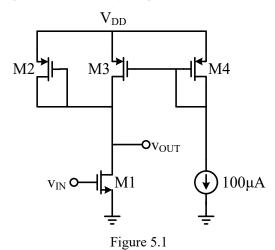
Table 5.1

		Typical Parameter Value		
Parameter Symbol	Parameter Description	n-Channel	p-Channel	Units
V_{T0}	Threshold	0.7	-0.8	V
	voltage(V _{BS} =0)			
K	Transconductance	134	50	μA/V²
	parameter(in			
	saturation)			
γ	Bulk threshold	0.45	0.4	$V^{1/2}$
	parameter			
λ	Channel length	0.1	0.2	V ⁻¹
	modulation parameter			
$2 \phi_F $	Surface potential at	0.9	0.8	V
	strong inversion			

$$K = \mu C_{OX}$$

5.1 Assume that W/L ratios of Figure 5.1 are $(W/L)_1 = 2\mu m/1\mu m$ and $(W/L)_2 = (W/L)_3 = (W/L)_4 = 1\mu m/1\mu m$. Find the dc value of v_{IN} that will give a dc current in M1 of 110 μ A. Calculate the small signal voltage gain and output resistance using the parameters of Table 5.1. Assume $\lambda = \gamma = 0$.



Answer:

$$I_{D1} = \frac{1}{2} K_N (\frac{W}{L})_1 (V_{in} - V_{TH1})^2$$

$$110\mu = \frac{1}{2} \times (134\mu) \times \frac{2}{1} \times (V_{in} - 0.7)^2$$

$$V_{in} = 1.61V$$

$$I_{D3} = I_{D4} = 100\mu A$$

$$I_{D2} = I_{D1} - I_{D3} = 10\mu A$$

$$A_v \cong -\frac{g_{m1}}{g_{m2}} = -\sqrt{\frac{K_N}{K_P} \frac{(W/L)_1}{(W/L)_2} \frac{I_{D1}}{I_{D2}}} = -\sqrt{\frac{134\mu}{50\mu} \times \frac{2}{1} \times \frac{110\mu}{10\mu}} = -7.68$$

$$R_{out} \cong \frac{1}{g_{m2}} = \frac{1}{\sqrt{2K_P(W/L)_2 I_{D2}}} = \frac{1}{\sqrt{2 \times 50 \times 10^{-6} \times 1 \times 10 \times 10^{-6}}} = 31.6K\Omega$$

- 5.2 Suppose the common-source stage of Fig 5.2 is to provide an output swing from 1V to 2.5V. Assume that $(W/L)_1 = 50/0.5$, $R_D = 2k\Omega$, $V_{DD} = 3V$ and $\lambda = 0$. Use model parameters in Table 5.1.
 - a) Calculate the input voltages that yield $V_{out} = 1V$ and $V_{out} = 2.5V$.
 - b) Calculate the drain current and the transconductance of M₁ for both cases.
 - c) How much does the small-signal gain, g_mR_D, vary as the output goes from 1V to 2.5V?

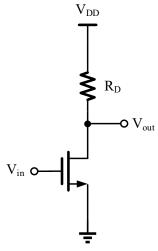


Figure 5.2

Answer:

a), b):

Vout=1V 时:

$$I_{D1} = \frac{V_{DD} - V_{out}}{R_D} = 1mA$$

$$V_{in} = V_{THN0} + \sqrt{\frac{2I_{D1}}{K_N \left(\frac{W}{L}\right)_1}} = 1.086V$$

$$g_{m1} = \sqrt{2K_N \left(\frac{w}{L}\right)_1} I_D = 5.18 \times 10^{-3} \text{S}$$

Vout=2.5V 时:

$$I_{D1} = \frac{V_{DD} - V_{out}}{R_D} = 0.25 mA$$

$$V_{in} = V_{THN0} + \sqrt{\frac{2I_{D1}}{K_N \left(\frac{W}{L}\right)_1}} = 0.893V$$

$$g_{m1} = \sqrt{2K_N \left(\frac{W}{L}\right)_1 I_D} = 2.588 \times 10^{-3} \text{S}$$

c):

$$\Delta g_m R_D = 5.18$$

- 5.3 Consider the circuit of Fig 5.3 with $(W/L)_1 = 50/0.5$ and $(W/L)_2 = 10/0.5$. Assume that $\lambda = \gamma = 0$, $V_{DD} = 3V$.
 - a) At what input voltage is M_1 at the edge of the triode region? What is the small-signal gain under this condition?
 - b) When V_{out} is 0.66 V, what is the small-signal gain under this condition?

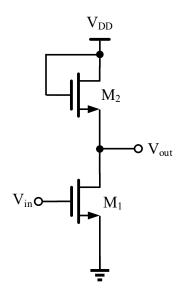


Figure 5.3

Answer:

a)

 M_1 at the edge of the triode region:

$$\begin{split} V_{out} &= V_{in} - V_{TH1} \\ I_{D1} &= I_{D2} = \frac{1}{2} K_N \left(\frac{W}{L}\right)_1 (V_{in} - V_{TH1})^2 = \frac{1}{2} K_N \left(\frac{W}{L}\right)_2 (V_{DD} - V_{out} - V_{TH2})^2 \\ V_{in} &= 1.41 V, \quad V_{out} = 0.71 V \\ A_V &= -\sqrt{\frac{2K_N \left(\frac{W}{L}\right)_1 I_{D1}}{2K_N \left(\frac{W}{L}\right)_2 I_{D2}}} = -2.236 \end{split}$$

b)

Vout=0.66V < 0.71V, M_1 is working in the triode region

$$\frac{1}{2}K_N \left(\frac{W}{L}\right)_2 (V_{DD} - V_{out} - V_{TH2})^2 = K_N \left(\frac{W}{L}\right)_1 \left[(V_{in} - V_{TH1})V_{out} - \frac{V_{out}^2}{2} \right]$$

$$V_{in} = 1.84V$$

$$I_{D} = K_{N} \left(\frac{W}{L}\right)_{1} \left[(V_{in} - V_{TH1}) V_{out} - \frac{V_{out}^{2}}{2} \right]$$

$$\frac{\partial I_{D}}{\partial V_{in}} = K_{N} \left(\frac{W}{L}\right)_{1} V_{out}$$

$$A_{V} = -\frac{g_{m1}}{g_{m2}} = -\frac{K_{N} \left(\frac{W}{L}\right)_{1} V_{out}}{K_{N} \left(\frac{W}{L}\right)_{2} (V_{DD} - V_{out} - V_{TH2})} = -2.015$$

5.4 In the circuit of Fig 5.4, $(W/L)_1 = 20/0.5$, $I_1 = 1$ mA, and $I_S = 0.75$ mA. Assuming $\lambda = 0$, $V_{DD} = 3$ V, calculate $(W/L)_2$ such that M_1 is at the edge of triode region. What is the small-signal voltage gain under this condition? Use model parameters in Table 5.1.

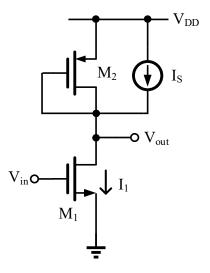


Figure 5.4

Answer:

M₁ at the edge of the triode region:

$$V_{out} = V_{in} - V_{TH1}$$

$$\frac{1}{2} K_P \left(\frac{W}{L}\right)_2 (V_{DD} - V_{out} - |V_{TH2}|)^2 + I_S = \frac{1}{2} K_N \left(\frac{W}{L}\right)_1 (V_{in} - V_{TH1})^2 = 10^{-3}$$

$$V_{in} = 1.311, \quad \left(\frac{W}{L}\right)_2 = 3.961$$

$$A_V = -\frac{g_{m1}}{g_{m2}} = -\sqrt{\frac{K_N \left(\frac{W}{L}\right)_1 I_1}{K_P \left(\frac{W}{L}\right)_2 I_2}} = -10.4$$

- 5.5 Consider the circuit of Fig 5.5 with $(W/L)_1 = 50/0.5$, $R_D = 2k\Omega$, and $R_S = 200\Omega$, $V_{DD} = 3V$. Use model parameters in Table 5.1.
 - a) Calculate the small-signal voltage gain if $I_D = 0.5 \text{mA}$.
 - b) Assuming that $\lambda = \gamma = 0$, calculate the input voltage that places M1 at the edge of the triode region. What is the gain under this condition?

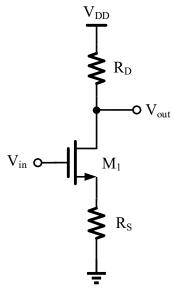


Figure 5.5

Answer:

a):

$$V_{S} = R_{S}I_{D} = 0.1V$$

$$V_{TH1} = V_{TH1,0} + \gamma \left(\sqrt{2|\varphi_{F}|} + V_{SB} - \sqrt{2|\varphi_{F}|} \right) = 0.7 + 0.45 \left(\sqrt{0.9 + 0.1} - \sqrt{0.9} \right) = 0.723$$

$$V_{out} = V_{DD} - R_{D}I_{D} = 2V$$

$$V_{DS} = 2 - 0.1 = 1.9V$$

$$g_{m} = \sqrt{2K_{N} \left(\frac{W}{L} \right)_{1} (1 + \lambda V_{DS})I_{D}} = 3.993 \times 10^{-3}$$

$$A_{V} = -\frac{g_{m}R_{D}}{1 + g_{m}R_{S}} = -4.44$$

b):

M1 at the edge of the triode region

$$\begin{split} V_{out} &= V_{in} - V_{TH1} \\ V_{in} &= V_{GS1} + R_S I_D \\ V_{DD} - R_D I_D &= V_{out} \\ V_{DD} - (R_S + R_D) I_D &= V_{GS1} - V_{TH1} \\ I_D &= \frac{1}{2} K_N \left(\frac{W}{L}\right)_1 (V_{GS1} - V_{TH1})^2 = \frac{1}{2} K_N \left(\frac{W}{L}\right)_1 [V_{DD} - (R_S + R_D) I_D]^2 \\ I_{D1} &= 1.58 mA \ (\text{V}_{GS} < \text{V}_{TH}, \text{ eliminate}), \ I_{D2} = 1.17 mA \\ V_{in} &= V_{DD} - R_D I_D + V_{TH1} = 1.36 V \\ g_{m1} &= \sqrt{2 K_N \left(\frac{W}{L}\right)_1 I_D} = 5.60 \times 10^{-3} \end{split}$$

$$G_m = \frac{g_{m1}}{1 + g_{m1}R_S} = 2.642 \times 10^{-3}$$

 $A_V = -G_m R_D = -5.283$