

FASURA: A Scheme for Quasi-Static Massive MIMO Unsourced Random Access Channels

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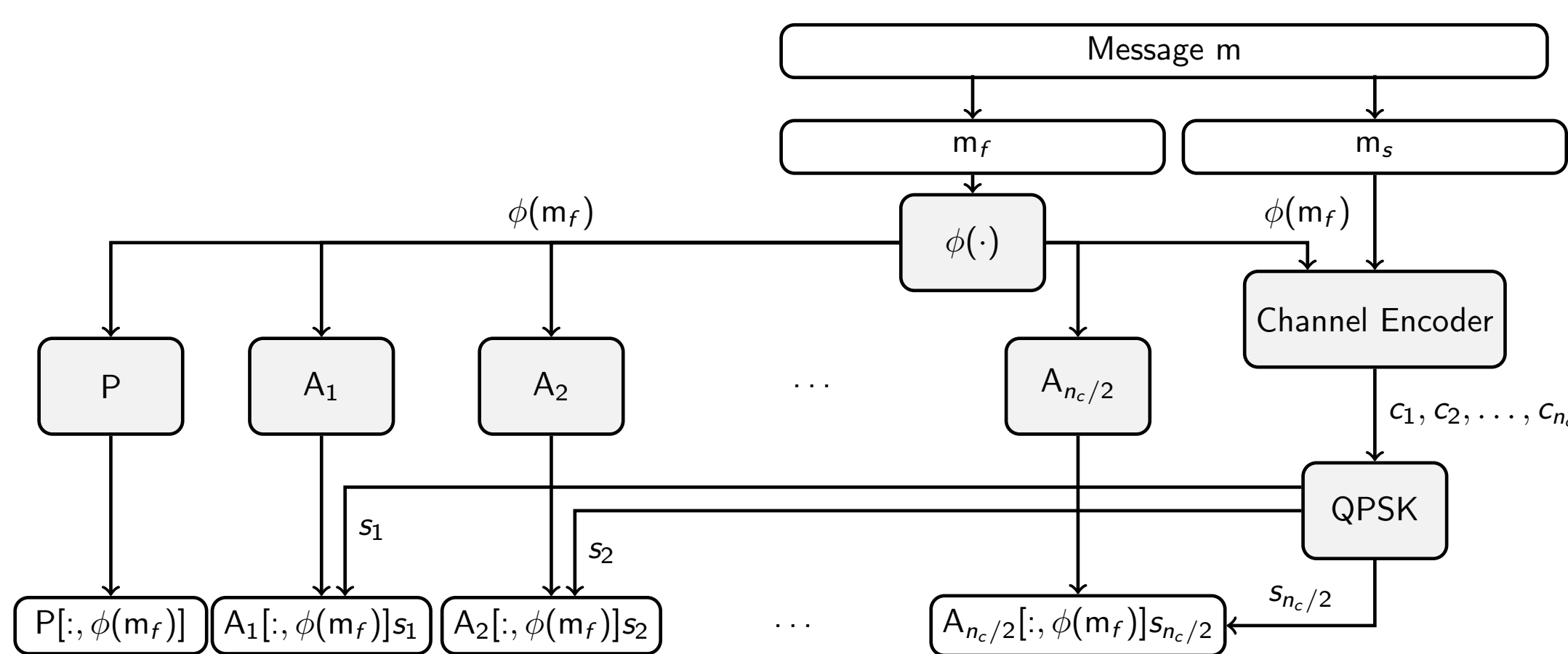
Problem Statement

- Fading Spread Unsourced Random Access (FASURA) is a scheme for unsourced random access (URA) channel
- URA models aim to capture the sporadic transmission of IoT devices
- Consider the massive MIMO URA problem on a quasi-static Rayleigh fading channel
- Probability of error is defined as the sum of the probability of missed detection P_{md} and probability of false alarm P_{fa} , i.e. $P_e = P_{\text{md}} + P_{\text{fa}}$
- Several schemes exist with varying levels of complexity and probability of error
 - Pilot-based: Based on MMV-AMP and polar codes (Fengler et al. 2022)
 - Orthogonal Pilots: Slot based scheme with interesting pilot design (Ahmadi et al. 2022)

Goal

Construct a communication scheme, that minimizes the energy per bit to noise power spectral density ratio E_b/N_0 while also satisfying the constraint $P_e \leq \varepsilon$

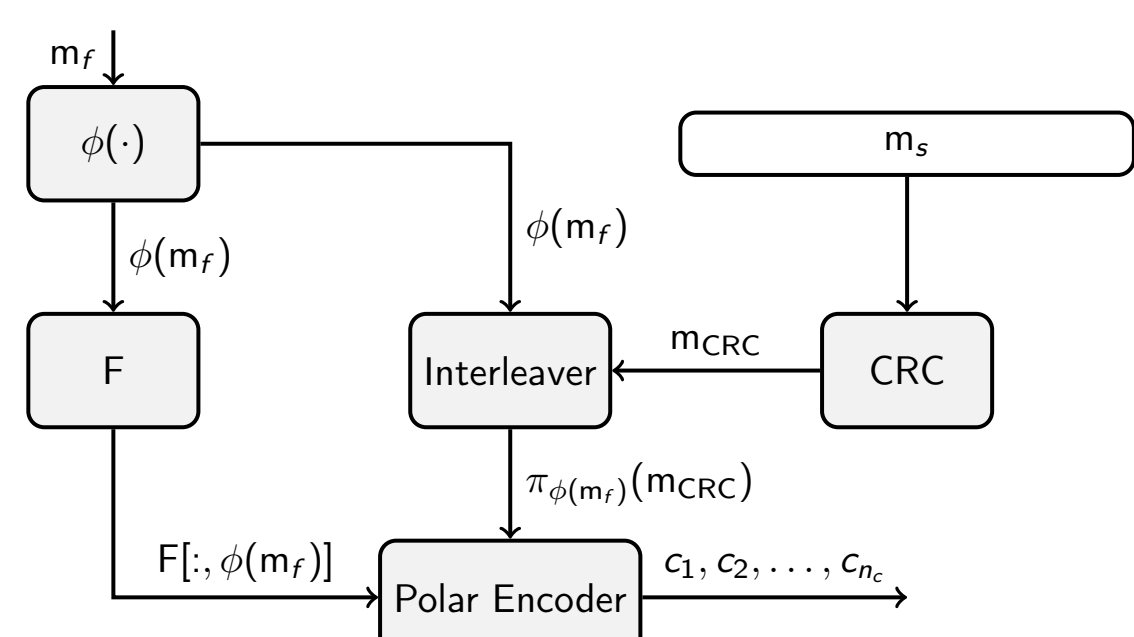
FASURA: Encoder



- Splits the message into two parts
- Computes the decimal representation of \mathbf{m}_f , i.e. $\phi(\mathbf{m}_f)$
- Picks the $\phi(\mathbf{m}_f)$ th row of $\mathbf{P}, \mathbf{A}_1, \dots, \mathbf{A}_{n_c/2}$
- Encodes \mathbf{m}_s , and modulate the coded bits, c_1, c_2, \dots, c_{n_c}
- Spreads the QPSK symbols $s_1, s_2, \dots, s_{n_c/2}$, and form the channel input
- Channel input

$$\mathbf{x} = \begin{bmatrix} \mathbf{P}^\top[:, \phi(\mathbf{m}_f)] & s_1 \mathbf{A}_1^\top[:, \phi(\mathbf{m}_f)] & s_2 \mathbf{A}_2^\top[:, \phi(\mathbf{m}_f)] & \dots & s_{n_c/2} \mathbf{A}_{n_c/2}^\top[:, \phi(\mathbf{m}_f)] \end{bmatrix}^\top = \begin{bmatrix} \mathbf{p}^\top(\mathbf{m}_f) & \mathbf{q}^\top(\mathbf{m}_f, \mathbf{m}_s) \end{bmatrix}^\top$$

Channel Encoder



- Uses the decimal representation of \mathbf{m}_f to select values for the frozen positions and an interleaver
- Encodes the second part of the message, \mathbf{m}_s , using a CRC encoder
- Permutates the CRC codeword using the pre-selected interleaver $\pi_{\phi(\mathbf{m}_f)}(\cdot)$
- Frozen values, $\mathbf{F}[:, \phi(\mathbf{m}_f)]$, and the permute CRC codeword act as an input to the Polar Encoder
- The codeword bits, c_1, c_2, \dots, c_{n_c} , are passed to the QPSK modulator

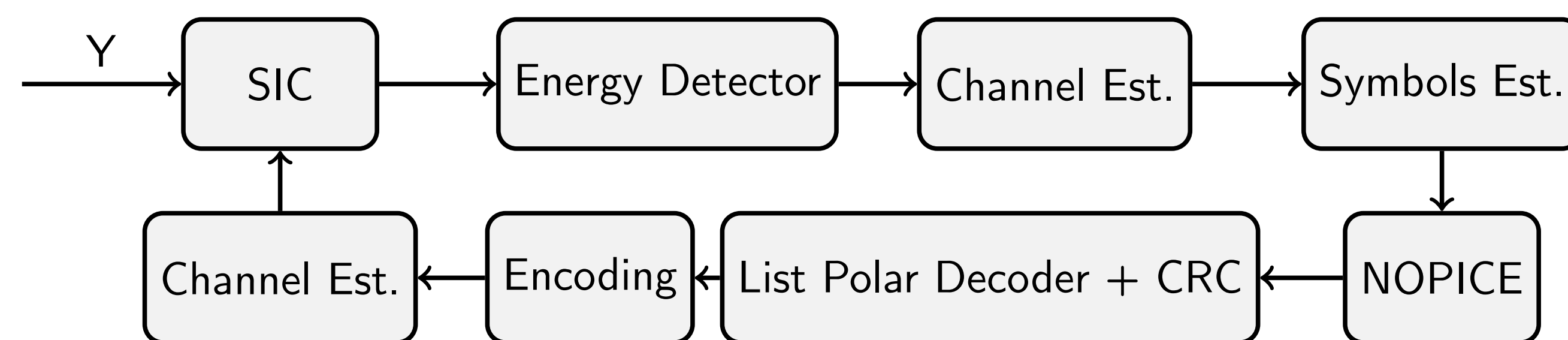
Massive MIMO URA Channel

- Consider a network with K active devices with single antenna
- Access point is equipped with M antennas
- The signal at the receiver is

$$\begin{bmatrix} \mathbf{Y}_p \\ \mathbf{Y}_q \end{bmatrix} = \begin{bmatrix} \mathbf{P}_a \\ \mathbf{Q}_a \end{bmatrix} \mathbf{H} + \begin{bmatrix} \mathbf{Z}_p \\ \mathbf{Z}_q \end{bmatrix}, \quad \mathbf{Y} = \mathbf{X}\mathbf{H} + \mathbf{Z}$$

where $\mathbf{Y} := \begin{bmatrix} \mathbf{Y}_p \\ \mathbf{Y}_q \end{bmatrix} \in \mathbb{C}^{n \times M}$, $\mathbf{H} \in \mathbb{C}^{K \times M}$ and $\mathbf{Z} := \begin{bmatrix} \mathbf{Z}_p \\ \mathbf{Z}_q \end{bmatrix} \in \mathbb{C}^{n \times M}$ is AWGN

FASURA: Decoder



- Energy Detector: Compute the statistic

$$\lambda_j = \|\mathbf{P}^*[:, j] \mathbf{Y}_p\|^2 + \sum_{t=1}^{n_c/2} \|\mathbf{A}_t^*[:, j] \mathbf{Y}_q[\mathbf{n}_t, :]\|^2, \forall j \in [J], \text{ where } \mathbf{n}_t = [(t-1)L + 1: tL]$$

- Output the indices correspond to the largest K values
- $$\mathcal{K} = \text{argmax}(\{\lambda_1, \lambda_2, \dots, \lambda_J\}, K)$$

- Channel Estimation

$$\hat{\mathbf{H}} = (\sigma_z^2 \mathbf{I}_K + \hat{\mathbf{P}}^* \hat{\mathbf{P}})^{-1} \hat{\mathbf{P}}^* \mathbf{Y}_p, \text{ where } \hat{\mathbf{P}} = \mathbf{P}[:, \mathcal{K}]$$

- Symbol Estimation

- The received signal of length L , for each symbol QPSK symbol, is given by

$$\begin{aligned} \mathbf{Y}_q[\mathbf{n}_t, m] &= \mathbf{A}_t \text{diag}(\mathbf{r}_t) \mathbf{H}[:, m] + \mathbf{Z}[\mathbf{n}_t, m] \\ &= \mathbf{A}_t \text{diag}(\mathbf{H}[:, m]) \mathbf{r}_t + \mathbf{Z}[\mathbf{n}_t, m], \end{aligned}$$

- where $\mathbf{r}_t = (s_{t,1}, s_{t,2}, \dots, s_{t,K})$ are the symbols of the users at time t
- Stacking the columns of \mathbf{Y}_q

$$\underbrace{\begin{bmatrix} \mathbf{Y}_q[\mathbf{n}_t, 1] \\ \vdots \\ \mathbf{Y}_q[\mathbf{n}_t, M] \end{bmatrix}}_{\mathbf{y}_t \in \mathbb{C}^{LM \times 1}} = \underbrace{\begin{bmatrix} \mathbf{A}_t \text{diag}(\mathbf{H}[:, 1]) \\ \vdots \\ \mathbf{A}_t \text{diag}(\mathbf{H}[:, M]) \end{bmatrix}}_{\mathbf{B}_t \in \mathbb{C}^{LM \times K}} \underbrace{\begin{bmatrix} \mathbf{r}_t \\ \vdots \\ \mathbf{r}_t \end{bmatrix}}_{\mathbf{r}_t \in \mathbb{C}^{K \times 1}} + \underbrace{\begin{bmatrix} \mathbf{Z}[\mathbf{n}_t, 1] \\ \vdots \\ \mathbf{Z}[\mathbf{n}_t, M] \end{bmatrix}}_{\mathbf{z}_t \in \mathbb{C}^{LM \times 1}}$$

- As a consequence, we can apply an LMMSE estimator to the vectorized received signal

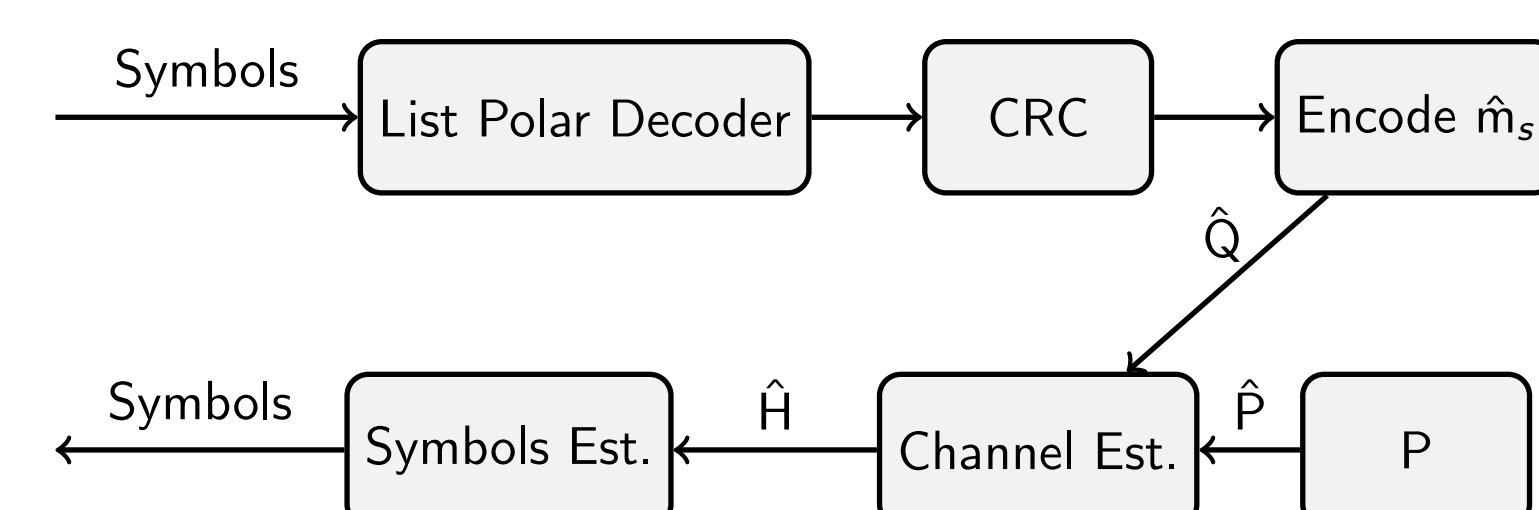
$$\mathbf{y}_t = \mathbf{B}_t \mathbf{r}_t + \mathbf{z}_t$$

- The estimated symbols of the active users at time t are given by

$$\hat{\mathbf{r}}_t = (\sigma_z^2 \mathbf{I}_K + \hat{\mathbf{B}}_t^* \hat{\mathbf{B}}_t)^{-1} \hat{\mathbf{B}}_t^* \mathbf{y}_t, \quad \forall \quad t = 1, 2, \dots, n_c/2$$

- NOPICE (Noisy Pilot Channel Estimation)

- Purpose:** Improve the symbols estimates
- Use the recovered symbols as pilots and re-estimate the channel
- Then re-estimate the QPSK symbols



- Channel Decoder
 - Includes list-polar decoder and a CRC step
- Channel Estimation
 - Use the pilot part and the data part to estimate the channel from the received signal,

$$\mathbf{Y}[:, m] = \hat{\mathbf{X}} \mathbf{H}[:, m] + \mathbf{Z}[:, m], \quad m = 1, 2, \dots, M$$

- LMMSE filter to estimate \mathbf{H}

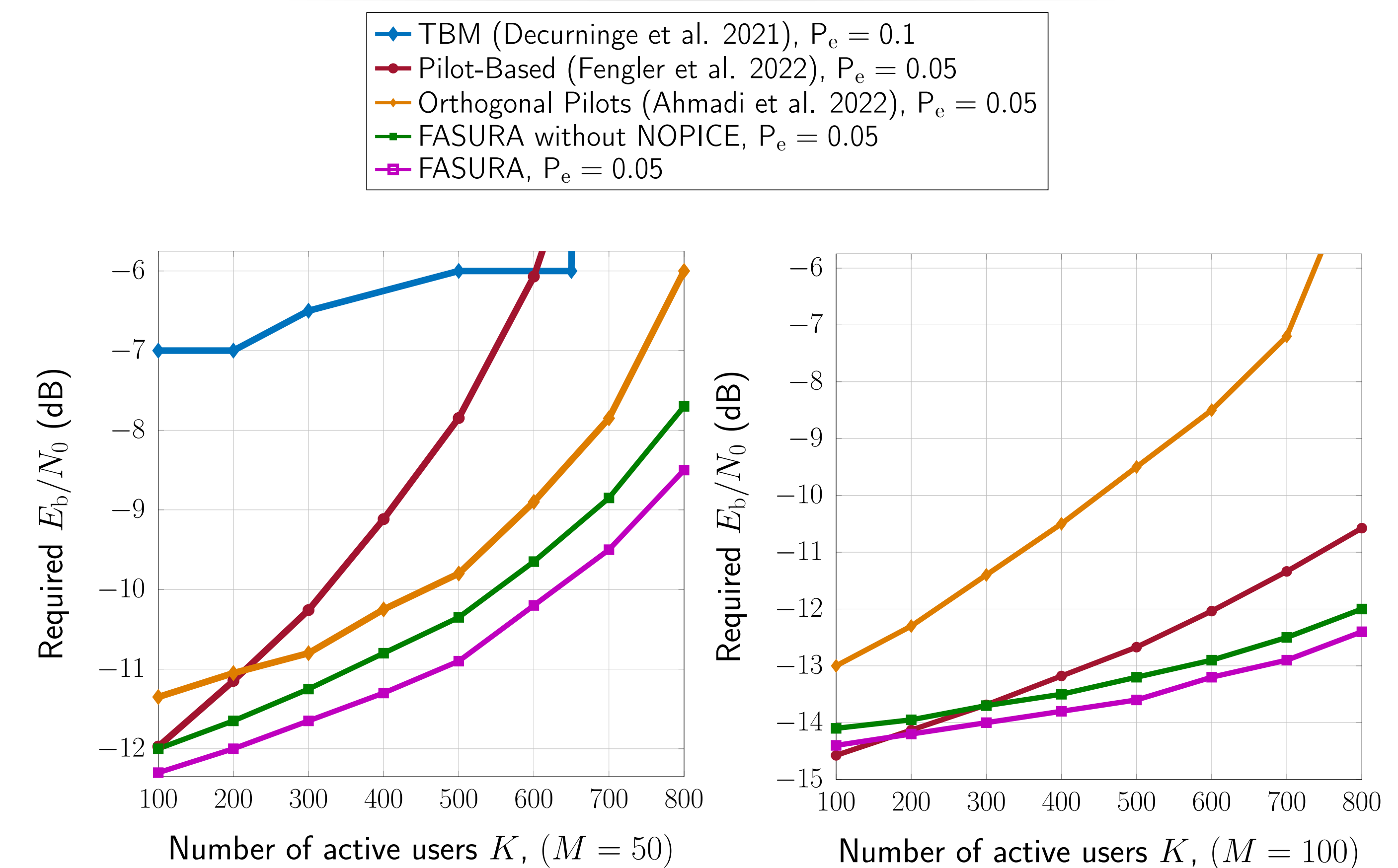
$$\mathbf{W}_2 = \hat{\mathbf{X}}^* (\hat{\mathbf{X}}^* \hat{\mathbf{X}} + \sigma_z^2 \mathbf{I}_K)^{-1}$$

- SIC (Successive Interference Cancellation)
 - Subtract the interference of the decoded users and proceed to the next round

$$\mathbf{Y}_{t+1} = \mathbf{Y}_t - \sum_{i \in \mathcal{K}} \hat{\mathbf{x}}_i \hat{\mathbf{h}}_i^\top$$

- Repeat steps 1-7 until the receiver is not able to decode a new user

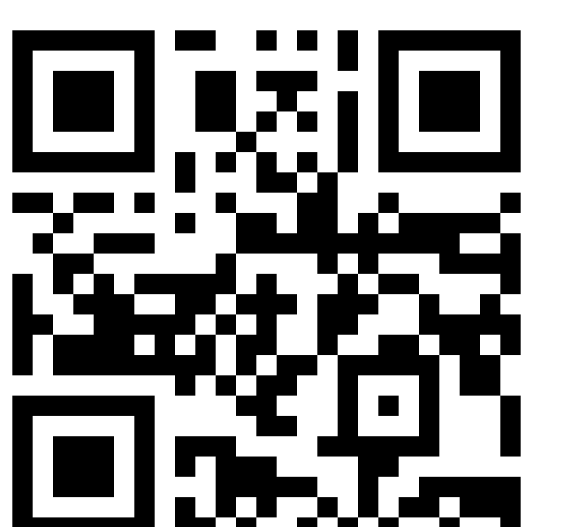
Numerical Results



E_b/N_0 as a function of the number of users, and number of antennas, $M = 50$ and $M = 100$

Selected References

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Paper



Source Code