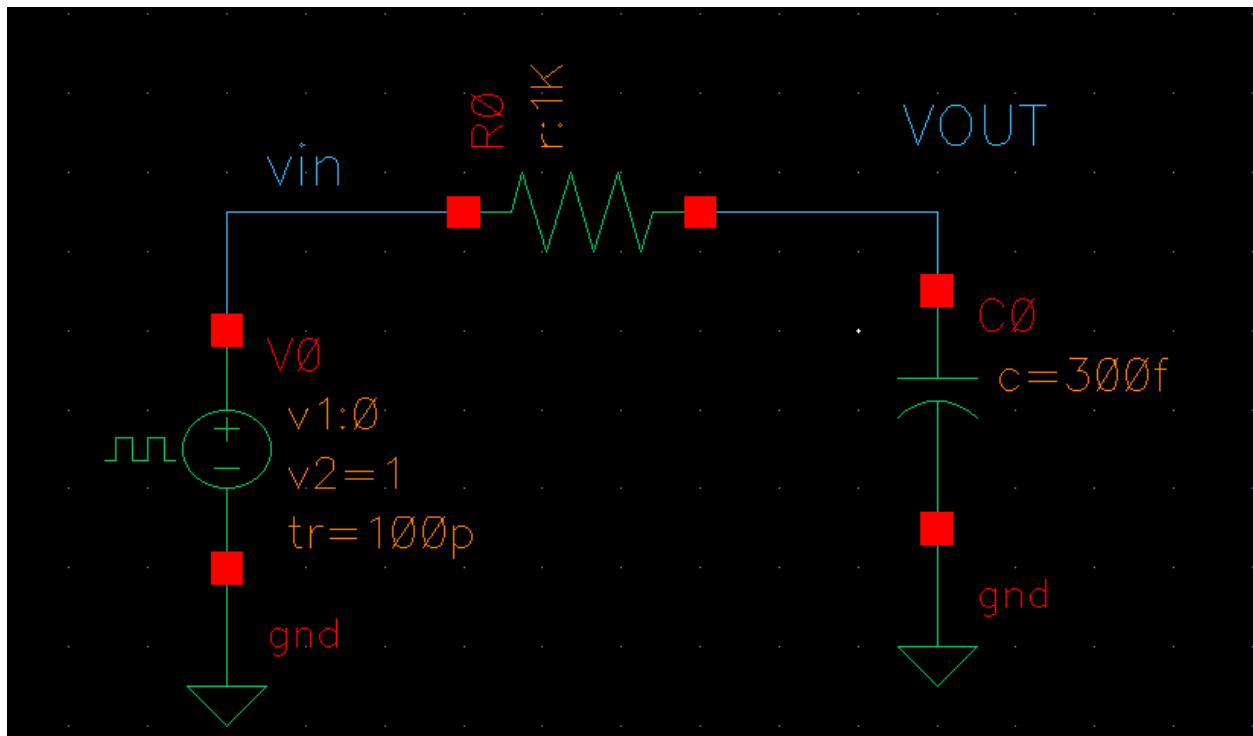


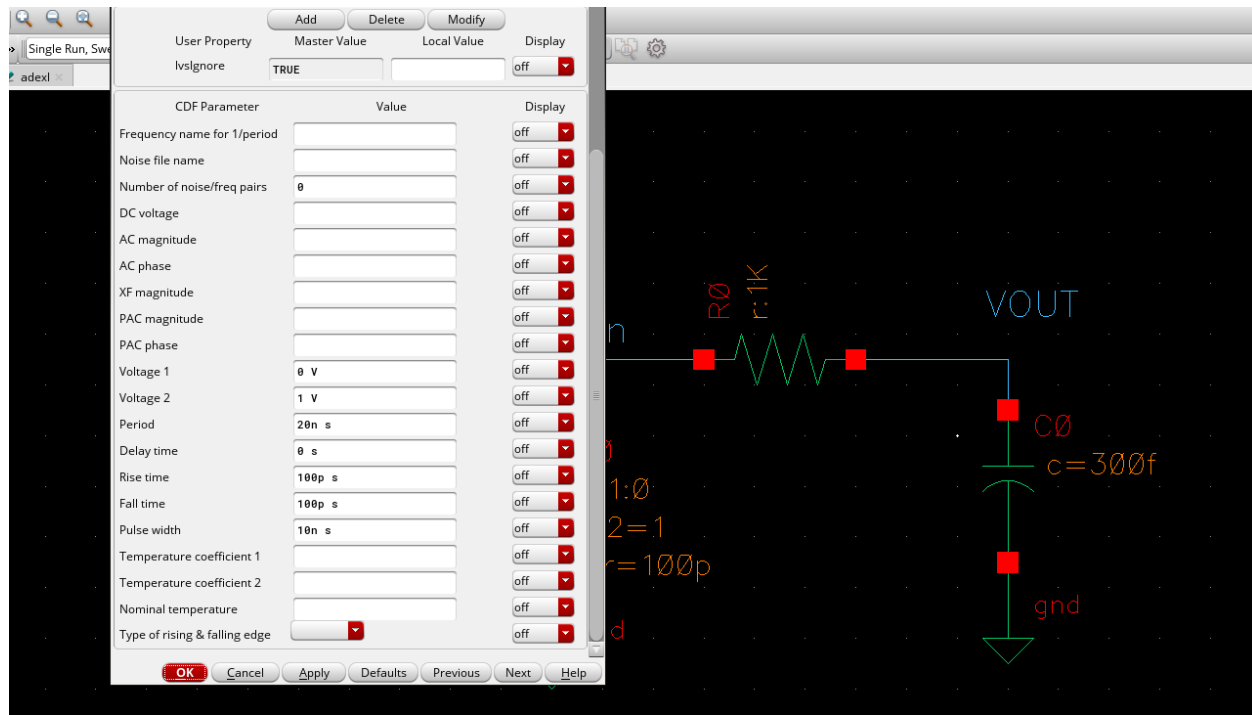
Lab 01 LPF Simulation and MOSFET Characteristics

PART 1: Low Pass Filter Simulation (LPF)

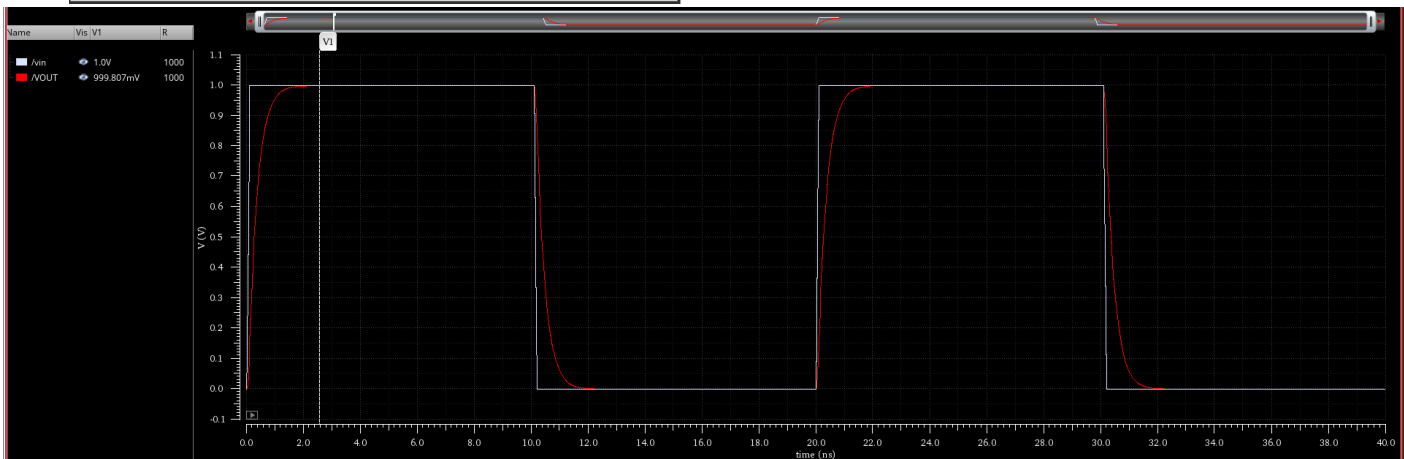
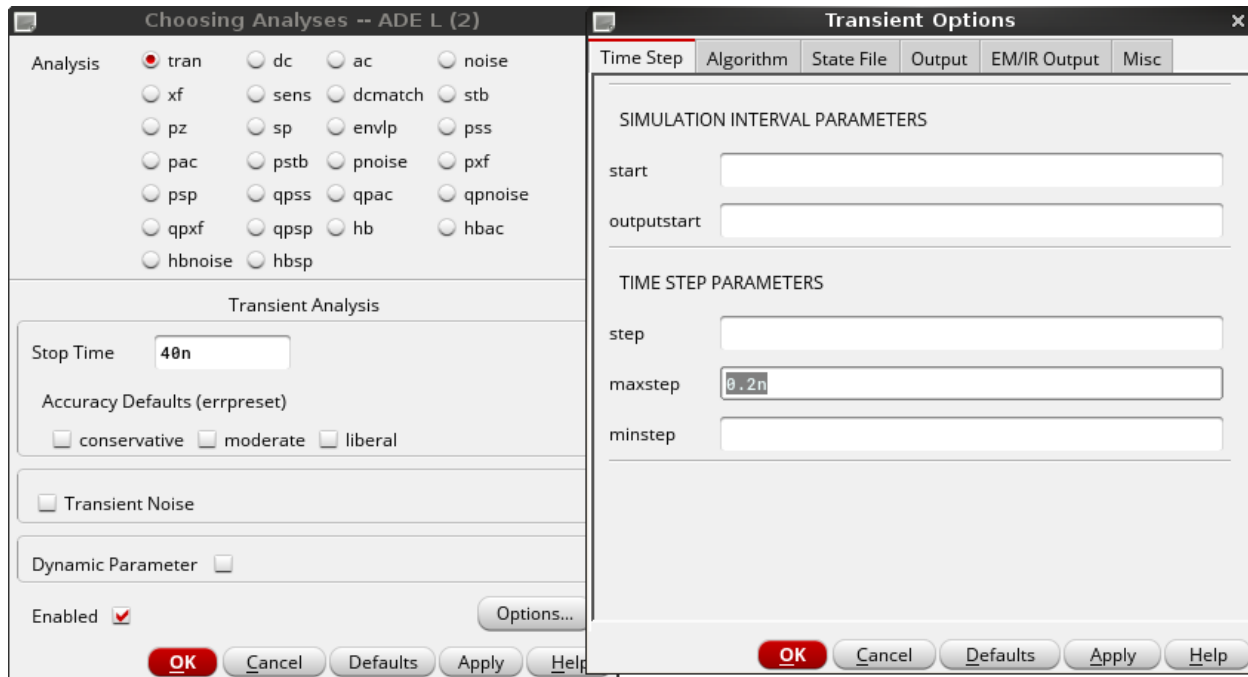
- 1 DESIGN A FIRST ORDER LOW PASS FILTER THAT HAS $R = 1k\Omega$ AND 0.3 ns (INSTEAD OF 1 ns) TIME CONSTANT.



2 APPLY A SQUARE WAVE INPUT WITH $T_{high} = \text{Pulse Width} = 10\text{ns}$,
 $T_{clk} = \text{Period} = 20\text{ns}$, AND $T_{rise} = T_{fall} = 100\text{ps}$.

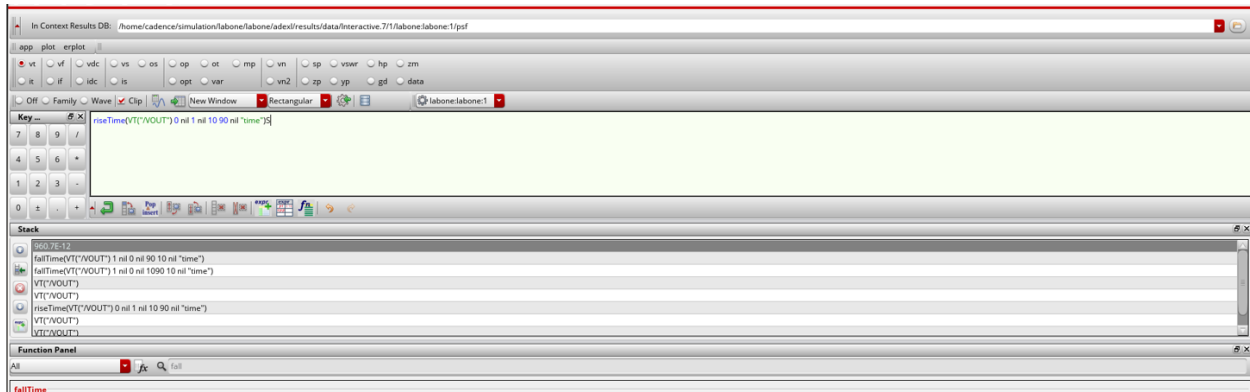


3 REPORT TRANSIENT ANALYSIS RESULTS FOR TWO PERIODS (USE MAX TIME STEP = $T_{clk}/100$).

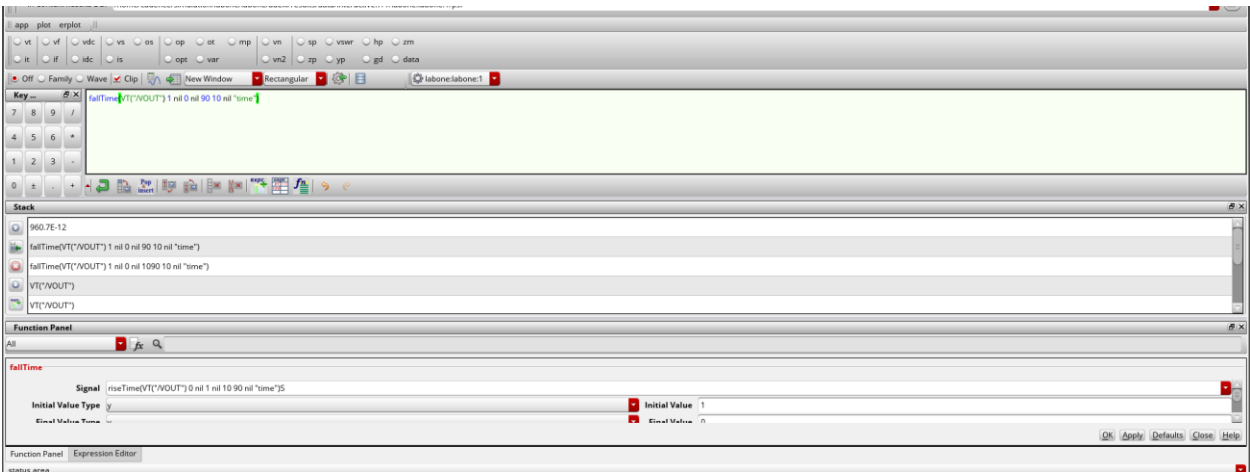


4 CALCULATE RISE AND FALL TIME (10% TO 90%) USING CADENCE CALCULATOR EXPRESSIONS. EXPORT THE EXPRESSIONS TO ADEXL.

4.1 RISE TIME



4.2 FALL TIME



Test	Name	Type	Details	EvalType	Plot	Save
labone:labone:1		expr		point	<input type="checkbox"/>	<input type="checkbox"/>
labone:labone:1	Trise	expr	riseTime(VT("/VOUT") 0 nil 1 nil 10 90 nil "time")	point	<input checked="" type="checkbox"/>	<input type="checkbox"/>
labone:labone:1	Tfall	expr	fallTime(VT("/VOUT") 1 nil 0 nil 90 10 nil "time")	point	<input checked="" type="checkbox"/>	<input type="checkbox"/>
labone:labone:1		signal	/vin	point	<input checked="" type="checkbox"/>	<input checked="" type="checkbox"/>
labone:labone:1		signal	/VOUT	point	<input checked="" type="checkbox"/>	<input checked="" type="checkbox"/>

Test	Output	Nominal	Spec	Weight	Pass/Fail
labone:labone:1	Trise	661.8p			
labone:labone:1	Tfall	661.8p			
labone:labone:1	/vin				
labone:labone:1	/VOUT				

5 COMPARE SIMULATION WITH ANALYTICAL RESULTS IN A TABLE.

	SIM	analytical
Trise	661.8p	659.17 p
Tfall	661.8p	659.17 p











$$0.9 = 1 - e^{-\frac{T_{90\%}}{0.3n}}, T_{90\%} = 690.77p$$

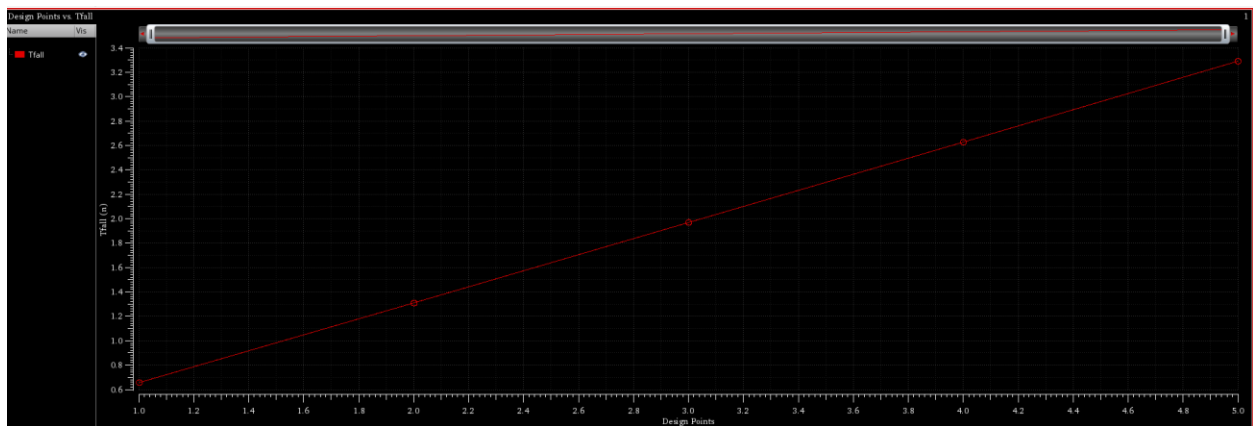
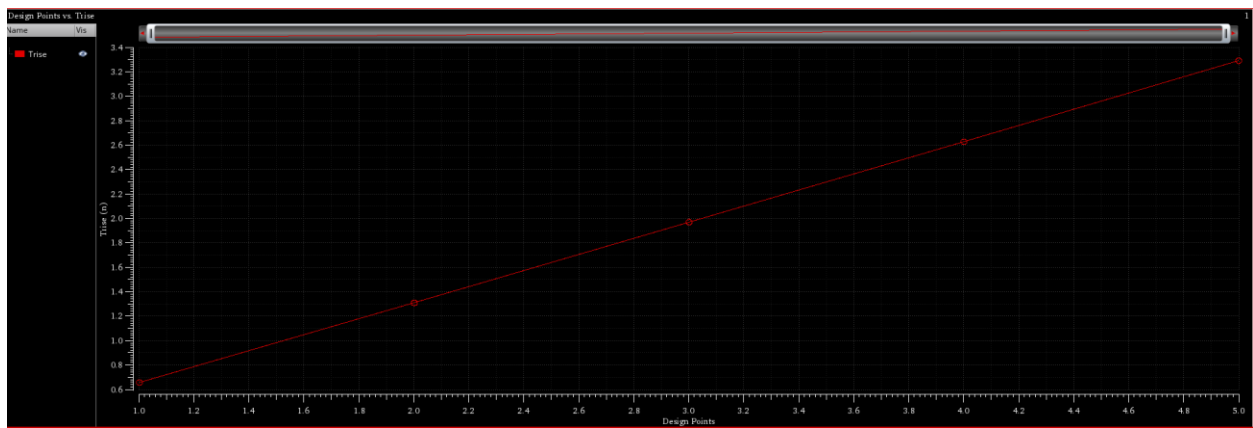
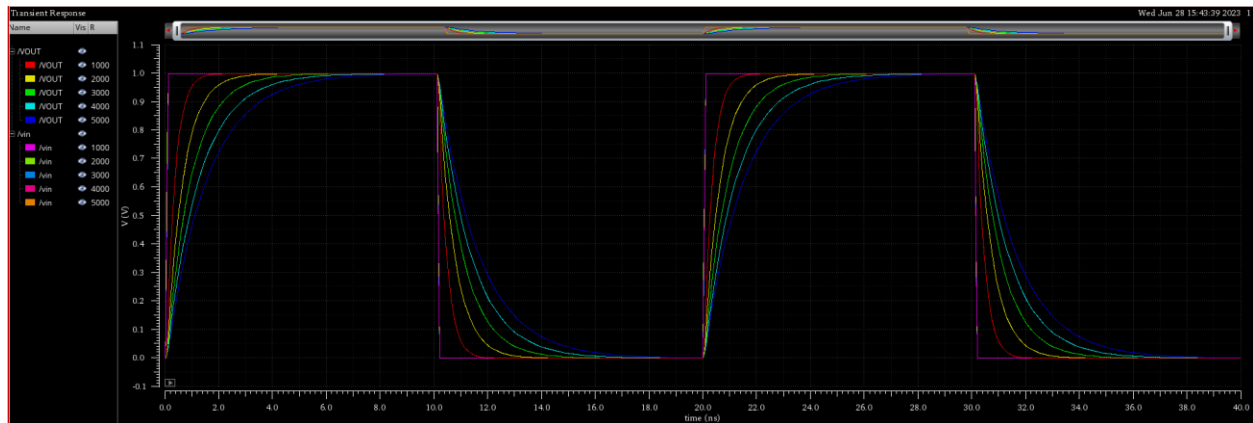
$$0.1 = 1 - e^{-\frac{T_{10\%}}{0.3n}}, T_{10\%} = 31.6p$$

$$T_{rise} = T_{90\%} - T_{10\%} = 690.77 - 31.6 = 659.17p$$

6 DO PARAMETRIC SWEEP FOR $R = 1: 1: 5k\Omega$. REPORT OVERLAID RESULTS. COMMENT ON THE RESULTS.

6.1

Point	Test	Output	Nominal	Spec	Weight	Pass/Fail
Parameters: R=1k						
1	labone:labone:1	Trise	661.8p			
1	labone:labone:1	Tfall	661.8p			
1	labone:labone:1	/vin				
1	labone:labone:1	/VOUT				
Parameters: R=2k						
2	labone:labone:1	Trise	1.312n			
2	labone:labone:1	Tfall	1.312n			
2	labone:labone:1	/vin				
2	labone:labone:1	/VOUT				
Parameters: R=3k						
3	labone:labone:1	Trise	1.974n			
3	labone:labone:1	Tfall	1.974n			
3	labone:labone:1	/vin				
3	labone:labone:1	/VOUT				
Parameters: R=4k						
4	labone:labone:1	Trise	2.632n			
4	labone:labone:1	Tfall	2.632n			
4	labone:labone:1	/vin				
4	labone:labone:1	/VOUT				
Parameters: R=5k						
5	labone:labone:1	Trise	3.294n			
5	labone:labone:1	Tfall	3.294n			
5	labone:labone:1	/vin				
5	labone:labone:1	/VOUT				



6.2 COMMENT :

As R increases the time constant of the filter increases $\tau = RC$.

AC ANALYSIS

Choosing Analyses -- ADE L (17) x

Analysis

<input type="radio"/> tran	<input type="radio"/> dc	<input checked="" type="radio"/> ac	<input type="radio"/> noise
<input type="radio"/> xf	<input type="radio"/> sens	<input type="radio"/> dcmatch	<input type="radio"/> stb
<input type="radio"/> pz	<input type="radio"/> sp	<input type="radio"/> envlp	<input type="radio"/> pss
<input type="radio"/> pac	<input type="radio"/> pstb	<input type="radio"/> pnoise	<input type="radio"/> pxf
<input type="radio"/> psp	<input type="radio"/> qpss	<input type="radio"/> qpac	<input type="radio"/> qpnoise
<input type="radio"/> qpxf	<input type="radio"/> qqsp	<input type="radio"/> hb	<input type="radio"/> hbac
<input type="radio"/> hbnoise	<input type="radio"/> hbsp		

AC Analysis

Sweep Variable

☒ Frequency

☐ Design Variable

☐ Temperature

☐ Component Parameter

☐ Model Parameter

☐ None

Sweep Range

☒ Start-Stop

Start Stop

☐ Center-Span

Sweep Type

▼

☒ Points Per Decade

☐ Number of Steps

Add Specific Points ☐

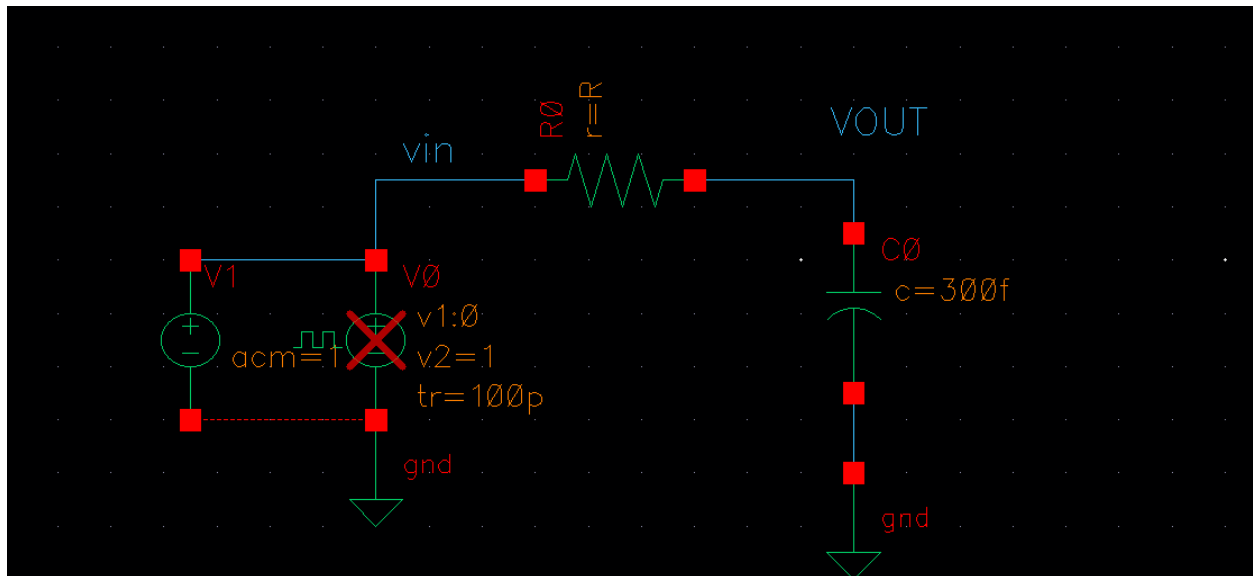
Specialized Analyses

▼

Enabled ☒

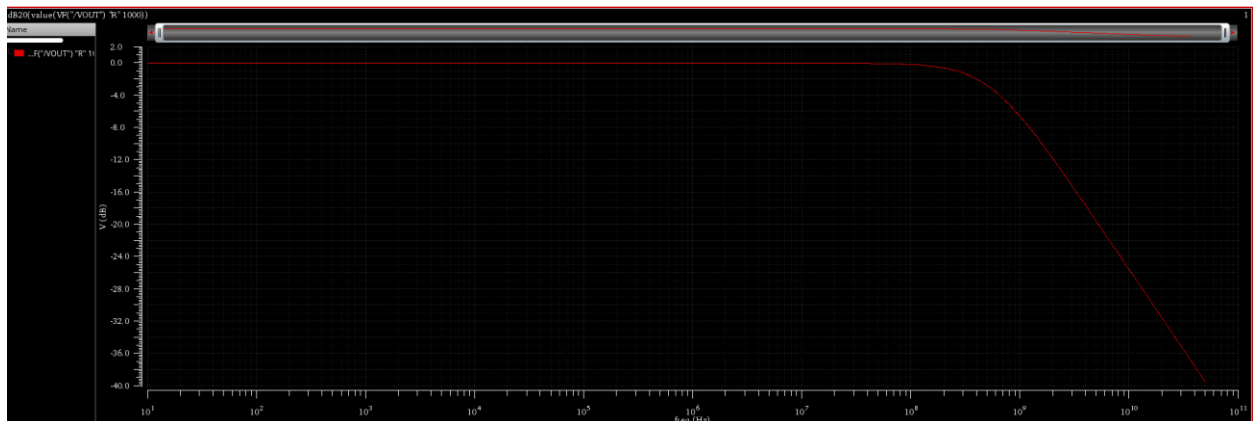
Options...

OK Cancel Defaults Apply Help

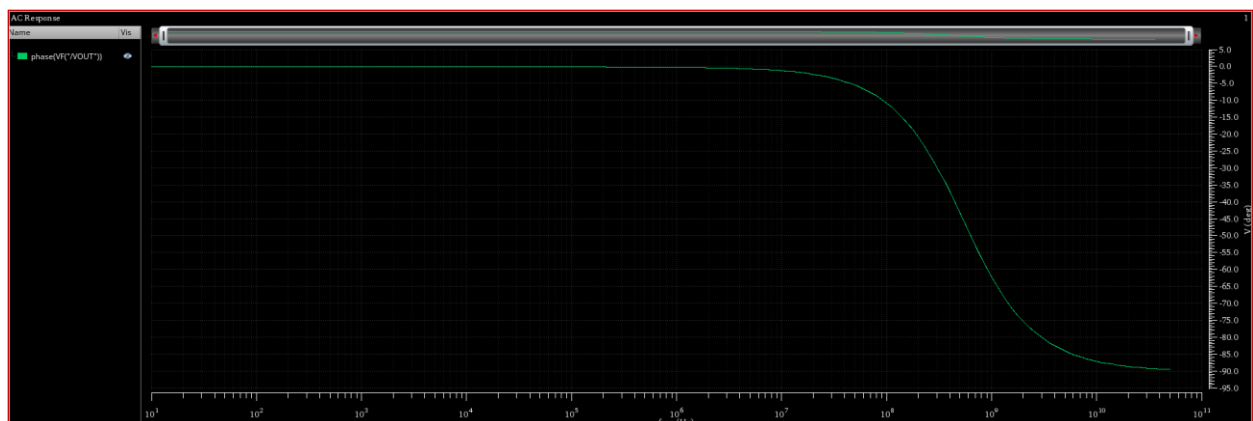


1 REPORT BODE PLOT (MAGNITUDE AND PHASE) FOR THE PREVIOUS LPF.

1.1 MANGNITUDE



1.2 PHASE



2 CALCULATE DC GAIN AND 3dB BANDWIDTH USING CADENCE CALCULATOR EXPRESSIONS. EXPORT THE EXPRESSIONS TO ADEXL.

Test	Output	Nominal	Spec	Weight	Pass/Fail
labone:labone:1	/VOUT				
labone:labone:1	/vin				
labone:labone:1	Trise	eval err			
labone:labone:1	Tfall	eval err			
labone:labone:1	BW	529.7M			
labone:labone:1	Ymax	1			

there is no transience, so Trise & Tfall give errors.

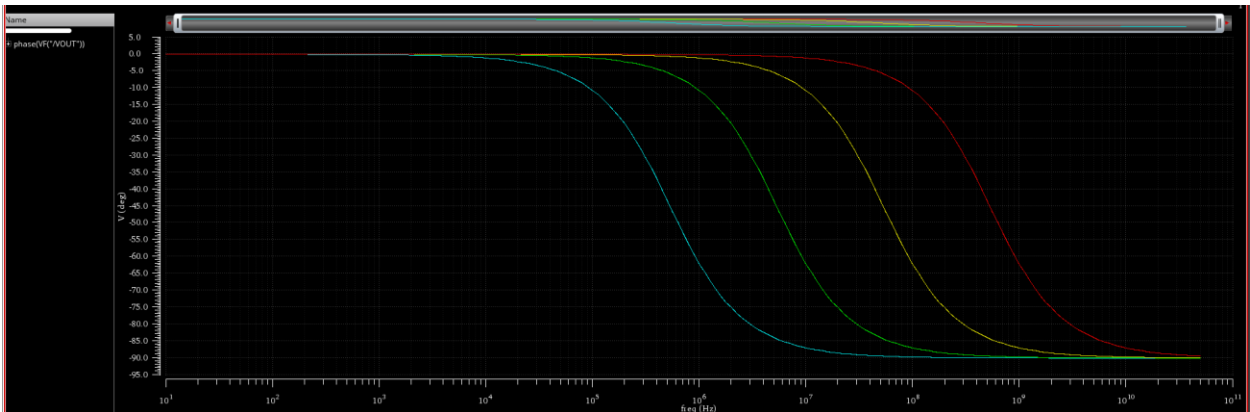
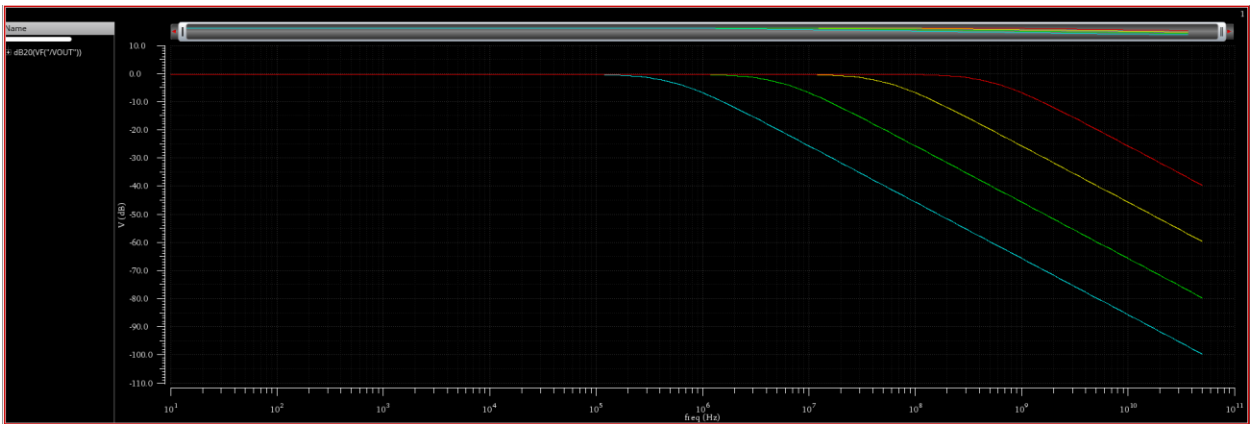
3 COMPARE SIMULATION WITH ANALYTICAL RESULTS IN A TABLE

	SIM	Analytical
BW	529.7M	$\frac{1}{0.3 * 10^{-9} * 2\pi} = 530.5M$
DCgain	1	1

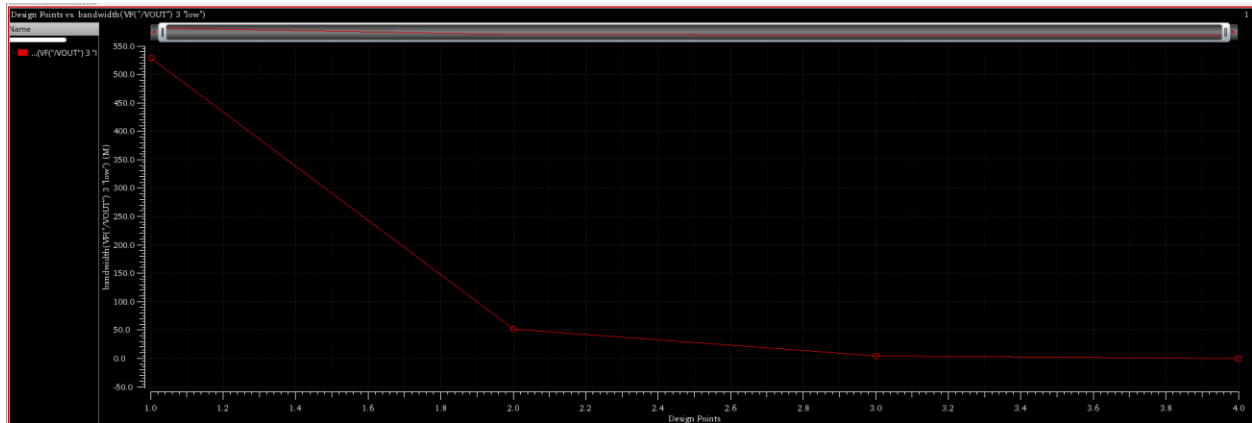
4 DO PARAMETRIC SWEEP FOR $R = 1, 10, 100, 1000k\Omega$. REPORT OVERLAID RESULTS. COMMENT ON THE RESULTS

4.1

Point	Test	Output	Nominal	Spec	Weight	Pass/Fail
Parameters: R=1k						
1	labone:labone:1	/VOUT				
1	labone:labone:1	/vin				
1	labone:labone:1	Trise	eval err			
1	labone:labone:1	Tfall	eval err			
1	labone:labone:1	BW	529.7M			
1	labone:labone:1	Ymax	1			
Parameters: R=10k						
2	labone:labone:1	/VOUT				
2	labone:labone:1	/vin				
2	labone:labone:1	Trise	eval err			
2	labone:labone:1	Tfall	eval err			
2	labone:labone:1	BW	52.97M			
2	labone:labone:1	Ymax	1			
Parameters: R=100k						
3	labone:labone:1	/VOUT				
3	labone:labone:1	/vin				
3	labone:labone:1	Trise	eval err			
3	labone:labone:1	Tfall	eval err			
3	labone:labone:1	BW	5.297M			
3	labone:labone:1	Ymax	1			
Parameters: R=1M						
4	labone:labone:1	/VOUT				
4	labone:labone:1	/vin				
4	labone:labone:1	Trise	eval err			
4	labone:labone:1	Tfall	eval err			
4	labone:labone:1	BW	529.7k			
4	labone:labone:1	Ymax	1			



4.2 COMMENT



As R increases Bandwidth decreases as $BW = \frac{1}{2\pi RC}$ but gain is the same.

[OPTIONAL] POLE ZERO ANALYSIS

1 REPORT POLE ZERO ANALYSIS RESULTS.

Choosing Analyses -- ADE L (1)

Analysis

<input type="radio"/> tran	<input type="radio"/> dc	<input type="radio"/> ac	<input type="radio"/> noise
<input type="radio"/> xf	<input type="radio"/> sens	<input type="radio"/> dcmatch	<input type="radio"/> stb
<input checked="" type="radio"/> pz	<input type="radio"/> sp	<input type="radio"/> envlp	<input type="radio"/> pss
<input type="radio"/> pac	<input type="radio"/> pstb	<input type="radio"/> pnoise	<input type="radio"/> pxf
<input type="radio"/> psp	<input type="radio"/> qpss	<input type="radio"/> qpac	<input type="radio"/> qpnoise
<input type="radio"/> qpxf	<input type="radio"/> qpsp	<input type="radio"/> hb	<input type="radio"/> hbac
<input type="radio"/> hbnoise	<input type="radio"/> hbsp		

Pole-Zero Analysis

Output

voltage

Positive Output Node /VOUT

Negative Output Node /gnd!

Input Source

voltage

Input Voltage Source /V0

Sweep Variable

Frequency ☐

Design Variable ☐

Temperature ☐

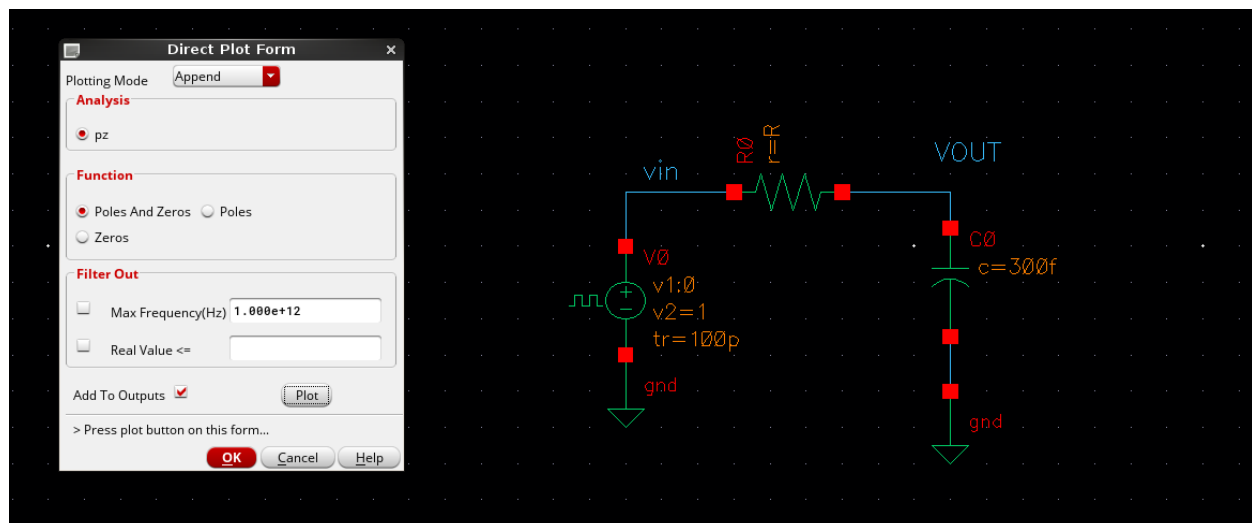
Component Parameter ☐

Model Parameter ☐

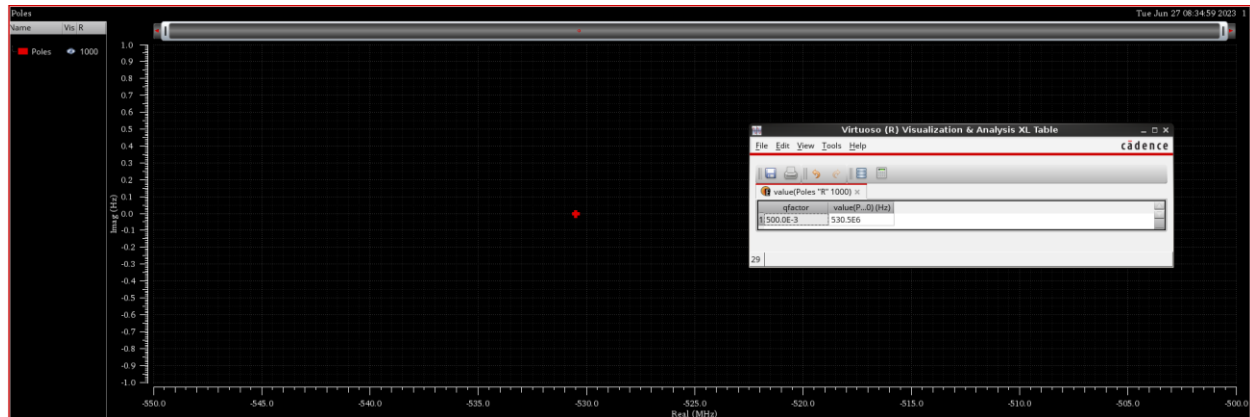
Component Eval Freq (Hz) 1

Enabled ☒

Options...



labone:labone:1	Poles				
labone:labone:1	Zeros	eval err			



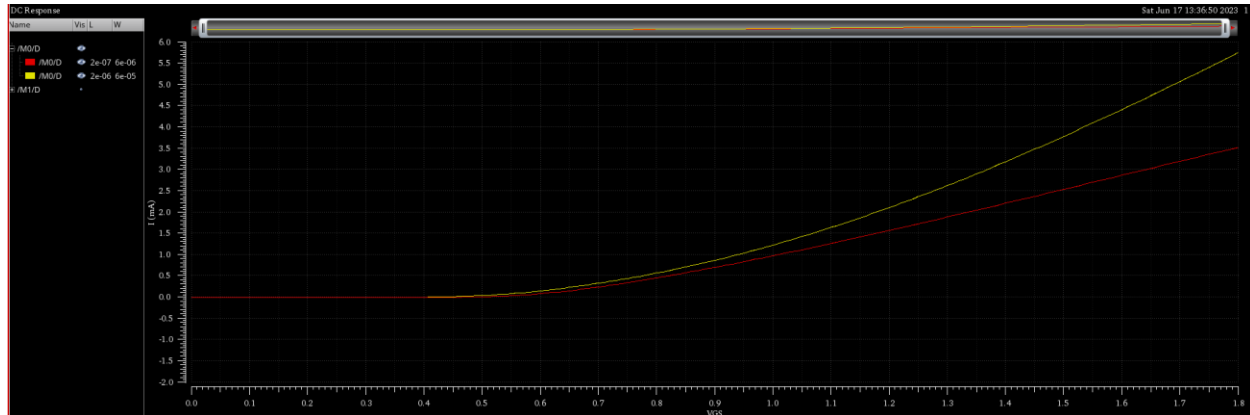
The pole frequency is 530.5M but the bandwidth calculated from AC analysis is 529.7m, so the pole frequency is more accurate and closer to the analytical solution which is also 530.5M.

PART 2: MOSFET Characteristics

1 ID vs VGS

1.1 PLOT $I_D - V_{GS}$ CHARACTERISTICS FOR NMOS AND PMOS DEVICES.

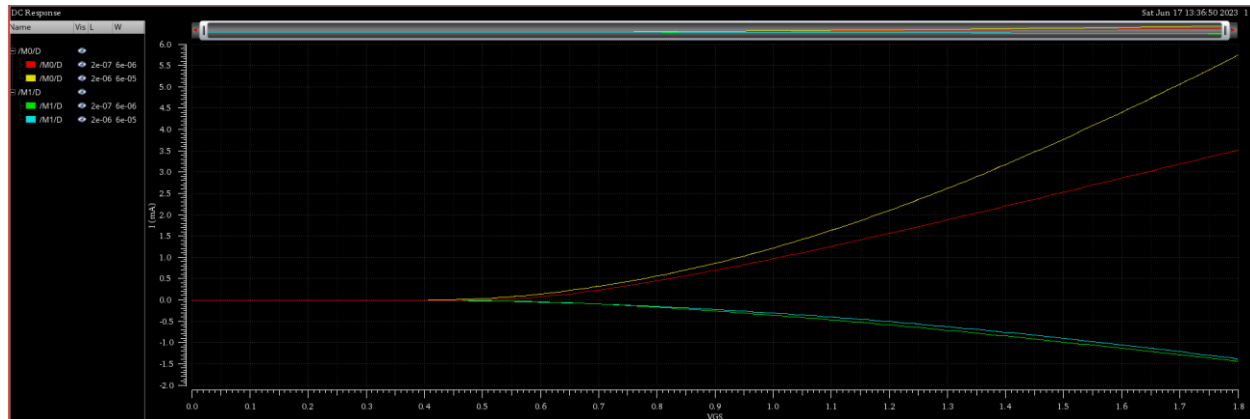
1.1.1 NMOS Small channel & Long channel



1.1.2 PMOS Small channel & Long channel



1.1.3 NMOS with Pmos



1.2 COMMENT ON THE DIFFERENCES BETWEEN SHORT CHANNEL AND LONG CHANNEL RESULTS.

1.2.1 Which one has higher current? Why?

Long channel has higher current, because of mobility degradation effect (increase of vertical electric field) & velocity saturation effect (increase of lateral electric field) when decreasing the channel length the electric field becomes very large to become critical electric field, so the velocity also saturate with specific Vds_{sat} even before the Pinch off saturation (Vov), so it affects the small channel current to become lower as it saturates quicker.

1.2.2 Is the relation linear or quadratic? Why?

In long channel it is quadratic (square law is applicable), but in short channel it is more like a linear after the effect happens with Vds_{sat} ($VGS = VDS_{sat} + Vth$) because of velocity saturation effect, as VDS is saturated so the relation between ID and Vov ($VGS - VTH$) becomes linear.

1.3 COMMENT ON THE DIFFERENCES BETWEEN NMOS AND PMOS :

1.3.1 Which one has higher current? Why?

NMOS has higher current because electrons has higher mobility than holes.

1.3.2 What is the ratio between NMOS and PMOS currents at $VGS = VDD$?

From calculator :

	NMOS	PMOS	RATIO
SHORT CHANNEL	3.532E-3	1.4269E-3	2.475
LONG CHANNEL	5.755E-3	1.373E-3	4.19

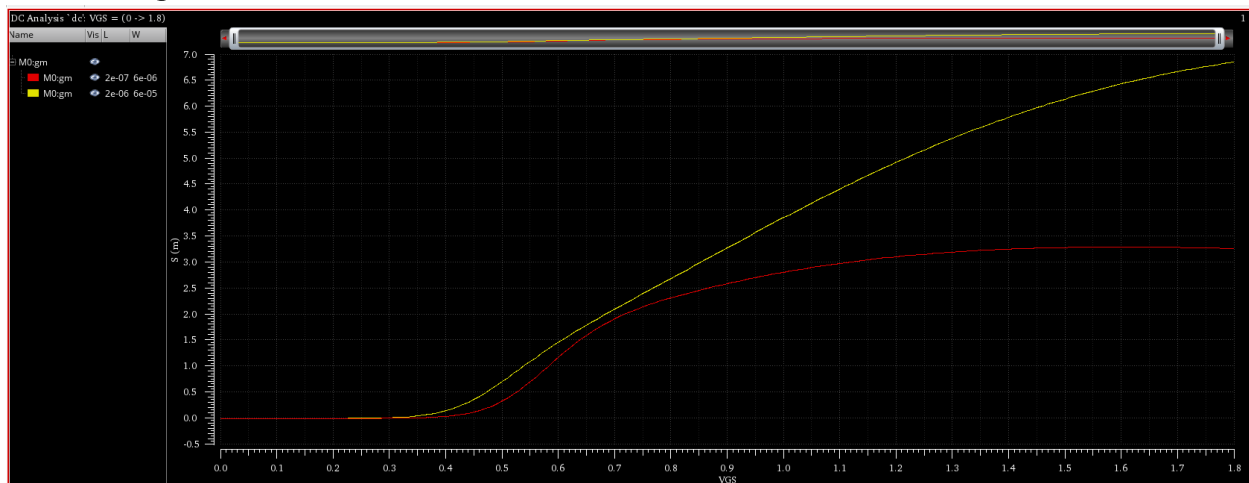
1.3.3 Which one is more affected by short channel effects?

NMOS is more affected because it has higher mobility than PMOS so it reaches velocity saturation faster,

$$\epsilon = \frac{v_{sat}}{\mu} = \frac{V_{sat}}{L}, V_{sat} = \frac{L \cdot v_{sat}}{\mu}.$$

2 gm vs V_{GS}

2.1 PLOT gm vs V_{GS} FOR NMOS DEVICE.



2.2 COMMENT ON THE DIFFERENCES BETWEEN SHORT CHANNEL AND LONG CHANNEL RESULTS.

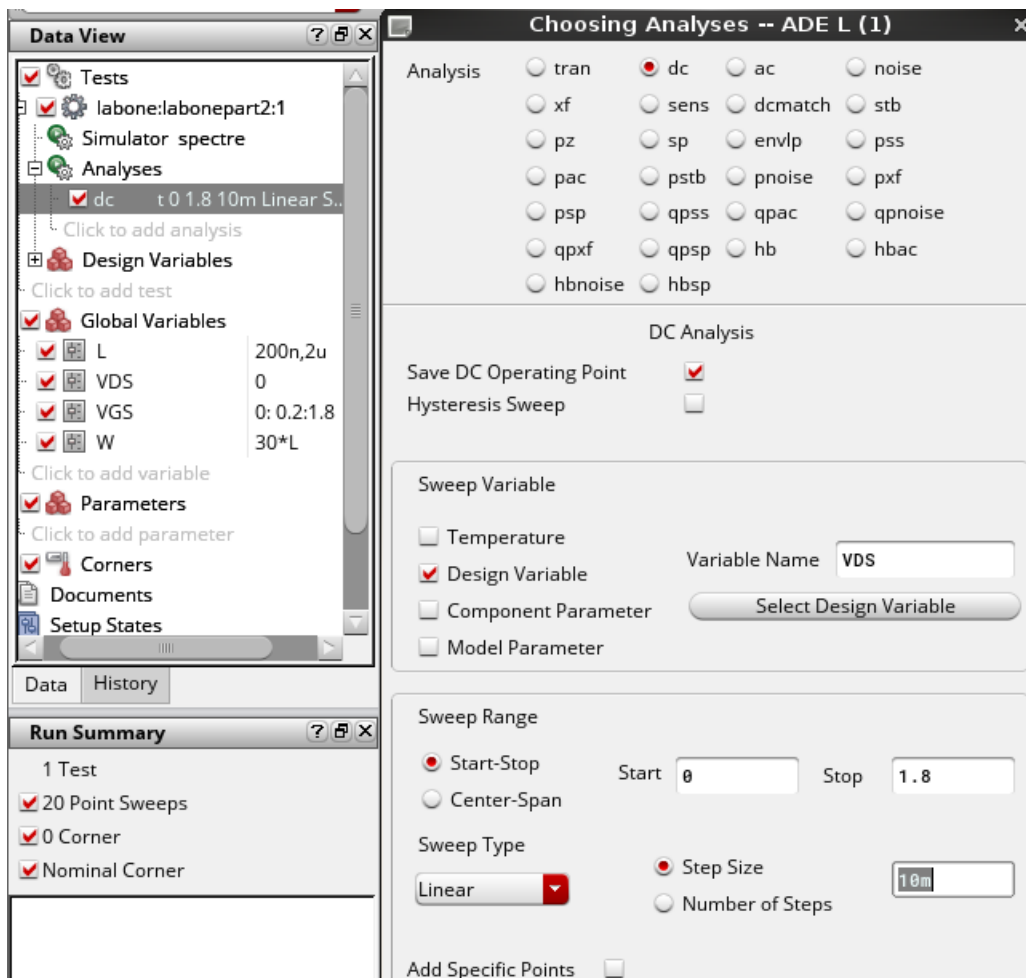
2.2.1 Does gm increase linearly? Why?

not ideally linear, but It kind of increasing linearly because the characteristic of I_d vs v_{gs} is quadratic and gm is the slope of it .

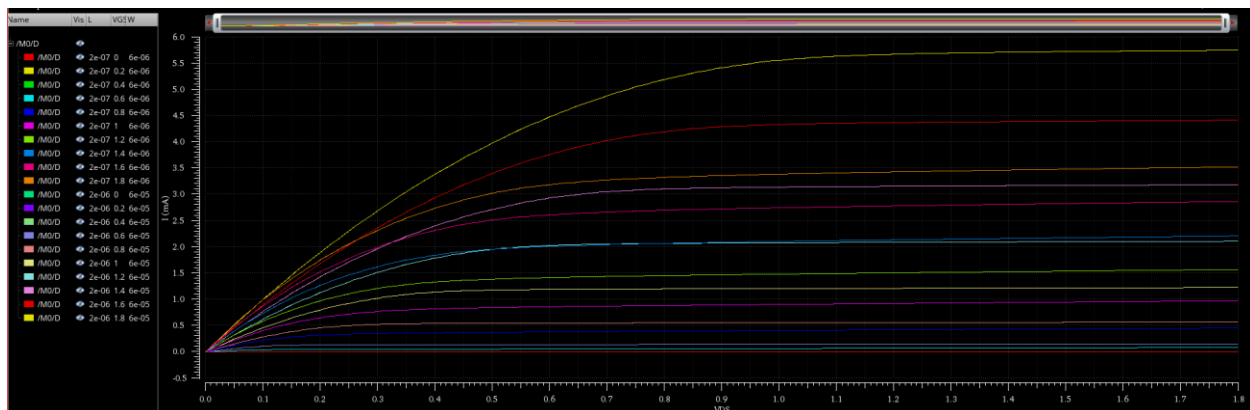
2.2.2 Does gm saturate? Why?

It saturates in short channel because the (i_d vs V_{gs}) becomes liner because of velocity saturation, mobility degradation and gm is the slope of this linear characteristic (gm is constant) but it doesn't saturate in large channel model.

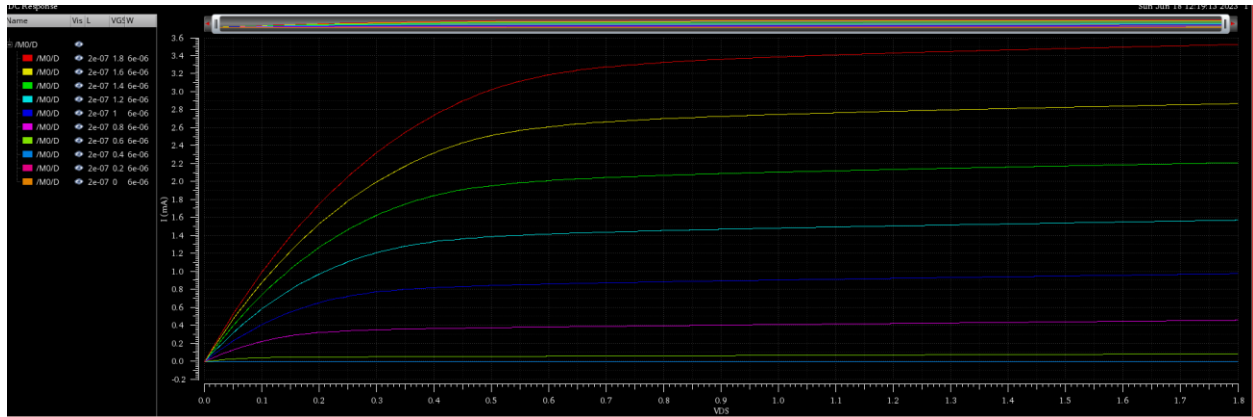
3 ID vs VDS



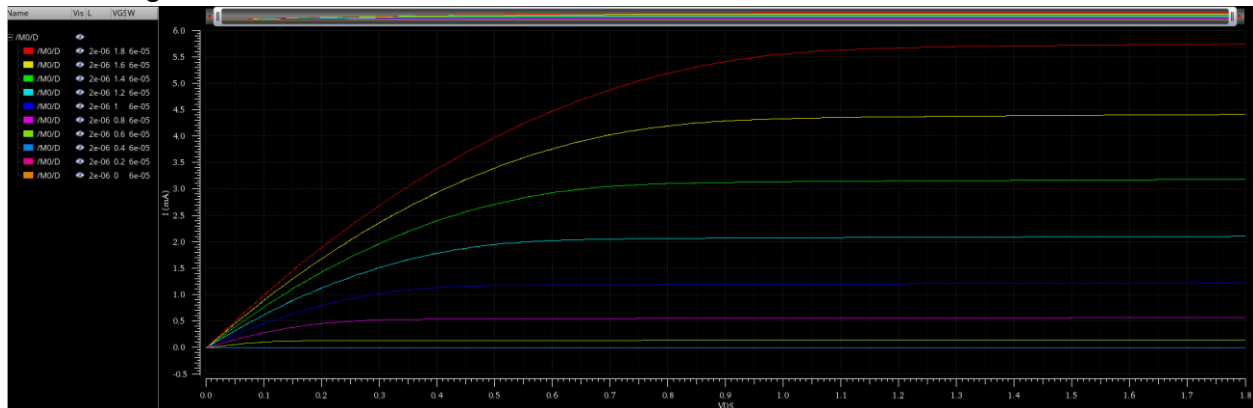
3.1 PLOT $I_D - V_{DS}$ CHARACTERISTICS FOR NMOS DEVICE



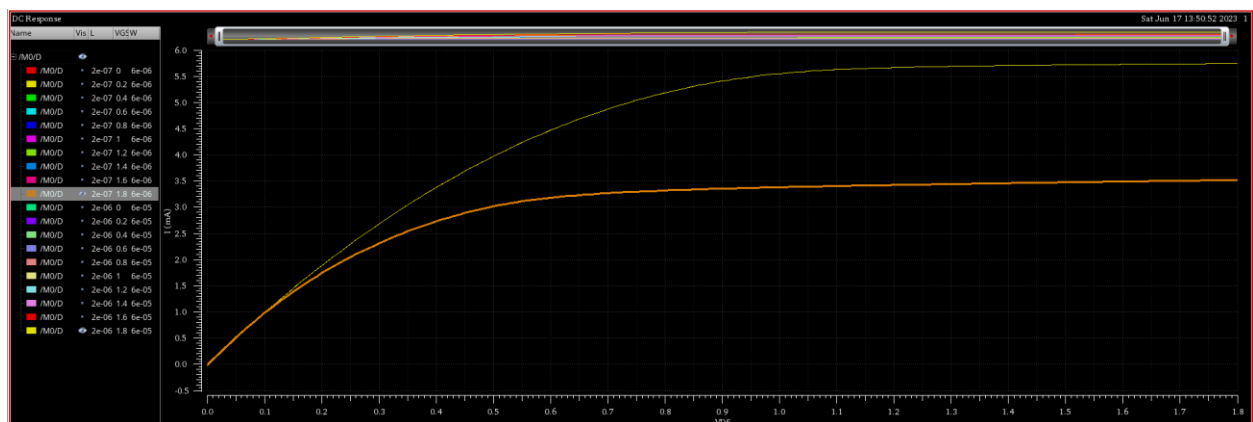
3.1.1 Short channel device:



3.1.2 Long channel device:



3.2 COMMENT ON THE DIFFERENCES BETWEEN SHORT CHANNEL AND LONG CHANNEL RESULTS.



Small channel current vs long channel current under same VGS

3.2.1 Which one has higher current? Why?

Long channel has higher current, because of velocity saturation (increase lateral electric field) and mobility degradation (increase vertical electric field which leads to decrease mobility) affect small channel current to become less & also I_D increases with V_{GS} in short and long channel, so Long channel with highest V_{GS} has the highest current.

3.2.2 Which one has higher slope in the saturation region? Why?

small channel current has higher slope because of DIBL & channel length modulation effect as it has higher λ because when L is small the pinch off point shift effect when increasing V_{ds} more than V_{ov} is higher than long channel case as L_{eff} clearly becomes less in short channel case, so it has smaller r_0 .