler approach in the implementation of control systems that generates such a gate signal by direct optimization of an error function between the desired control signals and those that can be achieved by semiconductor switches (see Chapters 6–7). In the second approach, there is no need to use the PWM technology; therefore it significantly reduces the complexity of controlling semiconductor switches.

In Chapter 6, in the  $d-q$  reference frame, finite control set (FCS) predictive controllers are used to directly optimize inverter states; as a result, PWMs are not required in the implementation of control systems, which simplifies the implementation procedure. The original FCS predictive control systems did not include integrators in their design and implementation. Consequently, there are steady-state errors

within control systems. The existence of steady-state errors affects closed-loop performance, particularly when there are parameter uncertainties in the system, which is the main reason why the majority of practical control systems have integrator in the controller structure. By analyzing the original FCS predictive control system without constraints, the discrete-time feedback controller gain and locations of closed-loop eigenvalues are revealed. To embed integrators in the FCS predictive controller, a cascade control system structure is proposed where the inner-loop system is controlled with the original FCS predictive controller and the outer-loop is by an integrated feedback control. There are perhaps many ways to include integrators in the FCS predictive control system; however, the proposed approach has kept the spirit of the original FCS predictive control system and maintained its simplicity both conceptually and computationally. Because the FCS predictive control systems are designed for current control, this chapter will also show how to design velocity and position control for AC drives when current controllers are FCS predictive controllers.

In Chapter 7, under investigation is the finite control set (FCS) predictive current control in the  $\alpha - \beta$  reference frame (or stationary frame). In the  $\alpha - \beta$  reference frame, the currents  $i_a(t)$  and  $i_b(t)$  are linear combinations of three phase currents  $i_a(t)$ ,  $i_b(t)$  and  $i_c(t)$ . Thus, they are sinusoidal functions. So are the voltage variables  $v_a(t)$  and  $v_b(t)$ . The current reference signals to FCS predictive control systems are sinusoidal signals, which differentiates current control systems in the  $\alpha - \beta$  reference frame from those in the  $d - q$  reference frame. It will be shown in this chapter that the original FCS predictive controllers are single-input and single-output controllers in exceptionally simple forms. However, in order to track sinusoidal current reference signals without steady-state errors, a controller with resonant characteristic is required in the  $\alpha - \beta$  reference frame. Extensive simulation and experimental results have been presented in these two chapters to show the outstanding closed-loop control performance of FCS predictive control systems.

The next two chapters of this book (see Chapters 8–9) apply the traditional model predictive control algorithms to AC machine drives and power converters. These predictive control algorithms were derived for general applications without those restrictions imposed on system dynamics. The MATLAB programs used in applications were given in Wang (2009). Although the traditional predictive control algorithms could be applied to current control, their advantages are perhaps lost to the simpler and more effective FCS predictive control approaches, also to simpler PI controllers. Therefore, in Chapters 8 and 9, velocity control in AC drives and DC voltage control in power converters are considered, and for these cases, traditional model predictive controllers offer the advantages of designing the control systems using multi-input and multi-output approaches in the presence of constraints.

The final chapter of this book will discuss the test beds used in the experimental evaluations of control systems. For those who wish to know how to perform real-time simulations using the physical models of drives and power converter, Simulink tutorials are given to show the model building process in a step-by-step manner.

## References

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## Asking for Feedback

We would like to ask our readers to contact us about any errors or suggestions for future improvement of our book.

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# List of Symbols and Acronyms

## Symbols

$\arg \min$	Minimizing argument
$A$	State matrix of state-space model
$B$	Input-to-state matrix of state-space model
$C$	State-to-output matrix of state-space model
$D$	Direct feed-through matrix of state-space model
$(A, B, C, D)$	State-space realization
$\Delta U$	Parameter vector for the control sequence in discrete time MPC
$\Delta u(k)$	Incremental control at sample $k$
$F_x, \Phi$	Pair of matrices used in the prediction equation $X = F_x x(k_i) + \Phi \Delta U$
$B_v$	Viscous friction coefficient in PMSM
$f_d$	Viscous friction coefficient in induction motor model
$G(s)$	Transfer function model
$\gamma$	Tuning parameter for PI controllers
$i_\alpha, i_\beta$	Currents of PMSM and power converter in $\alpha - \beta$ reference frame
$i_d, i_q$	Currents of PMSM and power converter in $d - q$ reference frame
$i_{sd}, i_{sq}$	Stator currents of induction motor in $d - q$ reference frame
$i_{s\alpha}, i_{s\beta}$	Stator currents of induction motor in $\alpha - \beta$ reference frame
$I_{q \times q}$	Identity matrix with appropriate dimensions
$J$	Performance index for optimization
$J_m$	Moment of inertia ( $\text{kg} \cdot \text{m}^2$ )
$K_c$	Proportional control gain
$K_{lqr}$	Feedback control gain using LQR
$K_{mpc}$	Feedback control gain using MPC
$K_{fcs}$	Feedback control gain using FCS predictive control in $d - q$ reference frame
$k_{fcs}^\alpha, k_{fcs}^\beta$	Feedback control gain using FCS predictive control in $\alpha - \beta$ reference frame
$K_{ob}$	Observer gain vector
$l_i(t)$	The $i$ th continuous-time Laguerre function
$L(t)$	Continuous-time Laguerre functions in vector form
$L_s$	Inductance of power converter and PMSM
$L_r, L_s$	Inductance of stator / rotor winding of induction motor
$L_h$	Machine mutual inductance of induction motor
$\lambda$	Lagrange multiplier
$\lambda_i(A)$	The $i$ th eigenvalue of matrix $A$

$\lambda^l, \lambda^m, \lambda^h$	Scheduling parameters
$\mu$	Disturbance vector
$N$	Number of terms used in Laguerre function expansion in continuous time
$N_c$	Control horizon
$N_p$	Prediction horizon
$\Omega_{mpc}, \Psi_{mpc}$	Pair of matrices in the cost of predictive control in either the continuous-time or discrete-time design, $J = \eta^T \Omega_{mpc} \eta + 2\eta^T \Psi_{mpc} x(t) + cons$
$\eta$	Parameter vector in the Laguerre expansion
$p$	Scaling factor for continuous-time Laguerre functions
$Q, R$	Pair of weight matrices in the cost function of predictive control
$R_s$	Resistance of stator in PMSM and induction motor, also grid resistance in power converter
$R_r$	Resistance of rotor winding.
$r(\cdot)$	Set-point signal
$q^{-i}$	Backward shift operator, $q^{-i}[f(k)] = f(k - i)$
$S(s)$	Sensitivity function
$S^i(s)$	Input sensitivity function
$S_i$	Switching state of inverter
$S_d, S_q$	Normalized voltage variables of converter's d-axis voltage $v_d$ and q-axis voltage $v_q$
$T(s)$	Complementary sensitivity function
$T_e$	Electromagnetic torque (N · m)
$T_L$	Load torque (N · m)
$T_p$	Prediction horizon in continuous-time
$\tau_D$	Derivative control time constant
$\tau_f$	Derivative control filter time constant
$\tau_I$	Integral control time constant
$\theta_r$	Mechanical position of motor shaft (radian)
$\theta_e$	Electrical position of motor shaft (radian)
$\theta_s$	Position of synchronous flux (radian)
$\psi_s$	Stator flux of induction motor (Wb)
$\psi_{rd}, \psi_{rq}$	Rotor flux of induction motor (Wb)
$u(\cdot)$	Control signal
$u_{s\alpha}, u_{s\beta}$	Stator voltages of induction motor (V) in $\alpha - \beta$ reference frame
$u_{sd}, u_{sq}$	Stator voltages of induction motor (V) in $d - q$ reference frame
$v_\alpha, v_\beta$	Voltages of PMSM and power converter (V) in $\alpha - \beta$ reference frame
$v_d, v_q$	Voltages of PMSM and power converter (V) in $d - q$ reference frame
$\omega_e$	Electrical motor speed (rad/s) (or RPM)
$\omega_m$ (or $\omega_r$ )	Mechanical motor speed (rad/s) (or RPM)
$\omega_s$	Speed of synchronous flux (rad/s) (or RPM)
$\omega_g$	Grid frequency (rad/s).
$\omega_{slip}$	Slip in induction motor
$u^{min}, u^{max}$	Minimum and maximum limits for $u$
$w_n$	Bandwidth or natural frequency in PI controller design (rad/s)
$x(\cdot)$	State vector
$x(t_i + \tau   t_i)$	Predicted state vector at time $\tau$ given current state vector $x(t_i)$
$\hat{x}(t)$	Estimated state vector in continuous-time
$\xi$	Damping coefficient in PI controller design
$Y$	Predicted output data vector
$Z_p$	Number of pole pairs

## Acronyms

CMPC	Continuous-time model predictive control
DMPC	Discrete-time model predictive control
MMF	Magnetic motive force
PMSM	Permanent magnetic synchronous machine
PLL	Phase-locked loop
PID	Proportional, integral and derivative
FCS	Finite control set
FCS-MPC	Finite control set predictive control
I-FCS-MPC	Integral finite control set predictive control