



Table 6.1

Parameter Symbol	Parameter Description	Typical Parameter Value		Units
		n-Channel	p-Channel	
V_{T0}	Threshold voltage($V_{BS}=0$)	0.7	-0.8	V
K	Transconductance parameter(in saturation)	134	50	$\mu A/V^2$
γ	Bulk threshold parameter	0.45	0.4	$V^{1/2}$
λ	Channel length modulation parameter	0.1	0.2	V^{-1}
$2 \phi_F $	Surface potential at strong inversion	0.9	0.8	V

$$K = \mu C_{OX}$$

6.1 Calculate the differential transconductance g_{md} and the differential voltage gain A_v of an n-channel input differential amplifier shown in Figure 6.1 , with the parameters shown in table 6.1. Consider $I_{SS}=100\mu A$ (the drain current of M5), and $W_1/L_1=W_2/L_2=W_3/L_3=W_4/L_4=1$. Assuming all the channel lengths are equal to $1\mu m$, and $V_{DD}=5V$. If $W_1/L_1=W_2/L_2=10W_3/L_3=10W_4/L_4=10$, repeat the calculation

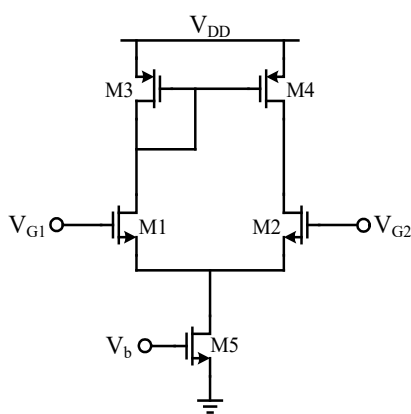


Figure 6.1

Answer:

$$a) \left(\frac{W}{L}\right)_1 = \left(\frac{W}{L}\right)_2 = \left(\frac{W}{L}\right)_3 = \left(\frac{W}{L}\right)_4 = 1$$

$$g_{md} = g_{m1} = g_{m2} = \sqrt{2 \times K_n \left(\frac{W}{L}\right)_1 \frac{I_{SS}}{2}} = 115.8 \mu S$$

$$A_v = \frac{g_{m2}}{r_{ds2} + r_{ds4}} = \frac{2g_{m2}}{(\lambda_2 + \lambda_4)I_{SS}} = 7.72$$

$$b) \left(\frac{W}{L}\right)_1 = \left(\frac{W}{L}\right)_2 = 10 \left(\frac{W}{L}\right)_3 = 10 \left(\frac{W}{L}\right)_4 = 10$$

$$g_{md} = g_{m1} = g_{m2} = \sqrt{2 \times K_n \left(\frac{W}{L}\right)_1 \frac{I_{SS}}{2}} = 366.1 \mu S$$

$$A_v = \frac{g_{m2}}{r_{ds2} + r_{ds4}} = \frac{2g_{m2}}{(\lambda_2 + \lambda_4)I_{SS}} = 24.4$$

6.2 Calculate the maximum($V_{IC(max)}$) and the minimum input common-mode voltages ($V_{IC(min)}$), and the input common mode voltage range (ICMR) of an n-channel input differential amplifier shown in Figure 6.1, with the parameters shown in table 6.1. Assume all MOSFETs are in saturation, all the $(W/L)_i$ are equal to $10\mu m/1\mu m$, $I_{SS}=10\mu A$, and $V_{DD}=5V$.

Answer:

The maximum input common-mode input is given by

$$V_{(IC)}(max) = V_{DD} + V_{T1} - V_{T3} - V_{GS3} = V_{DD} + V_{T1} - V_{T3} - \sqrt{\frac{I_{SS}}{K_P \left(\frac{W}{L}\right)_3}} = 4.76V$$

The minimum input common-mode input is given by

$$V_{(IC)}(min) = V_{SS} + V_{T1} + V_{GS3} + V_{GS5} = V_{SS} + V_{T1} + \sqrt{\frac{I_{SS}}{K_n \left(\frac{W}{L}\right)_3}} + \sqrt{\frac{2 \times I_{SS}}{K_n \left(\frac{W}{L}\right)_5}} = 0.91V$$

So, the input common-mode range becomes

$$ICMR = V_{(IC)}(max) - V_{(IC)}(min) = 3.85V$$

6.3 Find the value of the unloaded differential-transconductance, g_{md} , and the unloaded differential-voltage gain, A_v , for the p-channel input differential amplifier of Figure 6.2 when $I_{SS}=10\mu A$ and $I_{SS}=1\mu A$. What is the slew rate of the differential amplifier if a $100 pF$ capacitor is attached to the output? Assuming $W_1/L_1=W_2/L_2=W_3/L_3=W_4/L_4=1$, and all the channel lengths are equal to $1\mu m$.

Use the transistor parameters of Table 6.1.

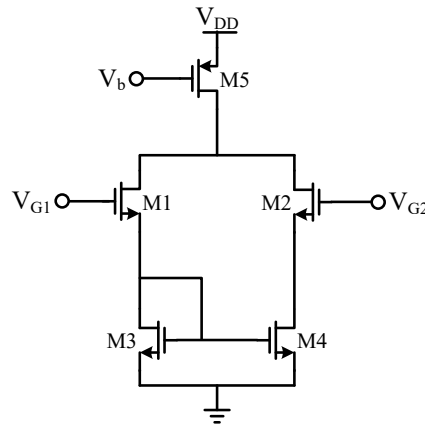


Figure 6.2

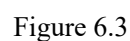
Given $I_{SS}=10\mu A$

$$A_v = \frac{g_{m1}}{g_{ds2} + g_{ds4}} = \frac{2g_{m1}}{(\lambda_1 + \lambda_2)I_{ss}} = 14.9$$

$$g_{md} = g_{m1} = g_{m2} = 7.07 \mu S$$

$$A_v = \frac{g_{m1}}{g_{ds2} + g_{ds4}} = \frac{2g_{m1}}{(\lambda_1 + \lambda_2)I_{SS}} = 47.13$$
$$SR = \frac{I_{ss}}{C_L}$$
$$SR = \frac{I_{ss}}{C_L} = 0.1V/\mu s$$
$$SR = \frac{I_{ss}}{C_L} = 0.01V/\mu s$$

(c) If I_{SS} requires a minimum voltage of 0.4V, what is the maximum differential output swing?



Answer:

$$a) A_V \approx -\frac{g_{m1}}{g_{m3}} = -\sqrt{\frac{K_n I_{D1}}{K_p I_{D3}}} = -\sqrt{\frac{134 \times 0.5 I_{SS}}{50 \times 0.2 \frac{I_{SS}}{2}}} = -3.66$$

$$b) I_{D5} = I_{D6} = 0.8 \frac{I_{SS}}{2} = 0.4 \text{ mA}, V_b = V_{DD} - V_{SG5} = V_{DD} - |V_{TH}| - \sqrt{\frac{2 I_{D5}}{K_p \frac{W}{L}}} = 1.8 \text{ V}$$

$$c) (V_{OUT1,2})_{max} = \min(V_b + |V_{TH,p}|, V_{DD} - |V_{TH,p}|) = \min(1.8 + 0.8, 3 - 0.8) = 2.2 \text{ V}$$

$$(V_{OUT1,2})_{min} = \max(V_{Imin} + V_{GS1}|_{I_D=0.6 I_{SS}} - V_{TH,N}, V_{DD} - V_{SG3}|_{I_D=0.2 I_{SS}})$$

$$V_{GS1}|_{I_D=0.6 I_{SS}} = V_{TH,N} + \sqrt{\frac{2 \times 0.6 I_{SS}}{K_n \frac{W}{L}}} = 0.7 + 0.299 = 0.999 \text{ V}$$

$$V_{SG3}|_{I_D=0.2 I_{SS}} = |V_{TH,p}| + \sqrt{\frac{2 \times 0.2 I_{SS}}{K_p \frac{W}{L}}} = 0.8 + 0.28 = 1.08 \text{ V}$$

$$(V_{OUT1,2})_{min} = \max(0.4 + 0.999 - 0.7, 3 - 1.08) = 1.92 \text{ V}$$

$$V_{out,swing} = 2 \times (2.2 - 1.92) = 0.56 \text{ V}$$

6.5 The circuit shown in Figure 6.4 called a folded-current mirror differential amplifier and is useful for low values of power supply. Assume that all W/L values of each transistor is 100. **Using the parameters shown in table 6.1,**

- Find the maximum input common mode voltage, $V_{IC(max)}$ and the minimum input common mode voltage, $V_{IC(min)}$. Keep all transistors in saturation for this problem.
- What is the input common mode voltage range, ICMR?
- Find the **small signal** voltage gain, v_{out}/v_{in} , if $v_{in} = v_1 - v_2$.

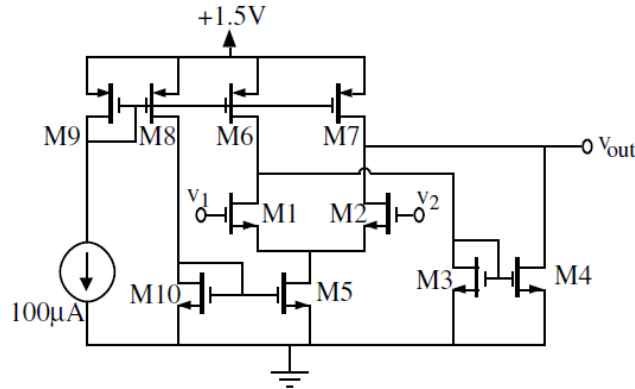


Figure 6.4

Answer:

$$a) v1(max) = V_{GS3} + 2 \times V_{TN} = 0.7 + \sqrt{\frac{2 \times 50}{134 \times 100}} + 0.7 = 1.486 \text{ V}$$

$$v1(min) = V_{SS} + V_{GS5} - V_{TN} + V_{GS1} = \sqrt{\frac{2 \times 100}{134 \times 100}} + \sqrt{\frac{2 \times 50}{134 \times 100}} + 0.7 = 0.908 \text{ V}$$

$$b) V_{ICMR} = v1(max) - v1(min) = 1.486 \text{ V} - 0.908 \text{ V} = 0.578 \text{ V}$$

c)

$$g_{md} = g_{m1} = g_{m2} = \sqrt{2 \times K_N \left(\frac{W}{L}\right)_1 \frac{I_{SS}}{2}} = 1157.58 \mu S$$

$$r_{o4} = r_{o2} = \frac{2}{\lambda_n I_{SS}} = 0.2 M\Omega$$

$$r_{o7} = \frac{1}{\lambda_p I_{SS}} = 0.05 M\Omega$$

$$A_v = g_{m1} \times (r_{o2} // r_{o4} // r_{o7}) = 38.586$$

6.6 In the circuit of Fig 6.5, assume that $I_{SS} = 0.5\text{mA}$, $V_{DD} = 3\text{V}$, $(W/L)_{1,2} = 50/0.5$ and $(W/L)_{3,4} = 10/0.5$. I_{SS} current is provided by NMOS, and its $W/L = 50/0.5$. Assuming $\lambda \neq 0$.

a) Calculate the range of input common mode voltage.

b) If $V_{in,CM} = 1.5\text{V}$, draw a sketch of the small signal differential voltage gain of the circuit when V_{DD} changes from 0 to 3V.

c) If the mismatch threshold voltage of M_1 and M_2 is 1mV , calculate CMRR.

d) If the $W_3 = 10\mu\text{m}$ and $W_4 = 11\mu\text{m}$, calculate CMRR.

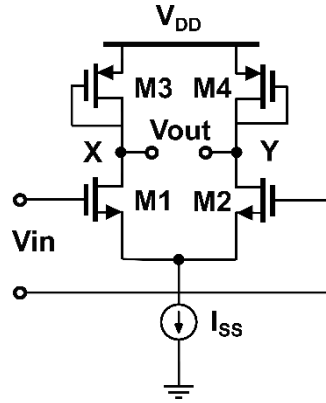


Figure 6.5

Answer:

$$a) (V_{in,cm})_{min} = V_{GS1} + V_{odSS} = V_{TH1} + \sqrt{\frac{2I_{D1}}{K_n \left(\frac{W}{L}\right)_1}} + \sqrt{\frac{2I_{SS}}{K_n \left(\frac{W}{L}\right)_{SS}}} = 0.7\text{V} + 0.193\text{V} + 0.273\text{V} =$$

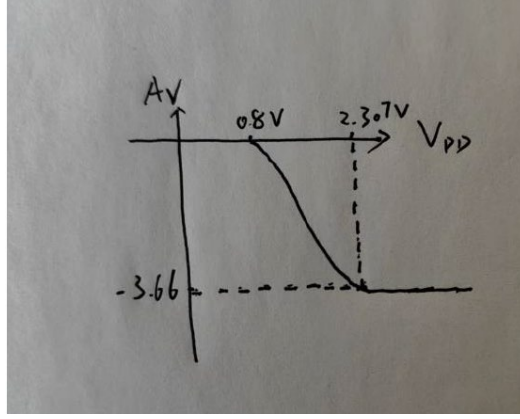
$$1.166\text{V}$$

$$(V_{in,cm})_{max} = V_{DD} - V_{od3} - V_{TH3} + V_{TH1} = V_{DD} - \sqrt{\frac{2I_{D3}}{K_p \left(\frac{W}{L}\right)_3}} - V_{TH3} + V_{TH1}$$

$$= 3\text{V} - 0.707\text{V} - 0.8\text{V} + 0.7\text{V} = 2.193\text{V}$$

$$ICMR = 2.193\text{V} - 1.166\text{V} = 1.027\text{V}$$

b) 三个标记点：开启电压为 0.8V ，增益是 -3.66 ，饱和点电压是 2.307V



c) 由于 M1 和 M2 的阈值电压失配, 因此有: $g_{m1} \neq g_{m2}, g_{m3} \neq g_{m4}$, 所以计算 A_{cm-dm} 有:

$$i_{D1} = g_{m1}(V_{in,cm} - V_p)$$

$$i_{D2} = g_{m2}(V_{in,cm} - V_p)$$

$$V_{out1} = -\frac{i_{D1}}{g_{m3}} = -\frac{g_{m1}(V_{in,cm} - V_p)}{g_{m3}}$$

$$V_{out2} = -\frac{i_{D2}}{g_{m4}} = -\frac{g_{m2}(V_{in,cm} - V_p)}{g_{m4}}$$

$$\frac{g_{m1}}{g_{m3}} = \frac{K_n \left(\frac{W}{L}\right)_{1,2}}{K_p \left(\frac{W}{L}\right)_{2,4}} = \frac{g_{m2}}{g_{m4}}$$

$$\therefore V_{out1} = V_{out2}$$

$$\therefore A_{cm-dm} = 0, CMRR = \infty$$

d) $A_{dm-dm} = -g_m R_D$

$$A_{cm-dm} = \frac{g_m R_D}{1 + 2g_m R_{SS}} - \frac{g_m(R_D + \Delta R_D)}{1 + 2g_m R_{SS}} = -\frac{g_m \Delta R_D}{1 + 2g_m R_{SS}}$$

$$\therefore CMRR = \left| \frac{A_{dm-dm}}{A_{cm-dm}} \right| = \frac{1 + 2g_m R_{SS}}{\Delta R_D / R_D}$$

$$\therefore R_{D1} = \frac{1}{g_{m3}}, R_{D2} = \frac{1}{g_{m4}}$$

$$\therefore \frac{\Delta R_D}{R_D} = \frac{R_{D1} - R_{D2}}{R_{D1}} = 1 - \frac{R_{D2}}{R_{D1}} = 1 - \frac{\sqrt{2K_p \left(\frac{W}{L}\right)_3 I_D}}{\sqrt{2K_p \left(\frac{W}{L}\right)_4 I_D}} = 1 - \sqrt{\frac{10}{11}} = 0.0465$$

$$g_m = \sqrt{2K_n \left(\frac{W}{L}\right)_1} i_{D1} = 2.588 \text{ m}\Omega^{-1}, \lambda = 2 \times 0.1 = 0.2, R_{SS} = \frac{1}{\lambda I_{SS}} = 20 \text{ k}\Omega$$

$$\therefore CMRR = 2270$$