### Z-transform

$$\begin{split} &\mathcal{Z}\left\{f_1(t) \pm f_2(t)\right\} = F_1(z) + F_2(z) & \text{Addition} \\ &\mathcal{Z}\left\{af(t)\right\} = aF(z) & \text{Multiplication by a Constant} \\ &\mathcal{Z}\left\{f(t-nT)\right\} = z^{-n}F(z) & \text{Shifting} \\ &\mathcal{Z}\left\{f(t+kT)\right\} = z^kF(z) - z^kf(0) - \cdots - zf(kT-T) & \text{Shifting (cont'd)} \\ &\mathcal{Z}\left\{e^{\mp at}f(t)\right\} = F(ze^{\pm at}) & \text{Complex Translation} \\ &\lim_{k \to \infty} f(kT) = \lim_{z \to 0} F(z) & \text{Initial Value Theorem} \\ \hline &\text{If } (1-z^{-1})F(z) \text{ has all singularities inside unit disk} & \text{Final Value Theorem} \end{split}$$

|z|=1, then

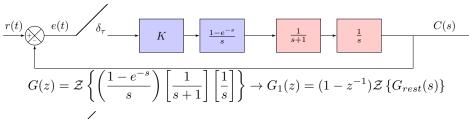
$$|z| = 1$$
, then  $\lim_{k \to \infty} f(kT) = \lim_{z \to 1} (1 - z^{-1}) F(z)$ 

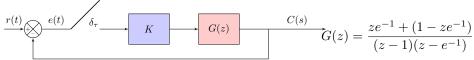
$$\mathcal{Z}\left\{\frac{\partial}{\partial a}f(t,a)\right\} = \frac{\partial}{\partial a}F(z,a)$$

Partial differentiation

## Sampling Theory

$$\mathcal{Z}\{G_1(s)G_2(s)\} = G_1G_2(z) = G_2G_1(z)$$
 In General  $G_1(z)G_2(z) \neq G_1G_2(z)$ 





ZOH(zero-hold-system) 
$$f^*(t) = \sum_{k=-\infty}^{\infty} f(kT)\delta(t-kT)$$
  $G_h(s) = \frac{1-e^{-Ts}}{s}$ 

Stability Test for Digital Systems 
$$P(z) = a_0 z^n + a_1 z^{n-1} + \dots + a_{n-1} z + a_n \quad G(z) = \frac{A(z)}{P(z)}$$

Stability Condition:  $P(z) \neq 0 \quad |z| \geq 1$  (Draw Unit Circle to test stability)

Routh-Stability in Digital Domain:  $s = \frac{z+1}{z-1}$   $z = \frac{s+1}{s-1}$ 

# Jury-Marden Table Uses function P of z

Necessary and Sufficient Condition for Stability

1. 
$$|a_n| < |a_0|$$
 Special Case  $n = 2$   $P(z) = a_0 z^2 + a_1 z + a_2$  2.  $P(1) > 0$   $z^0 z^1 z^2$ 

3. 
$$a_2 \quad a_1 \quad a_0$$
$$P(z) \neq 0 \text{ for } |z| \geq 1 \text{ if and only if}$$

$$P(-1) > 0$$
 for n even 1.  $|a_2| < |a_0|$   
< 0 for n odd 2.  $P(1) > 0$ 

4. 
$$b_{n-1} > |b_0|, |c_{n-2}| > |c_0|, \cdots |q_2| > |q_0|$$
 3.  $P(-1) > 0$   $(n=2)$ 

Root Locus presents the poles of the closed loop system when the gain K changes from zero to infinity.

## Construction of the Root Locus

Open loop transfer function KH (s)  $G(s) = K \frac{B(s)}{A(s)}$ 

m: the order of the **open-loop** numerator polynomial

n: the order of the **open-loop** denominator polynomial. q = n - m

Rule 1: number of branches equals the number of poles of the open-loop transfer function

Rule 2: If the total number of poles and zeros of the open-loop system to the right of the s-point on the real axis is odd, then this point lies on the locus.

Rule 3: The locus starting point (K=0) are at the open-loop poles and the locus ending points  $(K=\infty)$  are at the open loop zeros and n-m branches terminate at infinity.

Rule 4 and 5: Slope of asymptotes of root locus as 's' approaches infinity. Abscissa of the intersection between asymptotes of root locus and real-axis.

$$\sigma = \frac{\sum_{i=1}^{n} p_i - \sum_{i=1}^{m} z_i}{q}$$
  $\theta = \pm r \frac{180}{q}$  where r=1, 3, 5

Rule 6: Break-away and break-in points. From the characteristic equation

$$f(s) = A(s) + KB(s) = 0$$
 and  $K = -\frac{A(s)}{B(s)}$ 

The break-away and break-in points can be found from

$$\frac{dK}{ds} = -\frac{A'(s) B(s) - A(s) B'(s)}{B^{2}(s)} = 0$$

Rule 7: Angle of departure from complex poles or zeros. Subtract from 180° the sum of all angles from all other zeros and poles of the open-loop system to the complex pole (or zero) with appropriate signs.

Z-transform: Definition 
$$F(z) = Z[f(t)] - Z[f(kT)] = \sum_{k=0}^{\infty} f(kT)z^{-k}$$

$$e^*(\infty) = \lim_{z \to 1} (1 - z^{-1}) E(z)$$

$$K_p = \lim_{z \to 1} G(z), \quad e^*(\infty) = \frac{1}{1 + K_p}$$

$$e^*(\infty) = \frac{1}{K_v}, \quad K_v = \frac{1}{T} \lim_{z \to 1} (z - 1) G(z)$$

$$e^*(\infty) = \frac{1}{K_s}, \quad K_a = \frac{1}{T^2} \lim_{z \to 1} (z - 1)^2 G(z)$$

**Linear Factor Rule.** For each factor of O of the form  $(ax+b)^m$ , the partial fraction decomposition contains the following sum of m partial fractions:

$$\frac{A_1}{ax+b} + \frac{A_2}{(ax+b)^2} + \dots + \frac{A_m}{(ax+b)^m},$$

where the  $A_i$  are constants to be determined.

Quadratic Factor Rule. For each factor of Q of the form  $(ax^2 + bx + c)^m$ , where  $ax^2 + bx + c$  is an irreducible quadratic, the partial fraction decomposition contains the following sum of m partial fractions:

$$\frac{A_1x + B_1}{ax^2 + bx + c} + \frac{A_2x + B_2}{(ax^2 + bx + c)^2} + \dots + \frac{A_mx + B_m}{(ax^2 + bx + c)^m},$$

where the  $A_i$  and  $B_i$  are constants to be determined.

$$\begin{split} x(k+2) - \frac{3}{2}x(k+1) + \frac{1}{2}x(k) &= u(k), (x(0) = 1, x(1) = \\ [z^2X(z) - z^2x(0) - zx(1)] - \frac{3}{2}(zX(z) - zx(0)] + \frac{1}{2}X(z) \\ [z^2 - 1.5z + 0.5z]X(z) &= \frac{z}{z-1} + z^2 + (2.5 - 1.5)z \\ X(z) &= \frac{z[1 + (z+1)(z-1]}{(z-1)(z-0.5)} = \frac{z^3}{(z-1)^2(z-0.5)} \\ \frac{X(z)}{z} &= \frac{z^2}{(z-1)^2(z-0.5)} = \frac{A_{11}}{(z-1)^2} + \frac{A_{12}}{z-1} + \frac{A_{13}}{z-0.5} \end{split}$$

## **Example Partial Fractions**

$$\frac{20}{(s+3)(s^2+6s+25)} \to \frac{5}{4(s+3)} - \frac{\frac{5s}{4} + \frac{15}{4}}{s^2+6s+25}$$

$$\frac{5z}{4(z-e^{-3})} + \frac{5ze^3(\cos(4) - ze^3)}{4(e^6z^2 - 2\cos(4)e^3z + 1)} \text{ Z-table} = 17$$

$$u(kT)$$

$$u(KT)$$

$$kT$$

$$(kT)^{n}e^{-akT}$$

$$\sin \omega kT$$

$$\cos \omega kT$$

$$\cos \omega kT$$

$$e^{-akT}\sin \omega kT$$

$$e^{-akT}\cos \omega kT$$

$$\frac{z}{z-1}$$

$$\frac{Tz}{(z-1)^2}$$

$$\lim_{a\to 0} (-1)^n \frac{d^n}{da^n} \left[ \frac{z}{z-e^{-aT}} \right]$$

$$\frac{z}{z-e^{-aT}}$$

$$(-1)^n \frac{d^n}{da^n} \left[ \frac{z}{z-e^{-aT}} \right]$$

$$\frac{z}{z^2-2z\cos \omega T+1}$$

$$\frac{z(z-\cos \omega T)}{z^2-2z\cos \omega T+1}$$

$$\frac{z(z-2ze^{-aT}\cos \omega T+e^{-2aT})}{z^2-2ze^{-aT}\cos \omega T+e^{-2aT}}$$

F(s)

$$\frac{1}{s}$$
 $\frac{1}{s}$ 
 $\frac{1}{s^{2}}$ 
 $\frac{1}{s^{2}}$ 
 $\frac{1}{s^{4}+1}$ 
 $\frac{1}{s^{4}+1}$ 
 $\frac{n!}{s^{4}+1}$ 
 $\frac{n!}{s^{2}+2}$ 
 $\frac{n}{s^{2}+2}$ 
 $\frac{n}$ 

$$u(t)$$

$$u(t)$$

$$t^{n}$$

$$e^{-at}$$

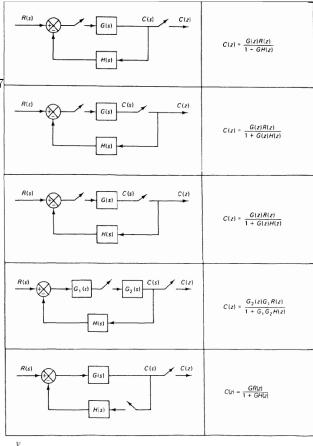
$$\sin \omega t$$

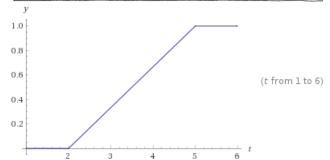
$$\cos \omega t$$

$$e^{-at} \sin \omega t$$

$$e^{-at} \cos \omega t$$

Ch.eqn = 
$$\Delta P(z) = z^2 + (K - 4)z + 0.8 = 0$$
  
Inputting  $z = 1$  and  $z = -1$ ,  $K = -0.8 - 1 + 4 = 2.2$ ,  $K = (1)^2 + 4 + 0.8 = 5.8$ , for stability  $-1 < K < 1$ .





$$\frac{X(z)}{z} = \frac{z^2}{(z-1)^2(z-0.5)} = \frac{A_{11}}{(z-1)^2} + \frac{A_{12}}{z-1} + \frac{A_{13}}{z-0.5}$$

$$\text{Ch.eqn} = \Delta P(z) = z^2 + (K-4)z + 0.8 = 0$$

$$\text{Ch.eqn} = \Delta P(z) = z^2 + (K-4)z + 0.8 = 0$$

$$\text{Geometric Sum } \sum_{k=-N}^{N} ar^{k-1} = a\frac{1-r^N}{1-r} \sum_{i=0}^{\infty} a^i = \frac{1}{1-a} \quad \text{Inputting } z = 1 \text{ and } z = -1, K = -0.8 - 1 + 4 = 2.2, K =$$

TABLE 2-1 TABLE OF z TRANSFORMS

X(z)		Z -k	$\frac{1}{1-z^{-1}}$	$\frac{1}{1-e^{-\epsilon T}z^{-1}}$	$\frac{Tz^{-1}}{(1-z^{-1})^2}$	$\frac{T^2z^{-1}(1+z^{-1})}{(1-z^{-1})^3}$	$\frac{T^3z^{-1}(1+4z^{-1}+z^{-2})}{(1-z^{-1})^4}$	$\frac{(1-e^{-a}r)z^{-1}}{(1-z^{-1})(1-e^{-aT}z^{-1})}$	$\frac{(e^{-aT} - e^{-bT})z^{-1}}{(1 - e^{-aT}z^{-1})(1 - e^{-bT}z^{-1})}$	$\frac{Te^{-aT}z^{-1}}{(1-e^{-aT}z^{-1})^2}$	$\frac{1 - (1 + aT)e^{-aT}z^{-1}}{(1 - e^{-aT}z^{-1})^2}$	$\frac{T^2 e^{-aT} (1 + e^{-aT} z^{-1}) z^{-1}}{(1 - e^{-aT} z^{-1})^3}$	$\frac{[(aT - 1 + e^{-aT}) + (1 - e^{-aT} - aTe^{-aT})z^{-1}]z^{-1}}{(1 - z^{-1})^2(1 - e^{-aT}z^{-1})}$	$\frac{z^{-1}\sin\omega T}{1-2z^{-1}\cos\omega T+z^{-2}}$	$\frac{1 - z^{-1} \cos \omega T}{1 - 2z^{-1} \cos \omega T + z^{-2}}$	$\frac{e^{-aT}z^{-1}\sin\omega T}{1-2e^{-aT}z^{-1}\cos\omega T+e^{-2aT}z^{-2}}$	$\frac{1 - e^{-aT}z^{-1}\cos\omega T}{1 - 2e^{-aT}z^{-1}\cos\omega T + e^{-2aT}z^{-2}}$	$\frac{1}{1-az^{-1}} \qquad \frac{\varepsilon}{\varepsilon}$	$\frac{z^{-1}}{1-az^{-1}}$	$\frac{z^{-1}}{(1-az^{-1})^2}$	$\frac{z^{-1}(1+az^{-1})}{(1-az^{-1})^3}$	$\frac{z^{-1}(1+4az^{-1}+a^2z^{-2})}{(1-az^{-1})^4}$	$\frac{z^{-1}(1+11az^{-1}+11a^2z^{-2}+a^3z^{-3})}{(1-az^{-1})^5}$	$\frac{1}{1+az^{-1}}$ $\frac{2}{2+\delta}$	$\frac{z^{-2}}{(1-z^{-1})^3}$
x(kT) or $x(k)$	Kronecker delta $\delta_0(k)$ 1, $k = 0$ 0, $k \neq 0$	$\delta_0(n-k)$ $1,  n=k$ $0,  n \neq k$	1(k)	e-akT	kT	$(kT)^2$	$(kT)^3$	$1 - e^{-akT}$	$e^{-akT} - e^{-bkT}$	kTe-akT	$(1 - akT)e^{-akT}$	$(kT)^2 e^{-akT}$	$akT - 1 + e^{-akT}$	$\sin \omega kT$	cos wkT	$e^{-akT}\sin\omega kT$	$e^{-akT}\cos \omega kT$	4 (19) 21(2) 32 A	$a^{k-1}$ $k = 1, 2, 3, \dots$	$ka^{k-1}$	$k^2a^{k-1}$	k3 a <sup>k-1</sup>	K4 a*-1	$a^k \cos k\pi$	$\frac{k(k-1)}{2!}$
x(t)		7 - 4 - 4 - 4 - 4 - 4 - 4 - 4 - 4 - 4 -	1(t)	e-a	1 3 to 2 +	f 7		$1 - e^{-at}$	$e^{-at} - e^{-bt}$	te-at	$(1-at)e^{-at}$	t <sup>2</sup> e <sup>-at</sup>	$at-1+e^{-at}$	sin wt	cos wt	$e^{-at} \sin \omega t$	e <sup>-at</sup> cos wt								
X(s)	7	179.14 1.17 (1.22)	$\frac{1}{s}$	$\frac{1}{s+a}$	$\frac{1}{s^2}$	$\frac{2}{s^3}$	S-4 0	$\frac{a}{s(s+a)}$	$\frac{b-a}{(s+a)(s+b)}$	$\frac{1}{(s+a)^2}$	$\frac{s}{(s+a)^2}$	$\frac{2}{(s+a)^3}$	$\frac{a^2}{s^2(s+a)}$	$\frac{\omega}{s^2 + \omega^2}$	$\frac{s}{s^2 + \omega^2}$	$\frac{\omega}{(s+a)^2+\omega^2}$	$\frac{s+a}{(s+a)^2+\omega^2}$				the section of desired side layer from the control of				Address of the Control of the Contro
	<b>:</b>	2.	3.	4.	5.	9.	7.	×.	9.	10.	11.	12.	13.	14.	15.	16.	17.	18.	19.	20.	21.	22.	23.	24.	25.