

Chapter 8 Frequency Response Methods

8.1 Introduction

Example:

If U_i =Asin ωt , determine $y(t)=U_o(t)$

$$\frac{Y(s)}{R(s)} = \frac{U_o(s)}{U_i(s)} = \frac{1}{RCs + 1}$$

$$Y(s) = \frac{1}{RCs + 1}R(s) = \frac{1}{RCs + 1} \cdot \frac{A\omega}{s^2 + \omega^2}$$

$$C = U_{o}$$

$$y(t) = \frac{A\tau\omega}{1 + \tau^2\omega^2} e^{-\frac{t}{\tau}} + \frac{A}{\sqrt{1 + \tau^2\omega^2}} \sin(\omega t - \arctan\tau\omega) \ (\tau = RC)$$

$$y(\infty) = \lim_{t \to \infty} y(t) = \frac{A}{\sqrt{1 + \tau^2 \omega^2}} \sin(\omega t - \arctan \omega)$$

8.1 Introduction

$$\frac{A(\omega)}{A} = \frac{1}{\sqrt{1 + \tau^2 \omega^2}}, \quad \Phi(\omega) = -\arctan\tau\omega$$
 amplitude phase

Both are the function of ω

$$G(s) = \frac{1}{\tau s + 1}$$
 Let $s = j\omega$, we get frequency transfer function

$$G(j\omega) = \frac{1}{j\tau\omega + 1}$$

$$|G(j\omega)| = \frac{1}{\sqrt{\tau^2 \omega^2 + 1}}$$
 $\angle G(j\omega) = -\arctan \omega \tau$ amplitude character phase character

$$\angle G(j\omega) = -\operatorname{arctan} \omega \tau$$
phase character

8.1 Introduction

Concepts:

The frequency response of a system is defined as the steady-state response of the system to a sinusoidal input signal

Note:

- ① the sinusoidal is a unique input signal;
- ② For a linear system, output signal is sinusoidal in the steady state. (Frequency method is for steady, not dynamic process)
- ③ Compared with the input sinusoidal signal Asin(ωt), the magnitude increases $|T(j\omega)|$, and the phase changes $\angle T(j\omega)$

Advantages:

- ① Frequency characteristic plot can be measured by experiment;
- 2 Plot is convenient for analysis and design.

System output response

$$Y(jw) = T(jw)R(jw) = \frac{G(jw)}{1 + G(jw)}R(jw)$$

System close-loop transfer function

$$T(jw) = \frac{G(jw)}{1 + G(jw)}$$

System open-loop transfer function

$$T(jw) = \frac{G(jw)}{1 + G(jw)} \Rightarrow G(jw)$$

System open-loop transfer function

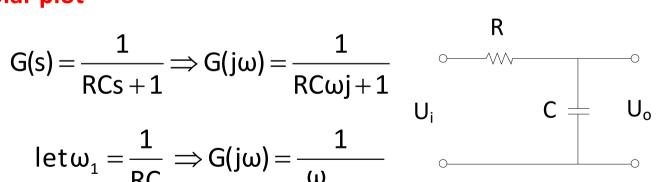
$$G(s) \stackrel{s=j\omega}{\Longrightarrow} G(j\omega) = R(\omega) + jX(\omega) = |G(j\omega)| e^{\angle G(j\omega)}$$

$$R(\omega) = Re[G(j\omega)], jX(\omega) = Im[G(j\omega)]$$

polar plot

$$G(s) = \frac{1}{RCs + 1} \Rightarrow G(j\omega) = \frac{1}{RC\omega i + 1}$$

$$let \omega_1 = \frac{1}{RC} \Rightarrow G(j\omega) = \frac{1}{\frac{\omega}{\omega_1} j + 1}$$

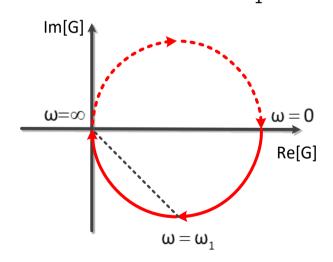


$$G(j\omega) = \frac{1}{1 + (\frac{\omega}{\omega_1})^2} - j \frac{\frac{\omega}{\omega_1}}{1 + (\frac{\omega}{\omega_1})^2}$$

$$|G(j\omega)| = \frac{1}{\sqrt{1 + (\frac{\omega}{\omega_1})^2}}$$

$$\angle G(j\omega) = -\arctan\frac{\omega}{\omega_1}$$

ω	$R(\omega)$	$X(\omega)$	$ G(j\omega) $	$\angle G(j\omega)$
0	1	0	1	0
ω_1	$\frac{1}{2}$	$-\frac{1}{2}$	$\frac{\sqrt{2}}{2}$	-45°
∞	0	0	0	-90°



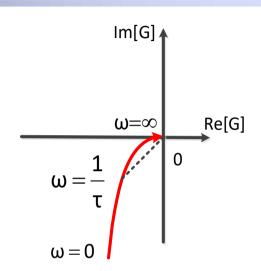
☐ Tips:

- $(1)R(\omega),X(\omega)$ or $|G(j\omega),\angle G(j\omega)$, choose one method for drawing
- ② $\omega(-\infty \to +\infty)$ so the plot is symmetric about the real axis; the regular polar plot is from 0 to $+\infty$, we use the plot of $\omega(-\infty \to +\infty)$ to verify the stability of system.

Example:

$$G(s) = \frac{1}{s(s+1)} \Rightarrow G(j\omega) = \frac{1}{jw(jw\tau + 1)}$$
$$|G(j\omega)| = \frac{1}{|jw||jw\tau + 1|} = \frac{1}{w\sqrt{(w\tau)^2 + 1}};$$
$$\angle G(j\omega) = -90^\circ - \tan^{-1}\tau\omega$$

ω	G(jω)	∠G(jω)
0	∞	−90°
$\frac{1}{-}$	1	-135°
$ au \infty$	√2 0	-180°



The limitations of polar plot:

- 1 The addition of poles and zeros to an existing system requires the recalculation of the frequency response.
- ② Calculating the frequency response in this manner is tedious and does not indicate the effect of the individual poles and zeros.

Unit: decibels(dB)

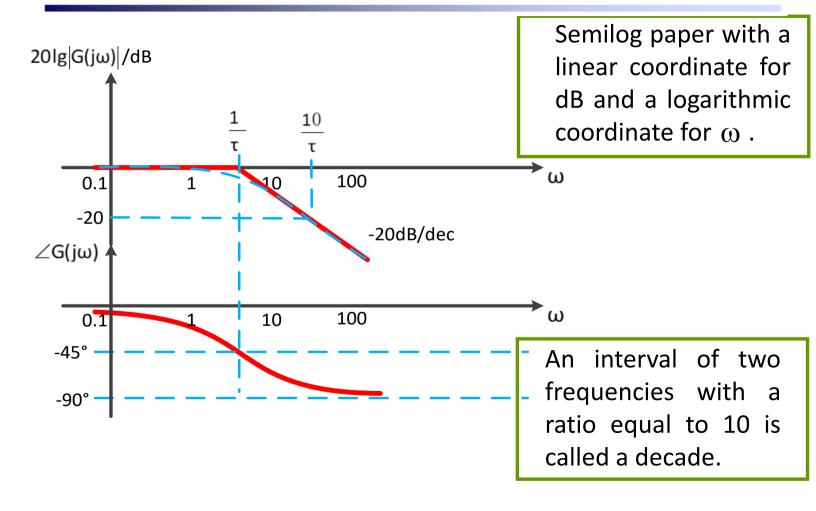
II logarithmic plots (Bode diagram)

Draw the diagram of $G(j\omega) \Rightarrow 20 \lg |G(j\omega)|$ and $\angle G(j\omega)$

Eg:
$$G(j\omega) = \frac{1}{j\omega\tau + 1}$$

 $20\lg|G(j\omega)| = 20\lg\frac{1}{\sqrt{\tau^2\omega^2 + 1}} = -10\lg(\tau^2\omega^2 + 1)$
 $\omega << \frac{1}{\tau}$ $20\lg|G(j\omega)| = 0$
 $\omega >> \frac{1}{\tau}$ $20\lg|G(j\omega)| = -20\lg\omega\tau$
 $\omega = \frac{1}{\tau}$ $20\lg|G(j\omega)| = -20\lg\frac{1}{\sqrt{2}} = -3dB$

Break frequency or corner frequency



Advantages of the logarithmic coordinate:

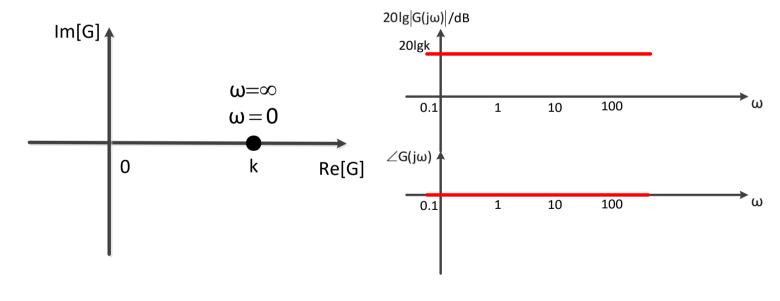
- The extension of the range of frequencies;
- ② Multiplicative factors are converted into additive factors.

Note: $\lg 0 = -\infty$, so there is no $\omega = 0$ point on the horizontal axis

Typical elements:

1 proportional element

$$G(s) = k, G(j\omega) = k, |G(j\omega)| = k, \angle G(j\omega) = 0$$



2inertial element

$$G(s) = \frac{1}{\tau s + 1}, G(j\omega) = \frac{1}{j\omega\tau + 1} = \frac{1}{\omega^2\tau^2 + 1} - j\frac{\omega\tau}{\omega^2\tau^2 + 1}$$

$$\lim[G]$$

$$\omega = \infty$$

$$\omega = 0$$

$$Re[G]$$

$$\log_{G(j\omega)}/dB$$

$$\frac{1}{\tau} \frac{10}{\tau}$$

$$\frac{1}{\tau} \frac{10}{\tau}$$

$$\frac{1}{\tau} \frac{10}{\tau}$$

$$\frac{1}{\tau} \frac{10}{\tau}$$

$$\frac{1}{\tau} \frac{10}{\tau}$$

The magnitude curve is asymptotic, the biggest error is 3dB at 1

-90°

$$\omega = \frac{1}{\tau}$$
:break frequency or corner frequency

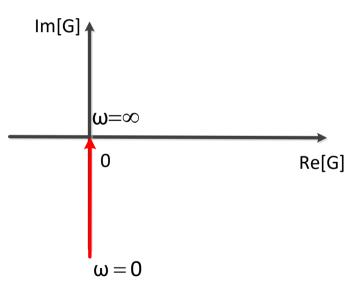
 $\omega = \omega_1$

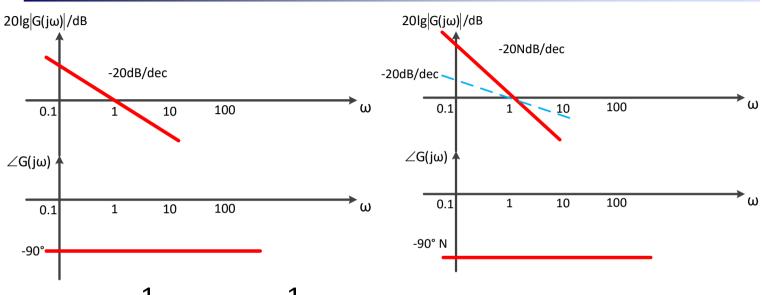
3integral element

$$G(s) = \frac{1}{s}, G(j\omega) = \frac{1}{j\omega} = -j\frac{1}{\omega}, |G(j\omega)| = \frac{1}{\omega}, \angle G(j\omega) = -90^{\circ}$$

$$\omega$$
 $|G(j\omega)|$ $\angle G(j\omega)$ 0 ∞ -90° ∞ 0 -90°

$$20 |g|G(j\omega)| = 20 |g^{\frac{1}{\omega}}| = -20 |g\omega|$$





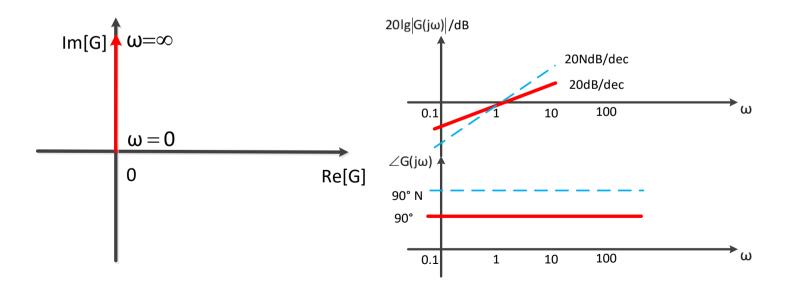
$$G(s) = \frac{1}{s^{N}}, G(j\omega) = \frac{1}{(j\omega)^{N}}, \angle G(j\omega) = \angle j\omega + \angle j\omega + \cdots = -90^{\circ} \times N$$

$$|G(j\omega)| = \frac{1}{\omega^N}$$
,20 $|g|G(j\omega)| = -20 N |g\omega|$

$oldsymbol{4}$ differential element

$$G(s) = s$$
, $G(j\omega) = j\omega$, $|G(j\omega)| = \omega$, $\angle G(j\omega) = 90^{\circ}$

$$20\lg|G(j\omega)| = 20\lg\omega$$



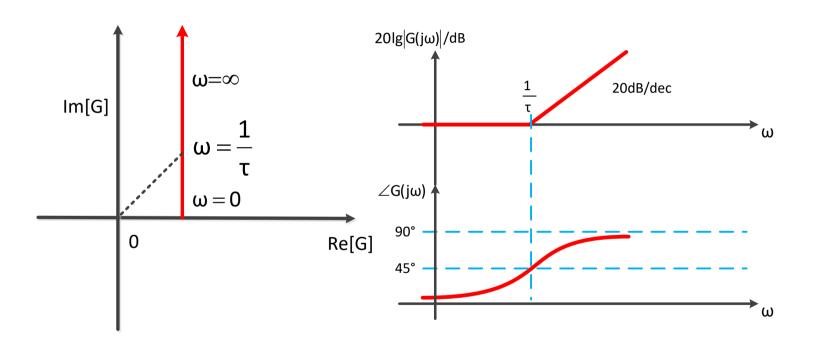
5 first-order differential element

$$G(s) = \tau s + 1, G(j\omega) = j\omega\tau + 1 \Rightarrow G(j\omega) = j\frac{\omega}{\omega_1} + 1, |G(j\omega)| = \sqrt{1 + (\frac{\omega}{\omega_1})^2}$$

$$\angle G(j\omega) = \arctan\frac{\omega}{\omega_1}, \ 20 |g|G(j\omega)| = 20 |g\sqrt{1 + (\frac{\omega}{\omega_1})^2}$$

ω	$20 \lg \sqrt{1 + (\frac{\omega}{\omega_1})^2}$
$\omega << \omega_1 = \frac{1}{\tau}$	0
$\omega >> \omega_1 = \frac{1}{\tau}$	∞

ω	G(jω)	∠G(jω)
0	1	0
$\begin{array}{c c} 1 \\ - \end{array}$	$\sqrt{2}$	45°
$egin{array}{c} au & & \ \infty & & \ \end{array}$	∞	90°



6 oscillating element

$$\begin{split} G(s) &= \frac{\omega_n^2}{s^2 + 2\xi\omega_n s + \omega_n^2} \overset{\omega_n = \frac{1}{\tau}}{\Longrightarrow} G(s) = \frac{1}{\tau^2 s^2 + 2\xi\tau s + 1} \\ G(j\omega) &= \frac{1}{-\tau^2 \omega^2 + 2\xi\tau\omega j + 1} \\ &= \frac{1 - \tau^2 \omega^2}{(1 - \tau^2 \omega^2)^2 + (2\xi\tau\omega)^2} - j \frac{2\xi\tau\omega}{(1 - \tau^2 \omega^2)^2 + (2\xi\tau\omega)^2} \end{split}$$

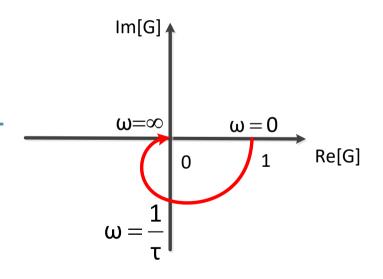
$$|G(j\omega)| = \frac{1}{\sqrt{(1-\tau^2\omega^2)^2 + (2\xi\tau\omega)^2}}$$

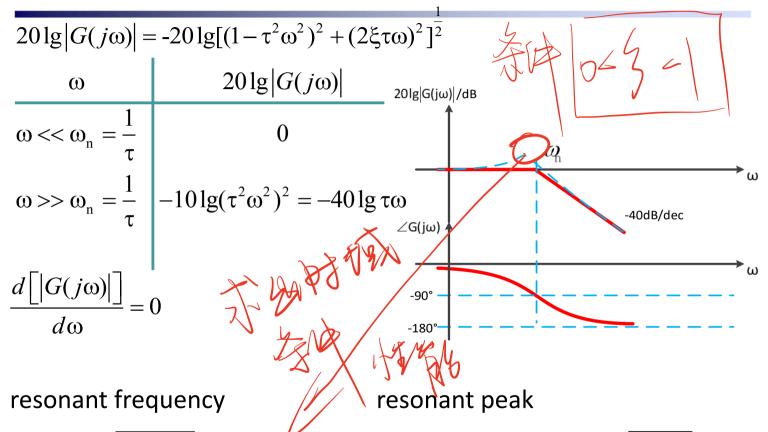
$$\angle G(j\omega) = -\arctan \frac{2\xi \tau \omega}{1 - \tau^2 \omega^2}$$

$$|G(j\omega)| = \frac{1}{\sqrt{(1-\tau^2\omega^2)^2 + (2\xi\tau\omega)^2}}$$

$$\angle G(j\omega) = -\arctan \frac{2\xi \tau \omega}{1 - \tau^2 \omega^2}$$

ω	$ G(j\omega) $	$\angle G(j\omega)$
0	1	0
$\omega = \omega_n = \frac{1}{\tau}$	$\frac{1}{2\xi}$	-90°
∞	0	-180°





 $M_r = |G(j\omega_r)| = (2\xi\sqrt{1-\xi^2})^{-1}$

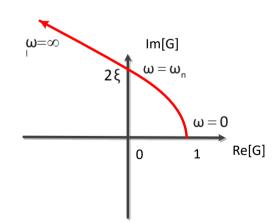
 $\omega_r = \omega_n \sqrt{1 - 2\xi^2} (\xi < 0.707)$

(7)second-order differential element

$$G(s) = \frac{s^2}{\omega_n^2} + 2\xi \frac{s}{\omega_n} + 1 = \frac{s^2 + 2\xi \omega_n s + \omega_n^2}{\omega_n^2} \Longrightarrow$$

$$G(j\omega) = \frac{-\omega^2 + 2\xi\omega_n\omega j + \omega_n^2}{\omega_n^2} = 1 - \frac{\omega^2}{\omega_n^2} + j\frac{2\xi\omega}{\omega_n}$$

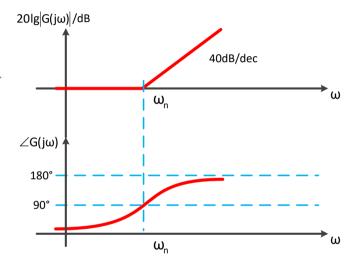
ω	Re	Im
0	1	0
ω_{n}	0	2ξ
∞	$-\infty$	∞



$$|G(j\omega)| = \sqrt{(1 - \frac{\omega^2}{\omega_n^2})^2 + (\frac{2\xi\omega}{\omega_n})^2}$$

$$|20 \lg |G(j\omega)| = 20 \lg \sqrt{(1 - \frac{\omega^2}{\omega_n^2})^2 + (\frac{2\xi\omega}{\omega_n})^2}$$

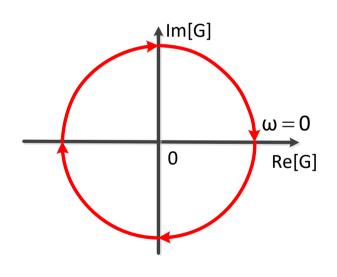
$$\angle G(j\omega) = \arctan \frac{2\xi \frac{\omega}{\omega_n}}{1 - \frac{\omega^2}{\omega_n^2}}$$

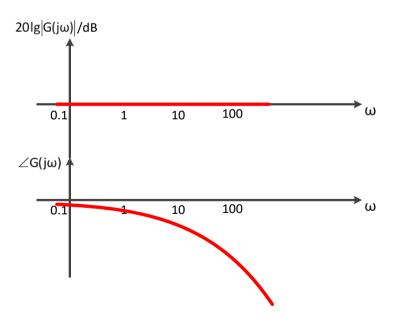


8 delay element

$$G(s) = e^{-\tau s} \stackrel{s=j\omega}{\Longrightarrow} G(j\omega) = e^{-\tau j\omega}, |G(j\omega)| = 1, \angle G(j\omega) = -\tau \omega$$

$$20 \lg |G(j\omega)| = 0$$



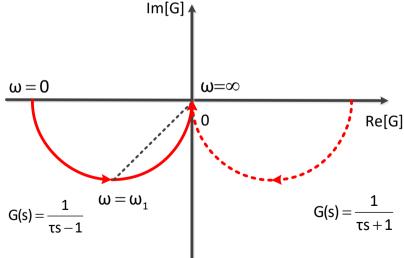


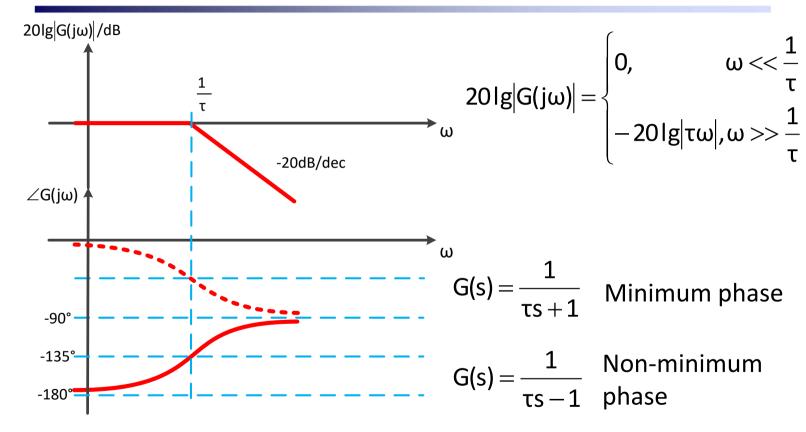
9 nonminimum phase element

$$G(s) = \frac{1}{\tau s - 1} \Longrightarrow G(j\omega) = \frac{1}{\tau \omega j - 1}, |G(j\omega)| = \frac{1}{\sqrt{\tau^2 \omega^2 + 1}}$$

$$\angle G(j\omega) = -\arctan\frac{\tau\omega}{-1} = -\pi + \arctan\tau\omega$$

ω	G(jω)	∠G(jω)
0	1	−180°
1	$\frac{1}{\sqrt{2}}$	-135°
$oldsymbol{ au}$	$\sqrt{2}$	-90°





The range of phase shift of a minimum phase transfer function is the minimum corresponding to a given amplitude curve.

Example

$$G(s) = \frac{100(0.1s + 1)}{s^2(0.01s + 1)}$$
 First we divide it into 4 typical elements

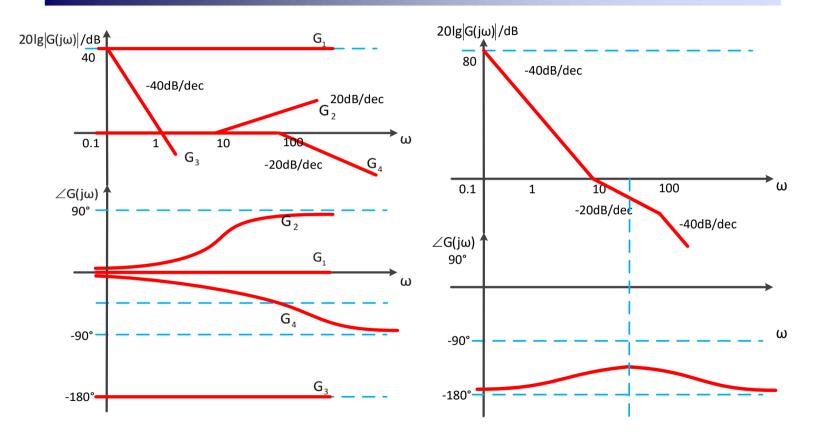
$$G_{1}(s) = 100, G_{1}(j\omega) = 100, |G_{1}| = 100, \angle G_{1} = 0$$

$$G_{2}(s) = 0.1s + 1, G_{2}(j\omega) = 0.1j\omega + 1, |G_{2}| = \sqrt{1 + 0.1^{2}\omega^{2}}, \ \angle G_{2} = tg^{-1}0.1\omega$$

$$G_{3}(s) = \frac{1}{s^{2}}, G_{3}(j\omega) = \frac{1}{-\omega^{2}}, |G_{3}| = \frac{1}{\omega^{2}}, G_{3} = -180^{\circ}$$

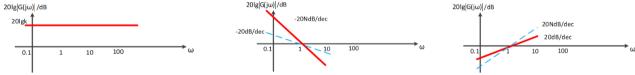
$$G_{4}(s) = \frac{1}{0.01s + 1}, G_{4}(j\omega) = \frac{1}{0.01j\omega + 1},$$

$$|G_4| = \frac{1}{\sqrt{1 + 0.01^2 \omega^2}}, \angle G_4 = -tg^{-1}0.01\omega$$



Steps:

- ①Locate break frequencies on the horizontal axis;
- ②In low frequency domain, the magnitude curve is determined by proportional element, integral element or differential element.



③Based on the low frequency domain, the slope changes at every break frequencies.

Inertial element(-20dB/dec); first order differential element(+20dB/dec) oscillating element(-40dB/dec); second order differential element(+40dB/dec)

crossover frequency ω_c

when
$$\omega = \omega_c$$
, $20 |g|G(j\omega_c) = 0$, $|G(j\omega_c)| = 1$

Examples:

es:
$$G(s) = \frac{100}{(s+1)(s+10)}$$

$$\int \int \int ds ds$$

$$G(s) = \frac{10(s+10)}{s(0.01s+1)}$$

$$G(s)|_{s=i\omega} = G(j\omega) = R(\omega) + jX(\omega) = |G(j\omega)|e^{\angle G(j\omega)}$$
 $\omega: 0 \to \infty$

- $\bigcirc 1 \quad \omega \rightarrow 0$
- $\bigcirc \omega \rightarrow \infty$
- 3 Analyze the tendency at the middle frequency segments
- 4 Key point (where $G(j\omega)$ intersects the negative real axis)

Example:

$$G(s) = \frac{k(\tau s + 1)}{s^{2}(Ts + 1)}$$

$$G(j\omega) = \frac{k(\tau j\omega + 1)}{-\omega^{2}(Tj\omega + 1)}, |G(j\omega)| = \frac{k\sqrt{\tau^{2}\omega^{2} + 1}}{\omega^{2}\sqrt{T^{2}\omega^{2} + 1}},$$

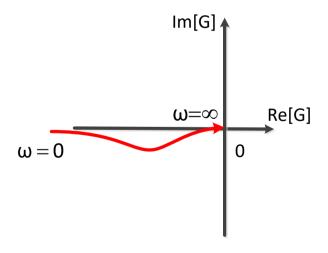
$$\angle G(j\omega) = \arctan \tau \omega - 180^{\circ} - \arctan T\omega$$

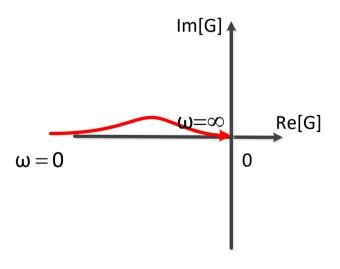
ω	G(jω)	∠G(jω)
0	∞	-180°
∞	0	-180°

Then find the tendency at the middle frequency segments

① $\tau > T$, arctanω $\tau >$ arctanTω

 2τ < T, arctanωτ < arctanTω





Example:

 ∞

$$G(j\omega) = \frac{k}{j\omega(T_{m}j\omega + 1)(T_{c}j\omega + 1)(T_{q}j\omega + 1)}$$

$$|G(j\omega)| = \frac{k}{\omega\sqrt{T_{m}^{2}\omega^{2} + 1}\sqrt{T_{c}^{2}\omega^{2} + 1}\sqrt{T_{q}^{2}\omega^{2} + 1}} \underset{lm[G]}{Im[G]}$$

$$\angle G(j\omega) = 0 - 90^{\circ} - \arctan T_{m}\omega$$

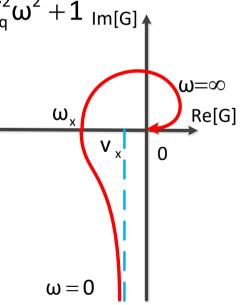
$$-\arctan T_{c}\omega - \arctan T_{q}\omega$$

$$\omega \quad |G(j\omega)| \quad \angle G(j\omega)$$

$$0 \quad \infty \quad -90^{\circ}$$

$$\infty \quad 0 \quad -360^{\circ}$$

Then determine v_x and ω_x



$$G(j\omega) = \frac{-k[\omega(T_m + T_c + T_q) - \omega^2 T_m T_c T_q]}{\omega(T_m^2 \omega^2 + 1)(T_c^2 \omega^2 + 1)(T_q^2 \omega^2 + 1)} + j \frac{-k[1 - \omega^2 (T_m T_c + T_c T_q + T_m T_q)]}{\omega(T_m^2 \omega^2 + 1)(T_c^2 \omega^2 + 1)(T_q^2 \omega^2 + 1)}$$

 ω_{x} is the most important Nyquist characteristic point

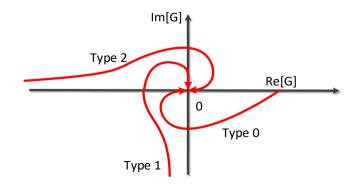
Tips: For a minimum phase system, the start phase is determined by the integral element

①
$$\omega = 0$$
 the start phase $\angle G(j0) = N \times (-90^\circ)$

② $\omega = \infty$ each non-zero pole adds the phase angle -90° each non-zero zero adds the phase angle +90°

$$n > m$$
, $\omega = \infty$, $|G(j\omega)| = 0$ and $\angle G(j\omega) = (n-m) \times (-90^\circ)$

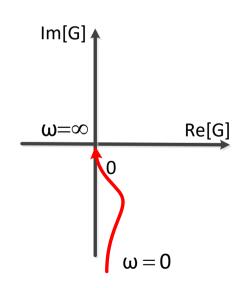
Example:
$$G(s) = \frac{10}{s^{N}(0.1s+1)^{2}}$$



Example:
$$G(s) = \frac{k(T_1 s + 1)}{s(T_2 s + 1)}(T_1 > T_2), \quad G(j\omega) = \frac{k(T_1 j\omega + 1)}{j\omega(T_2 j\omega + 1)}$$
$$|G(j\omega)| = \frac{k\sqrt{T_1^2 \omega^2 + 1}}{\omega\sqrt{T_2^2 \omega^2 + 1}}$$

$$\angle G(j\omega) = -90^{\circ} + tg^{-1}T_1\omega - tg^{-1}T_2\omega$$

ω	G(jω)	∠G(jω)
0	∞	-90°
∞	0	-90°



Example:
$$G(s) = \frac{k(s+1)}{s(s-1)}, \ G(j\omega) = \frac{k(j\omega+1)}{j\omega(j\omega-1)}, \ \left|G(j\omega)\right| = \frac{k\sqrt{\omega^2+1}}{\omega\sqrt{\omega^2+1}} = \frac{k}{\omega}$$

$$\angle G(j\omega) = -90^\circ + tg^{-1}\omega - (180^\circ - tg^{-1}\omega)$$

$$= -270^\circ + 2tg^{-1}\omega$$

$$\omega = 0 \text{ Im}[G]$$

$$\omega = |G(j\omega)| \ \angle G(j\omega)$$

$$0 \ \infty \ -270^\circ$$

$$\infty \ 0 \ -90^\circ$$
 Then determine k of the point ω_x yourself

□ Stability:

ightharpoonup Time domain: I $y(\infty) = \lim_{t \to \infty} y(t)$

II use Laplace transformation and observe the roots of CE

III Routh criterion

IV root locus method

Frequency domain:

use $G(j\omega)$ to sketch polar plot and then use Nyquist criterion