

Advanced Digital Signal Processing (ADSP)

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Part B: Adaptive signal processing

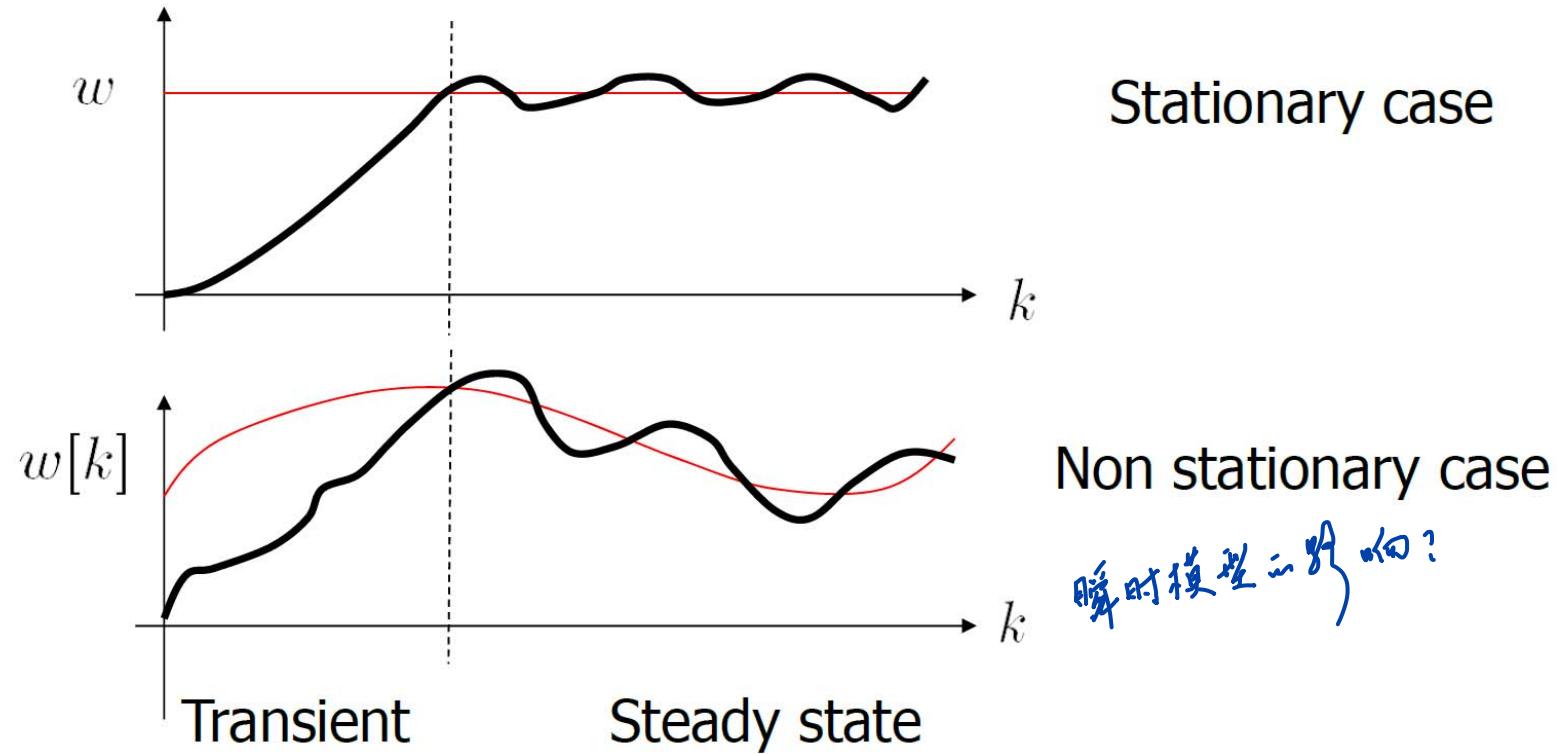


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Convergence LMS

Acquisition and tracking:



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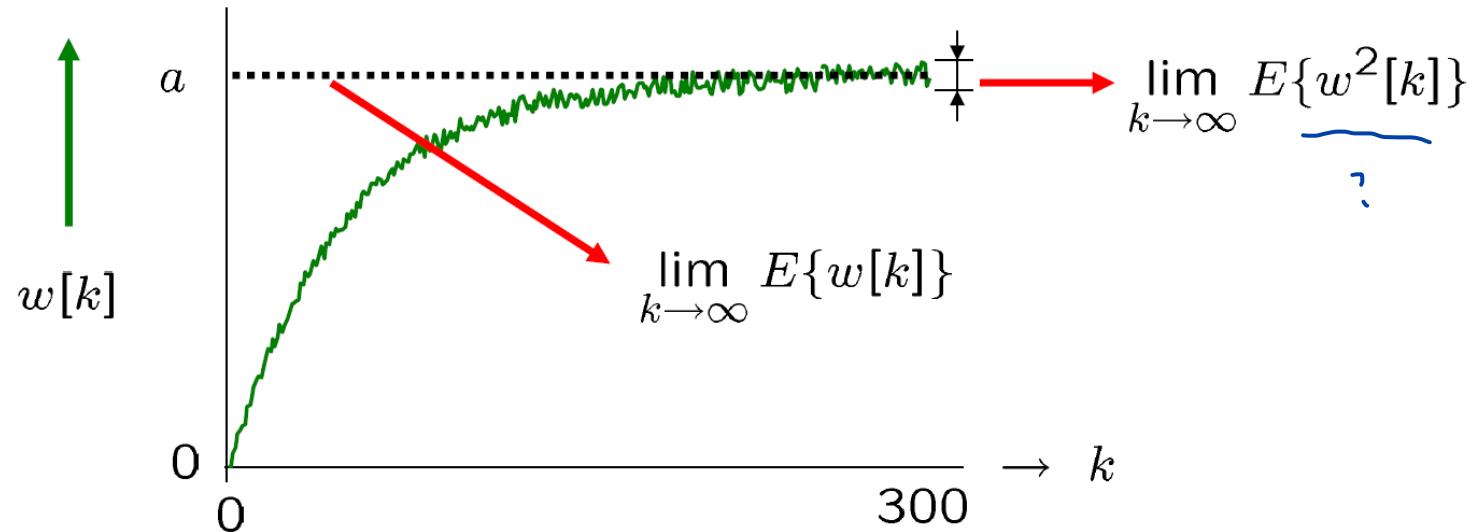
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Convergence LMS

Consequence of not using $E\{\cdot\}$? 不使用 $E\{\cdot\}$ 的推论

Example: LMS, $N = 1$, $w[0] = 0$ $w_o = a$



Questions about convergence:

$$\lim_{k \rightarrow \infty} E\{w[k]\} = w_o = a \text{ and } \lim_{k \rightarrow \infty} E\{w^2[k]\} < \infty?$$

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Convergence LMS

First compare difference with (optimal) Wiener weights:

$$\underline{d}[k] = \underline{w}[k] - \underline{w}_o \quad \text{with} \quad \underline{w}_o = \mathbf{R}_x^{-1} \cdot \underline{r}_{ex}$$

$$\left\{ \begin{array}{l} \underline{w}[k+1] = \underline{w}[k] + 2\alpha (\underline{x}[k]e[k] - \underline{x}[k]\underline{x}^t[k]\underline{w}[k]) \\ \underline{w}[k+1] - \underline{w}_o = (\mathbf{I} - 2\alpha \underline{x}[k]\underline{x}^t[k]) \underline{w}[k] - \underline{w}_o + 2\alpha \underline{x}[k]e[k] \end{array} \right.$$

$$\rightarrow \underline{d}[k+1] = (\mathbf{I} - 2\alpha \underline{x}[k]\underline{x}^t[k]) \underline{d}[k] + \underbrace{2\alpha \underline{x}[k]r_{min}[k]}_{- 2\alpha \underline{x}[k]\underline{x}^t[k]\underline{w}_o + 2\alpha \underline{x}[k]e[k]} + \underbrace{2\alpha \underline{x}[k](e[k] - \underline{x}^t[k]\underline{w}_o)}_{\hat{e}_{min}}$$

with $r_{min}[k] = e[k] - \underline{x}^t[k]\underline{w}_o$

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Convergence LMS

$$\text{迭代: SGD: } \underline{d}[k+1] = (\mathbf{I} - 2\alpha R_x) \underline{d}[k]; \quad \underline{d}[k] = (\mathbf{I} - 2\alpha R_x)^k \underline{d}[0]$$

Convergence in the mean: Why use mean $E[\cdot]$?

$$E\{\underline{d}[k+1]\} = E\{(\mathbf{I} - 2\alpha \underline{x}[k] \underline{x}^t[k]) \underline{d}[k]\}$$

$$+ 2\alpha \cdot (E\{\underline{x}[k] e[k]\} - E\{\underline{x}[k] \underline{x}^t[k]\} \underline{w}_o)$$

0

$r_{min} \perp \underline{x}$

With independence assumption:

$$E\{\underline{x}[k] \underline{x}^t[k] \underline{d}[k]\} \approx E\{\underline{x}[k] \underline{x}^t[k]\} \cdot E\{\underline{d}[k]\} \quad k \rightarrow \infty \quad (1 - 2\alpha \lambda_i)^k = 0$$

$$\Rightarrow E\{\underline{d}[k+1]\} = (\mathbf{I} - 2\alpha R_x) E\{\underline{d}[k]\} \quad \begin{matrix} 0 < |1 - 2\alpha \lambda_i| < 1 \\ \text{for all } i: \alpha < \frac{1}{\lambda_i} \Rightarrow \alpha < \frac{1}{\lambda_{max}} \end{matrix}$$

Average convergence behaviour LMS same as SGD

$$0 < \alpha < 1/\lambda_{max} : \lim_{k \rightarrow \infty} E\{\underline{w}[k]\} = \underline{w}_o; \quad \tau_{av,i} \approx 1/2\alpha\lambda_i$$

\Rightarrow Depends on coloration input process!

$\ln(e^{-\frac{k}{\tau_i}}) = \ln(1 - 2\alpha \lambda_i)^k$

白信号和有色信号不同点在于?

$\Rightarrow \tau_i = -\frac{1}{\ln(1 - 2\alpha \lambda_i)} \sim \frac{1}{2\alpha \lambda_i}$

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Convergence LMS

Mean-square convergence:

$$J_{LMS} = E\{r^2\} = E\{(e - \underline{\mathbf{w}}^t \underline{\mathbf{x}})(e - \underline{\mathbf{x}}^t \underline{\mathbf{w}})\}$$

With $\underline{\mathbf{d}} = \underline{\mathbf{w}} - \underline{\mathbf{w}}_o$ or $\underline{\mathbf{w}} = \underline{\mathbf{w}}_o + \underline{\mathbf{d}}$ \Rightarrow

$$J_{LMS} = E\{\left((e - \underline{\mathbf{w}}_o^t \underline{\mathbf{x}}) - \underline{\mathbf{d}}^t \underline{\mathbf{x}}\right) \left((e - \underline{\mathbf{x}}^t \underline{\mathbf{w}}_o) - \underline{\mathbf{x}}^t \underline{\mathbf{d}}\right)\}$$

$$\begin{aligned} J_{LMS} &= E\{(e - \underline{\mathbf{w}}_o^t \underline{\mathbf{x}})(e - \underline{\mathbf{x}}^t \underline{\mathbf{w}}_o)\} + E\{\underline{\mathbf{d}}^t \underline{\mathbf{x}} \underline{\mathbf{x}}^t \underline{\mathbf{d}}\} \\ &\quad - E\{\underline{\mathbf{d}}^t \underline{\mathbf{x}}(e - \underline{\mathbf{x}}^t \underline{\mathbf{w}}_o)\} - E\{(e - \underline{\mathbf{w}}_o^t \underline{\mathbf{x}})\underline{\mathbf{x}}^t \underline{\mathbf{d}}\} \end{aligned}$$

Independence assumption \Rightarrow

$$\begin{aligned} E\{\underline{\mathbf{d}}^t \underline{\mathbf{x}}(e - \underline{\mathbf{x}}^t \underline{\mathbf{w}}_o)\} &\approx E\{\underline{\mathbf{d}}^t\} \cdot (E\{\underline{\mathbf{x}}e\} - E\{\underline{\mathbf{x}}\underline{\mathbf{x}}^t\}\underline{\mathbf{w}}_o) \\ &= E\{\underline{\mathbf{d}}^t\} \cdot (\underline{\mathbf{r}}_{ex} - \mathbf{R}_x \underline{\mathbf{w}}_o) = 0 \end{aligned}$$

Similar $E\{(e - \underline{\mathbf{w}}_o^t \underline{\mathbf{x}})\underline{\mathbf{x}}^t \underline{\mathbf{d}}\} \rightarrow 0$

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Convergence LMS

Compare with MMSE expression

$$\Rightarrow J_{LMS} \approx E\{(e - \underline{\mathbf{w}}_o^t \underline{\mathbf{x}})(e - \underline{\mathbf{x}}^t \underline{\mathbf{w}}_o)\} + E\{\underline{\mathbf{d}}^t \underline{\mathbf{x}} \underline{\mathbf{x}}^t \underline{\mathbf{d}}\}$$

\uparrow \uparrow
Fixed *Adaptive*

Wiener error: $J_{min} = E\{(e - \underline{\mathbf{w}}_o^t \underline{\mathbf{x}})(e - \underline{\mathbf{x}}^t \underline{\mathbf{w}}_o)\} = E\{r_{min}^2\}$

$$\begin{aligned}\text{Excess error: } J_{ex} &= E\{\underline{\mathbf{d}}^t \underline{\mathbf{x}} \underline{\mathbf{x}}^t \underline{\mathbf{d}}\} \approx E\{\underline{\mathbf{d}}^t E\{\underline{\mathbf{x}} \underline{\mathbf{x}}^t\} \underline{\mathbf{d}}\} \\ &= E\{\underline{\mathbf{d}}^t \mathbf{R}_x \underline{\mathbf{d}}\}\end{aligned}$$

Dynamic behaviour adaptive filter: $\tilde{J}[k] = \frac{J_{ex}[k]}{J_{min}[k]}$?

Depends on coloration input process!



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Convergence LMS

Conclusion convergence LMS:

Convergence gradient based algorithms, like SGD heavily relates on correlation input process and initiation of adaptive weights!

This also follows from rewriting gradient as:

$$\underline{\nabla} = -2(\underline{r}_{ex} - \underline{R}_x \underline{w}[k])$$

$$\underline{\nabla} = -2\underline{R}_x \cdot (\underline{R}_x^{-1} \underline{r}_{ex} - \underline{w}[k])$$

⇒ Gradient (=update) depends on coloration input

Solution:

Alternative that decorrelates input → Newton

输入正相关



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Part B: Adaptive signal processing

Focus on **single channel** adaptive algorithms
using **FIR** structure

- Applications Adaptive Algorithms
- Minimum Mean Square Error (MMSE)
- Constrained MMSE
- Least Square (LS)
- Steepest Gradient Descent (SGD)
- Three LMS variants: NLMS, Complex LMS, Constrained LMS
- **Newton**
- Recursive Least Squares (RLS)
- Frequency Domain Adaptive Filter (FDAF)
- Summary

Part B: Adaptive signal processing



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Newton

取消彩色影响

Principle: Undo coloration effect SGD

SGD:

$$\underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] - \alpha \underline{\nabla} \text{ with } \underline{\nabla} = -2(\underline{\mathbf{r}}_{ex} - \mathbf{R}_x \underline{\mathbf{w}}[k])$$

Newton:

$$\underline{\mathbf{w}}[k + 1] = \underline{\mathbf{w}}[k] - \alpha \mathbf{R}_x^{-1} \underline{\nabla} \Rightarrow \text{自化}$$

$$\boxed{\underline{\mathbf{w}}[k + 1] = \underline{\mathbf{w}}[k] + 2\alpha \mathbf{R}_x^{-1} \cdot (\underline{\mathbf{r}}_{ex} - \mathbf{R}_x \underline{\mathbf{w}}[k])}$$

Note: General update rule $\underline{\mathbf{w}} = \underline{\mathbf{w}} - \alpha \underline{\mathbf{U}}$

$\underline{\mathbf{U}}$ must be such that each iteration J decreases

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Newton

Convergence Newton algorithm: 消除了 R_x 对收敛的影响

$$\underline{d}[k+1] = \underbrace{(\mathbf{I} - 2\alpha \mathbf{R}_x^{-1} \mathbf{R}_x)}_{\text{Conclusion: } \underline{d}[k+1] = (1-2\alpha)^k \underline{d}[0]} \underline{d}[k] = (1-2\alpha) \underline{d}[k]$$

Conclusion:

$$\underline{d}[k] = (1-2\alpha)^k \underline{d}[0]$$

For $|1-2\alpha| < 1 \Leftrightarrow 0 < \alpha < 1$

$$\lim_{k \rightarrow \infty} E\{\underline{d}[k]\} = \underline{0} \Leftrightarrow \lim_{k \rightarrow \infty} E\{\underline{w}[k]\} = \underline{w}_o = \mathbf{R}_x^{-1} \underline{r}_{ex}$$

用此迭代和“LMS”保持一致进行对比

\mathbf{R}_x^{-1} causes whitening of input x :

- All weights have same convergence!
- Equivalent to SGD with white noise input!

R_x 特征值相异?

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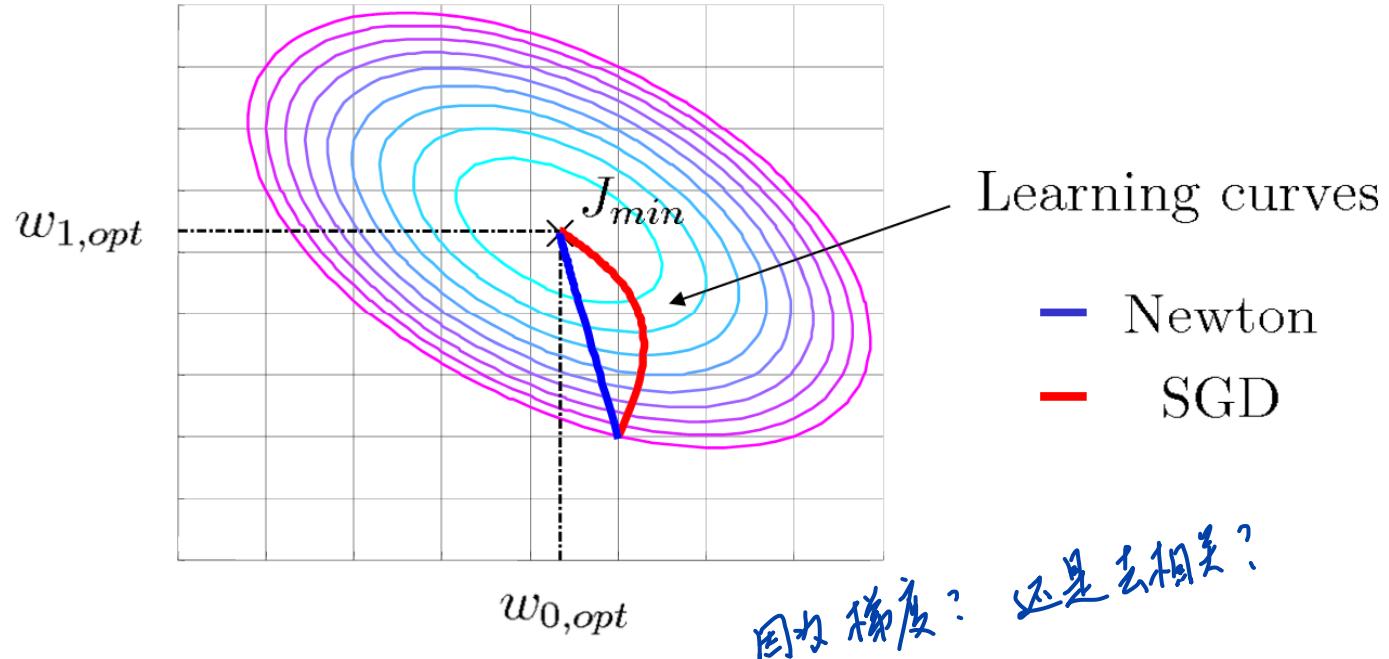
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Newton

Learning curves Newton vs. SGD in contour plot

Coloured input process with $\Gamma_x = \lambda_{max}/\lambda_{min} = 3$



Note: SGD-curve each iteration orthogonal to contourplot J

Newton-curve point each iteration towards J_{min}

自己设置了梯度？

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$$\textcircled{1} \text{ first view: } \underline{\nabla} = \underline{\nabla}_{SGD} = -2(\underline{r}_x - \underline{R}_x \underline{w})$$

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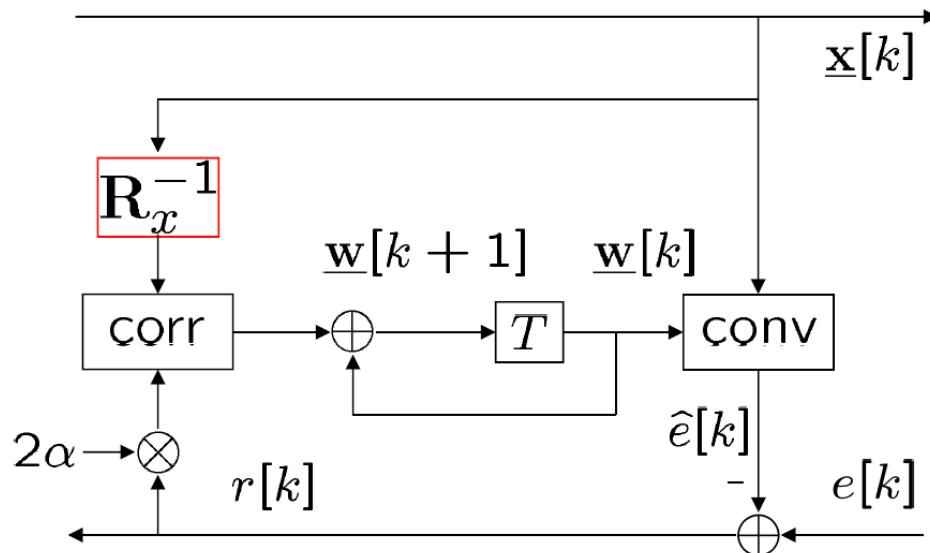
Newton

$$\textcircled{2} \text{ Another view: }$$

$$\underline{\nabla}_{LMS} = -2\underline{x}[k]\underline{r}[k]$$

By replacing $\underline{\nabla} \rightarrow \hat{\underline{\nabla}}_{LMS} = \underline{x}[k]\underline{r}[k]$

$$\Rightarrow \text{"LMS/Newton": } \underline{w}[k+1] = \underline{w}[k] + 2\alpha \underline{R}_x^{-1} \underline{x}[k] \underline{r}[k]$$



\underline{R}_x^{-1} causes whitening of input x

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Newton

收敛快(需要的步数少)
每一步计算复杂

Practical problems Newton:

Autocorrelation matrix \mathbf{R}_x :

- Not known in general
- May change in time (non-stationary process)
- Inversion is very expensive (many MIPS)

Complexity Newton: Huge

- ⇒ Need for efficient solution with estimate of \mathbf{R}_x
- ⇒ RLS; FDAF; etc.



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Part B: Adaptive signal processing

Focus on **single channel** adaptive algorithms
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- Recursive Least Squares (RLS) *递归最小二乘*
- Frequency Domain Adaptive Filter (FDAF)
- Summary



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Recursive Least Squares

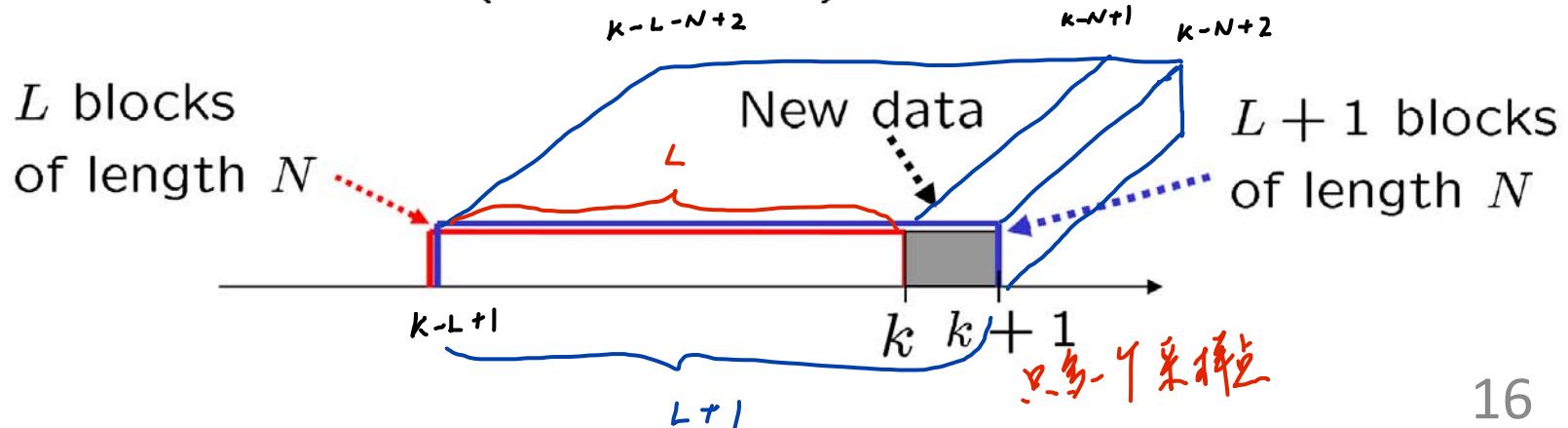
For L fixed, least squares problem becomes:

$$\min_{\underline{\mathbf{w}}} |\underline{\mathbf{e}} - \underline{\mathbf{X}} \cdot \underline{\mathbf{w}}|^2 \Rightarrow \underline{\mathbf{w}}_{LS} = (\underline{\mathbf{X}}^t \underline{\mathbf{X}})^{-1} \cdot (\underline{\mathbf{X}}^t \underline{\mathbf{e}})$$

$\underline{\mathbf{R}}_x^{-1}$ $\underline{\mathbf{r}}_{ex}$

RLS concept for $k \rightarrow k + 1$: 每次迭代使用“ $L+1$ ”blocks 数据 $(L+1) \cdot N$
除了 $k=0$ 时?

Find recursive (=adaptive) solution for LS problem



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Current solution (time k) :

Based on L data vectors, each of length N

$$\begin{aligned}\underline{\mathbf{w}}_{LS}^L[k] &= (\bar{\mathbf{R}}_x^L[k])^{-1} \cdot \underline{\mathbf{r}}_{ex}^L[k] \\ &= \left((\mathbf{X}^L[k])^t \mathbf{X}^L[k] \right)^{-1} \cdot (\mathbf{X}^L[k])^t \underline{\mathbf{e}}^L[k]\end{aligned}$$

$$\mathbf{X}^L[k] = \begin{pmatrix} \underline{\mathbf{x}}^t[k] \\ \underline{\mathbf{x}}^t[k-1] \\ \vdots \\ \underline{\mathbf{x}}^t[k-L+1] \end{pmatrix} \quad \underline{\mathbf{e}}^L[k] = \begin{pmatrix} e[k] \\ e[k-1] \\ \vdots \\ e[k-L+1] \end{pmatrix}$$

Similar result for $L+1$

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“RLS”
/ “LS” 中和迭代

Compute solution at time $k + 1$:

$$\begin{aligned}\underline{\mathbf{w}}_{LS}^{L+1}[k+1] &= (\bar{\mathbf{R}}_x^{L+1}[k+1])^{-1} \cdot \bar{\mathbf{r}}_{ex}^{L+1}[k+1] \\ &= \left((\mathbf{X}^{L+1}[k+1])^t \mathbf{X}^{L+1}[k+1] \right)^{-1} \cdot \\ &\quad \cdot (\mathbf{X}^{L+1}[k+1])^t \underline{\mathbf{e}}^{L+1}[k+1]\end{aligned}$$

$N \times 1$ $(L+1) \times N$ $N \times N$
 $N \times (L+1)$ $(L+1) \times 1$ $N \times 1$

With $\mathbf{X}^{L+1}[k+1] = \begin{pmatrix} \underline{\mathbf{x}}^t[k+1] \\ \underline{\mathbf{x}}^t[k] \\ \vdots \\ \underline{\mathbf{x}}^t[k-L+1] \end{pmatrix}_{(L+1) \times N}$

$$\underline{\mathbf{e}}^{L+1}[k+1] = (e[k+1], e[k], \dots, e[k-L+1])^t_{(L+1) \times 1}$$

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$$\bar{R}_x^t[k] = \underline{x}^L[k]^T \underline{x}[k] = \sum_{i=0}^{L-1} \underline{x}[k-i] \underline{x}^t[k-i]$$

Observe:

$$\begin{aligned} \bar{R}_x^{L+1}[k+1] &= \sum_{i=0}^L \underline{x}[k+1-i] \underline{x}^t[k+1-i] \text{ 共}(L+1)项 \\ &\stackrel{\substack{\sum_{i=1}^L \\ \leftarrow}}{=} \bar{R}_x^L[k] + \underline{x}[k+1] \cdot \underline{x}^t[k+1] \text{ } \xrightarrow[N \times N]{\text{矩阵}} \end{aligned}$$

$$\begin{aligned} \bar{r}_{ex}^{L+1}[k+1] &= \sum_{i=0}^L \underline{x}[k+1-i] e[k+1-i] \\ &= \bar{r}_{ex}^L[k] + \underline{x}[k+1] e[k+1] \end{aligned}$$

From matrix inversion lemma (see Appendix): \Rightarrow

$$\bar{R}_x^{-1}[k+1] = \bar{R}_x^{-1}[k] - \frac{\bar{R}_x^{-1}[k] \underline{x}[k+1] \underline{x}^t[k+1] \bar{R}_x^{-1}[k]}{1 + \underline{x}^t[k+1] \bar{R}_x^{-1}[k] \underline{x}[k+1]}$$

$N \cdot (L+1) \cdot (L+1) \cdot N^2$
 $N(L+1) \cdot (L+1) \cdot N$:

$$\text{Finally: } \underline{w}[k+1] = \bar{R}_x^{-1}[k+1] \cdot \bar{r}_{ex}[k+1]$$

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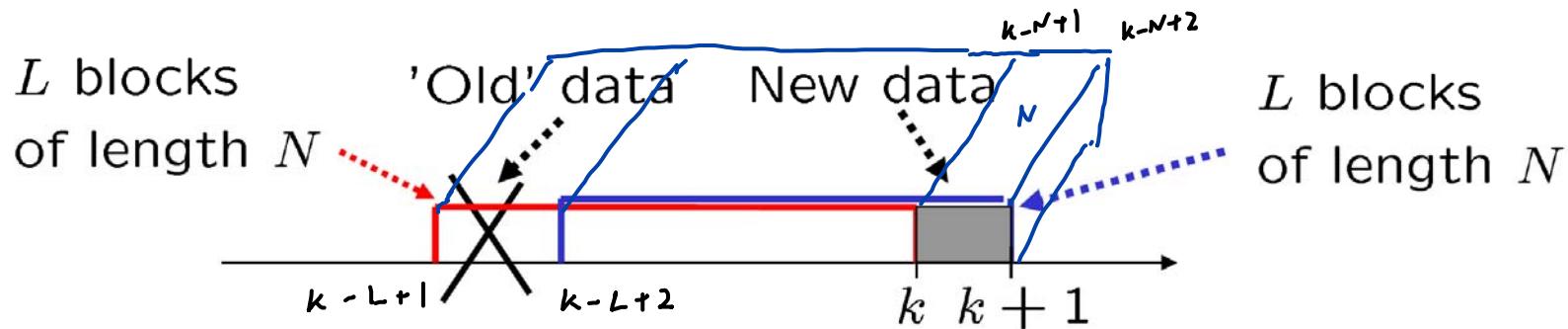
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RLS

For adaptivity → more effective data window

每次迭代使用“L”blocks数据

Sliding window: Keep window length L constant



Note: Now we can write for autocorrelation

$N \cdot L \cdot L \cdot N$

$$\bar{R}_x[k+1] = \bar{R}_x[k] - \underline{x}[k-L+1] \cdot \underline{x}^t[k-L+1] + \underline{x}[k+1] \cdot \underline{x}^t[k+1]$$

$$\bar{R}_x^{L+1}[k+1] = \bar{R}_x^L[k] + \underline{x}[k+1] \underline{x}^t[k+1]$$

$N \cdot (L+1) \cdot (L+1) \cdot N$ ⇒ Still very complex

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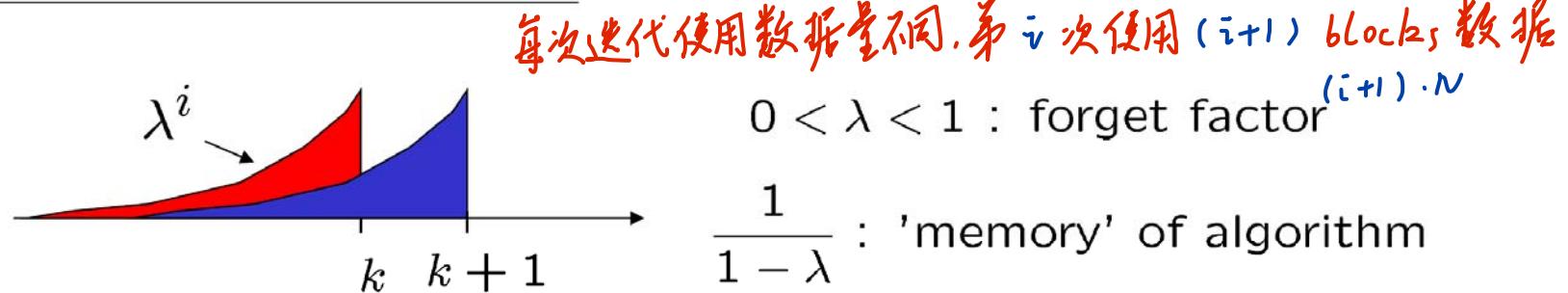


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Exponential window: Scale down data by factor λ



$$\text{X}[k] = \begin{pmatrix} \lambda^0 \underline{\mathbf{x}}^t[k] \\ \lambda^1 \underline{\mathbf{x}}^t[k-1] \\ \vdots \\ \lambda^k \underline{\mathbf{x}}^t[0] \end{pmatrix}_{(k+1) \cdot N ?} \quad \underline{\mathbf{e}}[k] = \begin{pmatrix} \lambda^0 e[k] \\ \lambda^1 e[k-1] \\ \vdots \\ \lambda^k e[0] \end{pmatrix}_{(k+1) \cdot 1}$$

所有累积量均含有
其中

$$J[k] = \sum_{i=0}^k \lambda^i r^2[k-i] = (\underline{\mathbf{e}}^t[k] - \underline{\mathbf{w}}^t[k] \mathbf{X}^t[k]) (\underline{\mathbf{e}}[k] - \mathbf{X}[k] \underline{\mathbf{w}}[k])$$

$$\bar{R}_x[k] = \sum_{i=0}^k \lambda^i \underline{\mathbf{x}}[k-i] \cdot \lambda^i \underline{\mathbf{x}}^t[k-i] = \sum_{i=0}^k \lambda^{2i} \underline{\mathbf{x}}[k-i] \underline{\mathbf{x}}^t[k-i]$$



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$$\bar{R}_x[k] = \sum_{i=0}^k \lambda^{2i} \underline{x}[k-i] \underline{x}^t[k-i]$$

Observe:

$$\begin{aligned}\bar{R}_x[k+1] &= \sum_{i=0}^{k+1} \lambda^{2i} \underline{x}[k+1-i] \underline{x}^t[k+1-i] = \sum_{i=1}^{k+1} [\cdot] + \sum_{i=0}^k \cdot \\ &= \lambda^2 \bar{R}_x[k] + \lambda^0 \underline{x}[k+1] \underline{x}^t[k+1]\end{aligned}$$

$$\bar{R}_x[k+1] = \lambda^2 \bar{R}_x[k] + \underline{x}[k+1] \cdot \underline{x}^t[k+1]$$

$$\bar{r}_{ex}[k+1] = \lambda^2 \bar{r}_{ex}[k] + \underline{x}^t[k+1] e[k+1]$$

From matrix inversion theorem (see Appendix):

$$\bar{R}^{-1}[k+1] = \lambda^{-2} (\bar{R}^{-1}[k] - \underline{g}[k+1] \cdot \underline{x}^t[k+1] \bar{R}^{-1}[k])$$

$$\text{with gain vector: } \underline{g}[k+1] = \frac{\bar{R}^{-1}[k] \underline{x}[k+1]}{\lambda^2 + \underline{x}^t[k+1] \bar{R}^{-1}[k] \underline{x}[k+1]}$$

New weight vector:

$$\underline{w}[k+1] = \bar{R}_x^{-1}[k+1] \cdot \bar{r}_{ex}[k+1]$$

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RLS algorithm

Initialization:

$$\underline{\mathbf{r}}_{ex}[0] = \underline{\mathbf{0}} ; \quad \overline{\mathbf{R}}_x^{-1}[0] = \delta \mathbf{I} \text{ with } \delta \text{ small}$$

For $k \geq 0$:

$$\underline{\mathbf{g}}[k+1] = \frac{\overline{\mathbf{R}}^{-1}[k]\underline{\mathbf{x}}[k+1]}{\lambda^2 + \underline{\mathbf{x}}^t[k+1]\overline{\mathbf{R}}^{-1}[k]\underline{\mathbf{x}}[k+1]}$$

$$\overline{\mathbf{R}}^{-1}[k+1] = \lambda^{-2} \left(\overline{\mathbf{R}}^{-1}[k] - \underline{\mathbf{g}}[k+1] \cdot \underline{\mathbf{x}}^t[k+1] \overline{\mathbf{R}}^{-1}[k] \right)$$

$$\underline{\mathbf{r}}_{ex}[k+1] = \lambda^2 \underline{\mathbf{r}}_{ex}[k] + \underline{\mathbf{x}}[k+1]e[k+1]$$

$$\underline{\mathbf{w}}[k+1] = \overline{\mathbf{R}}_x^{-1}[k+1] \cdot \underline{\mathbf{r}}_{ex}[k+1]$$

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Compare RLS with "LMS/Newton"

"LMS/Newton": $\underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] + 2\alpha \mathbf{R}_x^{-1} \underline{\mathbf{x}}[k] r[k]$

Update RLS can be rewritten as (see Appendix):

$$\begin{aligned}\underline{\mathbf{w}}[k+1] &= \underline{\mathbf{w}}[k] + \underline{\mathbf{g}}[k+1] (e[k+1] - \underline{\mathbf{x}}^t[k+1] \underline{\mathbf{w}}[k]) \\ &= \underline{\mathbf{w}}[k] + \underline{\mathbf{g}}[k+1] r[k+1]\end{aligned}$$

with gain vector:

$$\underline{\mathbf{g}}[k+1] = \frac{\overline{\mathbf{R}}^{-1}[k] \underline{\mathbf{x}}[k+1]}{\lambda^2 + \underline{\mathbf{x}}^t[k+1] \overline{\mathbf{R}}^{-1}[k] \underline{\mathbf{x}}[k+1]}$$

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Notes on RLS:

$$\underline{\mathbf{w}}[\infty] = \underline{\mathbf{w}}_o$$

Complexity: $O(N^2)$ per time update

Window length increases when time increases!

Exhibits unstable roundoff error accumulation

RLS basis for many practical algorithms

Decorrelation takes place in algorithm

去相关

误差的累积
通过平均的方法
消除累积误差



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FDAF

Frequency Domain Adaptive Filter

Alternative for LMS/Newton and RLS

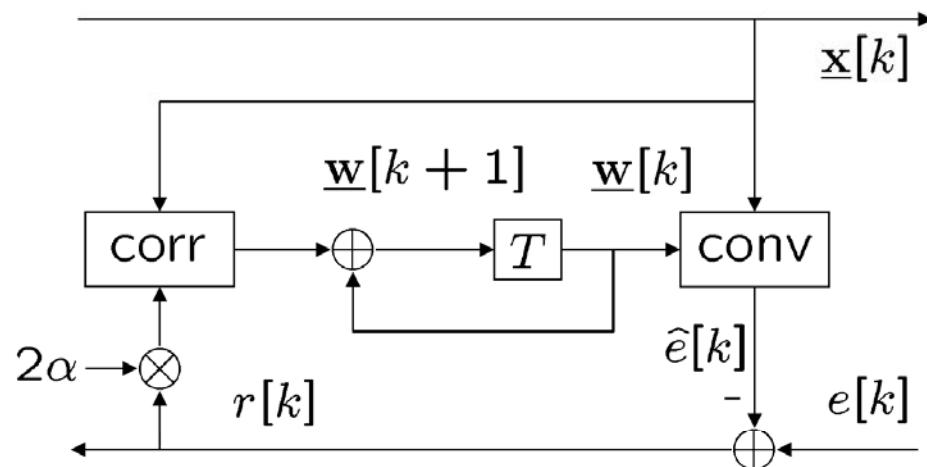


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FDAF

Frequency Domain Adaptive Filter

First translate LMS to frequency domain:



LMS weight update:

$$\underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] + 2\alpha \underline{\mathbf{x}}[k]r[k]$$

Filter output:

$$\hat{e}[k] = \underline{\mathbf{x}}^t[k] \cdot \underline{\mathbf{w}}[k]$$

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Apply filter operation in frequency domain:

$$\mathbf{F} \cdot \underline{\mathbf{x}}[k] = \underline{\mathbf{X}}[k] = (X_0[k], X_1[k], \dots, X_{N-1}[k])^t$$

$$\mathbf{F}^{-1} \cdot \underline{\mathbf{w}}[k] = \underline{\mathbf{W}}[k] = (W_0[k], W_1[k], \dots, W_{N-1}[k])^t$$

$$Note: \mathbf{F}^{-1} = \frac{1}{N} \mathbf{F}^*$$

Filter output:

$$\begin{aligned}\hat{e}[k] &= \sum_{i=0}^{N-1} x[k-i] w_i[k] = \underline{\mathbf{x}}^t[k] \cdot \underline{\mathbf{w}}[k] \\ &= \underline{\mathbf{x}}^t[k] \mathbf{F} \cdot \mathbf{F}^{-1} \underline{\mathbf{w}}[k] = (\mathbf{F} \underline{\mathbf{x}}[k])^t \cdot (\mathbf{F}^{-1} \underline{\mathbf{w}}[k]) \\ &= \underline{\mathbf{X}}^t[k] \cdot \underline{\mathbf{W}}[k] = \sum_{l=0}^{N-1} X_l[k] W_l[k]\end{aligned}$$

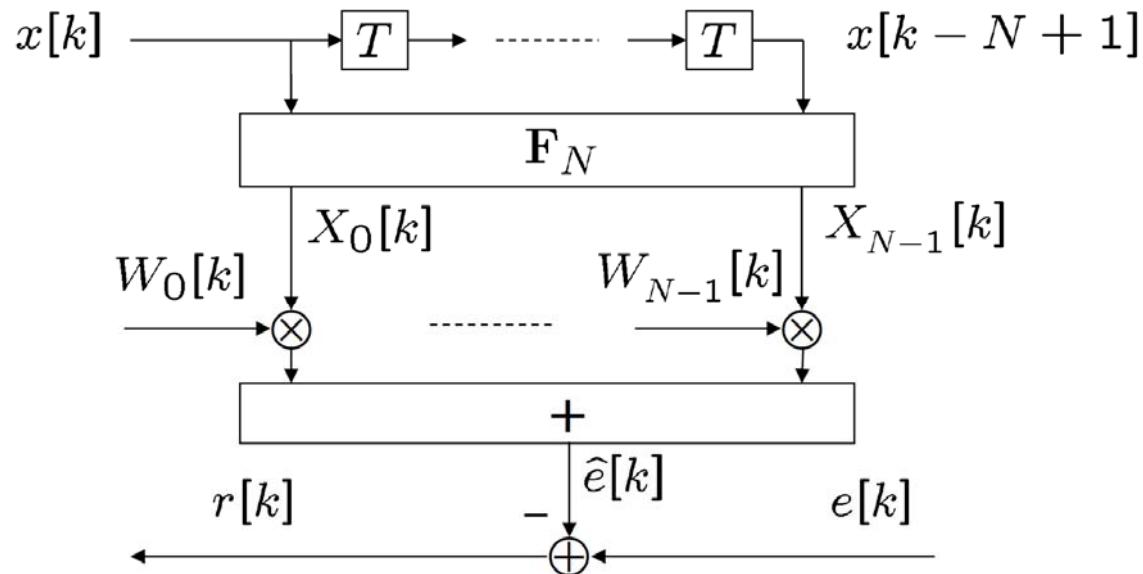
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Notes:

- Weights perform inverse transform
- Use DFT symmetry to reduce complexity
- Separate frequency bins ‘uncorrelated’ (large N)

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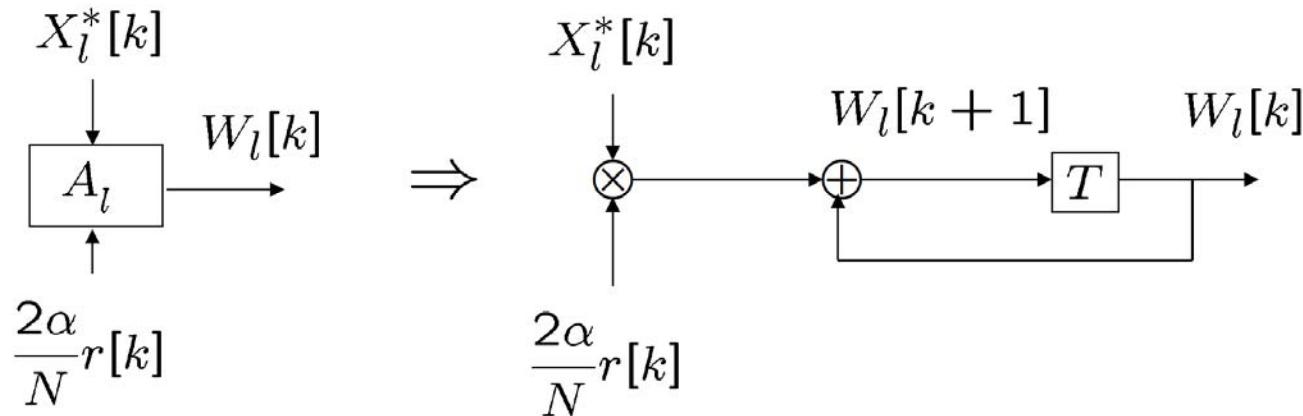
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Apply LMS update in frequency domain:

Multiply both sides by \mathbf{F}^{-1} \Rightarrow

$$\mathbf{F}^{-1}\underline{\mathbf{w}}[k+1] = \mathbf{F}^{-1}\underline{\mathbf{w}}[k] + 2\alpha\mathbf{F}^{-1}\underline{\mathbf{x}}[k]\mathbf{r}[k] \Rightarrow$$

$$\underline{\mathbf{W}}[k+1] = \underline{\mathbf{W}}[k] + \frac{2\alpha}{N}\underline{\mathbf{X}}^*[k]\mathbf{r}[k]$$



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Improve convergence properties easily by

Decorrelation by power normalization:

FDAF algorithm:

$$\underline{\mathbf{W}}[k+1] = \underline{\mathbf{W}}[k] + \frac{2\alpha}{N} \mathbf{P}^{-1} \underline{\mathbf{X}}^*[k] r[k]$$

$$\mathbf{P} = \text{diag}\{\underline{\mathbf{P}}\} \text{ with } P_l = \frac{1}{N} E\{|X_l[k]|^2\}$$

In practice (e.g.):

$$\hat{P}_l[k+1] = \beta \hat{P}_l[k] + (1 - \beta) \frac{|X_l[k]|^2}{N} \quad \forall l$$

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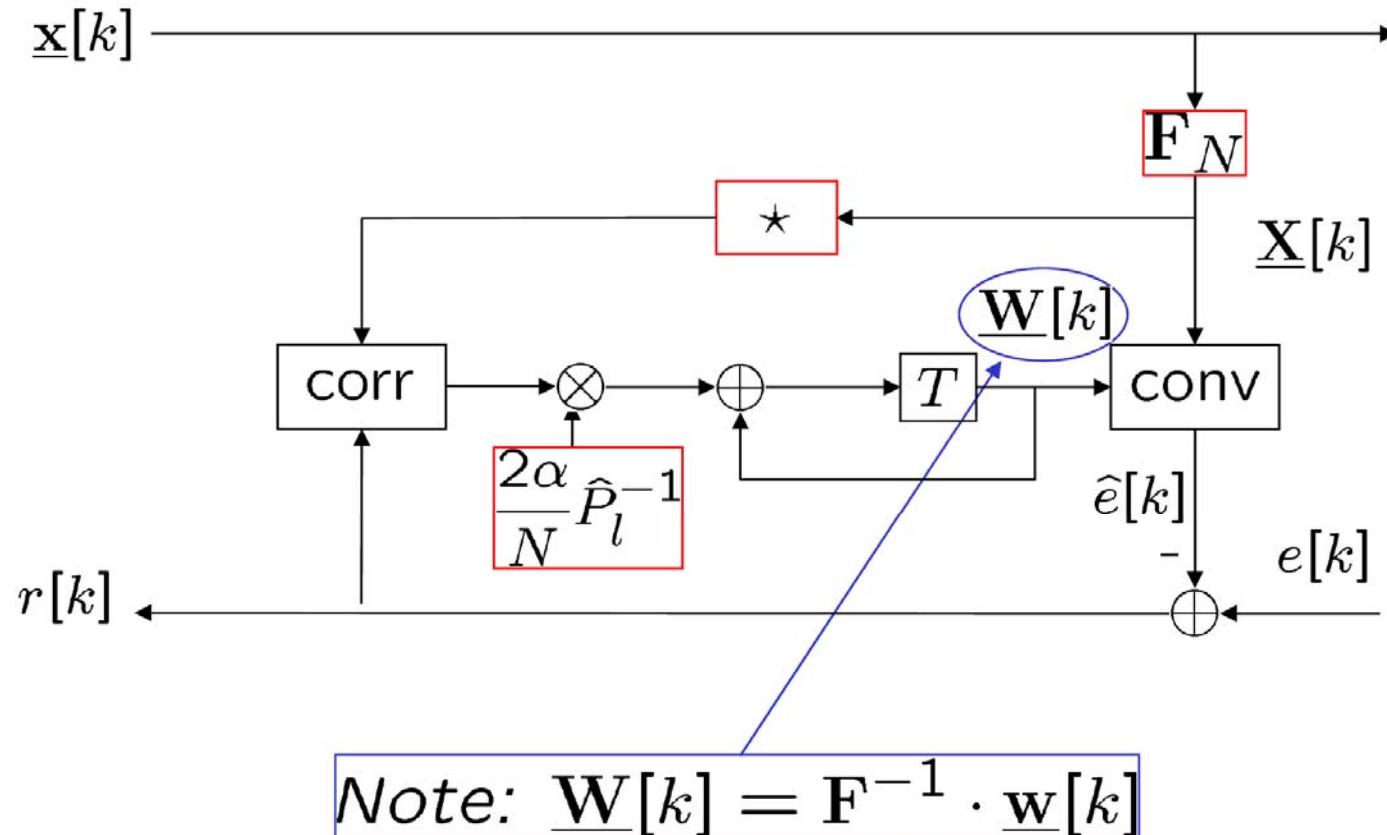


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Simplified realization scheme:



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Average behaviour FDAF:

$$\text{With } \underline{\mathbf{D}} = \mathbf{F}^{-1} \underline{\mathbf{d}} = \mathbf{F}^{-1} (\underline{\mathbf{w}} - \underline{\mathbf{w}}_o) = \underline{\mathbf{W}} - \underline{\mathbf{W}}_o$$

FDAF can be rewritten as:

$$\begin{aligned}\underline{\mathbf{D}}[k+1] &= \left(\mathbf{I} - \frac{2\alpha}{N} \mathbf{P}^{-1} \underline{\mathbf{X}}^*[k] \underline{\mathbf{X}}^t[k] \right) \underline{\mathbf{D}}[k] \\ &\quad + \frac{2\alpha}{N} \mathbf{P}^{-1} \underline{\mathbf{X}}^*[k] r_{min}[k]\end{aligned}$$

Different bins uncorrelated: $\Rightarrow \frac{E\{\underline{\mathbf{X}}^*[k] \underline{\mathbf{X}}^t[k]\}}{N} \approx \mathbf{P}$

Thus $E\{\underline{\mathbf{D}}[k+1]\} \approx (1 - 2\alpha) E\{\underline{\mathbf{D}}[k]\}$

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FDAF

Notes FDAF:

- Both update algorithm and filter are transformed
- FFT (or DFT) is fixed transform --> easy but not exact

Conclusions FDAF:

$$\lim_{k \rightarrow \infty} E\{\underline{\mathbf{D}}[k]\} = \underline{\mathbf{0}} \Leftrightarrow \lim_{k \rightarrow \infty} E\{\underline{\mathbf{W}}[k]\} = \underline{\mathbf{W}}_o = \mathbf{F}^{-1} \underline{\mathbf{w}}_o$$

\mathbf{P}_x^{-1} : update of each bin is power normalized

All weights have (in average) similar convergence!

Equivalent to NLMS with white noise input!



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Part B: Adaptive signal processing

Focus on **single channel** adaptive algorithms
using **FIR** structure

- Applications Adaptive Algorithms
- Minimum Mean Square Error (MMSE)
- Constrained MMSE
- Least Square (LS)
- Steepest Gradient Descent (SGD)
- Three LMS variants: NLMS, Complex LMS, Constrained LMS
- Newton
- Recursive Least Squares (RLS)
- Frequency Domain Adaptive Filter (FDAF)
- **Summary**

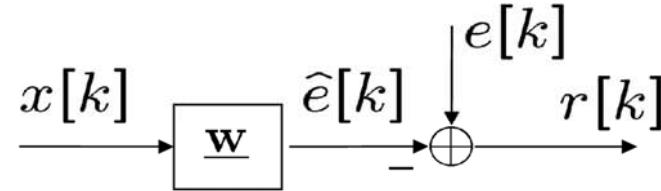
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Summary



	MMSE	LS
Auto correlation	$\underline{\mathbf{R}}_x = E\{\underline{\mathbf{x}}[k] \cdot \underline{\mathbf{x}}^t[k]\}$	$\overline{\mathbf{R}}_x = \mathbf{X}^t \cdot \mathbf{X}$
Cross correlation	$\underline{\mathbf{r}}_{ex} = E\{e[k] \cdot \underline{\mathbf{x}}[k]\}$	$\overline{\mathbf{r}}_{ex} = \mathbf{X}^t \cdot \underline{\mathbf{e}}$
Error J	$E\{r^2[k]\}$	$\sum_{i=0}^{L-1} r^2[k-i]$
Criterion	$\min_{\underline{\mathbf{w}}} \{E\{r^2[k]\}\}$	$\min_{\underline{\mathbf{w}}} \underline{\mathbf{e}} - \mathbf{X} \cdot \underline{\mathbf{w}} ^2$
Opt. solution $\underline{\mathbf{w}}_o$	$\mathbf{R}_x^{-1} \cdot \underline{\mathbf{r}}_{ex}$	$\overline{\mathbf{R}}_x^{-1} \cdot \overline{\mathbf{r}}_{ex}$
Min. error J_{min}	$E\{e^2\} - \underline{\mathbf{r}}_{ex}^t \mathbf{R}_x^{-1} \underline{\mathbf{r}}_{ex}$	$\underline{\mathbf{e}}^t \underline{\mathbf{e}} - \overline{\mathbf{r}}_{ex}^t \overline{\mathbf{R}}_x^{-1} \overline{\mathbf{r}}_{ex}$



Set of constraints: $\mathbf{C}^t \cdot \underline{\mathbf{w}} = \underline{\mathbf{f}}$

Solution for $N \geq M$: $\underline{\mathbf{w}}^c = \mathbf{C} (\mathbf{C}^t \mathbf{C})^{-1} \underline{\mathbf{f}}$

Solution for $N > M$ with MMSE:

$$\underline{\mathbf{w}}_o^c = \underline{\mathbf{w}}_o + \mathbf{R}_x^{-1} \mathbf{C} (\mathbf{C}^t \mathbf{R}_x^{-1} \mathbf{C})^{-1} (\underline{\mathbf{f}} - \mathbf{C}^t \underline{\mathbf{w}}_o)$$

Equivalently: $\underline{\mathbf{w}}_o^c = \mathbf{R}_x^{-1} \mathbf{C} (\mathbf{C}^t \mathbf{R}_x^{-1} \mathbf{C})^{-1} \underline{\mathbf{f}}$

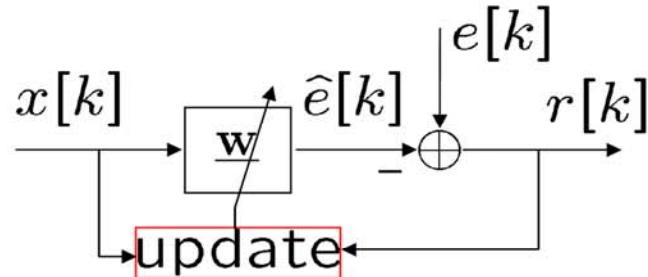
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Summary



Simple adaptive algorithms (no decorrelation):

$$\text{SGD} : \underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] + 2\alpha (\underline{\mathbf{r}}_{ex} - \mathbf{R}_x \underline{\mathbf{w}}[k])$$

$$(\text{Complex})\text{(N)LMS} : \underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] + \frac{2\alpha}{\hat{\sigma}_x^2} \underline{\mathbf{x}}[k] r^*[k]$$

Constrained LMS : $\mathbf{C}^t \cdot \underline{\mathbf{w}} = \underline{\mathbf{f}}$

$$\underline{\mathbf{w}}[k+1] = \tilde{\mathbf{P}} \cdot (\underline{\mathbf{w}}[k] + 2\alpha \underline{\mathbf{x}}[k] r[k]) + \underline{\mathbf{w}}[0]$$

$$\tilde{\mathbf{P}} = \mathbf{I} - \mathbf{C} \left(\mathbf{C}^t \mathbf{C} \right)^{-1} \mathbf{C}^t \text{ and } \underline{\mathbf{w}}[0] = \mathbf{C} \left(\mathbf{C}^t \mathbf{C} \right)^{-1} \underline{\mathbf{f}}$$



Algorithms with improved convergence:

"LMS/Newton" : $\underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] + 2\alpha \mathbf{R}_x^{-1} \cdot \underline{\mathbf{x}}[k]r[k]$

Newton : $\underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] + 2\alpha \mathbf{R}_x^{-1} \cdot (\underline{\mathbf{r}}_{ex} - \mathbf{R}_x \underline{\mathbf{w}}[k])$

RLS : $\underline{\mathbf{w}}[k+1] = \underline{\mathbf{w}}[k] + \underline{\mathbf{g}}[k+1]r[k+1]$

with $\underline{\mathbf{g}}[k+1] = \frac{\bar{\mathbf{R}}^{-1}[k]\underline{\mathbf{x}}[k+1]}{\lambda^2 + \underline{\mathbf{x}}^t[k+1]\bar{\mathbf{R}}^{-1}[k]\underline{\mathbf{x}}[k+1]}$

FDAF : $\underline{\mathbf{W}}[k+1] = \underline{\mathbf{W}}[k] + \frac{2\alpha}{N} \mathbf{P}^{-1} \underline{\mathbf{X}}^*[k]r[k]$

etc.

Part B: Adaptive signal processing



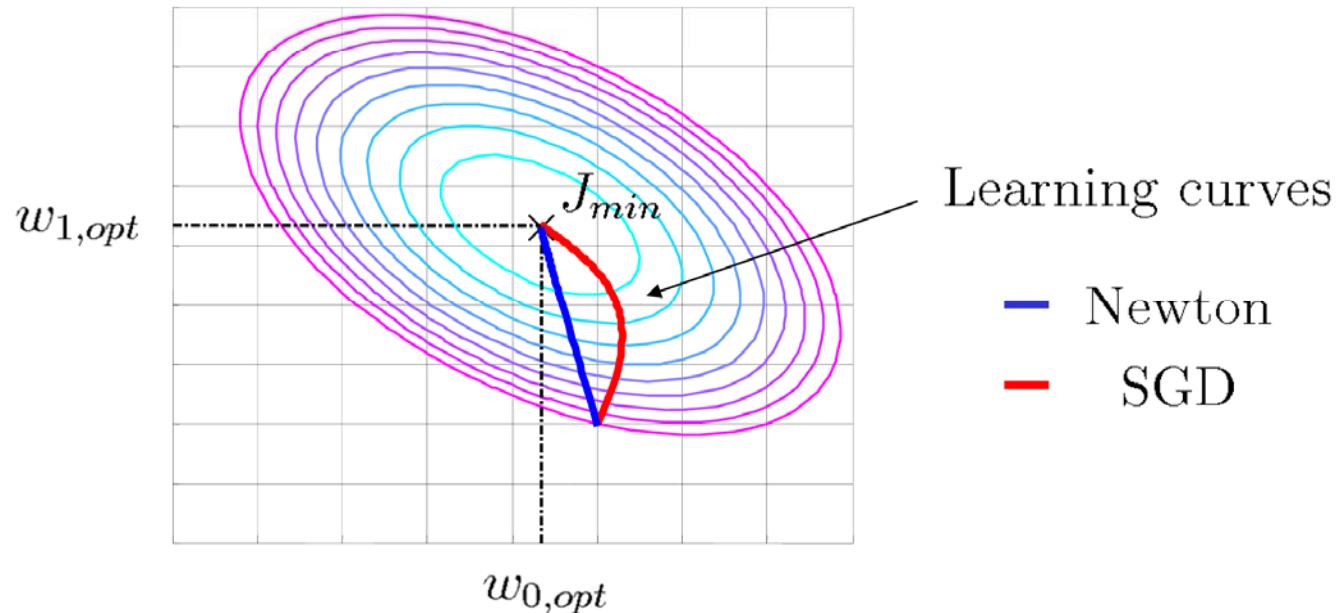
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Summary

Learning curves Newton vs. SGD in contour plot

Coloured input process with $\Gamma_x = \lambda_{max}/\lambda_{min} = 3$



Note: SGD-curve each iteration orthogonal to contourplot J

Newton-curve point each iteration towards J_{min}