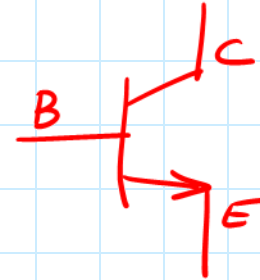
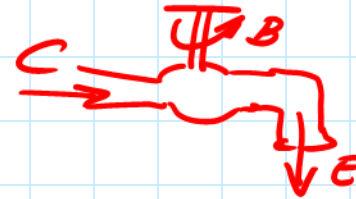


1.5 The bipolar transistor. Foundations

BJT: Bipolar Junction Transistor



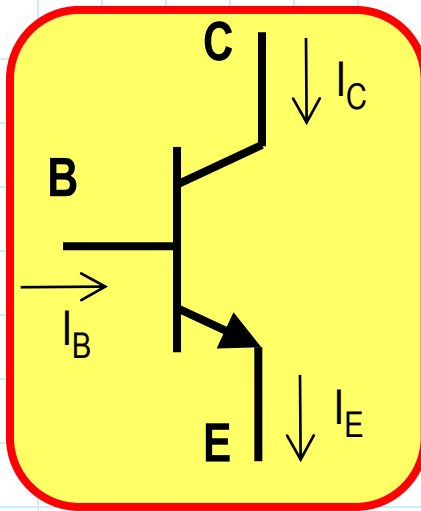
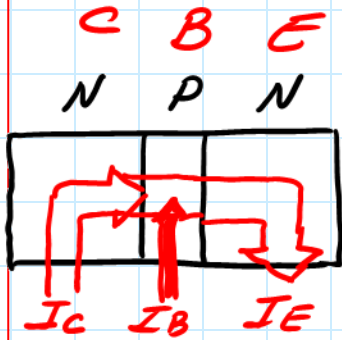
- A **Bipolar Junction Transistor** is a three-terminal device that, in most logic circuits, acts like a **current-controlled switch**. *Analogy with a tap:*



- If we put a small current into one of the terminals, called the **base**, then the switch is **ON**:
Current may flow between the other two terminals, called the **emitter** and the **collector**.
- If no current is put into the base, then the switch is **OFF**:
No current flows between the emitter and the collector.

1.5 The bipolar transistor. Foundations

N-P-N TRANSISTOR



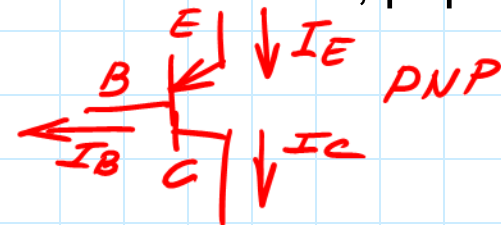
B: Base

C: Collector

E: Emitter

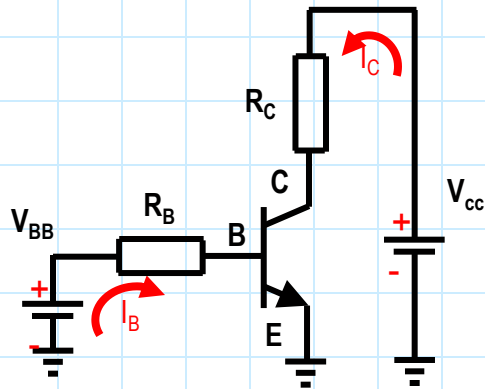
$$I_B + I_C = I_E$$

- Notice that the symbol contains an arrow in the direction of positive current flow (base-to-emitter junction, like in a diode)
- It is also possible to manufacture a **PNP** transistor. However, pnp transistors are seldom used in digital circuits.

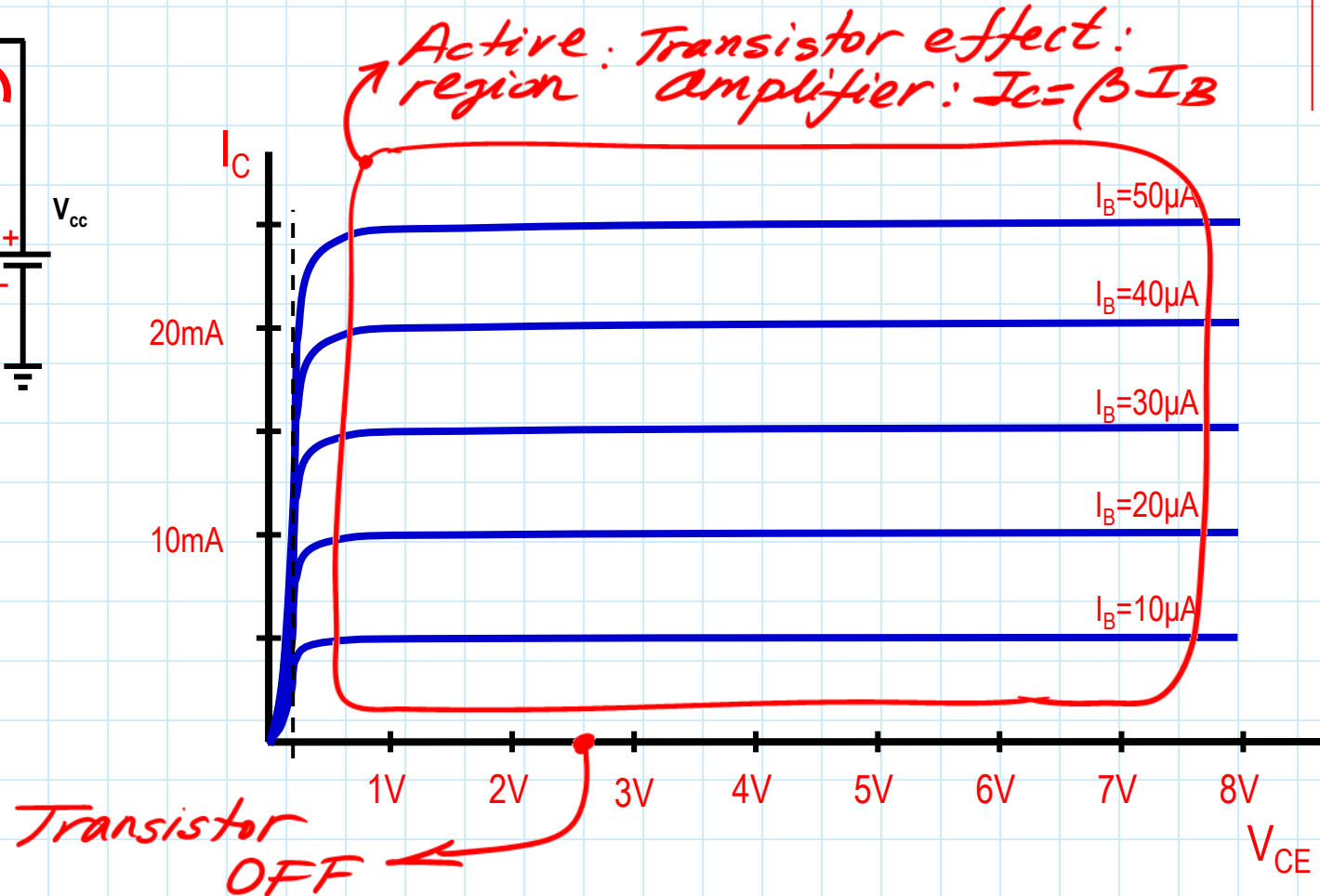


1.6 Output Characteristic Curves

Current I_C as a function of voltage V_{CE} and current I_B



Data:
 $\beta=500$
(gain of the
transistor)
 $R_B=100k\Omega$
 $R_C=0.4k\Omega$
 $V_{CC}=8V$



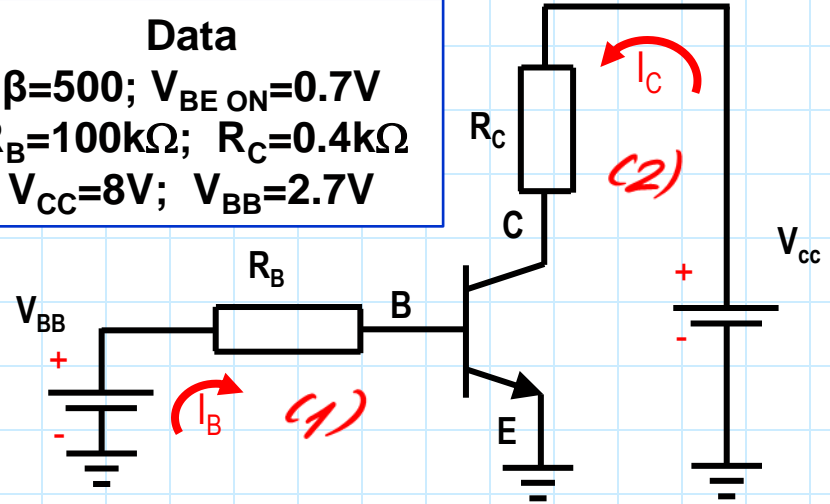
1.6 Static Load Line

(2): output loop
(1): Input loop

Static Load Line: $V_{CE} = V_{CC} - R_C \times I_C$ (1)

$$I_B = (V_{BB} - 0.7) / R_B = 20\mu A \quad (2)$$

Data
 $\beta = 500$; $V_{BE\text{ ON}} = 0.7V$
 $R_B = 100k\Omega$; $R_C = 0.4k\Omega$
 $V_{CC} = 8V$; $V_{BB} = 2.7V$

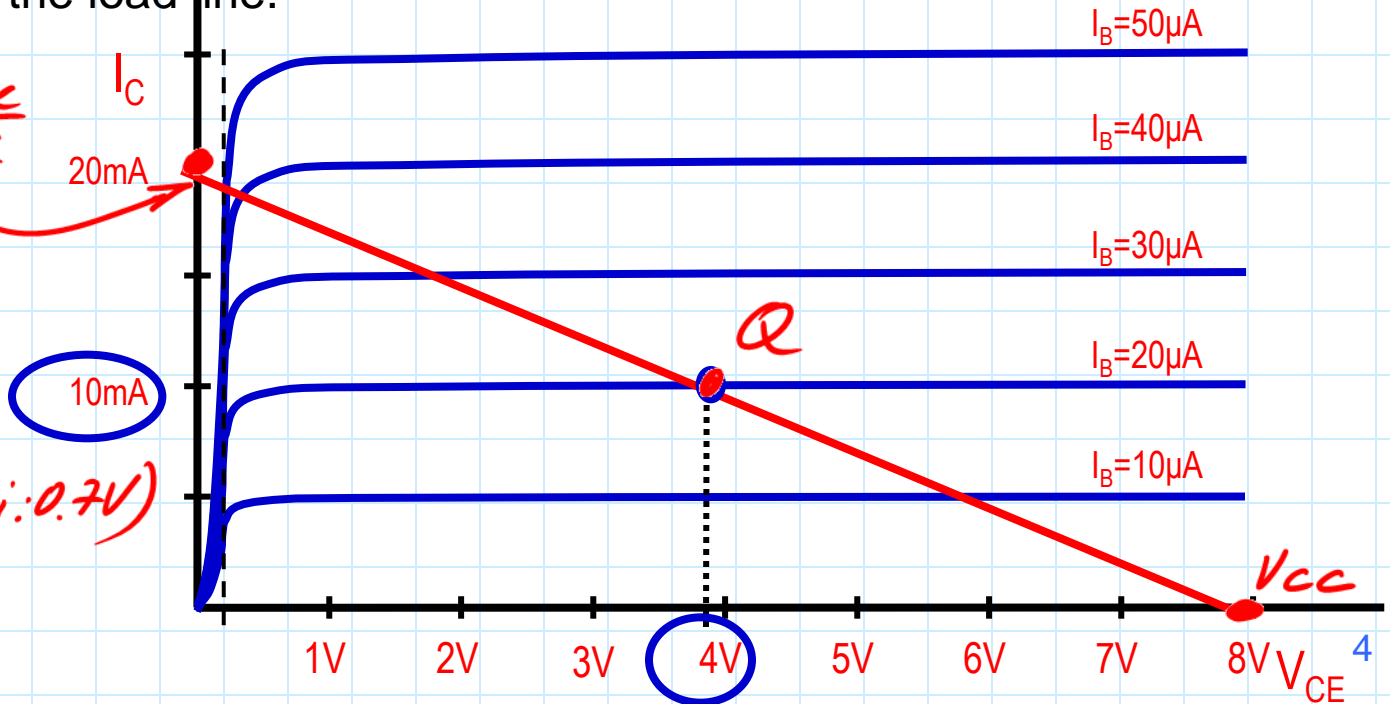


The **quiescent point** is determined by the intersection of the transistor output characteristic curve associated with $I_B = 20\mu A$ and the load line.

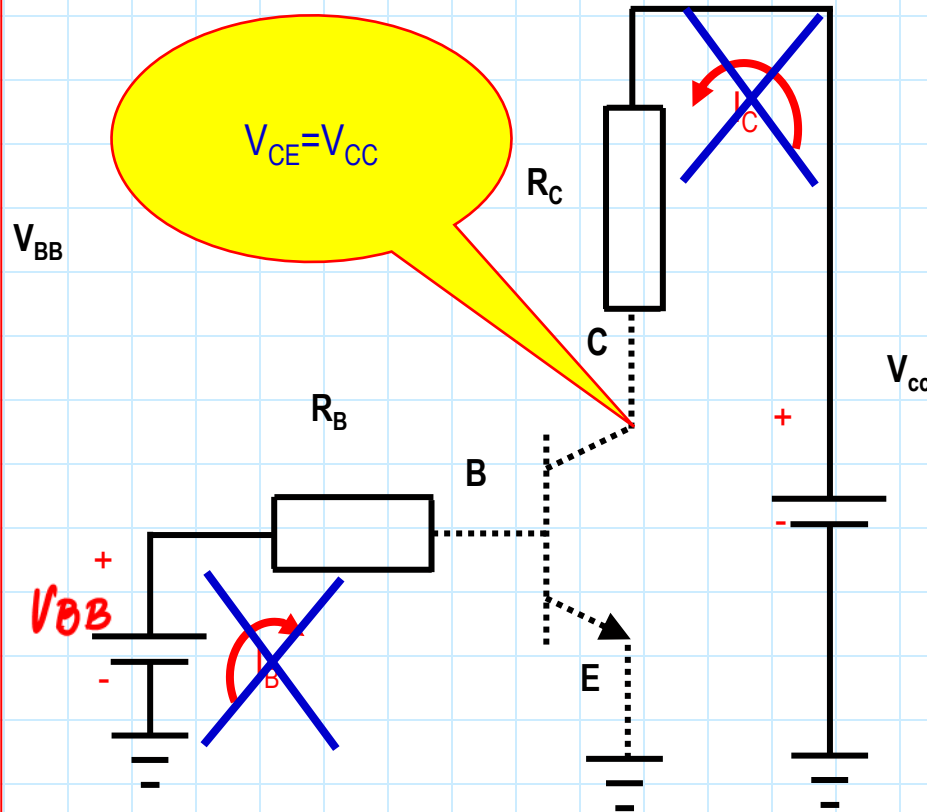
From (1):
 $V_{CE} = 0 \Rightarrow I_C = \frac{V_{CC}}{R_C}$

$I_C = 0 \Rightarrow V_{CE} = V_{CC}$

$V_{BE(ON)}$: V_{BE} for transistor ON (Si: 0.7V)



1.7 Regions of operation. Cut-off



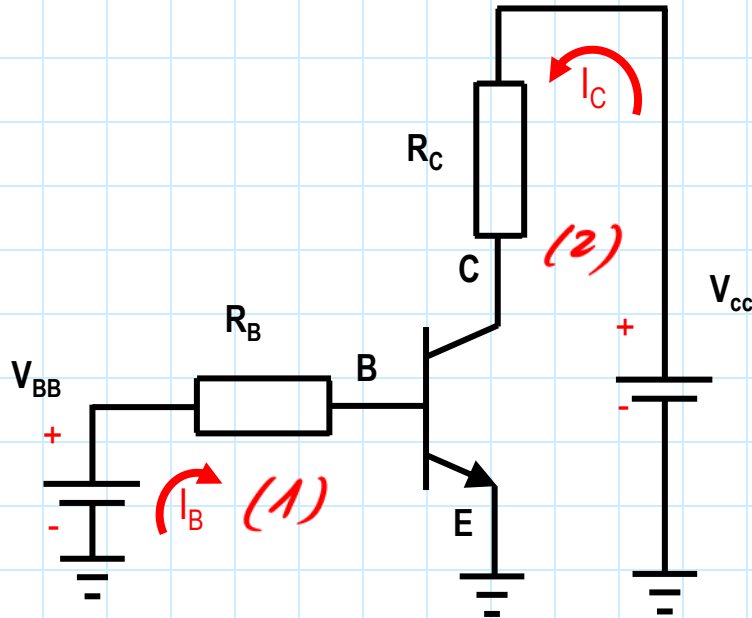
$$I_B=0, I_C=0, I_E=0$$

$$V_{BB} = V_{BE} < V_{BE(on)}$$

The switch is **OFF**

$$V_{CE} = V_{CC}$$

1.7 Operating regions. Active



$V_{BB} > V_{BE(ON)} = 0.7V$
 $\Rightarrow V_{BE} = V_{BE(ON)}$ and we have $I_B > 0$

$$I_B > 0 \rightarrow I_C = \beta I_B$$

I_C current is proportional to I_B
(linear zone)

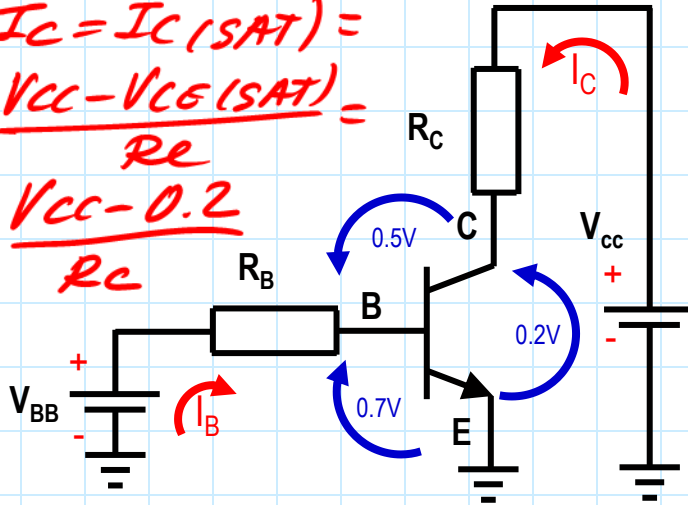
$$(1) I_B = \frac{V_{BB} - V_{BE(ON)}}{R_B} = \frac{V_{BB} - 0.7}{R_B}$$

$$(2) V_{CE} = V_{CC} - R_C I_C = V_{CC} - \beta I_B R_C$$

$I_B \uparrow \Rightarrow \begin{cases} I_C \uparrow \\ V_{CE} \downarrow \end{cases}$

1.7 Regions of operation. Saturation

$$I_C = I_C(\text{SAT}) = \frac{V_{CC} - V_{CE}(\text{SAT})}{R_C} = \frac{V_{CC} - 0.2}{R_C}$$



The switch is fully **ON**

In active if $I_B \uparrow \Rightarrow I_C \uparrow$; $V_{CE} \downarrow$, but V_{CE} has a lower limit: When $V_{CE} = V_{CE}(\text{SAT}) = 0.2V \Rightarrow$ Transist. saturated, and I_C doesn't increment more

- I_C can not increase more, the transistor is said to be **SATURATED**

$\rightarrow S_i$

$$I_C < \beta \cdot I_B ; V_{CE} \approx 0.2V ; \text{being } I_B \geq I_{B\text{minSAT}}$$

- Saturation occurs because the output circuit (V_{CC} and R_C) limits I_C to a maximum value.

BJT summary (NPN)

$$\begin{array}{l} \text{OFF} \\ V_{BE} \leq V_{BE(ON)} \end{array} \left\{ \begin{array}{l} V_{BE} \leq V_{BE(ON)} \\ I_B = 0; I_C = 0; V_{CE} = V_{CC} \end{array} \right\} \text{CUT OFF}$$

$$\begin{array}{l} \text{ON} \\ V_{BE} > V_{BE(ON)} \end{array} \left\{ \begin{array}{l} V_{BE} = V_{BE(ON)} \\ I_C = \beta I_B \\ V_{CE} = V_{CC} - R_C I_C \\ I_B \uparrow \Rightarrow I_C \uparrow; V_{CE} \downarrow \end{array} \right\} \text{ACTIVE (Linear)}$$
$$\left\{ \begin{array}{l} V_{BE} = V_{BE(ON)} \\ I_C < \beta I_B \\ V_{CE} = V_{CE(SAT)} \\ I_C = I_{C(SAT)} = (V_{CC} - V_{CE(SAT)}) / R_C \end{array} \right\} \text{SATURATION}$$

Si $V_{BE(ON)} = 0.7V; V_{CE(SAT)} = 0.2V$