

# Fault Detection of an Inverted Pendulum on a Cart

## Plant Model

- Control voltage to a DC motor delivers torque to the cart's drive wheel
- Position of the cart is measured using a circular coil potentiometer
- Velocity of the cart is measured using a tacho generator mounted on the drive wheel
- Angle of pendulum is determined via a potentiometer
- Angular velocity is not measured

$$\text{System States: } x = \begin{bmatrix} \text{cart position} \\ \text{cart velocity} \\ \text{pendulum angle} \\ \text{pendulum angular velocity} \end{bmatrix} = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix}$$

## Plant Parameters

```
M0 = 3.2;           % [kg]           Pendulum mass
M = 3.529;          % [kg]           Total mass pendulums and the cart
Theta = 0.072;      % [kg.m^2]       Moment of inertia of the pendulum rod
ls = 0.44;          % [m]           Rod length
C = 0.009;          % [kg.m^2/s]     Friction constant of the pendulum
g = 9.81;           % [m/s^2]       Acceleration of gravity
Fr = 6.2;           % [kg/s]        Static friction constant of the cart

% Kr = 2.6;

N = 0.1446;         % [kg.m]         Damping constant at the joints of the rod
N01_2 = 0.23315;   % [kg^2.m^2]     Damping constant at the joints of the cart
```

## Fault Model

We assume a possible fault in the actuator (DC motor drive)  $f_A$ , as well as a possible fault in each of the sensors  $f_{s1}$ ,  $f_{s2}$ , and  $f_{s3}$ .

$$f = \begin{bmatrix} f_{A1} \\ f_{s1} \\ f_{s2} \\ f_{s3} \end{bmatrix}$$

## Linearized Discrete-Tme Model

The system is linearized with a sampling time of  $0.03s$ , at the operating point:

$$x = \begin{bmatrix} r \\ \dot{r} \\ \phi \\ \dot{\phi} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}$$

The linearized model is valid under the following conditions:

- Control signal,  $|F| \leq 20N$
- Cart position,  $|r| \leq 0.5m$
- Cart velocity,  $|\dot{r}| \leq 1m/s$
- Pendulum angle,  $|\phi| \leq \frac{1}{18}\pi \text{ rad}$

### System Model

$$\dot{x} = Ax + Bu + E_d d + E_f f$$

$$y = Cx + F_f f$$

with:

- state vector,  $x$
- input vector,  $u$
- disturbance vector,  $d$
- fault vector,  $f$

```
A= [
    1, -0.0569,  0.0010,  0.0000;
    0,  0.9442, -0.0038,  0.0000;
    0,  0.0116,  1.0097,  0.0300;
    0,  0.7688,  0.6442,  1.0056
];

B= [ 0.0053; -0.1789; 0.0373; 2.4632];

E_d= B;

E_f= [B, zeros(4,3) ];

C= [diag([1,1,1]), zeros(3,1)];

D= zeros(3,1);

F_d= zeros(3,1);

F_f= [zeros(3,1), diag([1,1,1]) ];

% Sampling time
T_s= 0.03;
```

## State Feedback Control

Because the system is unstable at the linearization point  $\begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}$ , we stabilize using a state feedback controller.

We design the gain matrix  $K$  by placing the poles of  $A + BK$  as  $(0.21, 0.22, 0.23, 0.24)$ .

```
% State feedback controller
feed_poles= [ 0.21; 0.22; 0.23; 0.24 ];
K= place(A, B, feed_poles);
```

## Simulation

```
% No of states
n= size(A,1);
% No of inputs
m= size(B,2);
% No of disturbance inputs
n_d= size(E_d,2);
% No of fault inputs
n_f= size(E_f,2);
% No of outputs
p= size(C,1);

% Number of simulation steps
sim_steps= 15+1;

% Simulation time
sim_t= (0:sim_steps)*T_s;

% Pre-allocate matrices to hold simulation data
% Output matrix
y_k= zeros(p, sim_steps);
% State matrix
x_k= zeros(n, sim_steps+1);
% Input matrix
u_k= zeros(m, sim_steps);

% Set initial state into state matrix
x_k(:, 1)= [0; 0; 0.02; 0];

for idx= 1:sim_steps
    % Input
    u_k(:, idx)= -K*x_k(:,idx);
    % Next state
    x_k(:, (idx+1))= A*x_k(:,idx) + B*u_k(:,idx);
```

```

% Output
y_k(:, idx)= C*x_k(:,idx);
end

```

## Plotting

We plot the state estimation error to visualize the accuracy of the observed states

```

Fig_sfc= figure;

subplot(1,2,1);
plot(sim_t, x_k, 'LineWidth', 1);
grid on;
title('Fig. 1- State Feedback Control (States)');
ylabel('States', 'Interpreter', 'latex');
xlabel('time, sec', 'Interpreter', 'latex');
leg= legend(['$x_1$', '$x_2$', '$x_3$', '$x_4$']);
leg.Interpreter='latex';
leg.Location= 'southeast';

subplot(1,2,2);
plot(sim_t(1:end-1), u_k, 'LineWidth', 1);
grid on;
ylabel('Control Input', 'Interpreter', 'latex');
xlabel('time, sec', 'Interpreter', 'latex');
leg= legend(['$u$']); leg.Interpreter='latex';
leg.Location= 'southeast';

```

## Full Order Observer Design

We want to design a full-order observer such that:

$$\hat{\dot{x}} = A\hat{x} + Bu + L(y - \hat{y})$$

$$\hat{y} = C\hat{x}$$

The observer error dynamics is then governed by:

$$\dot{e} = (A - LC)e + E_d d + (E_f + LF_f)f$$

with:

- $e = x - \hat{x}$

## Pole Placement

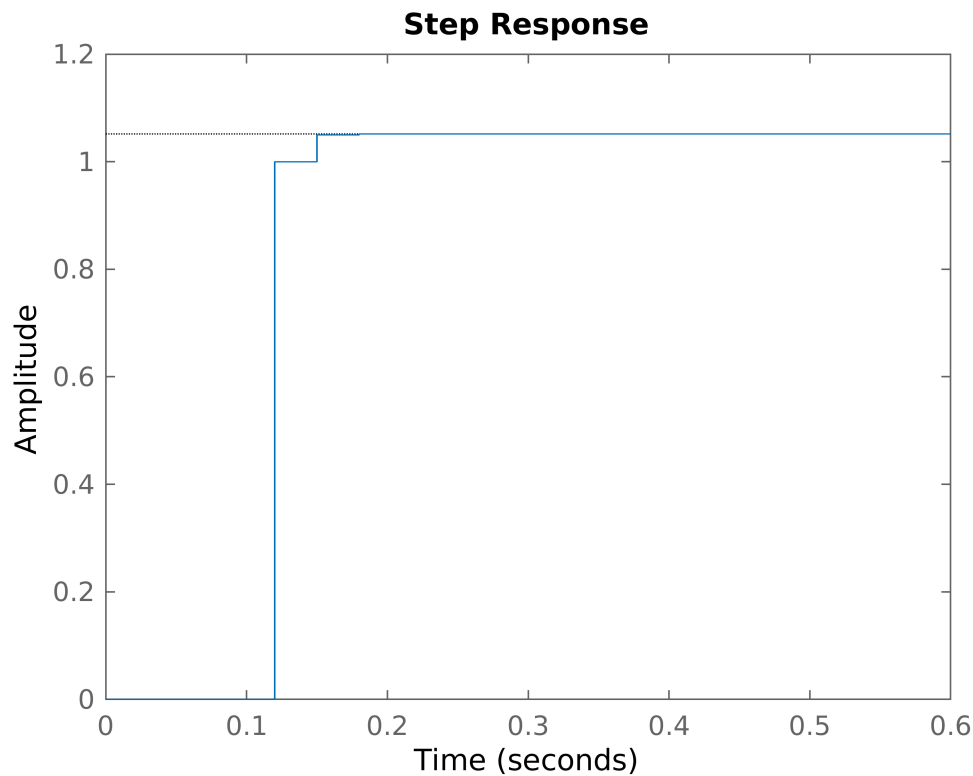
If we select our desired pole positions to be: (0.011, 0.012, 0.013, 0.014), we can examine the possible system response as:

```

obs_poles= [0.011, 0.012, 0.013, 0.014];
sys_temp= zpk([],obs_poles,1, T_s);

```

```
figure
step(sys_temp)
```



As can be seen, the response is quite satisfactory, with a response time of about  $2 \times 10^{-3}$ .

Now we compute the gain matrix  $L$  for our full order observer as:

```
L= place(A', C', obs_poles)'
```

```
L = 4x3
    0.9870   -0.0569    0.0010
         0    0.9302   -0.0038
    0.0000    0.0116    1.9923
    0.0000    0.7688   33.5854
```

## Full Order Observer Simulation

Here, we compare system states and output with the full order observer under zero disturbance and an initial

state of  $x = \begin{bmatrix} 0 \\ 0 \\ 0.02 \\ 0 \end{bmatrix}$ .

Because the system is unstable at the linearization point  $\begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \end{bmatrix}$ , we stabilize using a state feedback controller.

We design the gain matrix  $K$  by placing the poles of  $A + BK$  as  $(0.21, 0.22, 0.23, 0.24)$ .

```
% State feedback controller
feed_poles= [ 0.21; 0.22; 0.23; 0.24 ];
K= place(A, B, feed_poles);
```

## Simulation

```
% Number of simulation steps
sim_steps= 15+1;

% Simulation time
sim_t= (0:sim_steps)*T_s;

% Pre-allocate matrices to hold simulation data
% Output matrix
y_k= zeros(p, sim_steps);
% State matrix
x_k= zeros(n, sim_steps+1);
% Input matrix
u_k= zeros(m, sim_steps);

% Observed state matrix
x_k_hat= zeros(n, sim_steps+1);
% Observed output matrix
y_k_hat= zeros(p, sim_steps);

% State estimation error
e_k= zeros(n, sim_steps+1);

% Set initial state into state matrix
x_k(:, 1)= [0; 0; 0.02; 0];

% Initial state estimation error
e_k(:, 1)= x_k(:, 1) - x_k_hat(:, 1);

for idx= 1:sim_steps
    % Input
    u_k(:, idx)= -K*x_k(:,idx);
    % Next state
    x_k(:, (idx+1))= A*x_k(:,idx) + B*u_k(:,idx);
    % Output
    y_k(:, idx)= C*x_k(:,idx);

    % Observed output
    y_k_hat(:, idx)= C*x_k_hat(:,idx);
    % Observed state
    x_k_hat(:, (idx+1))= A*x_k_hat(:,idx) + B*u_k(:,idx) + ...
        L*( y_k(:, idx) - y_k_hat(:, idx) );

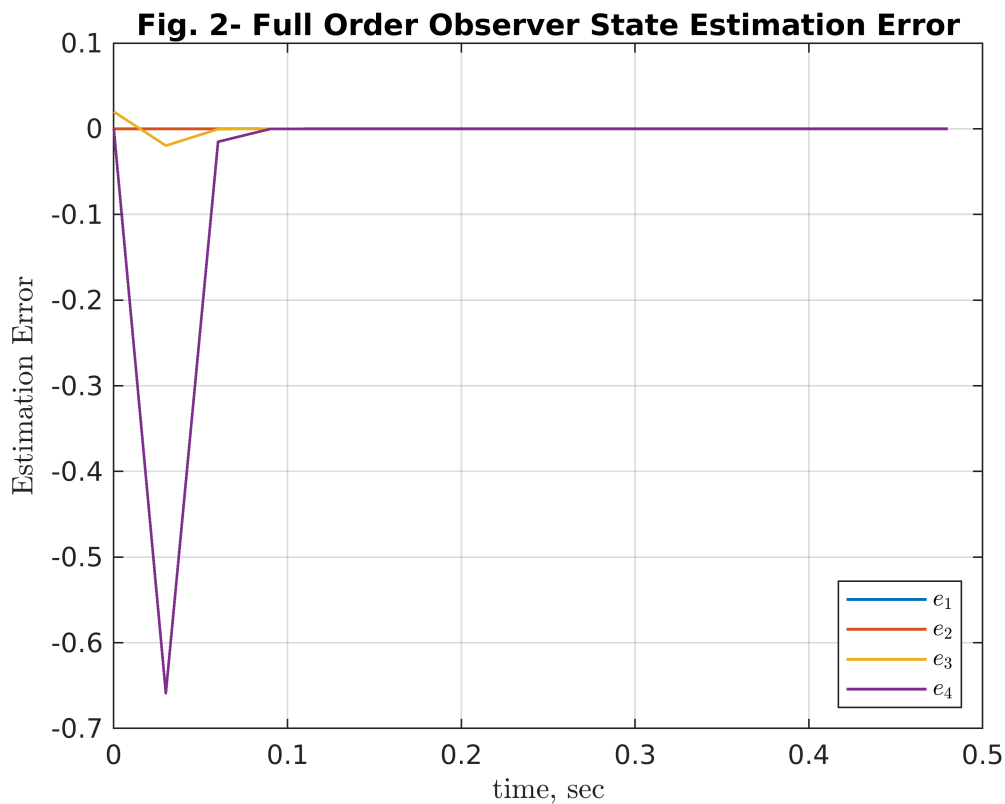
    % State estimation error
    e_k(:, idx+1)= x_k(:, idx+1) - x_k_hat(:, idx+1);
```

```
end
```

## Plotting

We plot the state estimation error to visualize the accuracy of the observed states

```
Fig_obs= figure;  
  
plot(sim_t, e_k, 'LineWidth', 1);  
grid on;  
title('Fig. 2- Full Order Observer State Estimation Error');  
ylabel('Estimation Error', 'Interpreter', 'latex');  
xlabel('time, sec', 'Interpreter', 'latex');  
leg= legend(['$e_1$', '$e_2$', '$e_3$', '$e_4$']); leg.Interpreter='latex';  
leg.Location= 'southeast';
```



## Fault Injection and Residual Generation

Now we inject faults into the system. We simulate the faults to happen at (3.5s, 5s, 7.5s, 12s) respectively.

```
fault_times= [3.5; 5; 7.5; 12];  
fault_points= round(fault_times./T_s);
```

## Simulation

```
% Number of simulation steps  
sim_steps= fault_points(end)+100;
```

```

% Simulatipn time
sim_t= (0:sim_steps)*T_s;

% Pre-allocate matrices to hold simulation data
% Output matrix
y_k= zeros(p, sim_steps);
% State matrix
x_k= zeros(n, sim_steps+1);
% Input matrix
u_k= zeros(m, sim_steps);
% Disturbance Matrix
d_k= 0.5*rand(n_d, sim_steps);
% Fault Matrix
f_k= zeros(n_f, sim_steps);
% Residual Matrix
r_k= zeros(p, sim_steps);
% Reisdual Filter to improve residual dynamics
W= diag([3,2,1]);

% Fault injection
for idx=1:n_f
    f_k(idx, fault_points(idx):end)= ones(1,sim_steps-fault_points(idx)+1);
end

% Observed state matrix
x_k_hat= zeros(n, sim_steps+1);
% Observed output matrix
y_k_hat= zeros(p, sim_steps);

% Set initial state into state matrix
x_k(:, 1)= [0; 0; 0.02; 0];

% Initial state estimation error
e_k(:, 1)= x_k(:, 1) - x_k_hat(:, 1);

for idx= 1:sim_steps
    % Input
    u_k(:, idx)= -K*x_k(:,idx);
    % Next state
    x_k(:, (idx+1))= A*x_k(:,idx) + B*u_k(:,idx) + E_d*d_k(:,idx) + E_f*f_k(:,idx);
    % Output
    y_k(:, idx)= C*x_k(:,idx) + F_f*f_k(:,idx);

    % Observed output
    y_k_hat(:, idx)= C*x_k_hat(:,idx);
    % Observed state
    x_k_hat(:, (idx+1))= A*x_k_hat(:,idx) + B*u_k(:,idx) + ...
        L*( y_k(:, idx) - y_k_hat(:, idx) );

    % Residual Generation
    r_k(:, idx)= W*( y_k(:,idx) - y_k_hat(:,idx) );

```

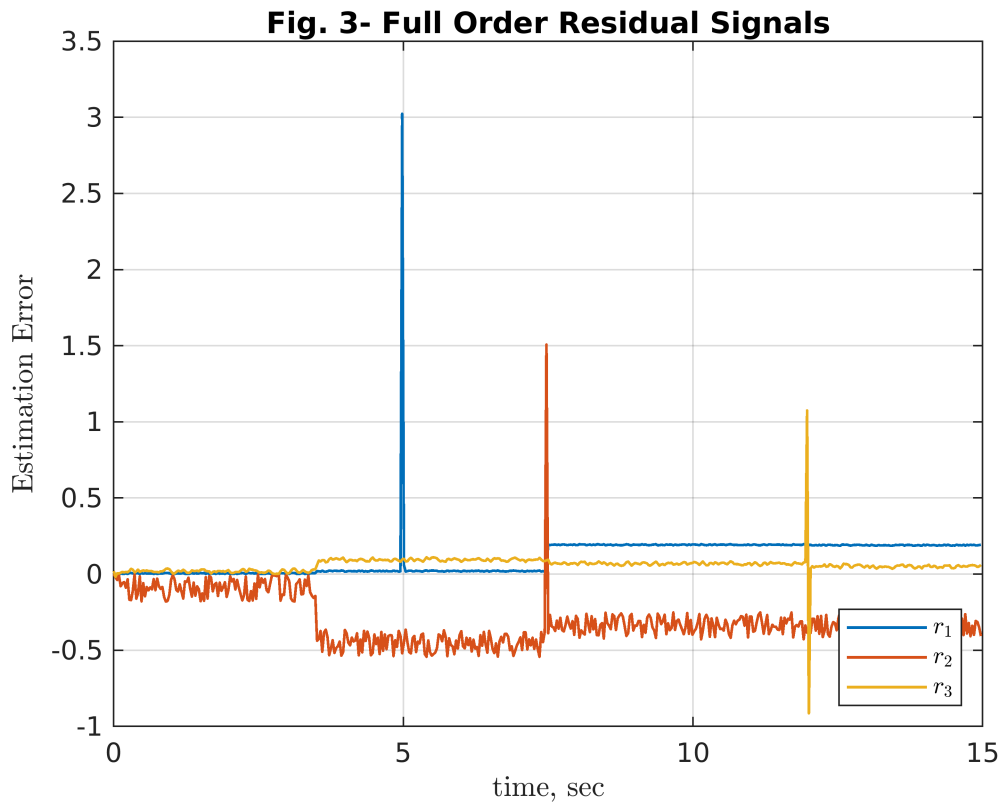


end

## Plotting

We plot the state estimation error to visualize the accuracy of the observed states

```
Fig_res= figure;  
  
plot(sim_t(1:end-1), r_k, 'LineWidth', 1);  
grid on;  
title('Fig. 3- Full Order Residual Signals');  
ylabel('Estimation Error', 'Interpreter', 'latex');  
xlabel('time, sec', 'Interpreter', 'latex');  
leg= legend(['$r_1$', '$r_2$', '$r_3$']); leg.Interpreter='latex';  
leg.Location= 'southeast';
```



## Full Decoupling using Unknown Input Observer

We try to eliminate the influence of the disturbance signals on the residual by using an Unknown Input Observer.

Assuming a fault-free operation, we have:

$$\dot{x} = Ax + Bu + E_d d$$

$$y = Cx$$

By taking the derivative of  $y$ , we have:

$$\dot{y} = C(Ax + Bu + E_d d)$$

We need  $CE_d$  to be left-invertible (of full column rank), such that:

$$d = CE_d^{-1}(\dot{y} - CAx - CBu)$$

With  $M = CE_d^{-1}$  we have:

$$\dot{x} = (A - E_d MCA)x + (B - E_d MCB)u + E_d M\dot{y}$$

To avoid using the derivative of the ooutput signal, we can perform a state transformation:

$$z = x - E_d My$$

Such that:

$$\dot{z} = (A - E_d MCA)x + (B - E_d MCB)u$$

for which we can then construct an observer.

```
% Checking if CE_d is left-invertible
C_E_d= C*E_d;
size(C_E_d)
```

```
ans = 1x2
      3      1
```

```
% Not square- i.e. not invertible
```

Hoeweever, for this plant,  $CE_d$  is not invertible, thus this approach cannot be used.

## Parity Space Approach

With the parity space approach, we explore the inherent characteristics of the system to generate residual signals.

First, we examine the behaviour of the system over a window of length  $s$

$$y(k-s) = Cx(k-s) + Du(k-s) + F_d d(k-s) + F_f f(k-s)$$

$$y(k-s+1) = Cx(k-s+1) + Du(k-s+1) + F_d d(k-s+1) + F_f f(k-s+1)$$

$$y(k-s+1) = CAx(k-s) + CBu(k-s) + CE_d d + CE_f f + Du(k-s+1) + F_d d(k-s+1) + F_f f(k-s+1)$$

.

.

.

$$y(s) = CA^s x(k-s) + CA^{s-1}Bu(k-s) + \dots + CBu(k-1) + Du(k) + CA^{s-1}E_d d(k-s) + \dots + CE_d d(k-1) + F_d d(k) + CA^{s-1}E_f f(k-s) + \dots + CE_f f(k-1) + F_f f(k)$$

If we stack up all the equations into matrix form, we have

$$\begin{bmatrix} y(k-s) \\ y(k-s+1) \\ y(k-s+2) \\ \vdots \\ y(k-1) \\ y(k) \end{bmatrix} = \begin{bmatrix} C \\ CA \\ CA^2 \\ \vdots \\ CA^{s-1} \\ CA^s \end{bmatrix} x(k-s) + \begin{bmatrix} D & 0 & 0 & \cdot & \cdot & \cdot & 0 & 0 \\ CB & D & 0 & \cdot & \cdot & \cdot & 0 & 0 \\ CAB & CB & D & \cdot & \cdot & \cdot & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ CA^{s-2}B & CA^{s-3}B & CA^{s-4}B & \cdot & \cdot & \cdot & D & 0 \\ CA^{s-1}B & CA^{s-2}B & CA^{s-3}B & \cdot & \cdot & \cdot & CB & D \end{bmatrix} + \begin{bmatrix} F_d & 0 & 0 & \cdot & \cdot & \cdot & 0 & 0 \\ CE_d & F_d & 0 & \cdot & \cdot & \cdot & 0 & 0 \\ CAE_d & CE_d & F_d & \cdot & \cdot & \cdot & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ CA^{s-2}E_d & CA^{s-3}E_d & CA^{s-4}E_d & \cdot & \cdot & \cdot & F_d & 0 \\ CA^{s-1}E_d & CA^{s-2}E_d & CA^{s-3}E_d & \cdot & \cdot & \cdot & CE_d & F_d \end{bmatrix} + \begin{bmatrix} F_f & 0 & 0 & \cdot & \cdot & \cdot & 0 & 0 \\ CE_f & F_f & 0 & \cdot & \cdot & \cdot & 0 & 0 \\ CAE_f & CE_f & F_f & \cdot & \cdot & \cdot & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\ CA^{s-2}E_f & CA^{s-3}E_f & CA^{s-4}E_f & \cdot & \cdot & \cdot & F_f & 0 \\ CA^{s-1}E_f & CA^{s-2}E_f & CA^{s-3}E_f & \cdot & \cdot & \cdot & CE_f & F_f \end{bmatrix}$$

We can then compress this into

$$y_s(k) = H_{o,s}x(k-s) + H_{u,s}u_s(k) + H_{d,s}d_s(k) + H_{f,s}f_s(k)$$

Rearranging yields

$$y_s(k) - H_{u,s}u_s(k) = H_{o,s}x(k-s) + H_{d,s}d_s(k) + H_{f,s}f_s(k)$$

Since  $y_s(k)$  and  $u_s(k)$  are available online, we generate the residual signal as

$$r(k) = v_s(y_s(k) - H_{u,s}u_s(k))$$

The dynamics of the residual are now governed by

$$r(k) = v_s(H_{o,s}x(k-s) + H_{d,s}d_s(k) + H_{f,s}f_s(k))$$

The vector  $v_s$  is called the parity vector and must satisfy:

- $v_s H_{o,s} = 0$
- $v_s H_{f,s} \neq 0$
- $v_s H_{d,s} = 0$  (for full decoupling)

For full decoupling to be possible, the following rank condition must be satisfied:

$$\text{rank}[H_{o,s} \ H_{d,s}] < \text{rank}[H_{o,s} \ H_{d,s} \ H_{f,s}]$$

## Build up data matrices

```
% Window size
s= 5;

% H_os, H_us, H_ds, H_fs
H_os = zeros(s*p,n);
H_us= zeros(s*p,s*m);
H_ds= zeros(s*p,s*n_d);
H_fs= zeros(s*p,s*n_f);

for r = 1:s
    H_os((r-1)*p+1:r*p,1:n) = C*(A^(r-1));
    for c = 1:r
        if r == c
            H_us((r-1)*p+1:r*p,(c-1)*m+1:c*m) = D;
            H_ds((r-1)*p+1:r*p,(c-1)*n_d+1:c*n_d) = F_d;
            H_fs((r-1)*p+1:r*p,(c-1)*n_f+1:c*n_f) = F_f;
        else
            H_us((r-1)*p+1:r*p,(c-1)*m+1:c*m) = C*A^(r-c-1)*B;
            H_ds((r-1)*p+1:r*p,(c-1)*n_d+1:c*n_d) = C*A^(r-c-1)*E_d;
            H_fs((r-1)*p+1:r*p,(c-1)*n_f+1:c*n_f) = C*A^(r-c-1)*E_f;
        end
    end
end

% Check rank condition for full decoupling
rank([H_os H_ds]) < rank([H_os H_ds H_fs])

ans = logical
     1
```

Since the rank condition is satisfied, full decoupling can be achieved.

## Parity Vector

```
% Left null space of H_os H_ds
N_basis= null([H_os H_ds]')';

% Parity vector (can be selected as any of the rows)
v_s= N_basis(1, :)

v_s = 1x15
    -0.0727    0.0363   -0.1677    0.1874    0.1251   -0.2757   -0.0701    0.0076 ...
```

Pre-computation of  $v_s H_{u,s}$  to speed up online computation

```
rho_s= v_s*H_us
```

```

rho_s = 1x5
10-16 x
    0.1388    0.3469    0.1735    0.0694    0

```

## Simulation

```

% Number of simulation steps
sim_steps= fault_points(end)+100;

% Simulation time
sim_t= (0:sim_steps)*T_s;

% Pre-allocate matrices to hold simulation data
% Output matrix
y_k= zeros(p, sim_steps);
% State matrix
x_k= zeros(n, sim_steps+1);
% Input matrix
u_k= zeros(m, sim_steps);
% Disturbance Matrix
d_k= 0.5*rand(n_d, sim_steps);
% Fault Matrix
f_k= zeros(n_f, sim_steps);
% Residual Matrix
r_k= zeros(1, (sim_steps-s));

% Fault injection
for idx=1:n_f
    f_k(idx, fault_points(idx):end)= ones(1,sim_steps-fault_points(idx)+1);
end

% Set initial state into state matrix
x_k(:, 1)= [0; 0; 0.02; 0];

% Before time k
% We need prior data for k-s to be valid
for idx= 1:(s+1)
    % Input
    u_k(:, idx)= -K*x_k(:,idx);
    % Next state
    x_k(:, (idx+1))= A*x_k(:,idx) + B*u_k(:,idx) + E_d*d_k(:,idx);
    % Output
    y_k(:, idx)= C*x_k(:,idx) + D*u_k(:,idx) + F_d*d_k(:,idx);
end

% After Time k
for idx= (s+1):sim_steps
    % Input
    u_k(:, idx)= -K*x_k(:,idx);
    % Next state
    x_k(:, (idx+1))= A*x_k(:,idx) + B*u_k(:,idx) + E_d*d_k(:,idx) + E_f*f_k(:, idx);
    % Output
    y_k(:, idx)= C*x_k(:,idx) + D*u_k(:,idx) + F_d*d_k(:,idx) + F_f*f_k(:, idx);

```

```

% Residual Generation
% y_s
y_s_k= y_k(:, (idx-s+1):(idx));
y_s_k= y_s_k(:);
% u_s
u_s_k= u_k(:, (idx-s+1):(idx));
u_s_k= u_s_k(:);
% Residual
r_k(:, (idx-s))= v_s*y_s_k - rho_s*u_s_k;

end

```

## Plotting

We plot the state estimation error to visualize the accuracy of the observed states

```

Fig_ps_res= figure;

plot(sim_t(s+1:end-1), r_k, 'LineWidth', 1);
grid on;
title('Fig. 4- Parity Space Residual Signal');
ylabel('Residual Signal', 'Interpreter', 'latex');
xlabel('time, sec', 'Interpreter', 'latex');
leg= legend(['$r_1$']); leg.Interpreter='latex';
leg.Location= 'southeast';

```

As can be seen in the figure above, the generated residual is insensitive to the actuator fault, this is probably because  $B = E_d$ .

If the residual signal is decoupled from the disturbance (via  $E_d$ ), then it is effectively decoupled from the input (via  $B$ ).

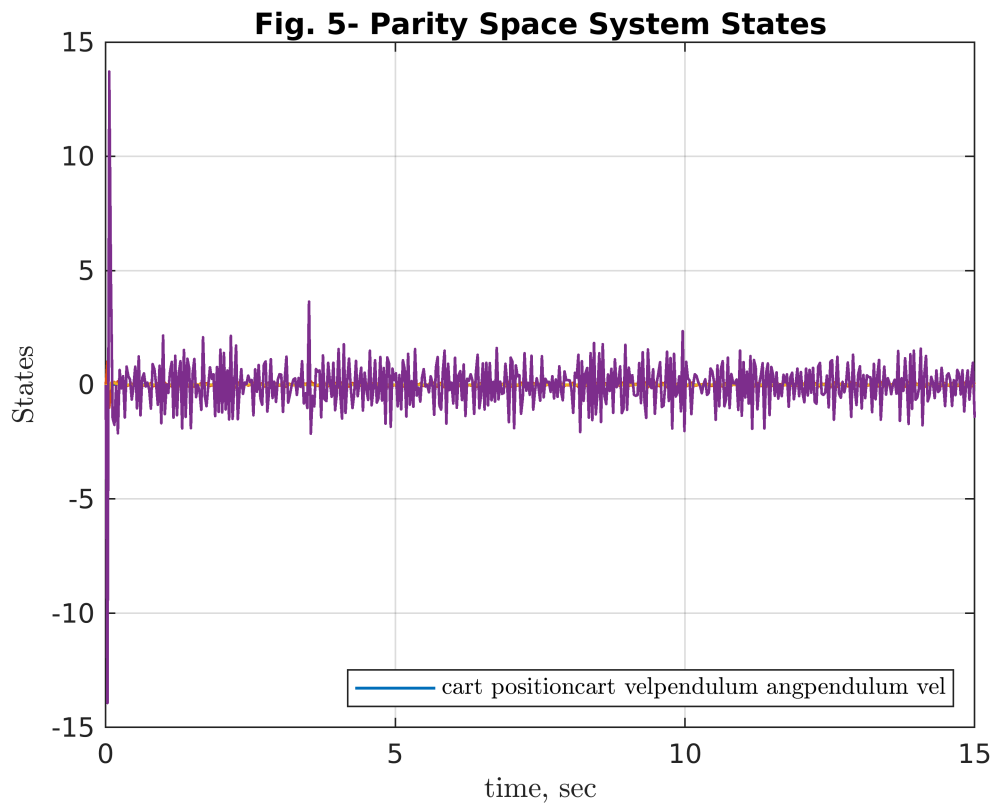
## System States

```

Fig_ps_states= figure;

plot(sim_t, x_k, 'LineWidth', 1);
grid on;
title('Fig. 5- Parity Space System States');
ylabel('States', 'Interpreter', 'latex');
xlabel('time, sec', 'Interpreter', 'latex');
leg= legend(['cart position', 'cart vel', 'pendulum ang', 'pendulum vel']); leg.Interpreter='latex';
leg.Location= 'southeast';

```



## References

[Fault Diagnosis and Fault Tolerant Systems Exercises, TU Kaiserslautern- Dina Martynova](#)