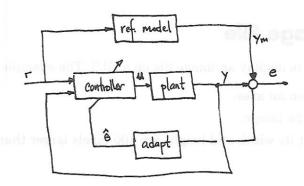
Model Reference Adaptive Control



System has four main components:

I Plant: known structure with unknown parameters

2] Reference Model: specifies ideal/desired response of the system to external commands. it is part of the adaptive control system design.

It should

- a) reflect performance specifications of the control task. (rise time, settle time, overshoot, etc.)
- b) should be achievable for the adaptive control system whits structural characteristics.

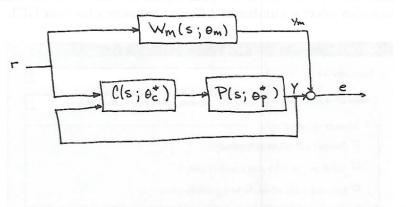
 (order, relative degree of regulated adapts. etc.)
- 3 Controller: feedback | feedforward control law with adjustable parameters should have perfect tracking ability if parameters exactly known.
- 4] Adaptation Law: adjust parameters in the control law.

GOAL: make tracking error, e, converge to zero.
must gaurantee stable controller meanwhile.

Direct & Indirect MRAC

- 1) direct adjust controller parameters during adaptation.
- 2) indirect estimate plant parameters for use by controller.

What is the main principle? ... Well, in the frequency domain lots examine the following model reference controller,



Om, Oc, Op - coefficients/parameters of the transfer function

Wm (5 ; 0m) - reference model achieving desired operating characteristics

Goal is to determine what of should be. This is equivalent to finding

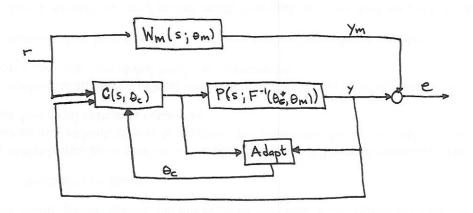
$$\theta_c^* = F(\theta_p^*, \theta_m)$$
 such that $\frac{Y(s)}{R(s)} = \frac{Y_m(s)}{R(s)}$

if
$$W_{m}(s;\theta_{m}) = \frac{N_{m}(s;\theta_{m})}{D_{m}(s;\theta_{m})}$$
 then find θ_{c}^{*} so that $\frac{Y(s)}{R(s)} = \frac{PC}{1+PC} = \frac{N(s;\theta_{p}^{*},\theta_{c}^{*})}{D(s;\theta_{p}^{*},\theta_{c}^{*})}$

satisfies: Nm(s;0m)= N(s;0p,0c*) & Dm(s;0m)= D(s;0p,0c*)

find $\theta_c = F(\theta_p, \theta_m)$ to achieve $N_m(s; \theta_m) = N(s; \theta_p^*, F(\theta_p^*, \theta_m))$ and $D_m(s; \theta_m) = D(s; \theta_p^*, F(\theta_p^*, \theta_m))$.

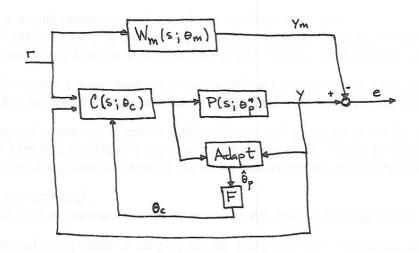
Direct Model Reference Adaptive Control:



recall that $\theta_c = F(\theta_p, \theta_m)$, so we define $\theta_p = F^{-1}(\theta_c, \theta_m)$.

goal is to estimate the controller parameters directly. ideally $\theta_c \rightarrow \theta_c^*$ as $t \rightarrow \infty$.

Indirect Model Reference Adaptive Control



come up with estimate $\hat{\Theta}_p$ of Θ_p , then choose controller parameters $\Theta_c = F(\Theta_p, \Theta_m)$. here, the goal is to have $\hat{\Theta}_p \to \Theta_p^+$ as $t \to \infty$.

Direct MRAC of First Order Systems

System | Plant:

xlo) = xo

unknown parameters, but sign(b) is known.

> essential; think of the phase of the control.

Model:

$$\dot{x}_m(t) = a_m x_n(t) + b_m r(t)$$

×mlo) = ×mio

4 L uniformly cts bdd input signal

am<0

for uniformly bounded input { r(t) E R | Ir(t) | 5 max }, define an adaptive feedback signal ult) such that the state x (t) tracks Xm(t) asymptotically with all signals bounded.

choose Controller:

u(t) = kx(t)x(t) + kr(t)r(t)

$$\dot{x}(t) = (a+bk_x(t))x(t) + bk_r(t)r(t)$$

(in order for controller to adapt to a solution, a unique solution must exist

$$a + bk_{x}^{*} = am$$

MATCHING CONDITIONS

bkr = bm

· need unique solution to the matching conditions. we may not know it, but we need to know it's possible for adaptive control problem to be well-posed.

Define the error and error dynamics
$$e(t) = x(t) - x_{ml}t)$$

$$\dot{e}(t) = \dot{x}(t) - \dot{x}_{m}(t)$$

$$= (a + bk_{x}(t))x(t) + bk_{r}(t)r(t) - a_{m}x_{m}(t) - b_{m}r(t)$$

$$= (a + bk_{x}^{*} + bk_{x}(t) - bk_{x}^{*})x(t) - a_{m}x_{m}(t)$$

$$+ (bk_{r}(t) - bk_{r}^{*} + bk_{r}^{*})r(t) - b_{m}r(t)$$

$$= a_{m}(x(t) - x_{m}(t)) + b(k_{x}(t) - k_{x}^{*})x(t) + b(k_{r}(t) - k_{r}^{*})r(t)$$

$$\dot{e}(t) = a_{m}e(t) + bAk_{x}(t)x(t) + bAk_{r}(t)r(t)$$

where
$$\Delta k_x(t) \equiv k_x(t) - k_x^{\dagger}$$

 $\Delta k_r(t) \equiv k_r(t) - k_r^{\dagger}$

Now, we are going to show asymptotic convergence with the candidate Lyapunov function,

$$V(e, \Delta k_x, \Delta k_r) = e^2(t) + |b|(Y_x^{-1} \Delta k_x^2(t) + Y_r^{-1} \Delta k_r^2(t))$$

$$\dot{V} = 2e(t)\dot{e}(t) + 2|b|\chi^{-1}_{x}\Delta k_{x}(t)\Delta k_{x}(t) + 2|b|\chi^{-1}_{r}\Delta k_{r}(t)\Delta k_{r}(t)$$

$$= 2e(t)\left[a_{m}e(t) + b_{x}\Delta k_{x}(t)x(t) + b_{x}\Delta k_{r}(t)r(t)\right]$$

$$+ 2|b|\chi^{-1}_{x}\Delta k_{x}(t)\Delta k_{x}(t) + 2|b|\chi^{-1}_{r}\Delta k_{r}(t)\Delta k_{r}(t)$$

=
$$2ame^{2}(t) + 2|b| \Delta k_{x}(t) \left[\chi_{x}^{-1} \Delta k_{x}(t) + sign(b) e(t) x(t) \right]$$

+ $2|b| \Delta k_{x}(t) \left[\chi_{x}^{-1} \Delta k_{x}(t) + sign(b) e(t) x(t) \right]$

best if these were to vanish.

unknown

$$\Delta k_{x}(t) = k_{x}(t) \equiv - sign(b) \delta_{x} e(t) \times (t)$$

$$\Delta k_r(t) = k_r(t) \equiv -sign(b) \delta_r e(t) r(t)$$

Adaptation Low.

=>

$$\dot{V} = -2|a_m|e^2(t) \leq c$$

€ 0 negative semi-definite.

D

Equilibrium is stable, the signals elt, skx lt), skx lt) are bounded

=>

X(t) is bounded since ells and xm(t) are bounded.

Can we conclude asymptotic stability? Well, ...

V = - 4 |am | elt elt)

both bounder

->

V bounded

=> Barbolot + Corollary

Tracking error is asy. Stable, e(t) -> 0 as t- 00 (parameters are only governteed bounded)

- Of note: (1) needed a Lyapunov function whose second time derivative was bounded.

 adaptive control design is an inverse Lyapunov design methodology.
 - (2) sign(b) was needed. Essential to providing control in the proper direction.

Indirect MRAC of First Order Systoms

· Idea is to estimate the plant purameters and not the controller parameters. Setup is going to differ.

Plant/System:
$$\dot{x}(t) = ax(t) + bu(t)$$

whenous parameters,
but sign(b) is known,
and a conservative lower bound for 161 is known.

* without loss of generality, assume $b > \overline{b} > 0$.

Model:
$$\dot{x}_m(t) = a_m x_n(t) + b_m r(t)$$
 $\dot{x}_m(0) = \dot{x}_{m,0}$

We are going to try to derive the ad controller by massaging alt) & both into x(t) & then find ball/able) in x(t).

$$\dot{x}(t) = a \times (t) + b \cdot u(t)$$

$$= a \times (t) + \hat{b}(t) \cdot u(t) + (b - \hat{b}(t)) \cdot u(t)$$

$$= a \times (t) + \hat{b}(t) \cdot u(t) - \Delta b(t) \cdot u(t)$$

$$= a \times (t) + (a_m \times (t) + b_m \cdot v(t) - \hat{a}(t) \times (t)) - \Delta b(t) \cdot u(t)$$

$$= a \times (t) + (a_m \times (t) + b_m \cdot v(t) - \hat{a}(t) \times (t)) - \Delta b(t) \cdot u(t)$$

$$= a \times (t) + (a_m \times (t) + b_m \cdot v(t) - \hat{a}(t) \times (t))$$

$$= a \times (t) + (a_m \times (t) + b_m \cdot v(t) - \hat{a}(t) \times (t)$$

$$= a \times (t) + b \cdot u(t) + b \cdot u(t)$$

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$$= a \times (t$$

our controller should be 3. get dynamics. Sa = sb = 0
gives perfect tracking.

Controller:
$$u(t) = \frac{1}{b(t)} \left[a_m \times (t) + b_m r(t) - \hat{a} \times (t) \right]$$

- what's going on with b(t)?

for Ybult) elt) > 0 there are no problems since this increases both (away from zero), or does nothing.

for Youlthelth ≤0, get decrease of b(t).

Shouldn't go too far below lower bound.

The term b-b(t) starts to grow as by goes

below to and drives it back up, or at least takes it to the term solution to b(t) = 0.

It cannot drive bolt) below (b-E) since that is where the denominator blows up.

bitt) is bounded away from zero bitt) > b- € > 0

feedback is OK and all affected differential equations are Lipschitz cts => uniqueness is garranteed.

also means that $\hat{S}(t)$ is ets on domain of solution S(t).

may continue with next step: Lyopunov analysis!

OK, moving on let's define the error and error dynamics

Inspired by direct MRAC, consider

$$\Delta a = \hat{a} = V_a \times (t) e(t)$$

We can't quite do the same because of the constraint that bott=0 can never happen. We need to keep adaptation dynamics away from 3ero. The following signal should do the trick:

$$\Delta \dot{b}(t) = \hat{b}(t) = \begin{cases} \chi_{b} u(t) e(t) & \text{if } \hat{b}(t) > \bar{b} \\ \chi_{b} u(t) e(t) + \frac{\bar{b} - \hat{b}(t)}{\hat{b}(t) - \bar{b} + \epsilon} & \text{if } \hat{b}(t) < \bar{b} \end{cases}$$

Lo Y_a, Y_b adaptetion gains $\epsilon > 0$ tolerance. Should be small enough so that $\overline{b} - \epsilon > 0$ Adaptation

Define
$$V(cH), a_0H, a_0H) = e^2(t) + Y_a^{-1} a_0^a(t) + Y_b^{-1} a_0^b(t)$$

$$V = 2e(t) e(t) + 2Y_a^{-1} a_0H) a_0(t) + 2Y_0^{-1} b_0H a_0(t)$$

$$= 2e(t) [a_me(t) - a_0H) x(t) - a_0H a_0(t)] + 2X_0^{-1} a_0H x_0^a x_0^a(t) + 2Y_0^{-1} b_0H a_0(t)$$

$$= 2a_me^2(t) + 2a_0(t) (e(t) x(t) - eH x(t)) + 2a_0(t) (Y_0^{-1} a_0(t) - u(t) e(t))$$

$$V = 2a_me^a(t) + 2a_0(t) (Y_0^{-1} a_0(t) - u(t) e(t))$$

$$V = \begin{cases} 2a_me^a(t) + 2a_0(t) \frac{1}{b(t)} \frac{1}{b(t)} - \frac{1}{b(t)} \frac{1}{b(t)$$

- choice of E kind of important.

More First Order, Scalar Direct MRAC:

$$\dot{x} = ax + b(u + f(x))$$

sign(b) is known.

suppose that fix) can be linearly parametrized in terms of N bounded basis functions

$$f(x) = \sum_{i=1}^{N} \alpha_{i} \varphi_{i}(x) = \alpha \cdot \overline{\Phi}(x) = \alpha^{T} \overline{\Phi}(x)$$

regressor vector

Reference Model:

Controller:

substitution leads to the same matching conditions as before:

$$a+bk_{x}^{*}=a_{m}$$

$$bk_{x}^{*}=b_{m}$$

* matching conditions not needed for α since complete system of equations exist for the α_i .

$$\dot{V} = 2e(t) \left[a_m e(t) + b \left[\Delta k_x(t) \times (t) + \Delta k_r(t) - (t) - \Delta \alpha^T(t) \oplus (x(t)) \right] \right]$$

$$+ 2|b| \left[Y_x^T \Delta k_x(t) \Delta k_x(t) + Y_r^T \Delta k_r(t) \Delta k_r(t) + \Delta \alpha^T(t) T_\alpha^{-1} \Delta \alpha^T(t) \right]$$

=
$$2ame^{z}(t) + 2|b|\Delta k_{x}(t) \left[e(t)x(t)sign(b) + Y_{x}^{-1}\Delta k_{x}(t)\right]$$

+ $2|b|\Delta k_{r}(t) \left[e(t)r(t)sign(b) + Y_{r}^{-1}\Delta k_{x}(t)\right]$
+ $2|b|\Delta x^{T}(t) \left[-e(t)sign(b) \Phi(x(t)) + \Gamma_{x}^{-1}\Delta x(t)\right]$

want these to vanish.

1

$$\Delta \dot{k}_{x}(t) = \dot{k}_{x}(t) \equiv -\text{sign(b)} \, \delta_{x} \, e(t) \, x(t)$$

$$\Delta \dot{k}_{r}(t) = \dot{k}_{r}(t) \equiv -\text{sign(b)} \, \delta_{r} \, e(t) \, r(t)$$

$$\Delta \dot{k}_{r}(t) = \hat{\alpha}(t) \equiv -\text{sign(b)} \, \delta_{r} \, e(t) \, r(t)$$

$$\Delta \dot{k}_{r}(t) = \hat{\alpha}(t) \equiv -\text{sign(b)} \, e(t) \, r(t)$$

\$

$$\dot{V} = -2|a_m|e^2(t) \le 0$$
 negative seni-definite.

and following the same logic as before, ett) -0 as t->00.

* radial unboundedness of $V \Rightarrow \text{globally asymptotically stable to } E$. $E = \left\{ (e, \Delta k_x, \Delta k_r, \Delta \alpha) \middle| \dot{V} = 0 \right\}$ $E = \left\{ (e, \Delta k_x, \Delta k_r, \Delta \alpha) \middle| e = 0 \right\}$

The same can be done w/indirect MRAC:

Plant | System:

$$\dot{x}(t) = ax(t) + b(u(t) + f(x(t)))$$

X(0) = X0

Reference Model:

Xm(0)= Xm10

Controller:

$$u(t) = \frac{1}{\hat{h}(t)} \left[a_m x(t) + b_m r(t) - \hat{a}(t) x(t) - \hat{\alpha}^T(t) \Phi(x(t)) \right]$$

Adaptation:

$$\hat{b}(t) = \begin{cases} x_a & \text{ett.} & \text{if } \hat{b}(t) > \bar{b} \\ x_a & \text{ett.} & \text{if } \hat{b}(t) < \bar{b} \end{cases}$$

$$\begin{cases} x_a & \text{ett.} & \text{if } \hat{b}(t) < \bar{b} \end{cases}$$

$$\hat{\alpha}(t) = \Gamma_{\alpha} \Phi(x(t)) e(t)$$

* other update methodo can be used for b(t) so long as there is an appropriate Lyapunov function of one can governtee no 3lro crossings.