

Cairo University - Faculty of Engineering Computer Engineering Department



M-ARY AMPLITIUDE SHIFT MODULATION

Subject: Digital Communication

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0.1 Part 1: Digital Communication

0.1.1 Problem 1

Figure 1 below showing the comparison between simulated BER and theoritical (analytical) BER VS the Eb/N0 in db.

Please notice, you'll have to input the no. of bits you wish to be transmitted, and it has to be divisible by 3.

0.1.2 Problem 2

The constellation of the 8-ary with decision region pf each symbol.

Boundaries are at:

$$-6\sqrt{E}, -4\sqrt{E}, -2\sqrt{E}, 0, 2\sqrt{E}, 4\sqrt{E}, 6\sqrt{E}$$

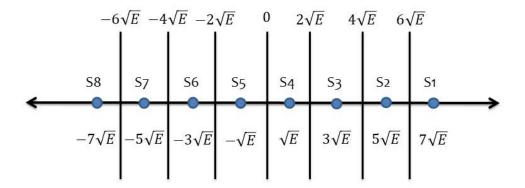


Figure 1: Symbols Boundary

0.1.3 Problem 3

The derivation of theoritical bit error rate using 8-ary.

$$Pe = \frac{1}{8} \sum_{i=0}^{7} P(e|Si)$$
 (1)

$$Pe(e|S0) = Pe(e|S7) \tag{2}$$

$$Pe(e|S1) = Pe(e|S2) = Pe(e|S3) = Pe(e|S4) = Pe(e|S5) = Pe(e|S6)$$
 (3)

Using Union bound S0, S7 only one neighbour and S1, S2,...S6 has two neighbours.

$$Pe(e|S0) = \frac{1}{2} erfc(\frac{\sqrt{E}}{\sqrt{N}}) \tag{4}$$

$$Pe(e|S1) = \frac{1}{2}erfc(\frac{\sqrt{E}}{\sqrt{N}}) + \frac{1}{2}erfc(\frac{\sqrt{E}}{\sqrt{N}})$$
 (5)

$$Pe(e|S1) = erfc(\frac{\sqrt{E}}{\sqrt{N}})$$
 (6)

$$Pe = \frac{1}{8*3} \left(2*\frac{1}{2}erfc(\frac{\sqrt{E}}{\sqrt{N}}) + 6*erfc(\frac{\sqrt{E}}{\sqrt{N}})\right)$$
 (7)

$$Pe = \frac{7}{24} (erfc(\frac{\sqrt{E}}{\sqrt{N}})) \tag{8}$$

0.1.4 Probelm 4

Figure 1 below showing the comparison between simulated BER and theoritical (analytical) BER VS the Eb/N0 in db.

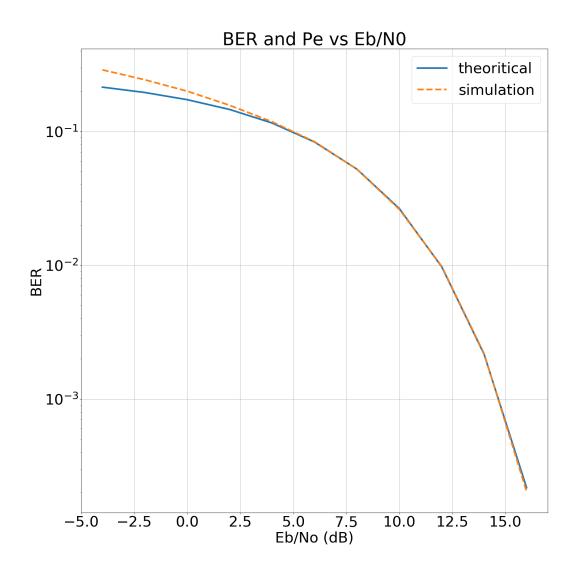


Figure 2: BER vs Eb/N0

0.1.5 Probelm 5

The answer is NO, We can't transmit at Rate 1 Mbps with bandwidth 0.5 MHz in passband transmition.

The minimum M required: 16 Only bit by bit transmittion is allowed. GIVEN:

$$Bt = 2Rs$$

$$Rb = 1Mbps$$

$$BW = 0.5MHz$$

$$M = 3$$

REQUIRED:

$$BW = ?2*Rs$$

$$BW = ?2*\frac{Rb}{\log_2 M}$$

$$BW = ?1Mbps*\frac{2}{3}$$

The Answer is: NO it can not be transmitted

$$0.5MHz = ?1Mbps * \frac{2}{\log_2 M}$$

$$4MHz = ?\log_2 M$$

and Minimum M alowed is:

$$M = 16$$

0.1.6 Probelm 6

Both of them satisfy the Gray Encoding criterion.

Because at the two examples only one bit is changed in each transition from symbol to the next one.

0.2 Part 2: Information Theory

Cyclic Codes.

0.2.1 Definition of Cyclic Codes

A Cyclic Codes is a block codes that follow the property of linearity where the circular is shift to the left (or n-1 shift to the right) always result in a word that belongs to the code words.

They are defined as C(x) = (n, k) it means the message has k bits and the code vector is n bits.

C(x) are defined with the help of generator polynomial g(x).

The degree of g(x) is equal to the number of parity-check digits of the code.

There are total of 2^k code polynomials in C(x).

They are error-correction codes, used earlier to transmit images of planets.

They have algebric properties that help detecting and correcting errors.

Cyclic codes can be used to correct errors, it can be generalized to correct burst of errors, not just one bit.

For example the set of [000, 1111, 0110, 1001]:

They follow the Linearity property, but not the Cyclic Shift property, therefore they can't be considered a valide Cyclic Codes.

Another Example the set of [0000, 1111, 0101, 1010]:

They follow both the linearity and cyclic shift property. Because, when adding any two of them it will result in another (third) codeword that lies in the finite list.

0.2.2 Systematic CodeWords:

In Systematic Codewords C(x) = [message, parity] this means, each generated codeword's first k bits are the message that got us that codeword while the remaining n - k bits are the parity check of the codeword.

Systematic Codewords follow the following conditions:

$$C(x) = x^{n-k}m(x) + p(x)$$

Where:

$$p(x) = Rem \left[\frac{x^{n-k}m(x)}{g(x)} \right]$$

and,

C(x) is codeword polynomial

m(x) is message polynomial

g(x) is generator polynomial

Example: Constructing a Systematic Cyclic Codes (7,4), using generator polynomial $g(x) = x^3 + x^2 + 1$, with message [1010].

Answer:

$$m(x) = x^{3} + x$$

$$p(x) = Rem \left[\frac{x^{3} * (x^{3} + x)}{(x^{3} + x^{2} + 1)} \right]$$
From division:
$$p(x) = 1$$
therefore,
$$C(x) = x^{3} * (x^{3} + x) + 1$$

$$C(x) = x^{6} + x^{4} + 1$$

$$C(x) = [1010001]$$

You may notice that the message m(x) = 1010 and the first k bits in the generated codeword is also 1010

Will the rest of the codeword is the parity check.

0.2.3Relation between generator polynomial and generator matrix:

At the last section we've introduced the difinition of generator polynomial such that $g(x) = x^3 + x^2 + 1$, this polynomial is unique and we can derive all the other codewords from it by multiplying with the various messages allowed.

In cyclic code C, the generator matrix's dimensions is [n,k] n number of columns and k number of rows.

Generator Matric [G] is composed of [I, P], I is the identity matrix with dimensions of [k, k] while P is the parity matrix with the dimensions of [k, (n-k)]

So, the main core of our problem is identifying the parity matrix, and it is identified as follows:

$$k^{th}Row = Rem\left[\frac{x^{n-k}}{g(x)}\right]$$

continuing on the previous example the generator matrix will be something like this:

$$G(x) = \begin{bmatrix} 1 & 0 & 0 & 0 & - & - & - \\ 0 & 1 & 0 & 0 & - & - & - \\ 0 & 0 & 1 & 0 & - & - & - \\ 0 & 0 & 0 & 1 & - & - & - \end{bmatrix}$$

so, to calculate the parity matrix we'll solve the following equations:

$$1^{st}Row = Rem \left[\frac{x^6}{x^3 + x^2 + 1} \right]$$
$$1^{st}Row = x^2 + 1 = [101]$$

$$2^{nd}Row = Rem\left[\frac{x^5}{x^3 + x^2 + 1}\right]$$
$$2^{st}Row = x^2 + x + 1 = [111]$$

$$2^{st}Row = x^2 + x + 1 = [111]$$

$$3^{rd}Row = Rem\left[\frac{x^4}{x^3 + x^2 + 1}\right]$$

$$3^{rd}Row = x^2 + 1 = [101]$$

$$4^{th}Row = Rem \left[\frac{x^3}{x^3 + x^2 + 1} \right]$$
$$4^{th}Row = x + 1 = [011]$$

therefore the final matrix will be:

$$G(x) = \begin{bmatrix} 1 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 1 & 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 1 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 1 & 0 & 1 & 1 \end{bmatrix}$$

0.3 Appendix: Main Code for part 1

Listing 1: Main Code for part 1

```
import math
import random
import matplotlib.pyplot as plt
import numpy as np
from scipy.special import erfc
def generateRandomProcess (elementCount):
    This function returns random process of streaming bits 0 an 1.
    eg. [1, 0, 0, 1, \ldots]
    and we know for sure that the main function is checking that the element Count
    is divisible by 3
    XofT = []
    for in range (element Count):
        XofT.append(math.floor(0.5 + random.uniform(0, 1)))
    return XofT
\mathbf{def} \ \ \mathbf{generateTimeSteps} \ (\ \mathbf{elementCount} \ , \ \ \mathbf{width} \ ) :
    This function returns time steps
    return np.linspace(0, elementCount, elementCount, endpoint=False)
def mapper():
         Function:
             Here you can find the Mapper Code and Logic
             Since M = 3 and Eb = 1, then E0 = 1/7
         Logic:
             Simply reading an input from the user count of bits
             Making sure its divisible by three
             Generating random bits of that size
             Returns M-ary of size bits.size/3
        Returns:
             Bits: the randomly generated bits
             Mapped: the mapped M-ary values
     ,,,
    E0 = 1/7
    elemCount = 1
    while elemCount\%3 != 0:
        elemCount = int(input("Please_enter_the_desired_count_of_elements,_and_make_sure
    Bits = generateRandomProcess (elemCount)
    Symbols = [-7, -5, -3, -1, 1, 3, 5, 7]
    Symbols = [i * math.sqrt(E0) for i in Symbols]
    Bits to Symbols = []
    for i in range (0, elemCount, 3):
        stack of elements = ""
```

```
stack of elements += str(Bits[i])
        stack of elements += str(Bits[i+1])
        stack of elements += str(Bits[i+2])
        if stack of elements == "000":
            Bits to Symbols.append ([Symbols [4]])
        elif stack\_of\_elements == "001":
            Bits_to_Symbols.append([Symbols[5]])
        elif stack_of_elements == "010":
            Bits to Symbols.append([Symbols[7]])
        elif stack_of elements == "011":
            Bits to Symbols.append([Symbols[6]])
        elif stack of elements == "100":
            Bits_to_Symbols.append([Symbols[3]])
        elif stack\_of\_elements == "101":
            Bits to Symbols.append([Symbols[2]])
        elif stack\_of\_elements == "110":
            Bits_to_Symbols.append([Symbols[0]])
        elif stack\_of\_elements == "111":
            Bits to Symbols.append([Symbols[1]])
        else:
            print ("error_occured_in_mapping")
    return Bits, Bits to Symbols, elemCount
def Channel (mean, variance, length):
    Function:
        This Function is responsible for simulating the AWGN channel effect of adding re
    transmitted signal.
    Inputs:
        mean: the mean of awan channel
        variance: the varinace of awgn channel
        length: Length of the time steps
    Outputs:
        generated \ random \ noise
    return math.sqrt (variance/2)*np.random.randn(length)
    \# return np.random.normal(mean, np.sqrt(variance), length)
def DeMapper (Noisy Bits to Symbols):
        Demap/Decode the symbols into their actual bits
    Inputs:
        Noisy Bits to Symbols: Symbols representing the bits added to noise
    Output:
        Received Bits: The mapped bits
    Received Bits = []
    E0 = 1/7
    Symbols boundry = [-6, -4, -2, 0, 2, 4, 6]
    Symbols boundry = [i * math.sqrt(E0) for i in Symbols boundry]
    for symbole in Noisy_Bits_to_Symbols:
        if symbole <= Symbols_boundry[0]:</pre>
            Received_Bits.append(1)
            Received_Bits.append(1)
            Received Bits.append(0)
```

```
elif symbole > Symbols boundry[0] and symbole <= Symbols boundry[1]:
            Received Bits.append(1)
            Received Bits.append(1)
            Received Bits.append(1)
        elif symbole > Symbols boundry[1] and symbole <= Symbols boundry[2]:
            Received \_ Bits.append(1)
            Received_Bits.append(0)
            Received_Bits.append(1)
        elif symbole > Symbols_boundry[2] and symbole <= Symbols_boundry[3]:
            Received Bits.append(1)
            Received Bits.append(0)
            Received Bits.append(0)
        elif symbole > Symbols boundry[3] and symbole <= Symbols boundry[4]:
            Received Bits.append(0)
            Received_Bits.append(0)
            Received_Bits.append(0)
        elif symbole > Symbols boundry[4] and symbole <= Symbols boundry[5]:
            Received Bits.append(0)
            Received Bits.append(0)
            Received Bits.append(1)
        elif symbole > Symbols boundry [5] and symbole <= Symbols boundry [6]:
            Received_Bits.append(0)
            Received_Bits.append(1)
            Received Bits.append(1)
        elif symbole > Symbols boundry [6]:
            Received Bits.append(0)
            Received_Bits.append(1)
            Received \_ Bits.append (0)
        else:
            print("error_occured_in_Demapping")
    return Received Bits
def BER (Bits, Received Bits):
    Function:
        Calculate the BER for each bit sent
    Inputs:
        Bits: The \ actual \ transmitted \ bits
        Received Bits: The received bits
    Output:
        actual ber: the sum of all errors occurred
    error = 0
    \#Converting them to ndarray
    for i in range (len(Received Bits)):
        if (Bits[i] != Received_Bits[i]):
            error += 1
    return error/len (Received Bits)
if name == " main ":
    ###VARIABLES###
    Eb No dB Min = -4 # min E/No alowed in db
    Eb No dB Max = 16 \# max E/No alowed in db
        Eb No dB: Array of values from Eb No dB Min to Eb No dB Max
        with step size = 2
        To plot(simulate) the vertical access for BER and Theoritical BER (Pe)
```

```
,,,
Eb No dB = np. arange (start=Eb No dB Min, stop=Eb No dB Max+1, step=2)
\# \ Linearize \ Eb/N0
    Liniarizing Eb/No in order to get the Variance of AWGN at this point
Eb No = 10**(Eb \text{ No } dB/10.0)
\#Theoritcal and Actual Errors:
\#Bit error rates of different Eb/N0
BERs = []
\#Probability of errors of different Eb/N0
PEs = []
\#1- Mapper is always the same as E0=1/7 all the time
Bits, Bits_to_Symbols, length = mapper()
mean = 0
for E N0 in Eb No:
    \#2-Channel
    \# variance = math.sqrt((1/E N0)/2)
    variance = (1/E N0)
    \#length//3 ==> floor(length/3)
    Noise = Channel (mean, variance, length //3)
    #Adding the noise to Symbolic Bits ELEMENT WISE
    Noisy Bits to Symbols = []
    for i in range (len(Bits_to_Symbols)):
        Noisy_Bits_to_Symbols.append(Bits_to_Symbols[i] + Noise[i])
    #3- Demapping
    Received Bits = DeMapper(Noisy Bits to Symbols)
    #4- BER
    actual ber = BER(Bits, Received Bits)
    BERs. append (actual ber)
    PEs. append ((7/8)* (erfc (math. sqrt (E N0/7))) * (1/3))
\#Plotting1/3)
plt.rcParams["figure.figsize"] = (20,20)
plt.rcParams.update({'font.size': 35})
plt.semilogy\,(Eb\ No\ dB,\ PEs, linestyle = 'solid', linewidth = 4)
plt.semilogy(Eb_No_dB, BERs, linestyle = 'dashed', linewidth=4)
plt.grid(True)
plt.legend(('theoritical', 'simulation'))
plt.xlabel('Eb/No_(dB)')
plt.ylabel('BER')
plt.title("BER_and_Pe_vs_Eb/N0")
\# plt.show()
plt.savefig("Figures/Figure 1")
print ("figure_is_saved_at_Figures/Figure_1.png")
```