

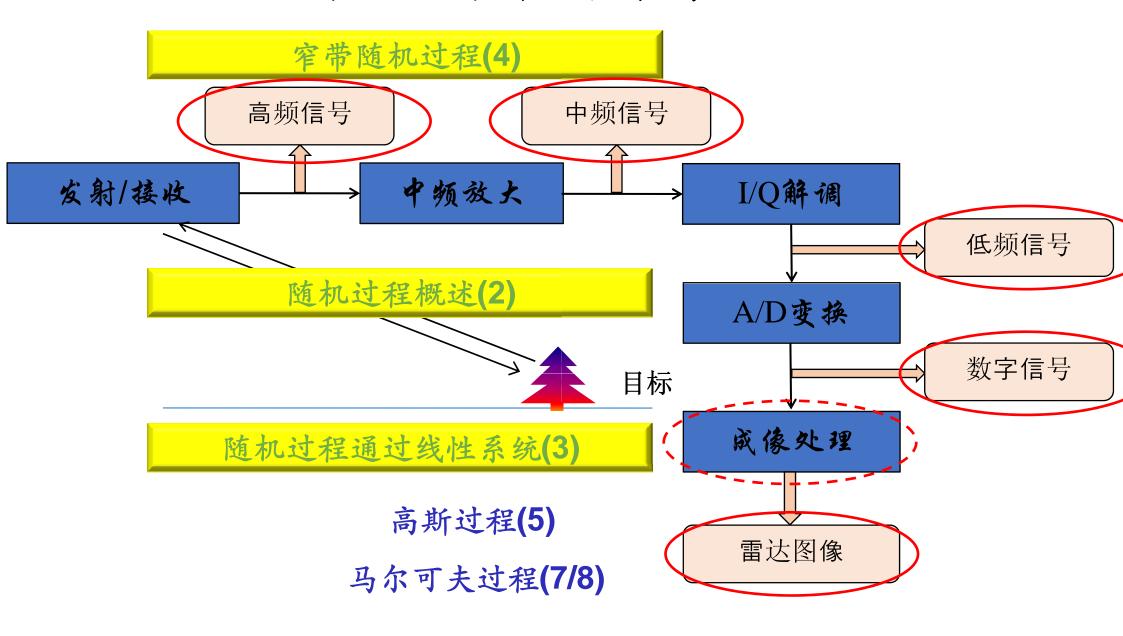


# 随机过程理论

Stochastic process theory

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### 微波成像雷达获取与处理





4.4 复随机过程

4.2 确定性信号的复表示

4.5 窄带实平稳随机过程的数字特征

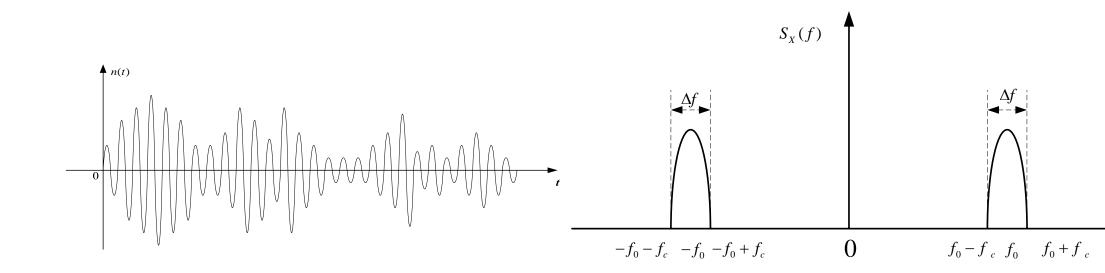
4.3 希尔伯特变换





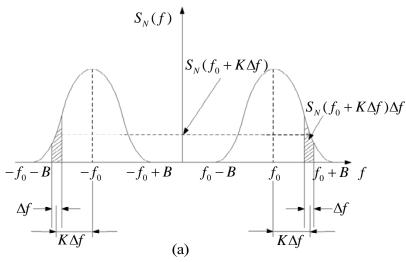
#### 1、窄带过程的描述

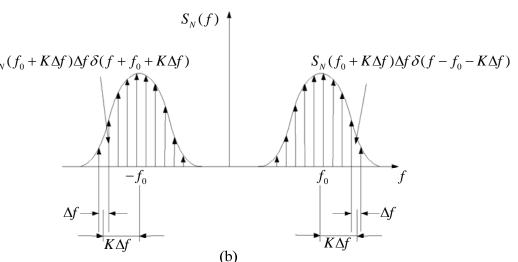
$$B \ll f_0(10 倍以上)$$



#### 1、窄带过程的描述

$$S_{N}(f) = \lim_{\Delta f \to 0} \sum_{k} S_{N}(f_{0} + k \cdot \Delta f) \cdot \left[ \delta(f + f_{0} + k \cdot \Delta f) + \delta(f - f_{0} - k \cdot \Delta f) \right] \cdot \Delta f$$





$$\begin{cases} X(t) = A\cos(2\pi f_0 t + \Theta) \\ S_X(f) = \frac{A^2}{2 \times 2} [\delta(f - f_0) + \delta(f + f_0)] \end{cases}$$
$$A_k = \sqrt{4S_N(f_0 + k \cdot \Delta f)\Delta f}$$

$$n(t) = \lim_{\Delta f \to 0} \sum_{k} A_k \cos(2\pi (f_0 + k\Delta f)t + \Theta_k)$$

$$= \lim_{\Delta f \to 0} \sum_{k} A_k \cos(2\pi k \Delta f t + \Theta_k) \cos(2\pi f_0 t)$$

$$-\lim_{\Delta f \to 0} \sum_{k} A_{k} \sin(2\pi k \Delta f t + \Theta_{k}) \sin(2\pi f_{0} t)$$

#### 1、窄带过程的描述

$$S_{N}(\omega) = \lim_{\Delta\omega \to 0} \sum_{k} S_{N}(\omega_{0} + k \cdot \Delta\omega) \cdot \left[ \delta(\omega + \omega_{0} + k \cdot \Delta\omega) + \delta(\omega - \omega_{0} - k \cdot \Delta\omega) \right] \cdot \Delta\omega$$

$$X(t) = A \cos[\omega_{0}t + \Theta] \to R_{X}(\tau) = \frac{1}{2} A^{2} \cos\omega_{0}\tau$$

$$S_{X}(\omega) = A^{2}\pi \left[ \delta(\omega - \omega_{0}) + \delta(\omega + \omega_{0}) \right] / 2$$

因此, 窄带噪声表示成:

$$n(t) = \lim_{\Delta\omega \to 0} \sum_{k} A_{k} \cos[(\omega_{0} + k\Delta\omega)t + \Theta_{k}]$$

其中 
$$A_k = \left[\frac{2S_N(\omega_0 + k\Delta\omega)\Delta\omega}{\pi}\right]^{1/2}$$

#### 1、窄带过程的描述

$$n(t) = \lim_{\Delta \omega \to 0} \sum_{k} A_{k} \cos[(\omega_{0} + k\Delta\omega)t + \Theta_{k}]$$

展开为

$$n(t) = \lim_{\Delta \omega \to 0} \sum_{k} A_{k} [\cos(k\Delta\omega t + \Theta_{k})\cos\omega_{0}t - \sin(k\Delta\omega t + \Theta_{k})\sin\omega_{0}t]$$
$$= n_{c}(t)\cos\omega_{0}t - n_{s}(t)\sin\omega_{0}t$$

这里

#### 1、窄带过程的描述

$$n_c(t) = \lim_{\Delta\omega \to 0} \sum_{k} A_k \cos(k\Delta\omega t + \Theta_k)$$

$$n_s(t) = \lim_{\Delta\omega \to 0} \sum_k A_k \sin(k\Delta\omega t + \Theta_k)$$

#### 显然有:

$$\sigma_n^2 = \frac{1}{2\pi} \int_{-\infty}^{+\infty} S_N(\omega) d\omega = R_n(0) = \lim_{\Delta \omega \to 0} \sum_{k} \frac{A_k^2}{2}$$

$$\sigma_{n_c}^2 = \sigma_{n_s}^2 = \lim_{\Delta\omega\to 0} \sum_{k} \frac{A_k^2}{2} = \lim_{\Delta\omega\to 0} \sum_{k} \frac{S(\omega_0 + k \cdot \Delta\omega)}{\pi} \cdot \Delta\omega$$
$$= \frac{1}{2\pi} \int_{-\infty}^{+\infty} S_N(\omega) d\omega = R_n(0)$$

2、两正交分量  $n_c(t)$ 、 $n_s(t)$  的性质

✓ 具有低频特性

$$\checkmark E[n_c(t)] = E[n_s(t)] = 0$$

$$\checkmark \quad \sigma_{n_c}^2 = \sigma_{n_s}^2 = \sigma_n^2$$

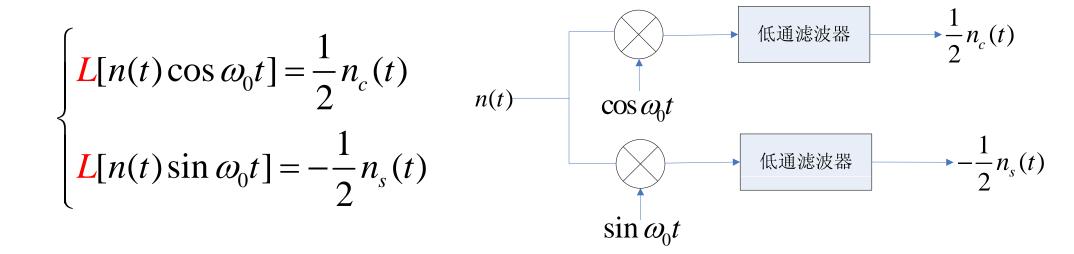
$$\checkmark S_{n_c}(\omega) = S_{n_s}(\omega) = \begin{cases} S(\omega + \omega_0) + S(\omega - \omega_0) & |\omega| < B \\ 0 & \sharp \dot{\Xi} \end{cases}$$

$$\checkmark E[n_c(t)n_s(t)] = 0$$

2、两正交分量  $n_c(t)$ 、 $n_s(t)$  的性质

2、两正交分量  $n_c(t)$ 、 $n_s(t)$  的性质  $n(t) = n_c(t)\cos\omega_0 t - n_s(t)\sin\omega_0 t$ 

同相分量  $n_c(t)$  和正交分量  $n_s(t)$  的获取





#### 1、余弦信号的复信号表示

$$\begin{split} s(t) &= a \cos \left[ \omega_0 t + \varphi \right] \qquad \hat{s}(t) = a \sin(\omega_0 t + \varphi) \\ \tilde{s}(t) &= a \cdot e^{j(\omega_0 t + \varphi)} = a \cdot \cos(\omega_0 t + \varphi) + j \cdot a \cdot \sin(\omega_0 t + \varphi) = s(t) + j \hat{s}(t) \\ \tilde{a} &= a e^{j\varphi} \qquad 则有 \\ S(\omega) &= \pi \left[ \tilde{a} \delta(\omega - \omega_0) + \tilde{a}^* \delta(\omega + \omega_0) \right] \\ \hat{S}(\omega) &= -j\pi \left[ \tilde{a} \delta(\omega - \omega_0) - \tilde{a}^* \delta(\omega + \omega_0) \right] \end{split}$$

$$\tilde{S}(\omega) = S(\omega) + j\hat{S}(\omega) = 2\pi\tilde{a}\delta(\omega - \omega_0) = 2S(\omega) \cdot U(\omega) = \begin{cases} 2S(\omega) & \omega \\ 0 & \omega \end{cases}$$

其频谱只在正频域存在, 具有单边谱

#### 2、高频窄带信号 $f_0 >> \Delta f$

$$s(t) = a(t)\cos[\omega_0 t + \varphi(t)]$$

$$= \frac{1}{2} [\tilde{a}(t)e^{j\omega_0 t} + \tilde{a}^*(t)e^{-j\omega_0 t}]$$

$$\tilde{s}(t) = a(t)e^{j[\omega_0 t + \varphi(t)]} = \tilde{a}(t)e^{j\omega_0 t}$$

由 
$$\tilde{a}(t) \leftrightarrow \tilde{A}(\omega)$$

$$S(\omega) = \frac{1}{2} [\tilde{A}(\omega - \omega_0) + \tilde{A}^*(-\omega - \omega_0)]$$
 
$$\tilde{S}(\omega) = \tilde{A}(\omega - \omega_0) = 2S(\omega) \cdot U(\omega)$$

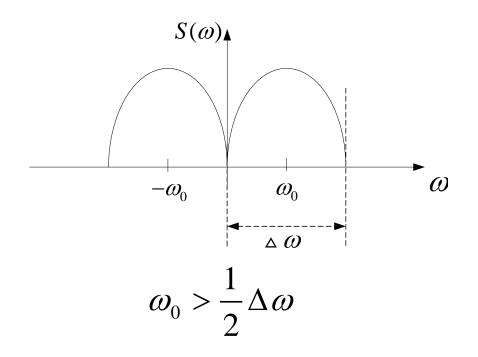
3、任意信号的复表示

设 x(t) 为任意实信号, 其复信号表达式为

$$\tilde{x}(t) = x(t) + j\hat{x}(t)$$

其与实信号的频谱关系为

$$\tilde{X}(\omega) = 2X(\omega)U(\omega)$$



#### 3、任意信号的复表示

$$\tilde{X}(\omega) = 2X(\omega)U(\omega)$$

$$2U(\omega) \Leftrightarrow \delta(t) + \frac{1}{-j\pi t}$$

利用傅里叶变换的相乘性质, 可得

$$\tilde{x}(t) = x(t) * [\delta(t) + \frac{J}{\pi t}]$$

$$= x(t) + jx(t) * \frac{1}{\pi t}$$

$$= x(t) + j\hat{x}(t)$$

$$\hat{x}(t) = x(t) * \frac{1}{\pi t}$$

$$\frac{1}{\pi t} \leftrightarrow -j \operatorname{sgn}(\omega)$$

$$\operatorname{sgn}(\omega) = \begin{cases} 1 & \omega > 0 \\ 0 & \omega = 0 \\ -1 & \omega < 0 \end{cases}$$

 $\hat{x}(t)$ 的频谱为

$$\hat{X}(\omega) = -j \operatorname{sgn}(\omega) X(\omega)$$

推荐一篇关于本部分内容的论文:

北京大学 程乾生 "希尔伯特变换与信号的包络、瞬时相位和瞬时频率"



✓ 好处:数据率不变,采样率减半

若在区间 $t \in R$ 中,给定实值函数x(t) ,则其复形式为

$$\tilde{x}(t) = x(t) + j\hat{x}(t)$$

 $\hat{x}(t)$ 为x(t)的希尔伯特变换,有

$$\hat{x}(t) = x(t) * \frac{1}{\pi t}$$

则  $\tilde{x}(t)$  的傅里叶变换为

$$F[\tilde{x}(t)] = 2X(\omega)U(\omega)$$

✓ 希尔伯特正变换

$$\hat{x}(t) = H[x(t)] = \frac{1}{\pi} \int_{-\infty}^{+\infty} \frac{x(\tau)}{t - \tau} d\tau = x(t) * \frac{1}{\pi t}$$

✓ 希尔伯特反变换

$$x(t) * \frac{1}{\pi t} \leftrightarrow X(\omega) [-j \operatorname{sgn}(\omega)] = F[\hat{x}(t)]$$

$$\therefore X(\omega) = F[\hat{x}(t)] \frac{1}{-j \operatorname{sgn}(\omega)} = F[\hat{x}(t)] j \operatorname{sgn}(\omega)$$

可得, 
$$x(t) = \hat{x}(t) * \left(-\frac{1}{\pi t}\right)$$

#### ✓ 希尔伯特变换性质

▶性质1 
$$F[\hat{x}(t)] = -j \operatorname{sgn}(\omega) X(\omega)$$

▶性质2 
$$H[\hat{x}(t)] = -x(t)$$

则 
$$\hat{y}(t) = \hat{x}(t) * v(t) = x(t) * \hat{v}(t)$$

✓希尔伯特变换性质 (续)

▶性质4

设实信号a(t) 傅里叶变换为  $A(\omega)$ 

$$A(\omega) = \begin{cases} A(\omega) & |\omega| \le B \\ 0 & \sharp \omega \end{cases}$$

刑  $H[a(t)\cos\omega_c t] = a(t)\sin\omega_c t$  $H[a(t)\sin\omega_c t] = -a(t)\cos\omega_c t$ 

式中  $\omega_c > B$ 

✓希尔伯特变换性质 (续)

》性质5 
$$\lim_{T \to \infty} \frac{1}{2T} \int_{-T}^{T} x^{2}(t) dt = \lim_{T \to \infty} \frac{1}{2T} \int_{-T}^{T} \hat{x}^{2}(t) dt$$
$$\left| F(\omega, T) \right|^{2} \quad \left| F(\omega, T) \cdot [-j \operatorname{sgn}(\omega)] \right|^{2}$$

》性质6 
$$\lim_{T \to \infty} \frac{1}{2T} \int_{-T}^{T} x(t) \hat{x}(t) dt = 0$$

即 x(t)和 $\hat{x}(t)$ 是正交的



#### 1、复随机变量

 $\checkmark$ 定义 如果X、Y为实随机变量,则Z=X+jY为复随机变量。

✓均値 
$$m_Z = E[Z] = E[X] + jE[Y] = m_X + jm_Y$$

イ方差
$$D(Z) = E[(Z - m_Z)(Z - m_Z)^*] = E[|Z - m_Z|^2]$$

$$= E\{|(X - m_X) + j(Y - m_Y)|^2\}$$

$$= E[(X - m_X)^2 + (Y - m_Y)^2]$$

$$= D[X] + D[Y]$$

$$= E[|Z|^2] - |m_Z|^2$$

#### 1、复随机变量

✓互相关 
$$E(Z_1Z_2^*)$$

✓协方差 
$$C_{Z_1Z_2} = \text{cov}(Z_1, Z_2)$$
 
$$= E\left\{ (Z_1 - m_{Z_1})(Z_2 - m_{Z_2})^* \right\}$$
 
$$= E(Z_1Z_2^*) - m_{Z_1}m_{Z_2}^*$$

#### 1、复随机变量

$$E(Z_1Z_2^*)=0$$

✓互不相关 
$$cov(Z_1, Z_2) = E\{(Z_1 - m_{Z_1})(Z_2 - m_{Z_2})^*\}$$
  
= 0

或有 
$$E[Z_1Z_2^*] = E[Z_1]E[Z_2^*]$$

$$f_{Z_1Z_2}(x_1, y_1; x_2, y_2) = f_{Z_1}(x_1, y_1) f_{Z_2}(x_2, y_2)$$

2、复随机过程

设复随机过程为
$$Z(t)=X(t)+jY(t)$$
,

✓均值

$$m_Z(t) = E[Z(t)] = m_X(t) + jm_Y(t)$$

イ方差 
$$\sigma_Z^2(t) = E[|Z(t) - m_Z(t)|^2] = \sigma_X^2(t) + \sigma_Y^2(t)$$
  
=  $E[|Z(t)|^2] - |m_Z(t)|^2$ 

✓均方値 
$$\psi_Z^2(t) = E[|Z(t)|^2] = E[Z(t)Z^*(t)]$$

#### 2、复随机过程

✓自相关函数

$$\begin{split} R_z(t_1, t_2) &= E[Z(t_1)Z^*(t_2)] \\ &= E\Big\{ \Big[ X(t_1) + jY(t_1) \Big] \Big[ X(t_2) - jY(t_2) \Big] \Big\} \\ &= R_X(t_1, t_2) + R_Y(t_1, t_2) + j[R_{YX}(t_1, t_2) - R_{XY}(t_1, t_2)] \end{split}$$

✓自协方差函数

$$\begin{split} C_Z(t_1, t_2) &= \text{cov}[Z(t_1), Z(t_2)] \\ &= E\{[Z(t_1) - m_Z(t_1)][Z(t_2) - m_Z(t_2)]^*\} \\ &= E[Z(t_1)Z^*(t_2)] - m_Z(t_1)m_Z^*(t_2) \end{split}$$

#### 2、复随机过程

✓互不相关、正交、相互独立

✓广义平稳

$$E[|Z(t)|^{2}] < \infty$$

$$m_{Z}(t) = E[Z(t)] = m_{Z}$$

$$R_{Z}(t, t - \tau) = R_{Z}(\tau)$$

$$= E[Z(t)Z^{*}(t - \tau)]$$

#### 2、复随机过程

✓ R(τ)的性质

$$(1) R_z(-\tau) = R_z^*(\tau)$$

(2) 
$$R_z(0) \ge 0$$

$$(3) |R_z(\tau)| \le R_z(0)$$

(4) R(t)为非负定的

$$\mathbb{P} \sum_{i,j=1}^{n} R_i(\tau_i - \tau_j) z_i z_j^* \ge 0$$

#### 3、实随机过程复表示

利用希尔伯特变换可以把实随机过程表示成复随机过程

$$\tilde{X}(t) = X(t) + j\hat{X}(t)$$

 $\checkmark$ 若X(t) 为平稳过程, $R_X(\tau)$ 存在

$$R_{\hat{X}}(\tau) = E[\hat{X}(t)\hat{X}(t-\tau)]$$
$$= R_{X}(\tau)$$

#### 3、实随机过程复表示

$$\angle R_{\hat{X}X}(\tau) = R_X(\tau) * \frac{1}{\pi \tau} = \hat{R}_X(\tau)$$

$$R_{X\hat{X}}(\tau) = R_X(\tau) * \frac{-1}{\pi \tau} = -\hat{R}_X(\tau)$$

$$\therefore R_{X\hat{X}}(0) = -\hat{R}_X(0) , R_{\hat{X}X}(0) = \hat{R}_X(0)$$

$$\therefore \hat{R}_X(0) = -\hat{R}_X(0) = 0$$

所以, 实部, 虚部在同一时刻是正交的

#### 3、实随机过程复表示

$$\begin{split} \checkmark \, R_{\tilde{X}}\left(\tau\right) &= E\{[X\left(t\right) + j\hat{X}\left(t\right)][X\left(t - \tau\right) - j\hat{X}\left(t - \tau\right)]\} \\ &= R_{X}\left(\tau\right) + R_{\hat{X}}\left(\tau\right) + j[R_{\hat{X}X}\left(\tau\right) - R_{X\hat{X}}\left(\tau\right)] \\ &= 2R_{X}\left(\tau\right) + j2\hat{R}_{X}\left(\tau\right) \end{split}$$

因为 $R_X(\tau)$ 偶函数,  $\hat{R}_X(\tau)$ 为奇函数, 故有

$$R_{\tilde{X}}(-\tau) = 2[R_X(\tau) - j\hat{R}_X(\tau)] = R_{\tilde{X}}^*(\tau)$$

#### 3、实随机过程复表示

$$\begin{aligned}
\checkmark & F[\hat{R}_X(\tau)] = -j \operatorname{sgn} S_X(\omega) \\
S_{\tilde{X}}(\omega) &= 2S(\omega)[1 + \operatorname{sgn}(\omega)] \\
&= \begin{cases}
0 & \omega < 0 \\
2S_X(\omega) & \omega = 0 \\
4S_X(\omega) & \omega > 0
\end{aligned}$$

由于 $\Delta \omega \ll \omega_0$ 

$$\therefore S_{\tilde{X}}(\omega) = 4S_X(\omega)U(\omega)$$



窄带过程

$$\omega_0 \gg B$$

$$n(t) = A(t)\cos[\omega_0 t + \phi(t)]$$
$$= n_c(t)\cos\omega_0 t - n_s(t)\sin\omega_0 t$$

$$\hat{n}(t) = n_c(t)\sin\omega_0 t + n_s(t)\cos\omega_0 t$$

$$n_c(t) = n(t)\cos\omega_0 t + \hat{n}(t)\sin\omega_0 t$$

$$n_{s}(t) = \hat{n}(t)\cos\omega_{0}t - n(t)\sin\omega_{0}t$$

$$n_c(t) = A(t)\cos\phi(t)$$
 ,  $n_s(t) = A(t)\sin\phi(t)$ 

$$A(t) = \sqrt{n_c^2(t) + n_s^2(t)} , \ \phi(t) = \tan^{-1} \frac{n_s(t)}{n_c(t)}$$

1、
$$R_{n_c}(\tau)$$
和 $R_{n_s}(\tau)$ 

$$\begin{split} R_{n_c}(\tau) &= E\{[n(t)\cos\omega_0 t + \hat{n}(t)\sin\omega_0 t][n(t-\tau)\cos\omega_0 (t-\tau) \\ &+ \hat{n}(t-\tau)\sin\omega_0 (t-\tau)]\} \\ &= R_n(\tau)\cos\omega_0 t\cos\omega_0 (t-\tau) + R_{\hat{n}}(\tau)\sin\omega_0 t\sin\omega_0 (t-\tau) \\ &+ R_{n\hat{n}}(\tau)\cos\omega_0 t\sin\omega_0 (t-\tau) + R_{\hat{n}n}(\tau)\sin\omega_0 t\cos\omega_0 (t-\tau) \\ &= R_n(\tau)\cos\omega_0 t\sin\omega_0 (t-\tau) + R_{\hat{n}n}(\tau)\sin\omega_0 t\cos\omega_0 (t-\tau) \\ &= R_n(\tau)\cos\omega_0 \tau + \hat{R}_n(\tau)\sin\omega_0 \tau \end{split}$$

长似,可得 
$$R_{n_s}(\tau) = R_n(\tau)\cos\omega_0\tau + \hat{R}_n(\tau)\sin\omega_0\tau = R_{n_c}(\tau)$$

$$R_{n_s}(0) = R_{n_c}(0) = R_n(0)$$

# 2、互相关函数 $R_{n_c n_s}(\tau)$

$$\begin{split} R_{n_c n_s}(\tau) &= E[n_c(t) n_s(t-\tau)] = E\{[n(t) \cos \omega_0 t + \hat{n}(t) \sin \omega_0 t] [\hat{n}(t-\tau) \cos \omega_0 (t-\tau) \\ &- n(t-\tau) \sin \omega_0 (t-\tau)]\} \\ &= E[n(t) \hat{n}(t-\tau) \cos \omega_0 t \cos \omega_0 (t-\tau)] - E[n(t) n(t-\tau) \cos \omega_0 t \sin \omega_0 (t-\tau)] \\ &- E[\hat{n}(t) n(t-\tau) \sin \omega_0 t \sin \omega_0 (t-\tau)] + E[\hat{n}(t) \hat{n}(t-\tau) \sin \omega_0 t \cos \omega_0 (t-\tau)] \\ &= -\hat{R}_n(\tau) [\cos \omega_0 t \cos \omega_0 (t-\tau) + \sin \omega_0 t \sin \omega_0 (t-\tau)] \\ &+ R_n(\tau) [-\cos \omega_0 t \sin \omega_0 (t-\tau) + \sin \omega_0 t \cos \omega_0 (t-\tau)] \\ &= R_n(\tau) \sin \omega_0 \tau - \hat{R}_n(\tau) \cos \omega_0 \tau \\ R_{n_s n_c}(\tau) &= -R_n(\tau) \sin \omega_0 \tau + \hat{R}_n(\tau) \cos \omega_0 \tau = -R_{n_c n_s}(\tau) = R_{n_c n_s}(-\tau) \end{split}$$

$$R_{n_c n_s}(0) = 0$$

3. 
$$R_n(\tau)$$
  

$$n(t) = n_c(t)\cos\omega_0 t - n_s(t)\sin\omega_0 t$$

$$I_n(\tau) = E[n(t)n(t-\tau)] = E\{[n_c(t)\cos\omega_0 t - n_s(t)\sin\omega_0 t]$$

$$[n_c(t-\tau)\cos\omega_0(t-\tau) - n_s(t-\tau)\sin\omega_0(t-\tau)]\}$$

$$= E[n_c(t)n_c(t-\tau)\cos\omega_0 t\cos\omega_0 t\cos\omega_0(t-\tau)] - E[n_c(t)n_s(t-\tau)\cos\omega_0 t\sin\omega_0(t-\tau)$$

$$- E[n_s(t)n_c(t-\tau)\sin\omega_0 t\cos\omega_0(t-\tau)] + E[n_s(t)n_s(t-\tau)\sin\omega_0 t\sin\omega_0(t-\tau)]$$

$$= R_{n_c}(\tau)[\cos\omega_0 t\cos\omega_0(t-\tau) + \sin\omega_0 t\sin\omega_0(t-\tau)] +$$

$$- R_{n_cn_s}(\tau)\cos\omega_0 t\sin\omega_0(t-\tau) - R_{n_sn_c}(\tau)\sin\omega_0 t\cos\omega_0(t-\tau)$$

$$= R_{n_c}(\tau)\cos\omega_0 \tau + R_{n_cn_s}(\tau)\sin\omega_0 \tau$$

#### 4、功率谱

$$\begin{split} S_{n_c}(\omega) &= F[R_{n_c}(\tau)] = F[R_n(\tau)\cos\omega_0\tau + \hat{R}_n(\tau)\sin\omega_0\tau] \\ &= \begin{cases} S_n(\omega - \omega_0) + S_n(\omega + \omega_0) & |\omega| \le B \\ 0 & \text{ i.i.} \end{cases} \end{split}$$

$$\begin{split} S_{n_c n_s}(\tau) &= F[R_{n_c n_s}(\tau)] = F[R_n(\tau) \sin \omega_0 \tau - \hat{R}_n(\tau) \cos \omega_0 \tau] \\ &= \frac{1}{2j} [S_n(\omega - \omega_0) - S_n(\omega + \omega_0)] + \frac{1}{2} [j \operatorname{sgn}(\omega - \omega_0) S_n(\omega - \omega_0) + j \operatorname{sgn}(\omega + \omega_0) S_n(\omega + \omega_0)] \\ &= \begin{cases} -j[S_n(\omega - \omega_0) - S_n(\omega + \omega_0)] & |\omega| \le B \\ 0 & \text{i.i.} \end{cases} \end{split}$$

## 作业

- ▶ 4.5, 4.6, 4.7, 4.8
- > 4.9, 4.11, 4.13, 4.19, 4.22