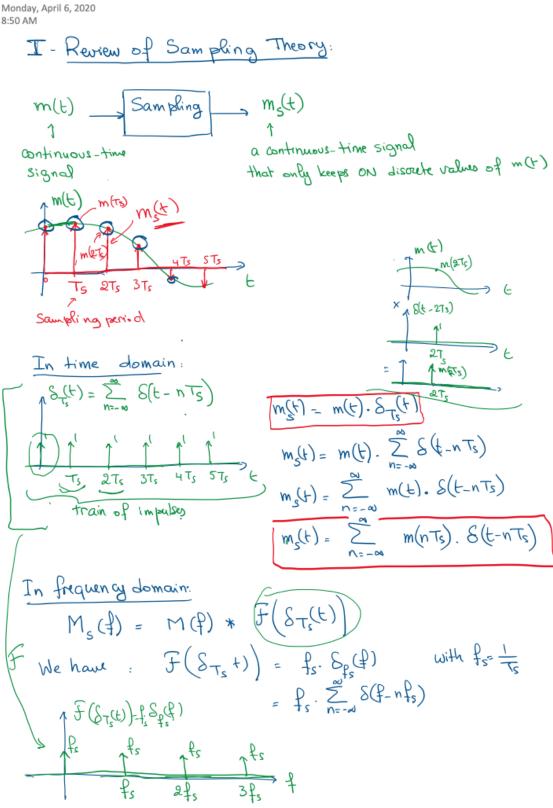
Online Lecture # 08 - Digital Baseband Modulation - Review of Sampling Theory

8:50 AM



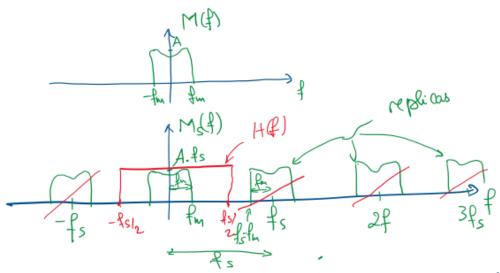
$$M_{S}(f) = M(f) * f_{S} \sum_{n=-\infty}^{\infty} S(f-nf_{S})$$

$$M_{S}(f) = f_{S} \sum_{n=-\infty}^{\infty} M(f) * S(f-nf_{S})$$

$$a shift of M(f) a nf_{S}$$

$$M_{S}(f) = f_{S} \sum_{n=-\infty}^{\infty} M(f-nf_{S})$$

For example if:

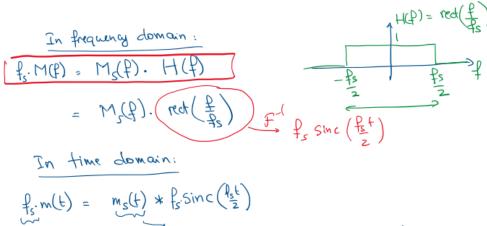


If $f_S > 2f_m$, there is no overlap between the replicas (no aliasing). In this case we can get back to M(f) and therefore m(f) using a low-pass filter.

In this case the sampling operation did not result in a loss of information.

* Reconstruction of sampled signa:

if $f_s > 2 f_m$: $m_s(t) \longrightarrow \begin{array}{c} LPF \\ H(f) \end{array}$ $f_s m(t)$



$$f_{s} \cdot m(t) = m_{s}(t) * f_{s} \cdot Sinc(f_{s}t)$$

$$= \sum_{n=-\infty}^{\infty} m(nT_{s}) S(t-nT_{s}) * f_{s} \cdot Sinc(f_{s}t)$$

$$\int_{S}^{\infty} w(t) = \sum_{N=-\infty}^{\infty} \int_{S}^{\infty} m(nT_{S}) \cdot \operatorname{Sinc}\left(\frac{1}{2}(t-nT_{S})\right)$$