

2.10-2. Consider the system described by

$$\mathbf{x}(k+1) = \begin{bmatrix} 0 & 1 \\ 0 & 3 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u(k)$$

$$y(k) = \begin{bmatrix} -2 & 1 \end{bmatrix} \mathbf{x}(k)$$

- (a) Find the transfer function $Y(z)/U(z)$.
- (b) Using any similarity transformation, find a different state model for this system.
- (c) Find the transfer function of the system from the transformed state equations.
- (d) Verify that \mathbf{A} given and \mathbf{A}_w derived in part (b) satisfy the first three properties of similarity transformations. The fourth property was verified in part (c).

Solution:

$$(a) \quad z\mathbf{I} - \mathbf{A} = \begin{bmatrix} z & -1 \\ 0 & z-3 \end{bmatrix}; \quad \Delta = |z\mathbf{I} - \mathbf{A}| = z(z-3) = \Delta$$

$$\begin{aligned}\frac{Y(z)}{U(z)} &= \mathbf{C}[z\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} = \frac{1}{\Delta}[-2 \quad 1] \begin{bmatrix} z-3 & 1 \\ 0 & z \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} \\ &= \frac{1}{\Delta}[-2z+6 \quad z-2] \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \frac{-z+4}{z(z-3)}\end{aligned}$$

$$(b) \quad \mathbf{P} = \begin{bmatrix} 1 & -1 \\ 1 & 1 \end{bmatrix}; \quad \mathbf{P}^{-1} = \begin{bmatrix} \frac{1}{2} & \frac{1}{2} \\ -\frac{1}{2} & \frac{1}{2} \end{bmatrix}$$

$$\begin{aligned}\mathbf{A}_w &= \mathbf{P}^{-1}\mathbf{A}\mathbf{P} = \begin{bmatrix} \frac{1}{2} & \frac{1}{2} \\ -\frac{1}{2} & \frac{1}{2} \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 0 & 3 \end{bmatrix} \begin{bmatrix} 1 & -1 \\ 1 & 1 \end{bmatrix} = \begin{bmatrix} \frac{1}{2} & \frac{1}{2} \\ -\frac{1}{2} & \frac{1}{2} \end{bmatrix} \begin{bmatrix} 1 & 1 \\ 3 & 3 \end{bmatrix} \\ &= \begin{bmatrix} 2 & 2 \\ 1 & 1 \end{bmatrix}\end{aligned}$$

$$\mathbf{B}_w = \mathbf{P}^{-1}\mathbf{B} = \begin{bmatrix} \frac{1}{2} & \frac{1}{2} \\ -\frac{1}{2} & \frac{1}{2} \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$\mathbf{C}_w = \mathbf{C}\mathbf{P} = [-2 \quad 1] \begin{bmatrix} 1 & -1 \\ 1 & 1 \end{bmatrix} = [-1 \quad 3]$$

$$\therefore \mathbf{w}(k+1) = \begin{bmatrix} 2 & 2 \\ 1 & 1 \end{bmatrix} \mathbf{w}(k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} \mathbf{u}(k)$$

$$\mathbf{y}(k) = [-1 \quad 3] \mathbf{w}(k)$$

$$(c) \quad z\mathbf{I} - \mathbf{A}_w = \begin{bmatrix} z-2 & -2 \\ -1 & z-1 \end{bmatrix}; \quad \Delta = |z\mathbf{I} - \mathbf{A}_w| = z^2 - 3z + 2 - 2 = z(z-3)$$

$$\begin{aligned}\frac{Y(z)}{U(z)} &= \mathbf{C}_w[z\mathbf{I} - \mathbf{A}_w]^{-1}\mathbf{B}_w = \frac{1}{\Delta}[-1 \quad 3] \begin{bmatrix} z-1 & 2 \\ 1 & z-2 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \\ &= \frac{1}{\Delta}[-1 \quad 3] \begin{bmatrix} z-1 \\ 1 \end{bmatrix} = \frac{-z+4}{z(z-3)}\end{aligned}$$

$$(d) \quad |z\mathbf{I} - \mathbf{A}| = \begin{vmatrix} z & 1 \\ 0 & z-3 \end{vmatrix} = z^2 - 3z; \quad |z\mathbf{I} - \mathbf{A}_w| = \begin{vmatrix} z-2 & -2 \\ -1 & z-1 \end{vmatrix} = z(z-3)$$

$$\therefore z_1 = 0, z_2 = 3$$

$$|\mathbf{A}| = \begin{vmatrix} 0 & 1 \\ 0 & 3 \end{vmatrix} = 0 = z_1 z_2; \quad |\mathbf{A}_w| = \begin{vmatrix} 2 & 2 \\ 1 & 1 \end{vmatrix} = 0$$

$$\text{tr } \mathbf{A} = 3 = z_1 + z_2; \quad \text{tr } \mathbf{A}_w = 3$$

2.10-3. Consider the system of Problem 2.10-2. A similarity transformation on these equations yields

$$\mathbf{w}(k+1) = \begin{bmatrix} d_1 & 0 \\ 0 & d_2 \end{bmatrix} \mathbf{w}(k) + \mathbf{B}_w u(k)$$

$$y(k) = \mathbf{C}_w \mathbf{x}(k)$$

- (a) Find d_1 and d_2 .
- (b) Find a similarity transformation that results in the \mathbf{A}_w matrix given. Note that this matrix is diagonal.
- (c) Find \mathbf{B}_w and \mathbf{C}_w .
- (d) Find the transfer functions of both sets of state equations to verify the results of this problem.

Solution:

- (a) Let z_1, z_2 be the characteristic value of \mathbf{A} . $d_1 = z_1, \quad d_2 = z_2$

$$z\mathbf{I} - \mathbf{A} = \begin{bmatrix} z & -1 \\ 0 & z-3 \end{bmatrix}, \quad \therefore |z\mathbf{I} - \mathbf{A}| = z(z-3); \quad \therefore z_1 = 0, \quad z_2 = 3$$

$$(b) \quad (z_1\mathbf{I} - \mathbf{A})\mathbf{m}_1 = \begin{bmatrix} 0 & -1 \\ 0 & -3 \end{bmatrix} \begin{bmatrix} m_{11} \\ m_{21} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \Rightarrow \begin{matrix} -m_{21} = 0 \\ -3m_{21} = 0 \end{matrix}$$

$$\therefore m_{21} = 0, \text{ let } m_{11} = 1, \quad \therefore \mathbf{m}_1 = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$(z_2\mathbf{I} - \mathbf{A})\mathbf{m}_2 = \begin{bmatrix} 3 & -1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} m_{12} \\ m_{22} \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \end{bmatrix} \Rightarrow 3m_{12} - m_{22} = 0$$

$$\therefore \text{let } m_{12} = 1, \quad m_{22} = 3, \quad \therefore \mathbf{m}_2 = \begin{bmatrix} 1 \\ 3 \end{bmatrix}$$

$$\therefore \mathbf{M} = \begin{bmatrix} 1 & 1 \\ 0 & 3 \end{bmatrix}, \quad |\mathbf{M}| = 3, \quad \mathbf{M}^{-1} = \begin{bmatrix} 1 & -1/3 \\ 0 & 1/3 \end{bmatrix}$$

$$\mathbf{M}^{-1}\mathbf{A}\mathbf{M} = \begin{bmatrix} 1 & -1/3 \\ 0 & 1/3 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 0 & 3 \end{bmatrix} \begin{bmatrix} 1 & 1 \\ 0 & 3 \end{bmatrix} = \begin{bmatrix} 1 & -1/3 \\ 0 & 1/3 \end{bmatrix} \begin{bmatrix} 0 & 3 \\ 0 & 9 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 0 & 3 \end{bmatrix}$$

1. The system is a discrete-time system.
 2. The input is a unit step function.
 3. The output is a unit step function.

$$(c) \quad \mathbf{B}_w = \mathbf{M}^{-1} \mathbf{B} = \begin{bmatrix} 1 & -1/3 \\ 0 & 1/3 \end{bmatrix} = \begin{bmatrix} 2/3 \\ 1/3 \end{bmatrix}$$

$$\mathbf{C}_w = \mathbf{C} \mathbf{M} = \begin{bmatrix} -2 & 1 \end{bmatrix} \begin{bmatrix} 1 & 1 \\ 0 & 3 \end{bmatrix} = \begin{bmatrix} -2 & 1 \end{bmatrix}$$

$$\therefore \mathbf{w}(k+1) = \begin{bmatrix} 0 & 0 \\ 0 & 3 \end{bmatrix} \mathbf{w}(k) + \begin{bmatrix} 2/3 \\ 1/3 \end{bmatrix} \mathbf{u}(k)$$

$$\mathbf{y}(k) = \begin{bmatrix} -2 & 1 \end{bmatrix} \mathbf{w}(k)$$

(d) See Problem 2.10-2(a) for the first transfer function.

$$z\mathbf{I} - \mathbf{A}_w = \begin{bmatrix} z & 0 \\ 0 & z-3 \end{bmatrix}; \quad |z\mathbf{I} - \mathbf{A}_w| = z(z-3) = \Delta$$

$$\frac{Y(z)}{U(z)} = \mathbf{C}_w [z\mathbf{I} - \mathbf{A}_w]^{-1} \mathbf{B}_w = \frac{1}{\Delta} \begin{bmatrix} -2 & 1 \end{bmatrix} \begin{bmatrix} z-3 & 0 \\ 0 & z \end{bmatrix} \begin{bmatrix} 2/3 \\ 1/3 \end{bmatrix}$$

$$= \frac{1}{\Delta} \begin{bmatrix} -2z+6 & z \end{bmatrix} \begin{bmatrix} 2/3 \\ 1/3 \end{bmatrix} = \frac{-\frac{4}{3}z + 4 + \frac{1}{3}z}{\Delta} = \frac{-z+4}{z(z-3)}$$

2.11-1. Consider a system with the transfer function

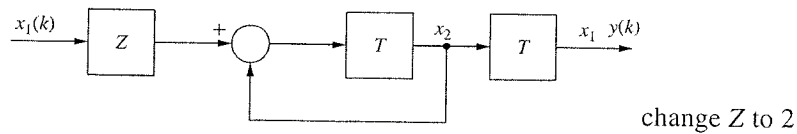
$$G(z) = \frac{Y(z)}{U(z)} = \frac{2}{z(z-1)}$$

- (a) Find three different state-variable models of this system.
 (b) Verify the transfer function of each state model in part (a), using (2-84).

Solution:

(a) $G(z) = G_1(z)G_2(z) = \frac{2}{z^2 - z} = \frac{2z^{-2}}{1 - z^{-1}}$

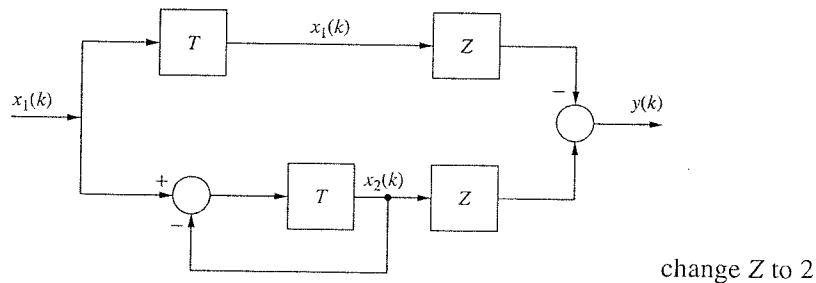
(1)



$$\mathbf{x}(k+1) = \begin{bmatrix} 0 & 1 \\ 0 & 1 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 0 \\ 2 \end{bmatrix} u(k)$$

$$y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}(k)$$

(2) $G(z) = \frac{2}{z(z-1)} = \frac{-2}{z} + \frac{2}{z-1} = G_1(z) + G_2(z)$

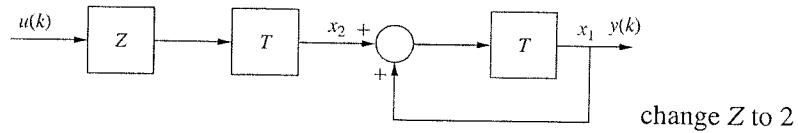


$$\mathbf{x}(k+1) = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 1 \\ 1 \end{bmatrix} u(k)$$

$$y(k) = \begin{bmatrix} -2 & 2 \end{bmatrix} \mathbf{x}(k)$$

Find a state-variable formulation for the system described by the coupled difference equation

(3)



$$\mathbf{x}(k+1) = \begin{bmatrix} 1 & 1 \\ 0 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 0 \\ 2 \end{bmatrix} u(k)$$

$$y(k) = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}(k)$$

(b) (1) $z\mathbf{I} - \mathbf{A} = \begin{bmatrix} z & -1 \\ 0 & z-1 \end{bmatrix}; |z\mathbf{I} - \mathbf{A}| = z^2 - z = \Delta$

$$G(z) = \mathbf{C}[z\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} = \frac{1}{\Delta} \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} z-1 & 1 \\ 0 & z \end{bmatrix} \begin{bmatrix} 0 \\ 2 \end{bmatrix} = \frac{1}{\Delta} \begin{bmatrix} z-1 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ 2 \end{bmatrix} = \frac{2}{z(z-1)}$$

(2) $z\mathbf{I} - \mathbf{A} = \begin{bmatrix} z & 0 \\ 0 & z-1 \end{bmatrix}; |z\mathbf{I} - \mathbf{A}| = \Delta = z^2 - z$

$$G(z) = \mathbf{C}[z\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} = \frac{1}{\Delta} \begin{bmatrix} -1 & 1 \end{bmatrix} \begin{bmatrix} z-1 & 0 \\ 0 & z \end{bmatrix} \begin{bmatrix} 2 \\ 2 \end{bmatrix} = \frac{1}{\Delta} \begin{bmatrix} -1 & 1 \end{bmatrix} \begin{bmatrix} 2z-2 \\ 2z \end{bmatrix} = \frac{2}{z(z-1)}$$

(3) $z\mathbf{I} - \mathbf{A} = \begin{bmatrix} z-1 & -1 \\ 0 & z \end{bmatrix}; |z\mathbf{I} - \mathbf{A}| = z^2 - z = \Delta$

$$G(z) = \mathbf{C}[z\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} = \frac{1}{\Delta} \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} z & 1 \\ 0 & z-1 \end{bmatrix} \begin{bmatrix} 0 \\ 2 \end{bmatrix} = \frac{1}{\Delta} \begin{bmatrix} z & 1 \end{bmatrix} \begin{bmatrix} 0 \\ 2 \end{bmatrix} = \frac{2}{z(z-1)}$$

2.11-2. Consider a system described by the coupled difference equation

$$y(k+2) - v(k) = 0$$

$$v(k+1) + y(k+1) = u(k)$$

where $u(k)$ is the system input.

(a) Find a state-variable formulation for this system. Consider the outputs to be $y(k+1)$ and $v(k)$.

Hint: Draw a simulation diagram first.

(b) Repeat part (a) with $y(k)$ and $v(k)$ as the outputs.

(c) Repeat part (a) with the single output $v(k)$.

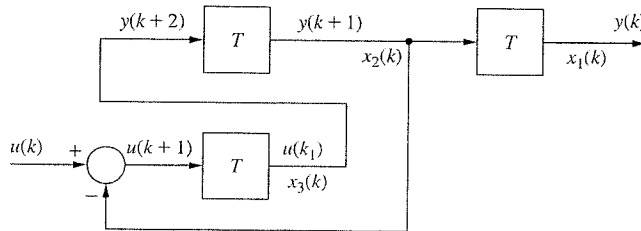
(d) Use (2-84) to calculate the system transfer function with $v(k)$ as the system output, as in part (c); that is, find $V(z)/U(z)$.

(e) Verify the transfer function $V(z)/U(z)$ in part (d) by taking the z-transform of the given system difference equations and eliminating $Y(z)$.

(f) Verify the transfer function $V(z)/U(z)$ in part (d) by using Mason's gain formula on the simulation diagram of part (a).

Solution:

(a)



$$\mathbf{x}(k+1) = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & -1 & 0 \end{bmatrix} \mathbf{x}(k) + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u(k)$$

$$y_0(k) = \begin{bmatrix} x_2(k) \\ x_3(k) \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \mathbf{x}(k); y_0(k) = \text{output}$$

(b) $\mathbf{x}(k+1)$ = same as (a)

$$y_0(k) = \begin{bmatrix} x_1(k) \\ x_3(k) \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 1 \end{bmatrix} \mathbf{x}(k)$$

(c) $\mathbf{x}(k+1)$ = same as (a)

$$y_0(k) = x_3(k) = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} \mathbf{x}(k)$$

$$(d) \quad z\mathbf{I} - \mathbf{A} = \begin{bmatrix} z & -1 & 0 \\ 0 & z & -1 \\ 0 & 1 & z \end{bmatrix}; |z\mathbf{I} - \mathbf{A}| = z^3 - (-z) = z^3 + z = \Delta$$

$$\text{Cof}[z\mathbf{I} - \mathbf{A}] = \begin{bmatrix} z^2 + 1 & z^2 & 0 \\ z & z^2 & z \\ 1 & z & z^2 \end{bmatrix}; [z\mathbf{I} - \mathbf{A}]^{-1} = \frac{1}{\Delta} \begin{bmatrix} z^2 + 1 & z & 1 \\ z^2 & z^2 & z \\ 0 & z & z^2 \end{bmatrix}$$

$$\therefore \frac{Y_0(z)}{U(z)} = \mathbf{C}[z\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} = \frac{1}{\Delta} \begin{bmatrix} 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} z^2 + 1 & z & 1 \\ z^2 & z^2 & z \\ 0 & z & z^2 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

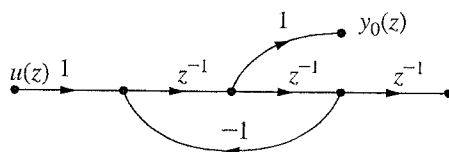
$$= \frac{1}{\Delta} \begin{bmatrix} 0 & z & z^2 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} = \frac{z^2}{z^3 - z} = \frac{z}{z^2 + 1}$$

$$(e) \quad z^2 Y(z) - V(z) = 0 \Rightarrow Y(z) = \frac{1}{z^2} V(z)$$

$$zV(z) + zY(z) = zV(z) + \frac{1}{z}V(z) = U(z)$$

$$\therefore \frac{V(z)}{U(z)} = \frac{Y_0(z)}{U(z)} = \frac{1}{z + \frac{1}{z}} = \frac{z}{z^2 + 1}$$

(f) From (a):



make u and y capital letters

$$\therefore \frac{Y_0(z)}{U(z)} = \frac{z^{-1}}{1 + z^{-2}} = \frac{z}{z^2 + 1}$$

3.4-2. Find $E^*(s)$ for each of the following functions. Express $E^*(s)$ in closed form.

$$(a) \quad e(t) = \varepsilon^{at} \quad (b) \quad E(s) = \frac{\varepsilon^{-2Ts}}{s-a}$$

$$(c) \quad e(t) = \varepsilon^{a(t-2T)} u(t-2T) \quad (d) \quad e(t) = \varepsilon^{a(t-T/2)} u(t-T/2)$$

Solution:

$$(a) \quad E^*(s) = 1 + \varepsilon^{aT} \varepsilon^{-Ts} + \varepsilon^{2aT} \varepsilon^{-2Ts} + \dots = 1 + \varepsilon^{(a-s)T} + [\varepsilon^{(a-s)T}]^2 + \dots$$

$$= \frac{1}{1 - \varepsilon^{(a-s)T}}$$

$$(b) \quad e(t) = \varepsilon^{a(t-2T)} u(t-2T)$$

$$E^*(s) = \varepsilon^{-2Ts} + \varepsilon^{aT} \varepsilon^{-3Ts} + \varepsilon^{2aT} \varepsilon^{-4Ts} + \dots$$

$$= \varepsilon^{-2Ts} (1 + \varepsilon^{aT} \varepsilon^{-Ts} + \varepsilon^{2aT} \varepsilon^{-2Ts} + \dots) = \frac{\varepsilon^{-2Ts}}{1 - \varepsilon^{(a-s)T}}$$

$$(c) \quad \text{From (b), } E^*(s) = \frac{\varepsilon^{-2Ts}}{1 - \varepsilon^{(a-s)T}}$$

$$(d) \quad E^*(s) = \varepsilon^{aT/2} \varepsilon^{-Ts} + \varepsilon^{3aT/2} \varepsilon^{-2Ts} + \varepsilon^{5aT/2} \varepsilon^{-3Ts} + \dots$$

$$E^*(s) = \varepsilon^{aT/2} \varepsilon^{-Ts} (1 + \varepsilon^{aT} \varepsilon^{-Ts} + \varepsilon^{2aT} \varepsilon^{-2Ts} + \dots)$$

$$= \frac{\varepsilon^{aT/2} \varepsilon^{-Ts}}{1 - \varepsilon^{(a-s)T}}$$

3.4-6. Find $E^*(s)$, with $T = 0.5$ s, for

$$E(s) = \frac{(1 - e^{-0.5s})^2}{0.5s^2(s+1)}$$

Solution:

Consider $E_1(s) = \frac{1}{s^2(s+1)}$; Then $E^*(s) = E_1^*(s)[2(1 - e^{-Ts})^2]$

$$\begin{aligned} (\text{residue})_{\lambda=0} &= \frac{d}{d\lambda} \left[\frac{1}{(\lambda+1)(1 - e^{-T(s-\lambda)})} \right]_{\lambda=0} \\ &= \frac{-(1 - e^{-T(s-\lambda)}) - (-Te^{-T(s-\lambda)})(\lambda+1)}{(\lambda+1)^2(1 - e^{-T(s-\lambda)})^2} \bigg|_{\lambda=0} = \frac{-1 + e^{-Ts} + Te^{-Ts}}{(1 - e^{-Ts})^2} \end{aligned}$$

$$(\text{residue})_{\lambda=-1} = \left[\frac{1}{\lambda^2(1 - e^{-T(s-\lambda)})} \right]_{\lambda=-1} = \frac{1}{1 - e^{-T(s+1)}}$$

$$\therefore E_1^*(s) = \frac{1}{1 - e^{-T(s+1)}} + \frac{-1 + e^{-Ts} + Te^{-Ts}}{(1 - e^{-Ts})^2}$$

$$\therefore E^*(s) = \frac{1 - e^{-Ts} - 1 + e^{-Ts} + Te^{-Ts}}{(1 - e^{-T(s+1)})(1 - e^{-Ts})^2} [2(1 - e^{-Ts})^2] = \frac{2Te^{-Ts}}{1 - e^{-T(s+1)}}$$

3.6-1.(a) Find $E^*(s)$, for $T = 0.1$ s, for the two functions below. Explain why the two transforms are equal, first from a time-function approach, and then from a pole-zero approach.

(i) $e_1(t) = \cos(4\pi t)$

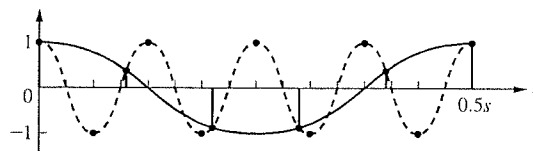
(ii) $e_2(t) = \cos(16\pi t)$

(b) Give a third time function that has the same $E^*(s)$.

Solution:

(a) (i) $\omega_1 = 4\pi$, $\omega_s = \frac{2\pi}{T} = 20\pi$ (ii) $\omega_2 = 16\pi = \frac{4\omega_s}{5} = 4\omega_1$

$$\omega_1 = \frac{\omega_s}{5}, T_1 = 0.5$$

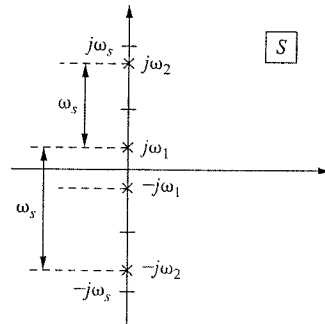


$$\cos 72^\circ = 0.309 \quad \cos(4 \times 72^\circ) = 0.309$$

$$\cos 144^\circ = -0.809 \quad \cos(4 \times 144^\circ) = -0.809$$

$$\cos 216^\circ = -0.809 \quad \cos(4 \times 216^\circ) = -0.809$$

$$\cos 288^\circ = 0.309 \quad \cos(4 \times 288^\circ) = 0.309$$



$$\omega_1 = \omega_s - \omega_2$$

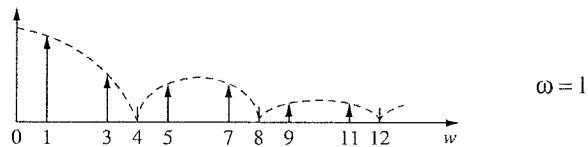
$$-\omega_1 = -\omega_s + \omega_2$$

(b) $\omega_3 = \omega_s + \omega_1 = 20\pi + 4\pi = 24\pi \Rightarrow e_3(t) = \cos(24\pi t)$

- 3.7-5. A signal $e(t) = 4\sin 7t$ is applied to a sampler/zero-order-hold device, with $\omega = 4 \text{ rad/s}$.
- What is the frequency component in the output that has the largest amplitude?
 - Find the amplitude and phase of that component.
 - Sketch the input signal and the component of part (b) versus time.
 - Find the ratio of the amplitude in part (b) to that of the frequency component in the output at $\omega = 7 \text{ rad/s}$ (the input frequency).

Solution:

- (a) Frequencies in $\bar{e}(t)$: 7 ; $4 \pm 7 = -3, 10$; $8 \pm 7 = 1, 15$; $12 \pm 7 = 5, 19$



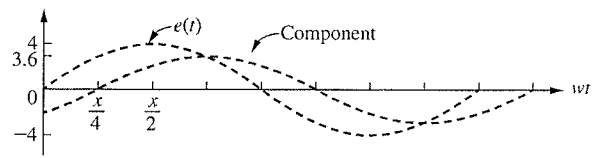
(b)

$$\frac{1}{T} G_{h0}(j1) = \frac{\sin\left(\frac{\pi(1)}{4}\right)}{\pi(1)/4} e^{-j\pi/4} = 0.900 \angle -45^\circ$$

\therefore component $= (4)(0.9) \sin(t - 45^\circ) = 3.6 \sin(t - 45^\circ)$

- (c)

(c) The magnitude of the component is 3.6. The magnitude of the input is 4. The ratio is 3.6/4 = 0.9. The ratio is 0.9.



$$(d) \quad \frac{1}{T} G_{h0}(j7) = \frac{\sin\left(\frac{\pi 7/4}{4}\right)}{\pi 7/4} e^{-j^{315^\circ}} = (-0.129)e^{-j^{315^\circ}} = 0.129 \angle -135^\circ$$

$$\therefore \text{component} = 4(0.129)\sin(7t - 135^\circ) = 0.516\sin(7t - 135^\circ)$$

$$\text{ratio} = \frac{3.6}{0.516} = 6.977 \approx 7$$