# THE SQUARE-ROOT UNSCENTED KALMAN FILTER FOR STATE AND PARAMETER-ESTIMATION

Rudolph van der Merwe and Eric A. Wan

Oregon Graduate Institute of Science and Technology 20000 NW Walker Road, Beaverton, Oregon 97006, USA {rvdmerwe,ericwan}@ece.ogi.edu

### **ABSTRACT**

Over the last 20-30 years, the extended Kalman filter (EKF) has become the algorithm of choice in numerous nonlinear estimation and machine learning applications. These include estimating the state of a nonlinear dynamic system as well estimating parameters for nonlinear system identification (e.g., learning the weights of a neural network). The EKF applies the standard linear Kalman filter methodology to a linearization of the true nonlinear system. This approach is sub-optimal, and can easily lead to divergence. Julier et al. [1] proposed the unscented Kalman filter (UKF) as a derivative-free alternative to the extended Kalman filter in the framework of state-estimation. This was extended to parameterestimation by Wan and van der Merwe [2, 3]. The UKF consistently outperforms the EKF in terms of prediction and estimation error, at an equal computational complexity of  $\mathcal{O}(L^3)^1$  for general state-space problems. When the EKF is applied to parameterestimation, the special form of the state-space equations allows for an  $\mathcal{O}(L^2)$  implementation. This paper introduces the squareroot unscented Kalman filter (SR-UKF) which is also  $\mathcal{O}(L^3)$  for general state-estimation and  $\mathcal{O}(L^2)$  for parameter estimation (note the original formulation of the UKF for parameter-estimation was  $\mathcal{O}(L^3)$ ). In addition, the square-root forms have the added benefit of numerical stability and guaranteed positive semi-definiteness of the state covariances.

#### 1. INTRODUCTION

The EKF has been applied extensively to the field of nonlinear estimation for both state-estimation and parameter-estimation. The basic framework for the EKF (and the UKF) involves estimation of the state of a discrete-time nonlinear dynamic system,

$$\mathbf{x}_{k+1} = \mathbf{F}(\mathbf{x}_k, \mathbf{u}_k) + \mathbf{v}_k$$
(1)  
$$\mathbf{y}_k = \mathbf{H}(\mathbf{x}_k) + \mathbf{n}_k,$$
(2)

$$\mathbf{y}_k = \mathbf{H}(\mathbf{x}_k) + \mathbf{n}_k, \tag{2}$$

where  $\mathbf{x}_k$  represent the unobserved state of the system,  $\mathbf{u}_k$  is a known exogenous input, and  $y_k$  is the observed measurement signal. The process noise  $\mathbf{v}_k$  drives the dynamic system, and the observation noise is given by  $\mathbf{n}_k$ . The EKF involves the recursive estimation of the mean and covariance of the state under a Gaussian assumption.

In contrast, parameter-estimation, sometimes referred to as system identification, involves determining a nonlinear mapping  $\mathbf{y}_k = \mathbf{G}(\mathbf{x}_k, \mathbf{w})$ , where  $\mathbf{x}_k$  is the input,  $\mathbf{y}_k$  is the output, and the nonlinear map,  $G(\cdot)$ , is parameterized by the vector  $\mathbf{w}$ . Typically, a training set is provided with sample pairs consisting of known input and desired outputs,  $\{\mathbf{x}_k, \mathbf{d}_k\}$ . The error of the machine is defined as  $\mathbf{e}_k = \mathbf{d}_k - \mathbf{G}(\mathbf{x}_k, \mathbf{w})$ , and the goal of learning involves solving for the parameters w in order to minimize the expectation of some given function of the error. While a number of optimization approaches exist (e.g., gradient descent and Quasi-Newton methods), parameters can be efficiently estimated on-line by writing a new state-space representation

$$\mathbf{w}_{k+1} = \mathbf{w}_k + \mathbf{r}_k \tag{3}$$

$$\mathbf{d}_k = \mathbf{G}(\mathbf{x}_k, \mathbf{w}_k) + \mathbf{e}_k, \tag{4}$$

where the parameters  $\mathbf{w}_k$  correspond to a stationary process with identity state transition matrix, driven by process noise  $\mathbf{r}_k$  (the choice of variance determines convergence and tracking performance). The output  $\mathbf{d}_k$  corresponds to a nonlinear observation on  $\mathbf{w}_k$ . The EKF can then be applied directly as an efficient "secondorder" technique for learning the parameters [4].

## 2. THE UNSCENTED KALMAN FILTER

The inherent flaws of the EKF are due to its linearization approach for calculating the mean and covariance of a random variable which undergoes a nonlinear transformation. As shown in shown in [1, 2, 3], the UKF addresses these flaws by utilizing a deterministic "sampling" approach to calculate mean and covariance terms. Essentially, 2L + 1, sigma points (L is the state dimension), are chosen based on a square-root decomposition of the prior covariance. These sigma points are propagated through the true nonlinearity, without approximation, and then a weighted mean and covariance is taken. A simple illustration of the approach is shown in Figure 1 for a 2-dimensional system: the left plot shows the true mean and covariance propagation using Monte-Carlo sampling; the center plots show the results using a linearization approach as would be done in the EKF; the right plots show the performance of the new "sampling" approach (note only 5 sigma points are required). This approach results in approximations that are accurate to the third order (Taylor series expansion) for Gaussian inputs for all nonlinearities. For non-Gaussian inputs, approximations are accurate to at least the second-order [1]. In contrast, the linearization approach of the EKF results only in first order accuracy.

The full UKF involves the recursive application of this "sampling" approach to the state-space equations. The standard UKF implementation is given in Algorithm 2.1 for state-estimation, and uses the following variable definitions:  $\{W_i\}$  is a set of scalar weights  $(W_0^{(m)} = \lambda/(L+\lambda), W_0^{(c)} = \lambda/(L+\lambda) + (1-\alpha^2+\beta)$ 

This work was sponsored in part by NSF under grant grant IRI-9712346, ECS-0083106, and DARPA under grant F33615-98-C-3516.

<sup>&</sup>lt;sup>1</sup>L is the dimension of the state variable.

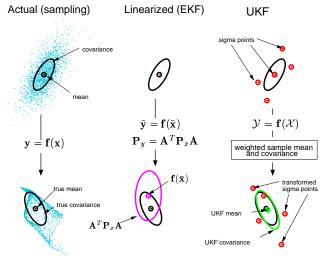


Figure 1: Example of mean and covariance propagation. a) actual, b) first-order linearization (EKF), c) new "sampling" approach (UKF).

,  $W_i^{(m)}=W_i^{(c)}=1/\{2(L+\lambda)\}$   $i=1,\ldots,2L$ ).  $\lambda=L(\alpha^2-1)$  and  $\eta=\sqrt{(L+\lambda)}$  are scaling parameters. The constant  $\alpha$  determines the spread of the sigma points around  $\hat{\mathbf{x}}$  and is usually set to  $1e-4\leq\alpha\leq1$ .  $\beta$  is used to incorporate prior knowledge of the distribution of  $\mathbf{x}$  (for Gaussian distributions,  $\beta=2$  is optimal). Also note that we define the linear algebra operation of adding a column vector to a matrix, i.e.  $\mathbf{A}\pm\mathbf{u}$  as the addition of the vector to each column of the matrix. The superior performance of the UKF over the EKF has been demonstrated in a number of applications [1, 2, 3]. Furthermore, unlike the EKF, no explicit derivatives (i.e., Jacobians or Hessians) need to be calculated.

## 3. EFFICIENT SQUARE-ROOT IMPLEMENTATION

The most computationally expensive operation in the UKF corresponds to calculating the new set of sigma points at each time update. This requires taking a matrix square-root of the state covariance  $\operatorname{matrix}^2, \mathbf{P} \in \mathbb{R}^{L \times L}$ , given by  $\mathbf{SS}^T = \mathbf{P}$ . An efficient implementation using a Cholesky factorization requires in general  $\mathcal{O}(L^3/6)$  computations [5]. While the square-root of  $\mathbf{P}$  is an integral part of the UKF, it is still the full covariance  $\mathbf{P}$  which is recursively updated. In the SR-UKF implementation,  $\mathbf{S}$  will be propagated directly, avoiding the need to refactorize at each time step. The algorithm will in general still be  $\mathcal{O}(L^3)$ , but with improved numerical properties similar to those of standard square-root Kalman filters [6]. Furthermore, for the special state-space formulation of parameter-estimation, an  $\mathcal{O}(L^2)$  implementation becomes possible.

The square-root form of the UKF makes use of three powerful linear algebra techniques<sup>3</sup>, *QR decomposition*, *Cholesky factor updating* and *efficient least squares*, which we briefly review below:

• *QR decomposition.* The QR decomposition or factorization of a matrix  $\mathbf{A} \in \mathbb{R}^{L \times N}$  is given by,  $\mathbf{A}^T = \mathbf{Q}\mathbf{R}$ , where  $\mathbf{Q} \in \mathbb{R}^{N \times N}$  is orthogonal,  $\mathbf{R} \in \mathbb{R}^{N \times L}$  is upper triangular and  $N \geq L$ . The upper triangular part of  $\mathbf{R}$ ,  $\tilde{\mathbf{R}}$ , is the transpose of the Cholesky factor of  $\mathbf{P} = \mathbf{A}\mathbf{A}^T$ , *i.e.*,

Initialize with:

$$\hat{\mathbf{x}}_0 = \mathbb{E}[\mathbf{x}_0] \quad \mathbf{P}_0 = \mathbb{E}[(\mathbf{x}_0 - \hat{\mathbf{x}}_0)(\mathbf{x}_0 - \hat{\mathbf{x}}_0)^T]$$
 (5)

For  $k \in \{1, \ldots, \infty\}$ ,

Calculate sigma points:

$$\mathcal{X}_{k-1} = \begin{bmatrix} \hat{\mathbf{x}}_{k-1} & \hat{\mathbf{x}}_{k-1} + \eta \sqrt{\mathbf{P}_{k-1}} & \hat{\mathbf{x}}_{k-1} - \eta \sqrt{\mathbf{P}_{k-1}} \end{bmatrix}$$
 (6)

Time update:

$$\boldsymbol{\mathcal{X}}_{k|k-1} = \mathbf{F}[\boldsymbol{\mathcal{X}}_{k-1}, \mathbf{u}_{k-1}] \tag{7}$$

$$\hat{\mathbf{x}}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(m)} \mathcal{X}_{i,k|k-1}$$
(8)

$$\mathbf{P}_k^- = \sum_{i=0}^{2L} W_i^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_k^-] [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_k^-]^T + \mathbf{R}^{\mathbf{v}}$$

$$\mathbf{\mathcal{Y}}_{k|k-1} = \mathbf{H}[\mathbf{\mathcal{X}}_{k|k-1}] \tag{9}$$

$$\hat{\mathbf{y}}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(m)} \mathcal{Y}_{i,k|k-1}$$
 (10)

Measurement update equations:

$$\mathbf{P}_{\hat{\mathbf{y}}_k\hat{\mathbf{y}}_k} = \sum_{i=0}^{2L} W_i^{(c)} [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_k^-] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_k^-]^T + \mathbf{R}^\mathbf{n}$$

$$\mathbf{P}_{\mathbf{x}_{k}\mathbf{y}_{k}} = \sum_{i=0}^{2L} W_{i}^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_{k}^{-}] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_{k}^{-}]^{T}$$
(11)

$$\mathcal{K}_k = \mathbf{P}_{\mathbf{x}_k \mathbf{y}_k} \mathbf{P}_{\tilde{\mathbf{y}}_k \tilde{\mathbf{y}}_k}^{-1}$$
 (12)

$$\hat{\mathbf{x}}_k = \hat{\mathbf{x}}_k^- + \mathcal{K}_k (\mathbf{y}_k - \hat{\mathbf{y}}_k^-) \tag{13}$$

$$\mathbf{P}_k = \mathbf{P}_k^- - \mathcal{K}_k \mathbf{P}_{\tilde{\mathbf{y}}_k \tilde{\mathbf{y}}_k} \mathcal{K}_k^T \tag{14}$$

where  $\mathbf{R}^{\mathbf{v}}$ =process noise cov.,  $\mathbf{R}^{\mathbf{n}}$ =measurement noise cov.

Algorithm 2.1: Standard UKF algorithm.

 $\tilde{\mathbf{R}} = \mathbf{S}^T$ , such that  $\tilde{\mathbf{R}}^T \tilde{\mathbf{R}} = \mathbf{A} \mathbf{A}^T$ . We use the shorthand notation  $\operatorname{qr}\{\cdot\}$  to donate a QR decomposition of a matrix where only  $\tilde{\mathbf{R}}$  is returned. The computational complexity of a QR decomposition is  $\mathcal{O}(NL^2)$ . Note that performing a Cholesky factorization directly on  $\mathbf{P} = \mathbf{A} \mathbf{A}^T$  is  $\mathcal{O}(L^3/6)$  plus  $\mathcal{O}(NL^2)$  to form  $\mathbf{A} \mathbf{A}^T$ .

- Cholesky factor updating. If  $\mathbf{S}$  is the original Cholesky factor of  $\mathbf{P} = \mathbf{A}\mathbf{A}^T$ , then the Cholesky factor of the rank-1 update (or downdate)  $\mathbf{P} \pm \sqrt{\nu}\mathbf{u}\mathbf{u}^T$  is denoted as  $\mathbf{S} = \text{cholupdate}\{\mathbf{S},\mathbf{u},\pm\nu\}$ . If  $\mathbf{u}$  is a matrix and not a vector, then the result is M consecutive updates of the Cholesky factor using the M columns of  $\mathbf{u}$ . This algorithm (available in Matlab as cholupdate) is only  $\mathcal{O}(L^2)$  per update.
- Efficient least squares. The solution to the equation
   (AA<sup>T</sup>)x = A<sup>T</sup>b also corresponds to the solution of the overdetermined least squares problem Ax = b. This can be solved efficiently using a QR decomposition with pivoting (implemented in Matlab's '/' operator).

The complete specification of the new square-root filters is given in Algorithm 3.1 for state-estimation and 3.2 for paramater-

<sup>&</sup>lt;sup>2</sup>For notational clarity, the time index k has been omitted.

<sup>&</sup>lt;sup>3</sup>See [5] for theoretical and implementation details.

estimation. Below we describe the key parts of the square-root algorithms, and how they contrast with the stardard implementations

**Square-Root State-Estimation:** As in the original UKF, the filter is initialized by calculating the matrix square-root of the state covariance once via a Cholesky factorization (Eqn. 16). However, the propagted and updated Cholesky factor is then used in subsequent iterations to directly form the sigma points. In Eqn. 20 the time-update of the Cholesky factor,  $\mathbf{S}^-$ , is calculated using a QR decompostion of the compound matrix containing the weighted propagated sigma points and the matrix square-root of the additive process noise covariance. The subsequent Cholesky update (or downdate) in Eqn. 21 is necessary since the the zero'th weight,  $W_0^{(c)}$ , may be negative. These two steps replace the time-update of  $\mathbf{P}^-$  in Eqn. 9, and is also  $\mathcal{O}(L^3)$ .

The same two-step approach is applied to the calculation of the Cholesky factor,  $\mathbf{S}_{\tilde{\mathbf{y}}}$ , of the observation-error covariance in Eqns. 24 and 25. This step is  $\mathcal{O}(LM^2)$ , where M is the observation dimension. In contrast to the way the Kalman gain is calculated in the standard UKF (see Eqn. 12), we now use two nested inverse (or *least squares*) solutions to the following expansion of Eqn. 12,  $\mathcal{K}_k(\mathbf{S}_{\tilde{\mathbf{y}}_k}\mathbf{S}_{\tilde{\mathbf{y}}_k}^T) = \mathbf{P}_{\mathbf{x}_k\mathbf{y}_k}$ . Since  $\mathbf{S}_{\tilde{\mathbf{y}}}$  is square and triangular, efficient "back-substitutions" can be used to solve for  $\mathcal{K}_k$  directly without the need for a matrix inversion.

Finally, the posterior measurement update of the Cholesky factor of the state covariance is calculated in Eqn. 29 by applying M sequential Cholesky downdates to  $\mathbf{S}_k^-$ . The downdate vectors are the columns of  $\mathbf{U} = \mathcal{K}_k \mathbf{S}_{\hat{\mathbf{y}}_k}$ . This replaces the posterior update of  $\mathbf{P}_k$  in Eqn. 14, and is also  $\mathcal{O}(LM^2)$ .

**Square-Root Parameter-Estimation:** The parameter-estimation algorithm follows a similar framework as that of the state-estimation square-root UKF. However, an  $\mathcal{O}(ML^2)$  algorithm, as opposed to  $\mathcal{O}(L^3)$ , is possible by taking advantage of the *linear* state transition function. Specifically, the time-update of the state covariance is given simply by  $\mathbf{P}_{\mathbf{w}_k}^- = \mathbf{P}_{\mathbf{w}_{k-1}} + \mathbf{R}_{k-1}^{\mathbf{r}}$ . Now, if we apply an exponential weighting on past data<sup>4</sup>, the process noise covariance is given by  $\mathbf{R}_k^{\mathbf{r}} = (\gamma^{-1} - 1)\mathbf{P}_{\mathbf{w}_k}$ , and the time update of the state covariance becomes,

$$\mathbf{P}_{\mathbf{w}_{k}}^{-} = \mathbf{P}_{\mathbf{w}_{k-1}} + (\gamma^{-1} - 1)\mathbf{P}_{\mathbf{w}_{k-1}} = \gamma^{-1}\mathbf{P}_{\mathbf{w}_{k-1}}.$$
 (15)

This translates readily into the factored form,  $\mathbf{S}_{\mathbf{w}_k}^- = \gamma^{-1/2} \mathbf{S}_{\mathbf{w}_{k-1}}$  (see Eqn. 32), and avoids the costly  $\mathcal{O}(L^3)$  QR and Cholesky based updates necessary in the state-estimation filter.

## 4. EXPERIMENTAL RESULTS

The improvement in error performance of the UKF over that of the EKF for both state and parameter-estimation is well documented [1, 2, 3]. The focus of this section will be to simply verify the equivalent error performance of the UKF and SR-UKF, and show the reduction in computational cost achieved by the SR-UKF for parameter-estimation. Figure 2 shows the superior performance of UKF and SR-UKF compared to that of the EKF on estimating the Mackey-Glass-30 chaotic time series corrupted by additive white noise (3dB SNR). The error performance of the SR-UKF and UKF are indistinguishable and are both superior to the EKF. The computational complexity of all three filters are of the same order but

Initialize with:

$$\hat{\mathbf{x}}_0 = \mathbb{E}[\mathbf{x}_0] \quad \mathbf{S}_0 = \text{chol}\left\{\mathbb{E}[(\mathbf{x}_0 - \hat{\mathbf{x}}_0)(\mathbf{x}_0 - \hat{\mathbf{x}}_0)^T]\right\}$$
 (16)

For  $k \in \{1, \ldots, \infty\}$ ,

Sigma point calculation and time update:

$$\boldsymbol{\mathcal{X}}_{k-1} = \begin{bmatrix} \hat{\mathbf{x}}_{k-1} & \hat{\mathbf{x}}_{k-1} + \eta \mathbf{S}_k & \hat{\mathbf{x}}_{k-1} - \eta \mathbf{S}_k \end{bmatrix}$$
(17)

$$\boldsymbol{\mathcal{X}}_{k|k-1} = \mathbf{F}[\boldsymbol{\mathcal{X}}_{k-1}, \mathbf{u}_{k-1}] \tag{18}$$

$$\hat{\mathbf{x}}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(m)} \mathcal{X}_{i,k|k-1}$$
(19)

$$\mathbf{S}_{k}^{-} = \operatorname{qr} \left\{ \left[ \sqrt{W_{1}^{(c)}} \left( \boldsymbol{\mathcal{X}}_{1:2L,k|k-1} - \hat{\mathbf{x}}_{k}^{-} \right) \quad \sqrt{\mathbf{R^{v}}} \right] \right\}$$

$$\mathbf{S}_{k}^{-} = \text{cholupdate} \left\{ \mathbf{S}_{k}^{-}, \ \mathcal{X}_{0,k} - \hat{\mathbf{x}}_{k}^{-}, \ W_{0}^{(c)} \right\}$$
(20)

$$\mathbf{\mathcal{Y}}_{k|k-1} = \mathbf{H}[\mathbf{\mathcal{X}}_{k|k-1}] \tag{22}$$

$$\hat{\mathbf{y}}_{k}^{-} = \sum_{i=0}^{2L} W_{i}^{(m)} \mathcal{Y}_{i,k|k-1}$$
(23)

Measurement update equations:

$$\mathbf{S}_{\tilde{\mathbf{y}}_k} = \operatorname{qr} \left\{ \left[ \sqrt{W_1^{(c)}} \left[ \mathbf{y}_{1:2L,k} - \hat{\mathbf{y}}_k \right] \right] \sqrt{\mathbf{R}_k^{\mathbf{n}}} \right] \right\}$$
 (24)

$$\mathbf{S}_{\hat{\mathbf{y}}_k} = \text{cholupdate} \left\{ \mathbf{S}_{\hat{\mathbf{y}}_k} , \ \mathcal{Y}_{0,k} - \hat{\mathbf{y}}_k , \ W_0^{(c)} \right\}$$
 (25)

$$\mathbf{P}_{\mathbf{x}_{k}\mathbf{y}_{k}} = \sum_{i=0}^{2L} W_{i}^{(c)} [\mathcal{X}_{i,k|k-1} - \hat{\mathbf{x}}_{k}^{-}] [\mathcal{Y}_{i,k|k-1} - \hat{\mathbf{y}}_{k}^{-}]^{T}$$
 (26)

$$\mathcal{K}_{k} = (\mathbf{P}_{\mathbf{x}_{h}, \mathbf{y}_{h}} / \mathbf{S}_{\tilde{\mathbf{y}}_{h}}^{T}) / \mathbf{S}_{\tilde{\mathbf{y}}_{h}}$$
(27)

$$\hat{\mathbf{x}}_k = \hat{\mathbf{x}}_k^- + \mathcal{K}_k(\mathbf{y}_k - \hat{\mathbf{y}}_k^-)$$

$$\mathbf{U} = \mathcal{K}_k \mathbf{S}_{\hat{\mathbf{y}}_k} \tag{28}$$

$$\mathbf{S}_k = \text{cholupdate} \left\{ \mathbf{S}_k^- , \mathbf{U} , -1 \right\}$$
 (29)

where  $\mathbf{R}^{\mathbf{v}}$ =process noise cov.,  $\mathbf{R}^{\mathbf{n}}$ =measurement noise cov.

Algorithm 3.1: Square-Root UKF for state-estimation.

the SR-UKF is about 20% faster than the UKF and about 10% faster than the EKF.

The next experiment shows the reduction in computational cost achieved by the square-root unscented Kalman filters and how that compares to the computational complexity of the EKF for parameter-estimation. For this experiment, we use an EKF, UKF and SR-UKF to train a 2-12-2 MLP neural network on the well known Mackay-Robot-Arm<sup>5</sup> benchmark problem of mapping the joint angles of a robot arm to the Cartesian coordinates of the hand. The learning curves (mean square error (MSE) vs. learning epoch) of the different filters are shown in Figure 3. Figure 4 shows how the computational complexity of the different filters scale as a function of the number of parameters (weights in neural network). While the standard UKF is  $\mathcal{O}(L^3)$ , both the EKF and SR-UKF are  $\mathcal{O}(L^2)$ .

 $<sup>^4 \</sup>text{This}$  is identical to the approach used in weighted recursive least squares (W-RLS).  $\gamma$  is a scalar weighting factor chosen to be slightly less than 1, *i.e.*  $\gamma=0.9995$ .

<sup>&</sup>lt;sup>5</sup>http://wol.ra.phy.cam.ac.uk/mackay

Initialize with:

$$\hat{\mathbf{w}}_0 = E[\mathbf{w}] \quad \mathbf{S}_{\mathbf{w}_0} = \text{chol}\left\{E[(\mathbf{w} - \hat{\mathbf{w}}_0)(\mathbf{w} - \hat{\mathbf{w}}_0)^T]\right\} \quad (30)$$

For  $k \in \{1, \ldots, \infty\}$ ,

Time update and sigma point calculation:

$$\hat{\mathbf{w}}_{k}^{-} = \hat{\mathbf{w}}_{k-1} \tag{31}$$

$$\mathbf{S}_{\mathbf{w}_k}^- = \gamma^{-1/2} \mathbf{S}_{\mathbf{w}_{k-1}} \tag{32}$$

$$\boldsymbol{\mathcal{W}}_{k|k-1} = \begin{bmatrix} \hat{\mathbf{w}}_k^- & \hat{\mathbf{w}}_k^- + \eta \mathbf{S}_{\mathbf{w}_k}^- & \hat{\mathbf{w}}_k^- - \eta \mathbf{S}_{\mathbf{w}_k}^- \end{bmatrix}$$
(33)

$$\mathcal{D}_{k|k-1} = \mathbf{G}[\mathbf{x}_k, \mathcal{W}_{k|k-1}]$$
(34)

$$\hat{\mathbf{d}}_k = \sum_{i=0}^{2L} W_i^{(m)} \mathcal{D}_{i,k|k-1}$$
 (35)

Measurement update equations:

$$\mathbf{S}_{\mathbf{d}_{k}} = \operatorname{qr} \left\{ \left[ \sqrt{W_{1}^{(c)}} \left[ \mathbf{\mathcal{D}}_{1:2L,k} - \hat{\mathbf{d}}_{k} \right] \quad \sqrt{\mathbf{R}^{\mathbf{e}}} \right] \right\}$$
(36)

$$\mathbf{S}_{\mathbf{d}_{k}} = \text{cholupdate} \left\{ \mathbf{S}_{\mathbf{d}_{k}}, \ \mathcal{D}_{0,k} - \hat{\mathbf{d}}_{k}, \ W_{0}^{(c)} \right\}$$
 (37)

$$\mathbf{P}_{\mathbf{w}_{k}\mathbf{d}_{k}} = \sum_{i=0}^{2L} W_{i}^{(c)} [\mathcal{W}_{i,k|k-1} - \hat{\mathbf{w}}_{k}^{-}] [\mathcal{D}_{i,k|k-1} - \hat{\mathbf{d}}_{k}]^{T}$$
 (38)

$$\mathcal{K}_k = (\mathbf{P}_{\mathbf{w}_k \mathbf{d}_k} / \mathbf{S}_{\mathbf{d}_k}^T) / \mathbf{S}_{\mathbf{d}_k}$$
 (39)

$$\hat{\mathbf{w}}_k = \hat{\mathbf{w}}_k^- + \mathcal{K}_k (\mathbf{d}_k - \hat{\mathbf{d}}_k) \tag{40}$$

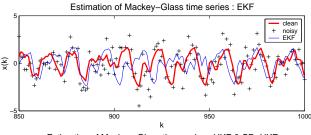
$$\mathbf{U} = \mathcal{K}_k \mathbf{S}_{\mathbf{d}_k} \tag{41}$$

(42)

$$\mathbf{S}_{\mathbf{w}_k} = \text{cholupdate} \left\{ \mathbf{S}_{\mathbf{w}_k}^- \; , \; \mathbf{U} \; , \; -1 \right\}$$

where  $\mathbf{R}^{\mathbf{e}}$ =measurement noise cov (this can be set to an arbitrary value, e.g., .5**I**.)

Algorithm 3.2: Square-Root UKF for parameter-estimation.



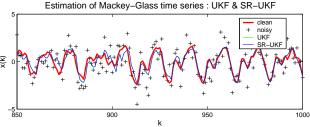


Figure 2: Estimation of the Mackey-Glass chaotic time-series with the EKF, UKF and SR-UKF.

### 5. CONCLUSIONS

The UKF consistently performs better than or equal to the well known EKF, with the added benefit of ease of implementation in

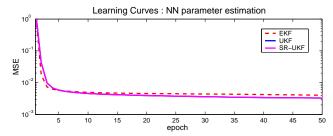


Figure 3: Learning curves for Mackay-Robot-Arm neural network parameter-estimation problem.

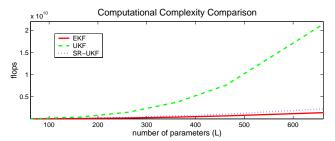


Figure 4: Computational complexity (flops/epoch) of EKF, UKF and SR-UKF for parameter-estimation (Mackay-Robot-Arm problem).

that no analytical derivatives (Jacobians or Hessians) need to be calculated. For state-estimation, the UKF and EKF have equal complexity and are in general  $\mathcal{O}(L^3)$ . In this paper, we introduced square-root forms of the UKF. The square-root UKF has better numerical properties and guarantees positive semi-definiteness of the underlying state covariance. In addition, for parameter-estimation an efficient  $\mathcal{O}(L^2)$  implementation is possible for the square-root form, which is again of the same complexity as efficient EKF parameter-estimation implementations. In this light, the SR-UKF is the logical replacement for the EKF in all state and parameter-estimation applications.

## 6. REFERENCES

- [1] S. J. Julier and J. K. Uhlmann, "A New Extension of the Kalman Filter to Nonlinear Systems," in *Proc. of AeroSense:* The 11th Int. Symp. on Aerospace/Defence Sensing, Simulation and Controls., 1997.
- [2] E. Wan, R. van der Merwe, and A. T. Nelson, "Dual Estimation and the Unscented Transformation," in *Neural Information Processing Systems* 12. 2000, pp. 666–672, MIT Press.
- [3] E. A. Wan and R. van der Merwe, "The Unscented Kalman Filter for Nonlinear Estimation," in *Proc. of IEEE Symposium* 2000 (AS-SPCC), Lake Louise, Alberta, Canada, Oct. 2000.
- [4] G.V. Puskorius and L.A. Feldkamp, "Decoupled Extended Kalman Filter Training of Feedforward Layered Networks," in *IJCNN*, 1991, vol. 1, pp. 771–777.
- [5] W. H. Press, S. A. Teukolsky, W. T. Vetterling, and B. P. Flannery, *Numerical Recipes in C: The Art of Scientific Computing*, Cambridge University Press, 2 edition, 1992.
- [6] A. H. Sayed and T. Kailath, "A State-Space Approach to Adaptive RLS Filtering," *IEEE Sig. Proc. Mag.*, pp. 18–60, July 1994.