FROM LAST TIME...

Solution of LTI State Equations

- State Transition Matrix
- Free Response
- Forced Response

$$x(t) = e^{At}x(0)$$

$$e^{At} = Te^{\lambda t}T^{-1} = T\begin{bmatrix} e^{\lambda_1 t} & 0 \\ & \ddots & \\ 0 & e^{\lambda_n t} \end{bmatrix}T^{-1}$$
 $e^{At} = \mathcal{L}^{-1}[(sI - A)^{-1}]$

$$x(t) = \Phi(t)x(0) + \int_0^t \Phi(t - \tau)Bu(\tau)d\tau$$

CONTROLLABILITY

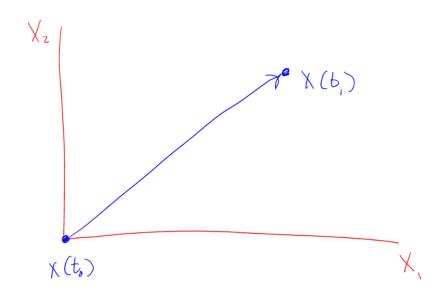
Topics

- Definition of Controllability
- Controllable Canonical Form
- Controllable Canonical Decomposition
- Stabilizability

At the end of this section, students should be able to:

- Determine controllability of a system.
- Transform a system into controllable canonical form.
- Decompose a system into controllable and uncontrollable subsystems.
- Determine if an uncontrollable system is stabilizable.

WHAT DOES IT MEAN FOR A SYSTEM TO BE CONTROLLABLE? Stotes - X, X



The system $\dot{x} = Ax + Bu$ is completely **state controllable** if a control u(t) exists that will transfer the state of the system from any $x(t_0) = x_0$ to any $x(t_1) = x_1$ in a finite time $t_1 - t_0$.

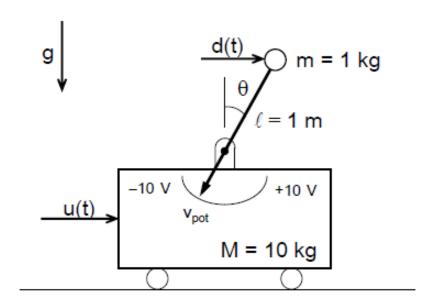
RECALL THE SEGWAY CASE STUDY

- Rider leans forward to start driving forward
- Rider leans backward to start driving backward
- Wheels move to prevent the rider from falling over (maintain stability)





MODEL THE (UNREALISTIC) SEGWAY AND RIDER AS AN INVERTED PENDULUM ON A CART



Is it possible to achieve zero position of both the cart and the rod with only a single control input u?

LET'S TRY A SIMPLE EXAMPLE USING NUMBERS

$$\dot{x} = \begin{bmatrix} 0 & 1 \\ -2 & -3 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

Is it completely state controllable? Yes

Transform to diagonal form:

$$\chi = -1_{3} - 2$$

$$T = \begin{bmatrix} 1 & 1 \\ -1 & -2 \end{bmatrix}$$

$$T = \begin{bmatrix} 2 & 1 \\ -1 & -1 \end{bmatrix}$$

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Controllability the combination of A (through T) and B.

DEFINITION OF CONTROLLABILITY

The system

$$\dot{x} = Ax + Bu$$

is completely state controllable iff the column vectors of the controllability matrix

$$W_C = \begin{bmatrix} B & AB & A^2B & \cdots & A^{n-1}B \end{bmatrix}$$

span the n-dimensional space (i.e., W_C has rank n).

$$\omega_{c} \begin{bmatrix} 1 & -1 \\ 1 & -2 \end{bmatrix}$$

FOR A CONTROLLABLE SYSTEM, AN INPUT MUST EXIST TO MOVE FROM

$$(t_0, x(Q))$$
 TO $(t_1, x(t_1))$

Assume u is a scalar:

$$x(t) = e^{At}x(0) + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau$$

Let
$$x(0) = 0$$
, $x(t_1) = x_1$

$$x(t_1) = x_1 = \int_0^{t_1} e^{A(t_1 - \tau)} Bu(\tau) d\tau$$

APPLYING THE SERIES REPRESENTATION OF THE STATE TRANSITION MATRIX...

$$e^{A(t_1-\tau)} = I + A(t_1-\tau) + \frac{1}{2!}A^2(t_1-\tau)^2 + \cdots$$

$$x_{1} = B \int_{0}^{t_{1}} u(\tau)d\tau + AB \int_{0}^{t_{1}} (t_{1} - \tau)u(\tau)d\tau + A^{2}B \int_{0}^{t_{1}} \frac{1}{2} (t_{1} - \tau)^{2}u(\tau)d\tau + \cdots$$

- This is an infinite series, it would be better to have a finite series
- We need the Cayley-Hamilton Theorem

CAYLEY-HAMILTON THEOREM: MATRIX A SATISFIES ITS OWN CHARACTERISTIC EQUATION

Some wothing
$$|\lambda I - A| = \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0 = 0$$

$$A^n + a_{n-1}A^{n-1} + \dots + a_1A + a_0 = 0$$

- From this, we can show that any function of A can be written as a sum of n-1 powers of A
- In particular, $e^{\mathbf{A}(t_1-\tau)} = \sum_{i=0}^{n-1} \alpha_i(t)A^i$

APPLYING CAYLEY-HAMILTON THEOREM THIS PROBLEM...

$$\mathbf{e}^{\mathbf{A}(t_1-\tau)} = \sum_{i=0}^{n-1} \alpha_i(t) A^i \qquad \mathbf{x}_1 = \mathbf{B} \int_0^{t_1} u(\tau) d\tau + \mathbf{A} \mathbf{B} \int_0^{t_1} (t_1-\tau) u(\tau) d\tau + \mathbf{A} \mathbf{B} \int_0^{t_1} \frac{1}{2} (t_1-\tau)^2 u(\tau) d\tau + \cdots$$

$$\chi_{i} = \sum_{i=0}^{h-1} A^{i} B \int_{0}^{b} \alpha_{i} \left(\frac{\omega}{L} \right) U \left(\frac{\omega}{L} \right) d \frac{\omega}{L}$$

$$\begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} = \begin{bmatrix} 3 & 0 \\ 1 & 2 \end{bmatrix} \begin{bmatrix} 3 & 0 \\ 3 & 0 \end{bmatrix}$$

$$\begin{bmatrix} 1 & 2 \\ 2 & 0 \end{bmatrix}$$

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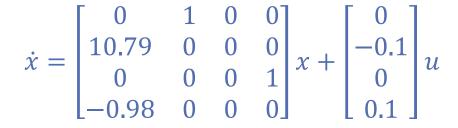
$$\begin{bmatrix} 3 & 0 \\ 3 & 0 \\ 3 & 0 \end{bmatrix}$$

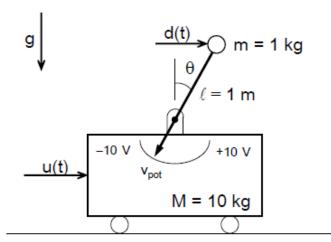
$$\begin{bmatrix} 3 & 0 \\ 3 & 0 \\ 3 & 0 \end{bmatrix}$$

• For this to work, $\begin{bmatrix} B & AB & A^2B & \cdots & A^{n-1}B \end{bmatrix}$ must be full rank!

RETURN TO THE INVERTED PENDULUM EXAMPLE

$$x = [\theta \quad \dot{\theta} \quad x \quad \dot{x}]^T$$





CHECK CONTROLLABILITY FOR INVERTED PENDULUM

$$W_C = \begin{bmatrix} B & AB & A^2B & A^3B \end{bmatrix}$$

$$W_C = \begin{bmatrix} 0 & -0.1 & 0 & -1.08 \\ -0.1 & 0 & -1.08 & 0 \\ 0 & 0.1 & 0 & 0.1 \\ 0.1 & 0 & 0.1 & 0 \end{bmatrix}$$
we eds to be
$$\begin{array}{c} v \in \partial S \\ S \in \partial S \\$$

For some X, & X.

U can be dosigned les it is state controllable

CONSIDER ANOTHER NUMERICAL EXAMPLE

$$\dot{x} = \begin{bmatrix}
-2 & 1 & 0 \\
0 & -2 & 0 \\
0 & 0 & a
\end{bmatrix} x + \begin{bmatrix}
0 \\
b
\end{bmatrix} u$$
This is in Jordan Canonical Form!

$$AB = \begin{bmatrix}
-2 & b \\
-2b \\
a
\end{bmatrix}$$

$$A^{7}B = \begin{bmatrix}
-4 & b \\
-4b \\
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$$A^{7}B = \begin{bmatrix}$$

WE CAN ALSO CONSIDER OUTPUT CONTROLLABILITY

The system

$$\dot{x} = Ax + Bu$$
$$y = Cx + Du$$

is output controllable if a control u exists that will transfer the output of the system from any $y(t_0)$ to any $y(t_1)$ in a finite time $t_1 - t_0$.

This is true iff

$$\operatorname{rank}[CB \quad CAB \quad CA^2B \quad \cdots \quad CA^{n-1}B \quad D] = m$$

It
$$D=0 \rightarrow conk [C[\omega_c]] = m$$

CONSIDER A 3RD ORDER CONTROLLABLE CANONICAL FORM

A CONTROLLABLE SYSTEM CAN ALWAYS BE TRANSFORMED INTO CONTROLLABLE CANONICAL FORM

$$\dot{x} = Ax + Bu \qquad \stackrel{x=Tz}{\Longrightarrow} \qquad \dot{z} = A_C z + B_C u$$

$$A_C = T^{-1} A T$$

$$B_C = T^{-1} B$$

$$W_C = \begin{bmatrix} B & AB & A^2B & \cdots & A^{n-1}B \end{bmatrix}$$

$$W_C^C = [B_C \quad A_C B_C \quad A_C^2 B_C \quad \cdots \quad A_C^{n-1} B_C]$$

$$= \left[\nabla^{-1} \mathcal{B} \quad \nabla^{-1} \mathcal{A} \nabla \nabla^{-1} \mathcal{B} \right]$$

related by

$$\nabla \omega_{i} = \lambda_{i}$$

$$\nabla - \omega_{i} \left[\lambda_{i} \right]^{-1}$$

$$\left[\lambda_{i} \right]^{-1} = \begin{bmatrix} \alpha_{i} & \alpha_{z} & 1 \\ \alpha_{z} & 1 & 6 \\ 1 & 0 & 6 \end{bmatrix}$$

Transforms original system to CCF

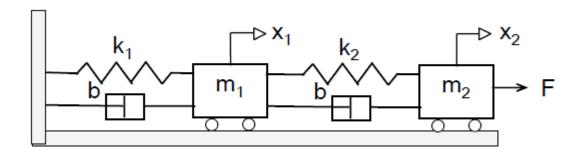
$$T = \begin{bmatrix} B & AB & A^2B \end{bmatrix} \begin{bmatrix} a_1 & a_2 & 1 \\ a_2 & 1 & 0 \\ 1 & 0 & 0 \end{bmatrix}$$

IN GENERAL, TO TRANSFORM TO CONTROLLABLE CANONICAL FORM:

$$T = W_C \begin{bmatrix} a_1 & a_2 & a_3 & \cdots & a_{n-1} & 1 \\ a_2 & a_3 & a_4 & \cdots & 1 & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ a_{n-2} & a_{n-1} & 1 & \cdots & 0 & 0 \\ a_{n-1} & 1 & 0 & \cdots & 0 & 0 \\ 1 & 0 & 0 & \cdots & 0 & 0 \end{bmatrix}$$

$$|\lambda I - A| = \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0$$

WHAT DOES AN UNCONTROLLABLE SYSTEM LOOK LIKE?



$$\sum F_{m_1} = -k_1 x_1 - b \dot{x}_1 + k_2 (x_2 - x_1) + b (\dot{x}_2 - \dot{x}_1) = m_1 \ddot{x}_1$$

$$\sum F_{m_2} = -k_2 (x_2 - x_1) - b (\dot{x}_2 - \dot{x}_1) + F = m_2 \ddot{x}_2$$

Assume that the block masses are negligible -> M, = Mz = 0 but a some file.

$$k_2(x_2 - x_1) + b(\dot{x}_2 - \dot{x}_1) = k_1x_1 + b\dot{x}_1 = F$$

CHECK CONTROLLAB

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} \frac{-k_1}{b} & 0 \\ -k_1 + k_2 & -k_2 \\ b \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \frac{1}{b} \\ \frac{2}{b} \end{bmatrix} F$$

$$x_2 = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

$$\omega_{c} = \begin{bmatrix} \frac{1}{b} & -\frac{k_{1}}{b^{2}} \\ \frac{2}{b} & -\frac{k_{1}}{b^{2}} - \frac{k_{2}}{b^{2}} \end{bmatrix}$$

$$controloble$$

$$controloble$$

b cout be 0 but in this system that cost really hoppen

CONVERT SYSTEM TO TRANSFER FUNCTION

$$X_2(s) = C(sI - A)^{-1}BU(s)$$

$$(sI - A) = \begin{bmatrix} s + \frac{k_1}{b} & 0 \\ -\frac{k_1 + k_2}{b} & s + \frac{k_2}{b} \end{bmatrix}$$

$$(sI - A)^{-1} = \begin{bmatrix} s + \frac{k_2}{b} & 0 \\ -\frac{k_1}{b} & s + \frac{k_2}{b} \end{bmatrix}$$

$$(sI - A)^{-1} = \begin{bmatrix} s + \frac{k_2}{b} & s + \frac{k_2}{b} \\ -\frac{k_1}{b} & s + \frac{k_2}{b} \end{bmatrix}$$

$$(sI - A)^{-1} = \begin{bmatrix} s + \frac{k_2}{b} & s + \frac{k_2}{b} \\ -\frac{k_1}{b} & s + \frac{k_2}{b} \end{bmatrix}$$

$$(sI - A)^{-1} = \begin{bmatrix} s + \frac{k_2}{b} & s + \frac{k_2}{b} \\ -\frac{k_1}{b} & s + \frac{k_2}{b} \end{bmatrix}$$

CONVERT SYSTEM TO TRANSFER FUNCTION

$$X_2(s) = C(sI - A)^{-1}BU(s)$$

$$\frac{X_2(s)}{U(s)} = \begin{bmatrix} 0 & 1 \end{bmatrix} \frac{\begin{bmatrix} s + \frac{k_2}{b} & 0 \\ -k_1 + k_2 & s + \frac{k_1}{b} \end{bmatrix}}{(s + \frac{k_1}{b})(s + \frac{k_2}{b})} \begin{bmatrix} \frac{1}{b} \\ \frac{2}{b} \end{bmatrix}$$

$$= \frac{2s + \frac{k_1 + k_2}{b}}{\left(s + \frac{k_1}{b}\right)\left(s + \frac{k_2}{b}\right)}$$

$$= \frac{1}{b} \frac{2s + \frac{k_1}{b}}{\left(s + \frac{k_2}{b}\right)}$$

$$= \frac{2}{b} \frac{1}{\left(s + \frac{k_2}{b}\right)}$$

$$= \frac{2}{b} \frac{1}{\left(s + \frac{k_2}{b}\right)}$$

$$If |z| = \frac{2}{b} \frac{1}{\left(s + \frac{k}{b}\right)}$$
then

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MEANINGS OF UNCONTROLLABLE STATES

- 1. If state space system is realized from input-output data, then the realization is redundant. Some states have no relationship to either the input or the output.
- 2. If states are meaningful (physical) variables that need to be controlled, then the design of the actuators are deficient.
- The effect of control is limited. There is also a possibility of instability

WHEN A SYSTEM IS NOT CONTROLLABLE, IT CAN BE PARTITIONED INTO CONTROLLABLE AND UNCONTROLLABLE PARTS.

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Given

$$\dot{x} = Ax + Bu$$

$$W_C = \begin{bmatrix} B & AB & A^2B & \cdots & A^{n-1}B \end{bmatrix}$$

If $\operatorname{rank}[W_C] = \ell < n$, then W_C has only ℓ linearly independent vectors, and $n - \ell$ states are uncontrollable.

CONTROLLABLE DECOMPOSITION

$$\dot{x} = Ax + Bu \qquad \stackrel{x=Tz}{\Longrightarrow} \qquad \dot{z} = \hat{A}z + \hat{B}u$$

$$\hat{A} = T^{-1}AT$$

$$\hat{B} = T^{-1}B$$

$$T = \begin{bmatrix} T_1 & T_2 \end{bmatrix}$$

$$T_1 = \begin{bmatrix} B & AB & A^2B & \cdots & A^{\ell-1}B \end{bmatrix}$$

$$T_2 = \text{any } n \times (n-\ell) \text{ matrix that makes T nonsingular}$$

$$\left[\frac{\dot{z}_{C}}{\dot{z}_{UC}}\right] = \left[\frac{\hat{A}_{C}}{0} \frac{\hat{A}_{CU}}{\hat{A}_{UC}}\right] \left[\frac{z_{C}}{z_{UC}}\right] + \left[\frac{\hat{B}_{C}}{0}\right] u$$

$$v_{Coutrolloble} = states$$

$$v_{C} = v_{C}$$

CONSIDER ANOTHER NUMERICAL EXAMPLE

$$\dot{x} = \begin{bmatrix} -2 & 1 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -2 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \\ 1 \end{bmatrix} u$$

This is in Jordan Canonical Form!

check it later

STABILIZABILITY

The system

$$\dot{x} = Ax + Bu$$

is **stabilizable** if any uncontrollable states are open-loop stable, i.e., the eigenvalues of the uncontrollable subsystem \hat{A}_{UC} have negative real part.

CONSIDER ANOTHER NUMERICAL EXAMPLE

$$\dot{x} = \begin{bmatrix} -2 & 1 & 0 \\ 0 & -2 & 0 \\ 0 & 0 & -2 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \\ 1 \end{bmatrix} u$$

$$\begin{bmatrix} \dot{z}_1 \\ \dot{z}_2 \\ \dot{z}_3 \end{bmatrix} = \begin{bmatrix} 0 & -4 & 0 \\ 1 & -4 & 0 \\ 0 & 0 & -2 \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \\ z_3 \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} u$$

COMING UP...

State Feedback Controller Design

- State Feedback Regulator
- Ackermann's Formula