

=> De dução: Considera-se que o campo elitrico e magnifico são da forma:

$$\overrightarrow{E}(x, y, z, t) = \overrightarrow{E}(x, y) e^{\int wt - f \beta z}$$

$$\overrightarrow{H}(x, y, z, t) = \overrightarrow{H}(x, y) e^{\int wt - f \beta z}$$

$$\vec{E}(x, y, z, t) = \vec{E}(x, y) \cos(wt - Bz)$$

 $\vec{H}(x, y, z, t) = \vec{H}(x, y) \cos(wt - Bz)$

B é o número de onda (ou constante de propagação) ao longo da dineção de propagação do guia. Pevido à direção de propação do guid é conveniente escrever:

$$\overline{E}(x,y) = \hat{x} E_{x}(x,y) + \hat{y} E_{y}(x,y) + \hat{z} E_{z}(x,y)$$
Thansversal Longitudinal

De modo similar,
$$\vec{\nabla}$$
 é decomposto
 $\vec{\nabla} = \hat{x} \hat{a} \times + \hat{y} \hat{a} y + \hat{z} \hat{a} z = \vec{\nabla} \hat{\tau} + \hat{z} \hat{a} z = \vec{\nabla} \hat{\tau} - \int \beta \hat{z}$

$$\frac{1}{\sqrt{2}} = -\int w \mu H \qquad (\nabla_{T} - \int \beta_{2}) \times (E_{T} + 2E_{2}) = -\int w \mu (H_{T} + 2H_{2}) = V_{T} + 2H_{2} =$$

$$\beta = \frac{\omega}{u} = \frac{z_{11}}{z_{1}} = \frac{z_{11}}{z_{1}}$$

As identidades vetoriais utilizadas nesta seção: 1 ($\nabla_{T} - J\beta\hat{z}$) \times ($E_{T} + \hat{z}E_{z}$) = $-J\omega\mu$ ($H_{T} + \hat{z}H_{z}$) 2 ($\nabla_{T} - J\beta\hat{z}$) \times ($H_{J} + \hat{z}H_{z}$) = $J\omega\xi$ ($E_{T} + \hat{z}E_{z}$) 3 ($\nabla_{T} - J\beta\hat{z}$) \times ($E_{T} + \hat{z}E_{z}$) = 0 $\hat{\mathbf{z}} \times (\hat{\mathbf{z}} \times \mathbf{A}_T) = -\mathbf{A}_T,$ $\hat{\mathbf{z}} \times (\nabla_T A_z \times \hat{\mathbf{z}}) = \nabla_T A_z,$ $\nabla_T \times \nabla_T A_z = 0,$ $\nabla_T \times (\hat{\mathbf{z}} \times \nabla_T A_z) = \hat{\mathbf{z}} \nabla_T^2 A_z,$ 4 (Vr - 132) X (HT + 2 Hz) = 0 $\nabla_T \cdot (\hat{\mathbf{z}} \times \nabla_T A z) = 0.$ Eq. 1: $(\hat{x}\hat{a}x + \hat{y}\hat{a}y) \times (\hat{x}\hat{E}_{x}(x,y) + \hat{y}\hat{E}_{y}(x,y)) = \hat{a}x\hat{E}y\hat{z}^{2} - \hat{a}y\hat{E}x\hat{z}^{2}$ $(\hat{x}\hat{a}x + \hat{y}\hat{a}y)x\hat{z}E_z = -\hat{d}xEz\hat{y} + \hat{d}yEz\hat{x}$ $-\beta\beta\hat{z}\times(\hat{x}E_{x}(x,y)+\hat{y}E_{y}(x,y))=-\beta\beta E_{x}\hat{y}+\beta\beta E_{y}\hat{x}$ axEyz-ayExz-axEzŷ+ayEzx-jBExý+jBEyx = -jwht-jwhtzź 2 (ax Ex - ay Ex) + PEz x 2 - JB 2 x Er = - JWHT - JWHZ 2 (*) $L_{7}(\hat{x} dx + \hat{y} dy) \times (\hat{x} Ex + \hat{y} Ey) = \hat{d} \times Ey\hat{z} - \hat{d} y Ex\hat{z}$ (*) $\vec{\nabla} \times \vec{E}_{T} + \vec{\nabla} \vec{E}_{2} \times \vec{z} - f \vec{B} \vec{z} \times \vec{E}_{T} = -f \omega \mu H_{T} - f \omega \mu H_{Z} \vec{z}$ C> Possuem componentes transversais (x, y) Comp. longitudinais. Assim $\nabla \times \vec{E_T} = - \int \omega \mu Hz \vec{2} = > \nabla \times \vec{E} + \int \omega \mu Hz \vec{2} = 0$ νέzx2-1β2xET = -1 wμHT • Eq. 2: (∇τ - 1β2) X (H̄τ + 2Hz) = Jw E(Ēτ + 2Ez) (#) $(\hat{x}\partial_x + \hat{y}\partial_y) \times (\hat{x} H_x(x,y) + \hat{y} H_y(x,y)) = \partial_x H_y \hat{z} - \partial_y H_x \hat{z} = \overline{\nabla} \times \overline{H_T} \left(Long. \right)$ $(\hat{x}\hat{a}_{x}+\hat{y}\hat{a}_{y})\times\hat{z}H_{z}=-\hat{a}_{x}H_{z}\hat{y}+\hat{a}_{y}H_{z}\hat{x}=\overline{\nabla}H_{z}\times\hat{z}$ (Transversal) - $f\beta\hat{z}\times(\hat{x}H_{x}(x_{1}y)+\hat{y}H_{y}(x_{1}y))=-f\beta H_{x}\hat{y}+f\beta H_{y}\hat{x}=f\beta H_{x}^{2}\times\hat{z}$ (Transversal)

 $(#) \overrightarrow{\nabla}_{x} \overrightarrow{H}_{x} + \overrightarrow{\nabla}_{H_{z}} \times 2 + f \overrightarrow{\beta}_{H_{x}} \times 2 = f \omega \varepsilon \overrightarrow{E}_{x} + f \omega \varepsilon \varepsilon z^{2}$ $(-) \overrightarrow{\nabla}_{x} \overrightarrow{H}_{x} - f \omega \varepsilon \varepsilon z^{2} = a$ $\overrightarrow{\nabla}_{H_{z}} \times 2 - f \overrightarrow{\beta}_{z} 2 \times \overrightarrow{H}_{x} = f \omega \varepsilon \varepsilon z^{2}$

• Eq. 3:
$$(\nabla_{\tau} - j\beta \hat{z}) \times (\bar{E}_{\tau} + \hat{z} E_{z}) = 0$$

$$(\nabla_{\tau} - j\beta \hat{z}) \times (\bar{E}_{\tau} + \hat{z} E_{z}) = 0$$

• Eq. 4:
$$(\nabla \vec{r} - j\beta \hat{z}) \times (\vec{H}_T + \hat{z} H_z) = 0$$

$$\nabla \vec{r} \cdot \vec{H}_T - j\beta H_z = 0$$

Com isso, as equações finais são:

$$\nabla E_{z} \times \hat{\mathbf{z}} - j\beta \,\hat{\mathbf{z}} \times \mathbf{E}_{T} = -j\omega\mu\mathbf{H}_{T} \qquad (11) \qquad \nabla \times \mathbf{E}_{T} + j\omega\mu H_{z} \,\hat{\mathbf{z}} = 0 \qquad (13)$$

$$\nabla H_{z} \times \hat{\mathbf{z}} - j\beta \,\hat{\mathbf{z}} \times \mathbf{H}_{T} = j\omega\varepsilon\mathbf{E}_{T} \qquad (12) \qquad \nabla \times \mathbf{H}_{T} - j\omega\varepsilon E_{z} \,\hat{\mathbf{z}} = 0 \qquad (14)$$

$$\nabla_{T} \cdot \mathbf{E}_{T} - j\beta E_{z} = 0 \qquad (15)$$

$$\nabla_{T} \cdot \mathbf{H}_{T} - j\beta H_{z} = 0 \qquad (16)$$

• Identidades vetonids viteis: $\hat{z}_x(\hat{z}_x\hat{A}T) = -\hat{A}T$ $\hat{z}_x(\hat{\nabla}T 4z \times \hat{z}) = \hat{\nabla}T Az$ • Expressando as componentos tranversais em função das longitudinais: Eq. (11):

$$\hat{z}_{x}(\vec{\nabla}_{z_{x}}\hat{z}) - \hat{z}_{x}(j\beta\hat{z}_{x}\vec{E}_{T}) = \hat{z}_{x}(-j\omega\mu(\vec{F}_{T}))$$

$$\overline{\nabla}_{r}E_{z} + j\beta E_{r} = -j\omega\mu(H_{x}\hat{g} - H_{y}\hat{x})$$

$$\overline{\nabla}_{r}E_{z} + j\beta \overline{E_{r}} = j\omega\mu H_{y}\hat{x} - j\omega\mu H_{x}\hat{y}$$

$$\overline{\nabla}_{r}E_{z} + j\beta \overline{E_{r}} = -j\omega\mu \hat{z} \times H_{r}$$
(1)

Eq. (12):

$$\hat{z} \times (\vec{\nabla} H_2 \times \hat{z} - f \vec{\beta} \hat{z} \times \vec{H_T}) = \hat{z} \times (f w \in \vec{E_T})$$

$$\overline{\nabla}_T H_2 + J \overline{B} \overline{H}_T = J W \varepsilon E_x \hat{y} - J W \varepsilon E_y \hat{x}$$

$$\overline{\nabla}_T H_2 + J \overline{B} \overline{H}_T = J W \varepsilon z \hat{x} E_T \qquad (2)$$

De (2):

$$\overrightarrow{H}\overrightarrow{i} = \underbrace{\omega \in \overrightarrow{z} \times \overrightarrow{E}\overrightarrow{i} + \overrightarrow{j} \overrightarrow{\nabla}\overrightarrow{i} H_{2}}_{\overrightarrow{B}}(3)$$

$$\overrightarrow{B}$$

$$\overrightarrow{B}$$

$$\overrightarrow{B}$$

$$\overrightarrow{A}$$

 $\frac{2}{2}x\overline{E_{T}} = -\frac{W\mu\overline{V}TH_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)} \cdot k = W\overline{V}\epsilon$ $\frac{1}{2}(\beta^{2} - W^{2}\mu\epsilon) - \frac{\beta \hat{z}x\overline{V_{T}}E_{2}}{J(\beta^{2} - W^{2}\mu\epsilon)}$

 $\frac{2^{2}xET = -\int \frac{B}{kc^{2}} \frac{2^{2}x\overline{V_{1}E_{2}} - \int \frac{WM}{kc^{2}} \overline{V_{1}H_{2}}}{kc^{2}}$

para determinar a outra equação, deve-se partir de (1):

$$\frac{\nabla T E_{2} + \int B E T = -\int w \mu z^{2} x H_{T}}{E T} = \left[-\int w \mu z^{2} x H_{T} - \nabla T E_{2} \right] / \int B$$

$$E_{T} = \int \nabla T E_{2} - \frac{w \mu}{B} z^{2} x H_{T} \qquad (4)$$

Tomanda (4) -> (3):

$$H_{\overline{I}} \left[1 - \frac{W^2 \cdot M \varepsilon}{\beta^2} \right] = \frac{1}{2} \overrightarrow{\nabla} H_z + \frac{1}{2} w \varepsilon \overrightarrow{Z} \times \overrightarrow{\nabla} E_z$$

$$H_{T}^{2} = \frac{1}{1} \frac{\beta}{\sqrt{1}} + \frac{1}{1} \frac{\omega \varepsilon}{\sqrt{1}} = \frac{2}{1} \frac{\sqrt{1}}{\sqrt{1}} \frac{1}{\sqrt{1}} = \frac{1}{1} \frac{\beta}{\sqrt{1}} + \frac{1}{1} \frac{\omega \varepsilon}{\sqrt{1}} = \frac{2}{1} \frac{\sqrt{1}}{\sqrt{1}} = \frac{1}{1} \frac{\omega \varepsilon}{\sqrt{1}} = \frac{2}{1} \frac{\sqrt{1}}{\sqrt{1}} = \frac{1}{1} \frac{\omega \varepsilon}{\sqrt{1}} = \frac{2}{1} \frac{\sqrt{1}}{\sqrt{1}} = \frac{1}{1} \frac{\omega \varepsilon}{\sqrt{1}} = \frac{2}{1} \frac{\omega \varepsilon}{\sqrt{1}}$$

Assim

<u>Podem sen eschitas como:</u>

AS COMO:

$$\hat{Z} \times \vec{Er} = -j \frac{B}{kc^{2}} \left[\hat{Z} \times \hat{\vec{Vr}} \vec{Ez} + \underbrace{WM} \hat{\vec{V}} \vec{Hz} \right]$$

$$\hat{HT} = -j \frac{B}{kc^{2}} \left[\underbrace{WE} \hat{Z} \times \hat{\vec{Vr}} \vec{Ez} + \hat{\vec{V}} \vec{Hz} \right]$$
(S)

Relação de dispensão;
$$k^2 = kc^2 + \beta^2 =$$
 $w^2 \mu \epsilon = w^2 \mu \epsilon + \beta^2$
 $w^2 = \frac{w^2 \mu \epsilon}{\mu \epsilon} - \frac{\beta^2}{\mu \epsilon} =$ $w^2 - \frac{1}{\mu \epsilon} \beta^2$

Velocidade de fase;
$$Vp = W = W$$
. $C^2 = C^2$

$$Vp = C$$

$$\sqrt{1 - (Wc/w)^2}$$

Velouidade de grupo:

$$V_{g} = \frac{\partial \omega}{\partial \beta} = \frac{\partial}{\partial \beta} \left[\left[w e^{2} + e^{2} \beta^{2} \right]^{1/2} \right] = \frac{1}{z} \cdot \left[w e^{2} + c^{2} \beta^{2} \right]^{-1/2} \cdot z e^{2} \beta^{2}$$

$$V_{f} = \frac{c^{2}\beta}{\left[\omega^{2}+c^{2}\beta^{2}\right]^{1/2}} = c^{2} \cdot \frac{1}{\left[\omega^{2}+c^{2}\beta^{2}\right]^{1/2}} = c^{2} \cdot \frac{1}{\left[\omega^{2}+c^{2}\beta^{2}\right]^{1/2}} = c^{2} \cdot \frac{1}{\left[\omega^{2}+c^{2}\beta^{2}\right]^{1/2}}$$

Petinindo as impedáncias transversais:

$$\eta_{TE} = \frac{WH}{P} = \frac{\eta W}{BC}, \quad \eta_{TM} = \frac{|B|}{WE} = \frac{\eta BC}{W}; \quad \text{onde } \eta = \frac{\sqrt{H/E}}{\sqrt{E}}$$

The property of the property

So que,
$$c = \frac{1}{\sqrt{\mu \epsilon}} = \sum_{n=1}^{\infty} \frac{1}{2} = \sum_{n=1}^{\infty} \frac{1$$

e também,
$$\frac{\eta}{c} = \mathcal{H} = \frac{1}{c^2 \varepsilon} = > \frac{\eta c}{c} = 1/\varepsilon$$

$$5abe-se: \\ w_{c} = \sqrt{w^{2} - c^{2}\beta^{z}} = > w_{c} = w - \sqrt{1 - c^{2}\beta^{z}} = > w_{c}^{2} = w^{2} \left[1 - c^{2}\beta^{2}\right]$$

$$\frac{c^{2}\beta^{2} = 1 - w_{c}^{2}}{w^{2}} = > \beta c = \sqrt{1 - w_{c}^{2}}$$

$$\frac{w^{2}}{w^{2}} = \sqrt{1 - w_{c}^{2}} = > w^{2} \left[1 - c^{2}\beta^{2}\right]$$

• Logo,
$$\eta_{TE} = \eta \underline{w} => \eta_{TE} = \eta \cdot \frac{1}{\sqrt{1 - wc^2/w^2}}$$

$$N_{TM} = \gamma \frac{Bc}{\omega} = > N_{TM} = \gamma . \sqrt{1 - \frac{Wc^2}{\omega^2}}$$

· Com essa definições as equações do sistema (5) são neesonitas:

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z + \cancel{WM}}{\overrightarrow{V}} \overrightarrow{V} H_Z \right]$$

$$\frac{4 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{w \cancel{E}}{\cancel{B}} \stackrel{2}{\cancel{V_T}} E_Z + \overrightarrow{V} H_Z \right]$$

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z + \cancel{V_T} E_Z + \cancel{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z + \cancel{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{E_T} = -j \cancel{B}}{kc^2} \left[\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z + \cancel{V_T} E_Z + \cancel{V_T} E_Z + \cancel{V_T} E_Z \right]$$

$$\frac{2 \times \overrightarrow{V_T} E_Z}{\cancel{V_T} E_Z} + \cancel{V_T} E_Z +$$

$$\overrightarrow{H_T} = -\int \frac{B}{k_c^2} \left[\frac{1}{\eta_{TM}} \hat{z}_X \hat{\nabla_T} E_Z + \hat{\nabla}^2 H_Z \right]$$

· Usando a indentidade vetonial zx(zx ET) = - ET, obtem-se:

$$-\frac{1}{E_{T}} = -\frac{1}{2} \frac{B}{k_{c}^{2}} \left[-\frac{1}{2} \frac{1}{E_{z}} + \frac{1}{2} \frac{1}{E_{z}} + \frac{1}{2} \frac{1}{E_{z}} \right]$$

$$= \frac{1}{2} \frac{B}{k_{c}^{2}} \left[-\frac{1}{2} \frac{1}{E_{z}} - \frac{1}{2} \frac{1}{E_{z}} - \frac{1}{2} \frac{1}{E_{z}} - \frac{1}{2} \frac{1}{E_{z}} \right]$$

logo, tem-se o sistema resultante:

Substituinda essas equação nas equações 13-16:

$$\nabla E_z \times \hat{\mathbf{z}} - j\beta \,\hat{\mathbf{z}} \times \mathbf{E}_T = -j\omega\mu\mathbf{H}_T \qquad (11) \qquad \nabla \times \mathbf{E}_T + j\omega\mu H_z \,\hat{\mathbf{z}} = 0 \qquad (13)$$

$$\nabla H_z \times \hat{\mathbf{z}} - j\beta \,\hat{\mathbf{z}} \times \mathbf{H}_T = j\omega\varepsilon\mathbf{E}_T \qquad (12) \qquad \nabla \times \mathbf{H}_T - j\omega\varepsilon E_z \,\hat{\mathbf{z}} = 0 \qquad (14)$$

$$\nabla_T \cdot \mathbf{E}_T - j\beta E_z = 0 \qquad (15)$$

$$\mathbf{\nabla}_T \cdot \mathbf{H}_T - j\beta H_z = 0 \tag{16}$$

Assim,
$$\frac{P}{I3}$$
:

$$\nabla_{x} \vec{E}_{T} = -\int_{B} \vec{\nabla}_{x} \vec{\nabla}_{x} \vec{E}_{z} - \eta_{TE} \vec{\nabla}_{x} \vec{z}_{x} \vec{\nabla}_{z} \vec{\nabla}_{z}$$

$$\nabla_{x} \vec{E}_{T} = -\int_{B} \vec{E}_{z} \vec{\nabla}_{x} \vec{\nabla}_{x} \vec{E}_{z} - \eta_{TE} \vec{\nabla}_{x} \vec{z}_{x} \vec{\nabla}_{z} \vec{\nabla}_{z}$$

$$\nabla_{x} \vec{E}_{T} = -\int_{B} \vec{E}_{z} \vec{\nabla}_{x} \vec{\nabla}_{x} \vec{E}_{z} - \eta_{TE} \vec{z}_{z} \vec{\nabla}_{x} \vec{\nabla}_{z}$$

$$\nabla_{x} \vec{E}_{T} + \int_{W} \mu \vec{H}_{z} \vec{z}_{z} = \int_{E} \vec{E}_{z} \eta_{TE} \vec{z}_{z} \vec{\nabla}_{x} \vec{H}_{z} + \int_{W} \mu \vec{H}_{z} \vec{z}_{z}$$

$$\nabla_{x} \vec{E}_{T} + \int_{W} \mu \vec{H}_{z} \vec{z}_{z} = \int_{E} \vec{E}_{z} \eta_{TE} \vec{z}_{z} \vec{\nabla}_{x} \vec{H}_{z} + \int_{W} \mu \vec{H}_{z} \vec{z}_{z}$$

$$\nabla_{x} \vec{E}_{T} + \int_{W} \mu \vec{H}_{z} \vec{z}_{z} = \int_{E} \vec{E}_{z} \eta_{TE} \vec{z}_{z} \vec{\nabla}_{x} \vec{H}_{z} + \int_{E} \mu \mu \vec{H}_{z} \vec{z}_{z}$$

$$\nabla \times \vec{E} + \int w \mu H_z \hat{z} = \int \frac{w \mu}{kc^2} \vec{z} \left[\nabla \vec{r} H_z + kc^2 H_z \right]$$

$$\frac{P}{I4}: \overline{\nabla_{I} \times H_{T}} = -\int_{K_{c}^{2}} \overline{B} \left[\overline{\nabla_{I} \times \nabla_{I} H_{Z}} + 1 \overline{\nabla_{I} \times 2} \times \overline{\nabla_{I}} E_{Z} \right]$$

$$\overline{\nabla_{I} \times H_{T}} = -\int_{K_{c}^{2}} \overline{B} \left[+ \underbrace{W \varepsilon}_{K_{c}^{2}} \overline{D} \overline{\nabla_{I}} E_{Z} \right]$$

$$\overline{\nabla_{I} \times H_{T}} - \int_{W \varepsilon} 2 E_{Z} = -\int_{W \varepsilon} 2 \overline{\nabla_{I}} E_{Z} - \int_{W \varepsilon} 2 E_{Z}$$

$$\overline{\nabla_{I} \times H_{T}} - \int_{W \varepsilon} 2 E_{Z} = -\int_{K_{c}^{2}} W \varepsilon 2 \overline{\nabla_{I}} E_{Z} - \int_{W \varepsilon} 2 E_{Z}$$

$$\nabla r \times HT - \mu \epsilon \hat{z} E_z = - \int \frac{\omega \epsilon}{k \epsilon^2} \hat{z} \left[\nabla r^2 E_z + k \epsilon^2 E_z \right]$$

$$\nabla_T \times \nabla_T A_z = 0,$$

$$\nabla_T \times (\hat{\mathbf{z}} \times \nabla_T A_z) = \hat{\mathbf{z}} \nabla_T^2 A_z,$$

$$\nabla_T \cdot (\hat{\mathbf{z}} \times \nabla_T A_z) = 0.$$

P/15:
$$\sqrt{r} \cdot E_{T} = -JB \left[\sqrt{r} \cdot \nabla T E_{z} - \gamma_{TE} \sqrt{r} \cdot Z \times \nabla H_{z} \right]$$

$$\frac{\nabla \vec{r} \cdot \vec{E} \cdot \vec{r} \cdot \vec{p} \vec{E}_{z}}{kc^{2}} = \int \frac{\vec{B}}{kc^{2}} \nabla \vec{r}^{2} \vec{E}_{z} - \int \vec{B} \vec{E}_{z} \\
\vec{\nabla} \vec{r} \cdot \vec{E} \vec{r} \cdot \vec{p} \vec{E}_{z} = \int \frac{\vec{B}}{kc^{2}} \left[\nabla \vec{r}^{2} \vec{E}_{z} - kc^{2} \vec{E}_{z} \right]$$

$$\overrightarrow{\nabla_1} \cdot \overrightarrow{H_T} = -j \underbrace{B}_{kc^2} \left[\overrightarrow{\nabla_1} \cdot \overrightarrow{\nabla_1} \overrightarrow{H_Z} + 1 \overrightarrow{\nabla_1} \cdot \overrightarrow{Z_X} \overrightarrow{\nabla_1} \overrightarrow{E_Z} \right]$$

$$\nabla T \cdot HT - JBHz = -JB \left[\nabla T^2 Hz + kc^2 Hz \right]$$

Resumo da equações;

$$\nabla_{T} \times E_{T} + j\omega\mu \,\hat{\mathbf{z}} H_{Z} = \frac{j\omega\mu}{k_{c}^{2}} \,\hat{\mathbf{z}} \left(\nabla_{T}^{2} H_{Z} + k_{c}^{2} H_{Z}\right)$$

$$\nabla_{T} \times H_{T} - j\omega\epsilon \,\hat{\mathbf{z}} E_{Z} = -\frac{j\omega\epsilon}{k_{c}^{2}} \,\hat{\mathbf{z}} \left(\nabla_{T}^{2} E_{Z} + k_{c}^{2} E_{Z}\right)$$

$$\nabla_{T} \cdot E_{T} - j\beta E_{Z} = -\frac{j\beta}{k_{c}^{2}} \left(\nabla_{T}^{2} E_{Z} + k_{c}^{2} E_{Z}\right)$$

$$\nabla_{T} \cdot H_{T} - j\beta H_{Z} = -\frac{j\beta}{k_{c}^{2}} \left(\nabla_{T}^{2} H_{Z} + k_{c}^{2} H_{Z}\right)$$

$$(3)$$

· Assim, as componente de campo Hz e Ez devem satisfazen as equações de Helmholtz

$$\nabla_1^2 E_z + K_c^2 E_z = 0$$

$$\nabla_1^2 H_z + K_c^2 H_z = 0$$

apois obten a solução pelas condições de fronteiras de Eze Hz as componentes de campo HT e ET são obtiglas pela equações apresentadas em (8). Pand obten o campo completo, (x, y, z, t), basta multiplican pelo fato edut-frz.

· Para o sistema de coordenades cartesiono:

$$E_{x}\hat{x} + E_{y}\hat{y} = -\int \frac{B}{K_{c}^{2}} \left[\partial_{x}E_{z}\hat{x} + \partial_{y}E_{z}\hat{y} - \eta_{TE}\partial_{x}H_{z}\hat{y} + \eta_{TE}\partial_{y}H_{z}\hat{x} \right]$$

logo,

$$\begin{aligned}
E_{x} &= -\int \frac{B}{K_{c}^{2}} \left[\partial_{x} E_{z} + \gamma_{TE} \partial_{y} H_{z} \right] \\
E_{y} &= -\int \frac{B}{K_{c}^{2}} \left[\partial_{y} E_{z} - \gamma_{TE} \partial_{x} H_{z} \right] \\
K_{c}^{2} \left[\partial_{y} E_{z} - \gamma_{TE} \partial_{x} H_{z} \right]
\end{aligned}$$

$$(\partial_x^2 + \partial_y^2)E_z + k_c^2 E_z = 0$$

$$(\partial_x^2 + \partial_y^2)H_Z + k_c^2 H_Z = 0$$



$$E_{x} = -\frac{j\beta}{k_{c}^{2}} \left(\partial_{x} E_{z} + \eta_{TE} \, \partial_{y} H_{z} \right)$$

$$E_{y} = -\frac{j\beta}{k_{c}^{2}} (\partial_{y} E_{z} - \eta_{TE} \partial_{x} H_{z})$$

$$H_x = -\frac{j\beta}{k_c^2} (\partial_x H_z - \frac{1}{\eta_{TM}} \partial_y E_z)$$

$$H_y = -\frac{j\beta}{k_c^2} (\partial_y H_z + \frac{1}{\eta_{TM}} \partial_x E_z)$$

· Soluções para diterentes modos;

ainda,
$$\eta_{fE} = \frac{WH}{B} = \eta \frac{W}{BC} = \eta \cdot \eta_{TM} = \frac{B}{WE} = \eta \cdot BC = \eta$$

$$(\partial_x^2 + \partial_y^2)E_z + k_c^2E_z = 0$$

$$(\partial_x^2 + \partial_y^2) H_z^2 + k_c^2 H_z^2 = 0$$

$$\nabla E_z \times \hat{\mathbf{z}} - j\beta \hat{\mathbf{z}} \times \mathbf{E}_T = -j\omega\mu\mathbf{H}_T$$

$$\mathbf{\nabla} \times \mathbf{E}_T + j\omega \mu H_z \,\hat{\mathbf{z}} = 0$$

$$\nabla H_z \times \hat{\mathbf{z}} - j\beta \hat{\mathbf{z}} \times \mathbf{H}_T = j\omega \varepsilon \mathbf{E}_T \setminus$$

$$\nabla \times \mathbf{H}_T - j\omega \varepsilon E_z \,\hat{\mathbf{z}} = 0$$

$$\mathbf{\nabla}_T \cdot \mathbf{E}_T - j\beta E_z = 0$$

$$\mathbf{\nabla}_T \cdot \mathbf{H}_T - j\beta H_z = 0$$

$$H_T = B 2xE_T = 1 2xE_T^2 = 1 , 2xE_T^2$$

WH

 η_{re}
 η_{re}

$$H_{1} = 1$$
, $2 \times E_{1}$

O campo eletrico pode sen de terminado pon

$$\nabla_{r} \times E_{r} = 0$$
 e $\nabla_{r} \cdot E_{r} = 0$

z) Modo TM: (Hz=Qe Ez +0)

$$(\partial_x^2 + \partial_y^2)E_z + k_c^2 E_z = 0$$

$$(\partial_x^2 + \partial_y^2)M_z^2 + k_z^2 M_z^2 = 0$$

$$\overline{E_{T}}^{2} = -\int_{\mathbb{R}}^{\mathbb{R}} \left[\overline{\nabla_{T}^{2}} E_{Z} - \gamma_{TE} \hat{z}_{X} \overline{\nabla_{T}^{2}} H_{Z} \right] = \sum_{K \in \mathbb{Z}}^{2} - \int_{\mathbb{R}}^{2} \overline{\nabla_{T}^{2}} E_{Z}$$

· Determinando HT:

$$\nabla E_{z} \times \hat{\mathbf{z}} - j\beta \,\hat{\mathbf{z}} \times \mathbf{E}_{T} = -j\omega\mu\mathbf{H}_{T} \qquad (11) \qquad \nabla \times \mathbf{E}_{T} + j\omega\mu H_{z} \,\hat{\mathbf{z}} = 0 \qquad (13)$$

$$\nabla H_{z} \times \hat{\mathbf{z}} - j\beta \,\hat{\mathbf{z}} \times \mathbf{H}_{T} = j\omega\varepsilon\mathbf{E}_{T} \qquad (12) \qquad \nabla \times \mathbf{H}_{T} - j\omega\varepsilon E_{z} \,\hat{\mathbf{z}} = 0 \qquad (14)$$

$$\nabla_{T} \cdot \mathbf{E}_{T} - j\beta E_{z} = 0 \qquad (15)$$

$$\nabla_{T} \cdot \mathbf{H}_{T} - j\beta H_{z} = 0 \qquad (16)$$

Pode-se esvieven (11) e (12) como;

$$\int \beta z^2 x \vec{E} \vec{T} - \int w \mu \vec{H} \vec{T} = \vec{\nabla} \vec{E}_z \times \vec{z} = \beta z^2 x \vec{E} \vec{T} - w \mu \vec{H} \vec{T} = - \int \vec{\nabla} \vec{E}_z \times \vec{z}$$

=> $\beta z^2 x \vec{E} \vec{T} - w \mu \vec{H} \vec{T} = 1 z^2 x \vec{\nabla} \vec{E}_z$

$$\beta \,\hat{\mathbf{z}} \times \mathbf{E}_T - \omega \mu \mathbf{H}_T = j\hat{\mathbf{z}} \times \nabla_T E_Z$$
$$\omega \epsilon \,\hat{\mathbf{z}} \times \mathbf{E}_T - \beta \mathbf{H}_T = -j \nabla_T H_Z$$

ainda
$$H\vec{1} = -\frac{1}{2} \vec{2} \times \vec{\nabla} \vec{\tau} \vec{E}_{2} + \underline{B} \vec{2} \times \vec{E} \vec{1} = -\frac{1}{2} \underline{1} \cdot \hat{2} \times \vec{\nabla} \vec{\tau} \vec{E}_{2} + \underline{1} \vec{2} \times \vec{E} \vec{1}$$

$$WH \qquad WH \qquad B \eta \vec{\tau} \vec{E} \qquad \eta \vec{\tau} \vec{E}$$

$$(aminho ehhado)$$

$$H_T - \frac{1}{\eta_{TM}} \hat{\mathbf{z}} \times \mathbf{E}_T = \frac{j}{\beta} \nabla_T H_Z$$

$$E_T - \eta_{TE} H_T \times \hat{\mathbf{z}} = \frac{j}{\beta} \nabla_T E_Z$$

$$\beta \hat{\mathbf{z}} \times \mathbf{E}_{T} - \omega \mu \mathbf{H}_{T} = j \hat{\mathbf{z}} \times \nabla_{T} \mathbf{E}_{Z}$$

$$\omega \epsilon \hat{\mathbf{z}} \times \mathbf{E}_{T} - \beta \mathbf{H}_{T} = -j \nabla_{T} \mathbf{H}_{Z}$$

$$= > \beta \mathbf{H}_{T} = \omega \epsilon \hat{\mathbf{z}} \times \mathbf{E}_{T}$$

$$= + \mathbf{I}_{T} \hat{\mathbf{z}} \times \mathbf{E}_{T}$$

$$= + \mathbf{I}_{T} \hat{\mathbf{z}} \times \mathbf{E}_{T}$$

$$\nabla_T^2 E_z + k_c^2 E_z = 0$$

$$\mathbf{E}_T = -j \frac{\beta}{k_c^2} \mathbf{\nabla}_T E_z$$

$$\mathbf{H}_T = \frac{1}{\eta_{\mathsf{TM}}} \hat{\mathbf{z}} \times \mathbf{E}_T.$$

Mode TM

3) Modo TE: (Ez=0 e Hz +0)

$$(\partial_x^2 + \partial_y^2)E_z^{7} + k_c^2 E_z^{7} = 0$$
$$(\partial_x^2 + \partial_y^2)H_z + k_c^2 H_z = 0$$

$$\overrightarrow{H_T} = -J \underbrace{B}_{kc^2} \left[\overrightarrow{\nabla_r} H_{Z} + \underbrace{1}_{\gamma_r} \widehat{z_x} \overrightarrow{\nabla_r} E_Z \right]$$

• H_T = - / B
$$\nabla T$$
 Hz

$$\beta \hat{\mathbf{z}} \times \mathbf{E}_T - \omega \mu \mathbf{H}_T = j \hat{\mathbf{z}} \times \nabla_T \mathbf{E}_Z$$
$$\omega \in \hat{\mathbf{z}} \times \mathbf{E}_T - \beta \mathbf{H}_T = -j \nabla_T \mathbf{H}_Z$$

$$\beta \vec{2} \times \vec{E_T} = \omega_{\mu} \vec{H_T} = \hat{2} \times \vec{E_T} = \eta_{TE} \vec{H_T} = \hat{-} \vec{E_T} = \eta_{TE} \hat{2} \times \vec{H_T}$$

• $\vec{E_T} = \eta_{TE} \vec{H_T} \times \hat{2}$

$$\begin{split} &\nabla_T^2 H_z + k_c^2 H_z = 0 \\ &\mathbf{H}_T = -j \frac{\beta}{k_c^2} \mathbf{\nabla}_T H_z \\ &\mathbf{E}_T = \eta_{\mathsf{TE}} \, \mathbf{H}_T \times \hat{\mathbf{z}}. \end{split}$$

Modo TE

4) Modos hi bnidos:

Análise - slab simétrico ($n_c = n_s$)

Etapas de análise (de acordo com a geometria do problema):

- 1. aplicação das equações de Maxwell;
- determinação das componentes dos campos;
- 3. aplicação das condições de contorno;
- 4. determinação da equação característica;
- a partir da solução da equação característica, determinar os parâmetros dos modos, como índice de refração e constante de propagação.

De acordo com as características do guia, como homogeneidade e simetria, é possível classificar a solução, como TEM ou TE, e determinar diretamente as componentes dos campos.

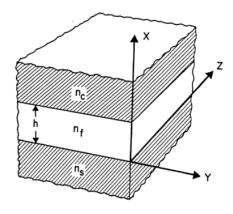


Figura 6: Guia dielétrico de placas planas assimétrico.

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A Fig. 6 ilustra um guia dielétrico retangular formado por três camadas dielétricas, cujo índice de refração é n_i (i=c,f,s). O guia é simétrico ($n_c=n_s$), em que $n_f>n_s$ e, portanto, satisfaz a condição de reflexão interna total. Assim, o campo propaga-se no dielétrico, cujo índice é igual a n_f . Tomando a Eq. 25 e Eq. 26 (geometria retangular) e considerando $\partial/\partial y=0$, estas equações são reescritas como:

$$\overrightarrow{E_T} = -\int_{\mathbb{R}} \underbrace{\begin{bmatrix} \overrightarrow{\nabla_T} E_z - \gamma_{TE} \overrightarrow{z} \times \overrightarrow{\nabla_T} H_z \end{bmatrix}}_{Kc^2}$$

$$\overrightarrow{H_T} = -\int_{\mathbb{R}} \underbrace{\begin{bmatrix} \overrightarrow{\nabla_T} H_z + 1 \ \overrightarrow{Z} \times \overrightarrow{\nabla_T} E_z \end{bmatrix}}_{\eta_{TM}}$$

$$H_{7} = -\int \frac{B}{k_{c^{2}}} \left[\frac{d}{dx} H_{z} \hat{x} + \frac{1}{\eta_{TM}} \frac{d}{dx} E_{z} \hat{y} \right]$$

$$Hx = -f \frac{B}{Ke^2} \frac{d}{dx} Hz = Hy = -f \frac{B}{Ke^2} \frac{1}{\eta TM} \frac{d}{dx} Ez$$

0 M

$$\begin{bmatrix} E_{X} \end{bmatrix} = -f \frac{B}{Kc} \begin{bmatrix} 1 & 0 \\ 0 & -\eta_{TE} \end{bmatrix} \begin{bmatrix} dE_{Z}/dx \\ dH_{Z}/dx \end{bmatrix} e \begin{bmatrix} H_{X} \\ H_{Y} \end{bmatrix} = -f \frac{B}{Kc^{2}} \begin{bmatrix} 1 & 0 \\ 0 & 1/\eta_{TM} \end{bmatrix} \begin{bmatrix} dH_{Z}/dx \\ dE_{Z}/dx \end{bmatrix}$$

As componentes longitudinais:
$$\frac{d^2Ez}{dx^2} + kc^2Ez = 0$$
 e $\frac{d^2Hz}{dx^2} + kc^2Hz = 0$

Observa-se que ha dois conjuntos de equações, pontanto, pode-se dividir em:

Então, podemos dividir as equações em dois grupos:

Modo TM:

$$E_x = -j\frac{\beta}{k_c^2} \frac{dE_z}{dx}$$

$$H_y = -j\frac{\omega\varepsilon}{k_c^2} \frac{dE_z}{dx}$$

$$\frac{d^2E_z}{dx^2} + k_c^2E_z = 0.$$

Modo TE:

$$E_y = j \frac{\omega \mu}{k_c^2} \frac{\mathrm{d}H_z}{\mathrm{d}x}$$

$$H_x = -j \frac{\beta}{k_c^2} \frac{\mathrm{d}H_z}{\mathrm{d}x}$$

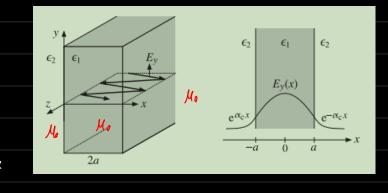
$$\frac{\mathrm{d}^2 H_z}{\mathrm{d}x^2} + k_c^2 H_z = 0.$$

· No intenion e extenion do guia as equações de Helmholtz são:

$$\frac{d^{2}Hz + kc1^{2}Hz = 0}{dx^{2}} e \frac{dHz + kc2^{2}Hz = 0}{dx^{2}}$$

$$- extenion$$

• Solucionanda d ED: $h^2 + k_{c1}^2 = 9 = 9 + 9 = \frac{1}{2} \int_{\mathbb{R}^2} k_{c1} dk_{c1} x = A\cos(k_{c1}x) + Bhen(k_{c1}x)$



• Pand o extenior, assume-se que $K_{cz} = -\infty c^2$ L_{0} L_{0}

$$logo, h^2 - \alpha c^2 = 0 = 7 h = \pm \alpha c$$

 $|x| > 0$ $H_Z(x) = Ce^{\alpha c x} + be^{-\alpha c x}$

Adotando: Kc1= kc e Kcz=-cc=> kc= ko²n²-β²; - cc² = ko²n²-β² cc² = β²-ko²n²²

O campo deve decain fona do cone exponecialmente. Assim:

$$\frac{d}{dx} H_{z(x)} + Kc^{2} H_{z(x)} = O(|x| \le \alpha)$$

$$\frac{d}{dx} H_{z(x)} - \alpha c H(x) = O(|x| \ge \alpha)$$

$$H_{x(x)} = \int \underline{B} \text{ KeEben } (\text{Kex} - \theta); |x| = a$$

$$K_{c^{2}} \qquad (E = H_{1}; c = H_{2}; P = H_{3})$$

$$H_{x(x)} = -\int \underline{B} c e^{-\alpha c x}; x > a$$

$$\alpha c$$

$$H_{x(x)} = \int \underline{B} p e^{\alpha c x}; x \leq a$$

$$H_{x(x)} = \int \underline{B} H_1 \operatorname{ben}(K_{cx} - \theta); |x| = a$$

$$K_{c}$$

$$H_{x(x)} = -\int \underline{B} H_2 e^{-\alpha_{cx}}; \quad x > a$$

$$\alpha_{c}$$

$$H_{x(x)} = \int \underline{B} H_3 e^{\alpha_{cx}}; \quad x \leq a$$

$$\alpha_{c}$$

· Para Ey(x) o modo TE se tem: Ey(x) = - MTE Hx(x)

$$\begin{cases} H_{x(x)} = \int \underline{B} H_1 \text{ ben } (k_{cx} - \theta); |x| \leq a \\ k_c \end{cases}$$

$$K_{x(x)} = -\int \underline{B} H_2 e^{-\alpha_{cx}}; |x| = a$$

$$K_{x(x)} = -\int \underline{B} H_2 e^{-\alpha_{cx}}; |x| \leq a$$

$$E_{y(x)} = -\int \underline{B} H_2 \eta_{TE} e^{-\alpha_{cx}}$$

$$E_{y(x)} = -\int \underline{B} H_2 \eta_{TE} e^{-\alpha_{cx}}$$

$$E_{y(x)} = -\int \underline{B} H_3 \eta_{TE} e^{\alpha_{cx}}$$

$$E_{y(x)} = -\int \underline{B} H_3 \eta_{TE} e^{\alpha_{cx}}$$

Logo,
$$\begin{cases} E_{y(x)} = E_{1} \operatorname{ben}(k(x-\theta)), |x| = a \\ E_{y(x)} = E_{2}e^{-\alpha cx}, |x| = a \\ E_{y(x)} = E_{3}e^{-\alpha cx}, |x| = a \end{cases}$$

The boundary conditions state that the tangential components of the magnetic and electric fields, that is, H_Z , E_Y , are continuous across the dielectric interfaces at x=-a and x=a. Similarly, the normal components of the magnetic field $B_X=\mu_0 H_X$ and therefore also H_X must be continuous. Because $E_Y=-\eta_{TE}H_X$ and η_{TE} is the same inboth media, the continuity of E_Y follows from the continuity of H_X . The continuity of H_Z at X=a and X=a implies that:

$$H_{x(x)} = \int \frac{B}{kc} H_1 \operatorname{ben}(kcx - \theta) \qquad \qquad \left(\frac{H_1 \operatorname{ben}(kcx - \theta) = -1}{kc} \right) \qquad H_2 e^{-\alpha_c x} \qquad (x=a)$$

$$H_{x(x)} = -\int \frac{B}{kc} H_2 e^{-\alpha_c x} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kcx - \theta) = -1}{kc} H_3 e^{\alpha_c x} \qquad (x=-a)$$

$$H_{x(x)} = \int \frac{B}{kc} H_3 e^{\alpha_c x} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_2 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad = \sum_{\alpha \in \mathbb{Z}} \frac{H_1 \operatorname{ben}(kc\alpha - \theta) = -1}{kc} H_3 e^{-\alpha_c \alpha} \qquad =$$

Po sistema (8) tem-se:
$$H_z = -H_z = -\frac{\alpha_c}{\alpha_c} H_1 e^{\alpha_c \alpha_0} ben(-k_c \alpha - \theta)$$

$$K_c$$

$$H_2 = -H_3 = \frac{\alpha_c}{\alpha_c} H_1 e^{\alpha_0 \alpha_0} ben(k_c \alpha + \theta)$$
(9)
$$K_c$$

$$H_{z(x)} = H_{1} \cos(k_{1}x - \theta) = H_{1} \cos(k_{2}x - \theta) = H_{2}e^{-\alpha z} (x = a)$$

$$H_{z(x)} = H_{2}e^{\alpha z} \cos(k_{2}x - \theta) + H_{1} \cos(k_{2}x - \theta) + H_{2} \cos(k_{2}x - \theta) + H_{1} \cos(k_{2}x - \theta)$$

$$H_{z(x)} = H_{3}e^{\alpha z} \cos(k_{2}x - \theta) + H_{1} \cos(k_{2}x - \theta) + H_{2} \cos(k_{2}x - \theta)$$

$$H_{z(x)} = H_{3}e^{\alpha z} \cos(k_{2}x - \theta) + H_{1} \cos(k_{2}x - \theta)$$

$$H_{z(x)} = H_{3}e^{\alpha z} \cos(k_{2}x - \theta) + H_{2} \cos(k_{2}x - \theta)$$

$$H_{z(x)} = H_{3}e^{\alpha z} \cos(k_{2}x - \theta)$$

$$H_{z(x)} = H_{3}e^{\alpha z} \cos(k_{2}x - \theta)$$

• (10) = (9) =>
$$H_1e^{\alpha_{c}a}$$
 cas $(k_{c}a - \theta) = \underline{\alpha_{c}} H_1e^{\alpha_{c}a}$ ben $(k_{c}a + \theta)$

$$\alpha c = k c \frac{c o s (k c a - b)}{b o n (k c a + b)}$$

Definição do guia

O guia dielétrico planar é formado por N camadas de material dielétrico de espessura h_i e índice de refração n_i (i=f,n,s). Nosso primeiro estudo consiste em um guia dielétrico planar de 3 (três) camadas, conforme ilustra a Fig. 7. A relação entre os índices é

$$0 < n_c < n_s < n_f. (79)$$

Por esta relação, devido à reflexão interna total, o campo é confinado majoritariamente no núcleo (índice n_f). Entretanto, há campo evanescente na casca (índice n_c) e no substrato (índice n_s). É importante destacar que μ_r (permeabilidade magnética) em todos os guias dielétricos é unitária.

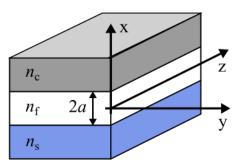


Figura 7: Ilustração do guia dielétrico planar e a escolha do sistema de coordenadas. Observe que a origem da coordenada \times é o centro da camada cujo índice é n_f .

$$\nabla_T^2 H_z + k_c^2 H_z = 0$$
$$\mathbf{H}_T = -j \frac{\beta}{k_c^2} \nabla_T H_z$$

$$\mathbf{E}_T = \eta_{\mathsf{TE}} \, \mathbf{H}_T imes \hat{\mathbf{z}}.$$

$$\frac{\partial}{\partial y} = 0$$

assim,
$$\frac{d^2Hz + kc^2Hz = 9}{dx^2}$$
; $Hx = -\int \frac{B}{kc^2} \frac{d}{dx} Hz$; $Ex = 0$

$$E_x = \eta_{TE} \cdot H_x \hat{x} \hat{x} \hat{z} = F_y = H_{TE} f \frac{B}{K_c^2} \frac{d}{dx}$$

$$\begin{cases} \frac{\mathrm{d}^2 H_z}{\mathrm{d}x^2} + k_i^2 H_z = 0 \\ H_x = -j \frac{\beta}{k_i^2} \frac{\mathrm{d}H_z}{\mathrm{d}x} \\ E_y = j \eta_{\mathrm{TE}} \frac{\beta}{k_i^2} \frac{\mathrm{d}H_z}{\mathrm{d}x} \end{cases}$$

· As componentes de Hz devem sen iguais, funções continuas, pana as três negições do guia. Assim, as condições de contonno senão satisfeitas.

1)
$$\frac{d^2H_{z(x)} + K_c^2H_{z(x)} = 0}{dx^2}$$
 $\times \otimes a = > h^2 + K_c^2 = 0 \sim h = \pm \int K$

Logo, $K_c^2 = -\infty e^2$
 $\times \otimes A = > h^2 + K_c^2 = 0 \sim h = \pm \int K$

Logo, $K_c^2 = -\infty e^2$

assim,
$$\frac{d}{dx}H_{z(x)} - \frac{\alpha c^2}{dx}H_{z(x)} = 0 \Rightarrow h^7 - \alpha z^7 = 0 \Rightarrow h = \pm \alpha_c$$

bontanto, $H_{z(x)} = A e^{\alpha c x} + A^7 e^{-\alpha c x} \quad (x > a)$

Z) Pana o cone:
$$\frac{d^2H_{z(x)} + k_{\sharp}^2H_{z(x)} = 0}{dx^2} \Rightarrow h = \frac{t}{f}k_{\sharp}$$

$$\log_{0} \frac{H_{z(x)} = B\cos(k_{\sharp}x) + C\sin(k_{\sharp}x)}{(|x| \leq a)}$$

3) Para o substrato:
$$\frac{d^2H_{z(x)} - \alpha_s^2H_{z(x)} = 0}{dx^2}$$
 $H_{z(x)} = De^{\alpha sx} + D^1e^{-\alpha sx} (x \le -a)$

$$\mathcal{R}_{e \leq \mu M \ 0}: H_z(x) = \begin{cases} Ae^{-\alpha_c x} + A'e^{\alpha_c x}, & x \geq a \\ B\cos(k_f x) + C\sin(k_f x), & |x| \leq a \\ De^{\alpha_s x} + D'e^{-\alpha_s x}, & x \leq -a \end{cases}$$

Para essas condições senem sortisteitas, tem-se:

$$H_z(x) = \begin{cases} Ae^{-\alpha_c x} + A'e^{\alpha_c x}, & x \ge a \\ B\cos(k_f x) + C\sin(k_f x), & |x| \le a \\ De^{\alpha_s x} + D'e^{-\alpha_s x}, & x \le -a \end{cases}$$

A henx + B cosx =
$$\sqrt{A^2 + B^2} \ln (x + t_g^{-1}(B_A))$$

$$H_z(x) = \begin{cases} Ae^{-\alpha_c x}, & x \geqslant a \\ H_0 \sin(k_f x + \phi), & |x| \leqslant a \\ De^{\alpha_s x}, & x \leqslant -a \end{cases}$$

Aplicando as condições de fronteira: Hobin(
$$k_{fa} + \emptyset$$
) = $A e^{-\alpha ca}$ (x= a)
$$A = Hobin(k_{fa} + \emptyset) e^{\alpha ca}$$

e

Hobin
$$(-k_{\beta}a+\alpha)=De^{-\alpha_{\beta}a}$$
 $(X=-\omega)$
 $D=Hosin(-k_{\beta}a+\alpha)e^{\alpha_{\beta}a}$

Desse modo,

$$H_{2}(x) = \begin{cases} H_{0} h_{0} \ln(k_{\beta} + \alpha) e^{\alpha_{\zeta} \alpha} e^{-\alpha_{\zeta} x} \\ H_{0} h_{0} \ln(k_{\beta} + \alpha) \end{pmatrix} e^{\alpha_{\zeta} \alpha} e^{\alpha_{\zeta} x} \\ -H_{0} h_{0} \ln(k_{\beta} - \alpha) e^{\alpha_{\zeta} \alpha} e^{\alpha_{\zeta} x} \\ \chi \leq \alpha \end{cases}$$

· Para deferminan Hx(x):

$$\begin{cases} \frac{\mathrm{d}^2 H_z}{\mathrm{d}x^2} + k_i^2 H_z = 0 \\ H_x = -j \frac{\beta}{k_i^2} \frac{\mathrm{d}H_z}{\mathrm{d}x} \\ E_y = j \eta_{\mathrm{TE}} \frac{\beta}{k_i^2} \frac{\mathrm{d}H_z}{\mathrm{d}x} \end{cases}$$

Casca:
$$H \times (x) = \int B \frac{\alpha_c}{\alpha_c} H_0 \sin(k_{\beta} a + \emptyset) e^{\alpha_c} a e^{-\alpha_c x}$$

$$- \alpha_c^2$$

$$H \times (x) = -\int H_0 \beta \alpha_c \sin(k_{\beta} a + \emptyset) e^{\alpha_c} a e^{-\alpha_c x} (x > a)$$

Cone:
$$H_{X}(x) = -\int \underline{B} k_{\sharp} H_{0} \cos(k_{\sharp}x + \emptyset)$$

$$k_{\sharp}^{2}$$

$$H_{X}(x) = -\int \underline{B} H_{0} k_{\sharp}^{-1} \cos(k_{\sharp}x + \emptyset) \quad (|x| \leq a)$$

Substrati:
$$H(x) = -\int \underline{B}$$
, $(-\infty + 10) \sin(k + \alpha - \alpha) e^{\alpha + \alpha} e^{\alpha + \alpha}$
 $-\alpha + \frac{3}{2}$
 $H(x) = -\int H_0 B \propto \sin(k + \alpha - \alpha) e^{\alpha + \alpha} e^{\alpha + \alpha}$ $(x + \alpha - \alpha)$

• Pana Ey(x):
$$E_y = + \eta_{TE} f \frac{B}{Kc^2} \frac{d}{dx}$$

$$H_{z(x)} = \begin{cases} H_0 \sin(k_{\sharp} a + \emptyset) e^{\alpha_{\zeta} a} e^{-\alpha_{\zeta} x}, & x > 0 \\ H_0 \sin(k_{\sharp} x + \emptyset), & |x| \leq a \end{cases}$$

$$-H_0 \sin(k_{\sharp} a - \emptyset) e^{\alpha_{\zeta} a} e^{\alpha_{\zeta} x}, & |x| \leq a \end{cases}$$

NTE = WM/B

· Em hesumo:

$$H_z(x) = H_0 \begin{cases} \sin(k_f a + \phi) e^{-\alpha_c(x-a)}, & x \geqslant a \\ \sin(k_f x + \phi), & |x| \leqslant a \\ -\sin(k_f a - \phi) e^{\alpha_s(x+a)}, & x \leqslant -a \end{cases}$$

$$H_{x}(x) = -jH_{0}\beta \begin{cases} \alpha_{c}^{-1}\sin(k_{f}a + \phi)e^{-\alpha_{c}(x-a)}, & x \geqslant a \\ k_{f}^{-1}\cos(k_{f}x + \phi), & |x| \leqslant a \\ \alpha_{s}^{-1}\sin(k_{f}a - \phi)e^{\alpha_{s}(x+a)}, & x \leqslant -a \end{cases}$$

$$E_{y}(x) = jH_{0}\omega\mu \begin{cases} \alpha_{c}^{-1}\sin(k_{f}a + \phi)e^{-\alpha_{c}(x-a)}, & x \geqslant a \\ k_{f}^{-1}\cos(k_{f}x + \phi), & |x| \leqslant a \\ \alpha_{s}^{-1}\sin(k_{f}a - \phi)e^{\alpha_{s}(x+a)}, & x \leqslant -a \end{cases}$$

· As condições de contonno de Ey(x) nesulta:

(1):
$$1 \cos(k_{\dagger} a + \emptyset) = 1 \sin(k_{\dagger} a + \emptyset) -> \tan(k_{\dagger} a + \emptyset) = \alpha c \quad (x=0)$$
 k_{\dagger}
 αc

(2):
$$\frac{1}{k_F} \cos(-\alpha k_F + \emptyset) = \underline{1} \sin(k_F \alpha - \emptyset) \rightarrow \frac{1}{k_F} \cos(k_F \alpha - \emptyset) = \frac{1}{k_F} \cos(k_F \alpha$$

Ob tenn-se:

$$\frac{\tan(k_{\sharp} a + \emptyset) = \alpha c}{k_{\sharp}} = \alpha c} = \lambda c = \lambda c + \beta = \alpha c + \delta c + \delta$$

Assim,
$$2kfa = anctan(\alpha e/kf) + anctan(\alpha s/kf)$$
 (11)
Sabondo que: $anctan(a) \pm anctan(b) = anctan(a \pm b)$

logo, de (11):
$$2 \text{ kpa} = \frac{1}{3} \left(\frac{\alpha \frac{1}{2} \text{ kp} + \frac{\alpha \frac{1}{2} \text{ kp}}{1 - \alpha \frac{1}{2} \frac{\alpha \frac{1}{2} \text{ kp}}{1 - \alpha \frac{1}{2} \frac{\alpha \frac{1}{2} \text{ kp}}{2}} \right) = 2 \text{ kan} (2 \text{ kpa}) = \frac{\alpha \frac{1}{2} + \alpha \frac{1}{2} \frac{\alpha \frac{1}{2} \frac{1}{2}}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2}}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2}}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{\alpha \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{1}{2} \frac{\alpha \frac{1}{2} \frac{1}{2}$$

$$\frac{\int dx}{dx} = \frac{k f(\alpha c + \alpha s)}{k f^{2} - \alpha c \alpha s} \qquad (12)$$

Para praticidade nas manipulações algébricas, são definidos os parâmetros modais $u=k_fh,\,v=\alpha_sh$ e $w=\alpha_ch$. Esses parâmetros são positivos (ou nulos) e adimensionais. A partir desses parâmetros modais, a Eq. (12) é reescrita como

$$fan(zu) = \frac{u/\alpha \left[v/\alpha + w/\alpha \right]}{(v/\alpha)^2 - vw/\alpha^2} = \frac{u \left[v + w \right]}{u^2 - vw}$$

$$Logg, \quad fan(zu) = \frac{u[V+W]}{u^2 - VW}$$
 (13)

assim:

$$\begin{cases}
U = \alpha k_f = \alpha k_0 \sqrt{n_f^2 - n_{eff}^2} \\
W = \alpha \alpha c = \alpha k_0 \sqrt{n_{eff}^2 - n_{c}^2}
\end{cases}$$

$$V = \alpha \alpha s = \alpha k_0 \sqrt{n_{eff}^2 - n_{s}^2}$$

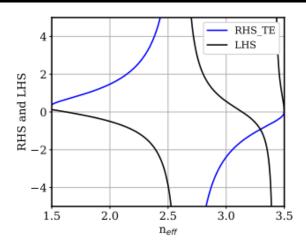


Figura 8: Curvas da equação característica, considerando $n_f=3.5,\,n_s=1.45,\,n_c=1.0,\,a=0,25~\mu\mathrm{m}$ e $\lambda=1550~\mathrm{nm}.$

=> De tenminando gutros panâmetros impontantes:

$$\tan(k_f a + \phi) = \frac{\alpha_c}{k_f}$$

$$\tan(k_f a - \phi) = \frac{\alpha_s}{k_f}$$

$$\tan(u - \phi) = \frac{\alpha_s}{k_f}$$

$$\tan(u - \phi) = \frac{v}{u}$$

$$\tan(u - \phi) = \frac{v}{u}$$

$$\frac{\tan(u+\phi) = \tan(u+\alpha+m\pi) = w}{u} = \frac{1}{2} u + \alpha + m\pi = \frac{\tan(w)}{u} (14)$$

$$\frac{\tan(u-\alpha) = \tan(u-\alpha+m\pi)}{u} = \frac{1}{2} u - \alpha + m\pi = \frac{\tan(w)}{u} (15)$$

$$\frac{\tan(u+\alpha) = \tan(u-\alpha+m\pi)}{u} = \frac{1}{2} u + \frac{\tan(w)}{u} (15)$$

$$\frac{\tan(u+\alpha) = \tan(u+\alpha+m\pi)}{u} = \frac{\tan(w)}{u} (15)$$

$$\frac{\tan(u+\alpha) = \tan(u+\alpha+m\pi)}{u} = \frac{\tan(w)}{u} (14)$$

$$\frac{\tan(u+\alpha) = \tan(u+\alpha+m\pi)}{u} = \frac{\tan(w)}{u} = \frac{\tan(w)}{u}$$

$$\frac{\tan(u+\alpha) = \tan(u+\alpha+m\pi)}{u} = \frac{\tan(u+\alpha+m\pi)}{u}$$

$$\frac{\tan(u+\alpha) = \tan(u+\alpha+m\pi)}{u}$$

$$\frac{\tan(u+\alpha) = \tan(u+\alpha+m\pi)}{u}$$

$$-(14)-(15)$$
: $-28 = m\pi - \tan^{-1}(\frac{w}{u}) + \tan^{-1}(\frac{v}{u})$
 $-28 = m\pi + \tan^{-1}(\frac{w}{u}) - \tan^{-1}(\frac{v}{u})$

Em hesumo:
$$2u = m\pi + \arctan\left(\frac{w}{u}\right) + \arctan\left(\frac{v}{u}\right)$$

$$2\phi = m\pi + \arctan\left(\frac{w}{u}\right) - \arctan\left(\frac{v}{u}\right)$$

$$\phi = m\pi + \arctan\left(\frac{w}{u}\right) - \arctan\left(\frac{v}{u}\right)$$

$$2u = \tan^{-1}\left(\frac{w}{u}\right) + \tan^{-1}\left(\frac{v}{u}\right) => 2u = \tan^{-1}a + \tan^{-1}b$$

$$2u = \tan^{-1}\left(\frac{a+b}{1-ab}\right) => \tan(2u) = \frac{a+b}{1-ab} = \frac{w/u + v/u}{1-wv/u^2}$$

$$\tan(2u) = \frac{u(w+v)}{u^2-wv}$$

Partindo do parâmetro modal u, temos (mostre!)

$$u = \sqrt{1 - \underbrace{\left(\frac{\beta^2 - k_0^2 n_s^2}{k_0^2 n_f^2 - k_0^2 n_s^2}\right)}_{b} \underbrace{\left[k_0 a \sqrt{\left(n_f^2 - n_s^2\right)}\right]}_{V},$$

C> 7. Demonstran

- b: Constante de propagação normalizada
 V: Frequencia normalizada

· Relação entre o parametro moda we os demais:

$$\left(\frac{w}{u}\right)^{2} - \left(\frac{x \cdot \alpha}{k + \alpha}\right)^{2} - \frac{\beta^{2} - k_{0}^{2}n_{c}^{2}}{k_{0}^{2}n_{f}^{2} - \beta^{2}} = \frac{\beta^{2} - k_{0}^{2}n_{c}^{2} + k_{0}^{2}n_{s}^{2} - k_{0}^{2}n_{s}^{2}}{k_{0}^{2}n_{f}^{2} - \beta^{2} + k_{0}^{2}n_{s}^{2} - k_{0}^{2}n_{s}^{2}} = \right)$$

$$\frac{\beta^{2} - k_{0}^{2} m s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n z^{2}}{k_{0}^{2} n \xi^{2} - k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} - \beta^{2}}$$

$$\frac{\beta^{2} - k_{0}^{2} m s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}}{k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}} = \frac{\beta^{2} - k_{0}^{2} m s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}}{k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}} = \frac{\beta^{2} - k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}}{k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}} = \frac{\beta^{2} - k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}}{k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2}} = \frac{\beta^{2} - k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} - k_{0}^{2} n s^{2}}{k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2}} = \frac{\beta^{2} - k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2}}{k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2}} = \frac{\beta^{2} - k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2} + k_{0}^{2} n s^{2}}{k_{0}^{2} n s^{2} + k_{0}^{2} n s^{$$

$$\frac{B^{2} - k_{0}^{2}ms^{2}}{k_{0}^{2}mf^{2} - k_{0}^{2}ms^{2}} + \frac{k_{0}^{2}ns^{2} - k_{0}^{2}mc^{2}}{k_{0}^{2}mf^{2} - k_{0}^{2}ns^{2}} = \frac{b + 6}{k_{0}^{2}mf^{2} - k_{0}^{2}ns^{2}} + \frac{k_{0}^{2}ns^{2} - k_{0}^{2}ns^{2}}{k_{0}^{2}nf^{2} - k_{0}^{2}ns^{2}} + \frac{k_{0}^{2}ns^{2} - k_{0}^{2}ns^{2}}{k_{0}^{2}nf^{2} - k_{0}^{2}ns^{2}}$$

Pontanto,
$$S = \frac{k_0^2 n s^2 - k_0^2 n c^2}{k_0^2 n \xi^2 - k_0^2 n s^2} = \frac{n s^2 - n c^2}{n \xi^2 - n s^2}$$

$$\delta = \frac{\eta_5^2 - \eta_c^2}{\eta_c^2 - \eta_s^2}$$
 denominado de parâmetro de assimetria. Caso $n_s = n_c$ (guia simétrico), então $\delta = 0$.

Desse modo,
$$\frac{w}{u} = \frac{b+5}{1-b} = \frac{w}{u} = \frac{b+5}{1-b} = \frac{w}{1-b} = \frac{\sqrt{b+5}}{\sqrt{1-b}}$$

$$W = \sqrt{b+\delta} V \Rightarrow W^2 = (b+\delta) V^2 \quad b = (V/V)^2$$

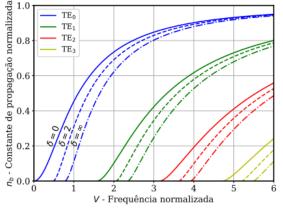
$$W^2 = (\frac{V^2}{V^2} + \delta) V^2 = V^2 + \delta V^2$$

$$W^2 - V^2 = \delta V^2, \quad 0 \leq 2 \leq 4, \quad \frac{w}{u} = \sqrt{\frac{b+\delta}{1-b}} \quad \Rightarrow \quad w^2 - v^2 = \delta V^2$$

· Como u= VI-bV a equação 2u= míi+tan-1(w)+tan-1(v)
pade sen escrita como:

$$2\sqrt{1-b}V = m\hat{i} + \lambda an^{-1} \left(\sqrt{\frac{b+\delta}{1-b}} \right) + \lambda an^{-1} \left(\sqrt{\frac{b+\delta}{1-b}} \right)$$

$$2V\sqrt{1-b} = m\pi + \arctan\left(\sqrt{\frac{b+\delta}{1-b}}\right) + \arctan\left(\sqrt{\frac{b}{1-b}}\right)$$



Curvas universais para diferentes valores do parâmetro de assimetria - modos TE

Frequência de corte

A frequência normalizada de corte V_c^m é determinada quando b=0 e, portanto, a partir da Eq. 101:

$$2V_c^m = m\pi + \arctan(\sqrt{\delta}) \to f_c^m = \frac{m\pi + \arctan(\sqrt{\delta})}{4\pi \frac{a}{c_0} \sqrt{n_f^2 - n_s^2}}$$
(102)

Observe que para $\delta=0$ (guia simétrico), as equações anteriores são reescritas como

$$V_c^m = m\frac{\pi}{2} \quad \text{ e} \quad f_c^m = \frac{m\pi}{4\pi\frac{a}{c_0}\sqrt{n_f^2-n_s^2}}$$

A quantidade de modos TE guiados pode ser obtida considerando que $V_c^m \leqslant V$ ou 6

$$m \leqslant \frac{2V - \arctan(\sqrt{\delta})}{\pi} \to M = \text{floor}\left(\frac{2V - \arctan(\sqrt{\delta})}{\pi}\right)$$
 (103)

em que M é o índice do modo de mais alta ordem e, portanto, haverá M+1 modos.

 $^{^{6}}$ floor(x) é o maior inteiro menor que x.

Para o modo TM:

$$E_z(x) = E_0 \begin{cases} \sin(k_f a + \phi) e^{-\alpha_c(x-a)}, & x \ge a \\ \sin(k_f x + \phi), & |x| \le a \\ -\sin(k_f a - \phi) e^{\alpha_s(x+a)}, & x \le -a \end{cases}$$

$$E_x(x) = -jE_0\beta \begin{cases} \alpha_c^{-1}\sin(k_f a + \phi)e^{-\alpha_c(x-a)}, & x \geqslant a \\ k_f^{-1}\cos(k_f x + \phi), & |x| \leqslant a \\ \alpha_s^{-1}\sin(k_f a - \phi)e^{\alpha_s(x+a)}, & x \leqslant -a \end{cases}$$

$$H_y(x) = -jE_0\omega \begin{cases} \frac{n_c^2}{\alpha_c} \sin(k_f a + \phi)e^{-\alpha_c(x-a)}, & x \geqslant a \\ \frac{n_f^2}{k_f} \cos(k_f x + \phi), & |x| \leqslant a \\ \frac{n_s^2}{\alpha_s} \sin(k_f a - \phi)e^{\alpha_s(x+a)}, & x \leqslant -a \end{cases}$$

· Aplicando de condições de fronteira para Hy(x):

$$\frac{nf^{2}\cos(k_{f}a+\alpha)}{nf^{2}\cos(k_{f}a+\alpha)} = \frac{nc^{2}\sin(k_{f}a+\alpha)}{nc} (x=a) = \frac{1}{2} \tan(k_{f}a+\alpha) = \frac{1}{2} \cot(k_{f}a+\alpha) = \frac{1}{2} \cot(k$$

$$2kfa = fan^{-1}\left(\frac{x_{c}p_{c}}{KF} + \frac{x_{5}p_{5}}{kF}\right)$$

$$\frac{x_{c}x_{5}}{kF^{2}}$$

$$fan(zkfa) = \frac{kf(xcbe + xsbs)}{kf^z - xexsbebs}$$

A solução de (A) são as constantes de propagação B dos modos TM.

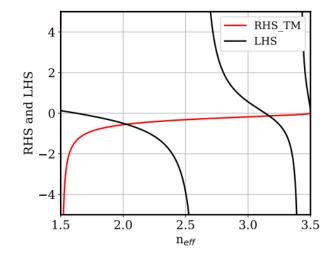
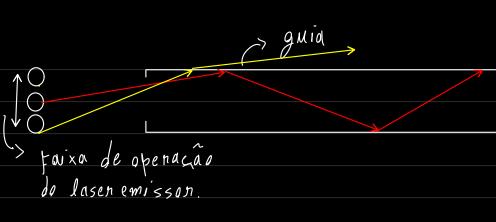
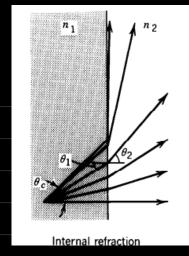


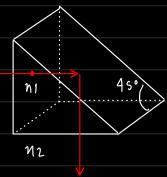
Figura 10: Curvas da equação característica, considerando $n_f=3.5,\,n_s=1.45,\,n_c=1.0,\,a=0,5\;\mu{\rm m}$ e $\lambda=1550$ nm.







chitical angle:
$$Az = 90^{\circ}$$
, $n_{1} \sin a_{1} = n_{2} \sin a_{2}$ (Lei de Snell)
$$A_{1} = \sin^{-1}\left(\frac{n_{2}}{n_{1}}\right)$$



$$Mz=1$$
 $\theta_1 = ancsin(1/\sqrt{z}) = 45^{\circ}$
 $M = \sqrt{z}$



