Advanced Digital Signal Processing (ADSP)

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ADSP

Advanced digital signal processing

Main content ADSP course

- ➤ Part A: Stochastic Signal Processing
- ➤ Part B: Adaptive Signal Processing
- ➤ Part C: Array Signal Processing (ASP) (including DOA)
- **▶** Part D: Adaptive Array Signal Processing (AASP)



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Contents

- Direction of Arrival (DOA) estimation
- Optimum (data-dependent) beamforming
 - Minimum mean squared error (MMSE)
 - Multiple sidelobe canceller (MSC)
 - Linearly constrained minimum variance (LCMV)
 - Minimum variance distortionless response (MVDR)
 - Generalized sidelobe canceller (GSC)
- Adaptive beamforming

Adaptive versions of the MMSE, MSC, MVDR, LCMV and GSC



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Introduction to DoA estimation

- Estimate direction of arrival of sources (desired and/or interfering) from noisy observations
- Applications

E.g., Tracking and presence detection for smart lighting;

Detecting active talker in an video-conference to automatically steer a video camera;

Surveillance applications, etc.



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Techniques to DoA estimation

- Maximize steered response power
- Using high resolution spectral estimation concepts
- Using time difference of arrival



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Techniques to DoA estimation

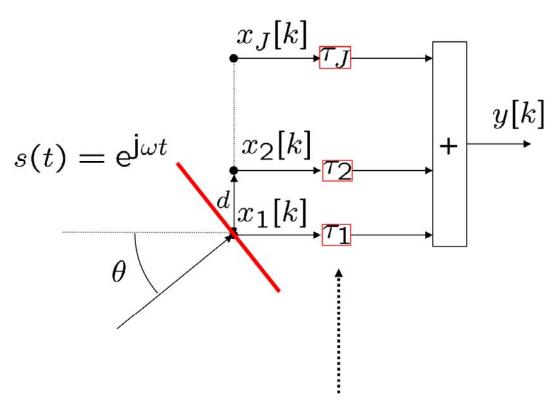
- Maximize steered response power
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Part C: Array signal processing



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Data independent beamforming



For ULA choose: $\tau_i = (i-1)\tau \leftrightarrow w_i^* = \mathrm{e}^{\mathrm{j}(i-1)\omega\tau}$



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(DoA) Maximizing steered response power

- Consider one source located at θ^* and J sensors
- Input signals: $\underline{\mathbf{x}}[k] = \underline{\mathbf{a}}(\theta^*) \cdot s[k] + \underline{\mathbf{n}}[k]$
- Autocorrelation: $\mathbf{R}_{x} = \sigma_{s}^{2} \underline{\mathbf{a}}(\theta^{*}) \underline{\mathbf{a}}^{h}(\theta^{*}) + \sigma_{n}^{2} \mathbf{I}$
- We know that a beamformer steered towards θ^* maximizes the SNR, with weights $\underline{\mathbf{w}} \equiv \underline{\mathbf{a}}(\theta^*)$ (matched filter)

$$P_y = \sigma_s^2 \cdot \underline{\mathbf{w}}^h (\underline{\mathbf{a}}(\theta^*) \cdot \underline{\mathbf{a}}^h(\theta^*)) \underline{\mathbf{w}} + \sigma_n^2 \cdot \underline{\mathbf{w}}^h \underline{\mathbf{w}}$$
$$= P_s + P_n$$

Maximum SNR \longrightarrow Maximum P_y



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(DoA) Maximizing steered response power

- Steer the beamformer to a number of candidate directions θ .
- Output power: $P_{v}(\theta) = E\{|y[k]|^{2}\} = \underline{\mathbf{w}}^{h} \cdot \mathbf{R}_{x} \cdot \underline{\mathbf{w}}$
- Spatial spectrum: $P(\theta) = \frac{P_y(\theta)}{|\mathbf{w}|^2} = \frac{\mathbf{a}^h(\theta)\mathbf{R}_x\mathbf{a}(\theta)}{J}$
- Clearly $P(\theta)$ attains its maximum when $\theta = \theta^*$. Thus, the peak in $P(\theta)$ is the DoA estimate



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(DoA) Maximizing steered response power

ULA, $d/\lambda = 1/2$, P = 3 source, J = 8 sensors

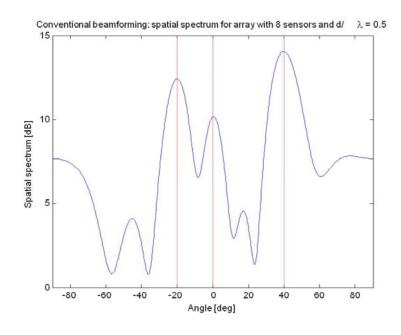


Figure: DoA estimation by steered response power method: sources at -20, 0 and 40 degrees



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(DoA) Maximizing steered response power

- In practice, the search space is discretized depending on the desired accuracy
- Efficient search strategies may be implemented

• The autocorrelation matrix may be computed as $\hat{\mathbf{R}}_{x} = \frac{1}{T} \sum_{t=1}^{T} \underline{\mathbf{x}}[t] \cdot \underline{\mathbf{x}}^{h}[t]$



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Techniques to DoA estimation

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(DoA) MUltiple Signal Classification: MUSIC

DoA based on high-resolution spectral estimation

MUSIC: MUltiple SIgnal Classification

Employs signal subspace method



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- Consider J sensors and P sources, P < J
- $x_i[k] = \sum_{p=1}^{P} a_i[\theta_p] s_p[k] + n_i[k], \quad i = 1 \dots J$
- $\bullet \ \underline{\mathbf{x}}[k] = \mathbf{A}\underline{\mathbf{s}}[k] + \underline{\mathbf{n}}[k]$
- $\bullet \ \mathbf{R}_{x} = E\{\underline{\mathbf{x}}\underline{\mathbf{x}}^{h}\} = \mathbf{A}\mathbf{R}_{s}\mathbf{A}^{h} + \mathbf{R}_{n}$
- $\mathbf{R}_s = \operatorname{diag}(\sigma_{s_1}^2, \dots, \sigma_{s_P}^2)$
- $\mathbf{AR}_s \mathbf{A}^h$ is a $J \times J$ matrix of rank P, and has J P zero eigenvalues



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- Let $\underline{\mathbf{u}}_i$ be an eigenvector of $\mathbf{AR}_s\mathbf{A}^h$ corresponding to one of the zero eigenvalues
- Then $\mathbf{AR}_s \mathbf{A}^h \mathbf{\underline{u}}_i = 0$
- $\Rightarrow \underline{\mathbf{u}}_{i}^{h} \mathbf{A} \mathbf{R}_{s} \mathbf{A}^{h} \cdot \underline{\mathbf{u}}_{i} = 0 \text{ or } (\mathbf{A}^{h} \underline{\mathbf{u}}_{i})^{h} \mathbf{R}_{s} (\mathbf{A}^{h} \underline{\mathbf{u}}_{i}) = 0$
- $\bullet \Rightarrow \mathbf{A}^h \underline{\mathbf{u}}_i = 0$, as \mathbf{R}_s is positive definite
- So, the eigenvectors corresponding to the *J - P* eigenvalues that are zero are orthogonal to the steering vectors.



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- Let U_n denote the $J \times J P$ matrix containing the J - P eigenvectors corresponding to the eigenvalues that are zero
- Define $P_{SM}(\theta) = \frac{1}{\sum_{i=1}^{J-P} |\underline{\mathbf{a}}^h(\theta)\underline{\mathbf{u}}_i|^2} = \frac{1}{\underline{\mathbf{a}}^h(\theta)\mathbf{U}_n\mathbf{U}_n^h\underline{\mathbf{a}}(\theta)}$
- If θ corresponds to a source direction, then the denominator of $P_{SM}(\theta)$ becomes zero
- So, the P largest peaks of $P_{SM}(\theta)$ provide the source directions. $P_{SM}(\theta)$ is referred to as the pseudo-spectrum



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- The analysis so far involved the EVD of $\mathbf{AR}_s \mathbf{A}^h$, which is not available in practice; only \mathbf{R}_x is. This turns out not to be a problem.
- For any $\underline{\mathbf{u}}_i \in \mathbf{U}_n$, $\mathbf{AR}_s \mathbf{A}^h \underline{\mathbf{u}}_i = \lambda \underline{\mathbf{u}}_i$.
- $\Rightarrow \mathbf{R}_{\times} \underline{\mathbf{u}}_{i} = (\mathbf{A}\mathbf{R}_{s}\mathbf{A}^{h} + \mathbf{R}_{n})\underline{\mathbf{u}}_{i} = \lambda\underline{\mathbf{u}}_{i} + \sigma^{2}\underline{\mathbf{u}}_{i} = (\lambda + \sigma^{2})\underline{\mathbf{u}}_{i}$
- So the eigen vectors of \mathbf{R}_{\times} are also the eigen vectors of $\mathbf{A}\mathbf{R}_{\times}\mathbf{A}^{h}!$ (note that we assumed the noise to be spatially white)



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(DoA) Multiple Signal Classification: MUSIC

- Compute/Estimate \mathbf{R}_{x}
- Perform EVD of \mathbf{R}_{x} ; determine \mathbf{U}_{n} as the matrix containing the eigenvectors corresponding to the J-P smallest eigenvalues
- Evaluate pseudo-spectrum:

$$P_{SM}(\theta) = \frac{1}{\underline{\mathbf{a}}^h(\theta)\mathbf{U}_n\mathbf{U}_n^h\underline{\mathbf{a}}(\theta)}$$

• Locate P sharpest peaks in $P_{SM}(\theta)$



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- The pseudo spectrum exhibits sharp peaks in the vicinity of the true DoAs
- P_{SM} averages J P pseudo spectra of individual noise sources. A large value of J – P results in sharper peaks
- The name pseudo is used because P_{SM} contains no information about power
- In practice, $\hat{\mathbf{R}}_{x} = \frac{1}{T} \sum_{k=1}^{T} \underline{\mathbf{x}}[k]\underline{\mathbf{x}}^{h}[k]$



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(DoA) Multiple Signal Classification: MUSIC

ULA, $d/\lambda = 1/2$, P = 3 sources, J = 4 sensors

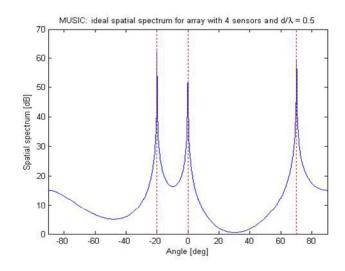


Figure: DoA estimation by Spectral MUSIC: sources at -20, 0 and 70 degrees



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(DoA) Multiple Signal Classification: MUSIC

ULA,
$$d/\lambda = 1/2$$
, $P = 3$ sources, $J = 4$ sensors $P(\theta) = \underline{\mathbf{a}}^h(\theta)\mathbf{R}_{\times}\underline{\mathbf{a}}(\theta)/J$

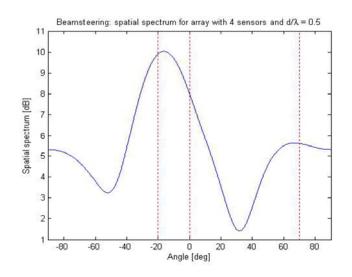


Figure: DoA estimation by beamsteering: sources at -20, 0 and 70 degrees



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(DoA) Multiple Signal Classification: MUSIC

ULA, $d/\lambda = 1/2$, P = 4 sources, J = 10 sensors Larger J - P in this case results in deeper minima

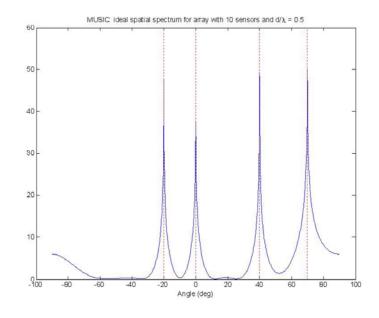


Figure: DoA estimation by Spectral MUSIC: sources at -20, 0, 40 and 70 degrees



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Data dependent beamforming

- The methods discussed so far compute beamformer weights regardless of data being processed
- Alternatively, weights can be based on the statistics of the data, obtained by optimizing a certain criterion



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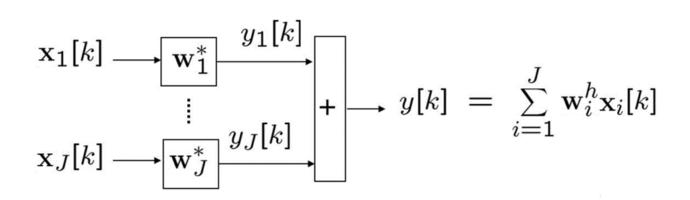
Data dependent beamforming

- Optimum: Assume knowledge of array data statistics. Optimal with respect to a certain optimization criterion (e.g., min MSE, max SINR)
- Adaptive: Estimate the required statistics as data becomes available (when statistics are unknown or time-varying). The derivation is done assuming known statistics. These statistics are estimated on-line.



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Data dependent beamforming: MMSE



Short notation:
$$y[k] = \underline{\mathbf{w}}^h \cdot \underline{\mathbf{x}}[k]$$

$$\underline{\mathbf{x}}[k] = (\mathbf{x}_1[k], \cdots \mathbf{x}_J[k])^t$$

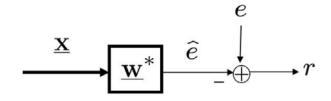
$$\underline{\mathbf{w}} = (\mathbf{w}_1, \cdots, \mathbf{w}_J)^t$$

$$\underline{\mathbf{x}}[k] \longrightarrow \underline{\mathbf{w}}^* \longrightarrow y[k]$$



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Data dependent beamforming: MMSE



- Cost function: $\xi = E\{|r|^2\}$
- Solution: $\underline{\mathbf{w}}_{\mathrm mse} = \mathrm{arg}\, \mathrm{min}_{\underline{\mathbf{w}}}\, \xi = \mathbf{R}_{\mathrm x}^{-1} \cdot \underline{\mathbf{r}}_{\mathrm xe^*}$

$$\mathbf{R}_{x} = E\{\underline{\mathbf{x}} \cdot \underline{\mathbf{x}}^{h}\} \qquad \underline{\mathbf{r}}_{xe^{*}} = E\{\underline{\mathbf{x}} \cdot e^{*}\}$$

• Need to know \mathbf{R}_{x} and $\underline{\mathbf{r}}_{xe^{*}}$ (from measurements)



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Data dependent beamforming: MMSE

Example: ULA, one source, narrowband, farfield

With
$$\underline{\mathbf{x}} = \underline{\mathbf{a}} \cdot s + \underline{\mathbf{n}}$$
 and $e = s$

$$\Rightarrow$$

$$\mathbf{R}_x = E\{\underline{\mathbf{x}} \cdot \underline{\mathbf{x}}^h\} = (\underline{\mathbf{a}} \cdot \underline{\mathbf{a}}^h) \cdot \sigma_s^2 + \sigma_n^2 \cdot \mathbf{I}$$
 and

$$\underline{\mathbf{r}}_{xe^*} = E\{\underline{\mathbf{x}} \cdot e^*\} = \underline{\mathbf{a}} \cdot \sigma_s^2$$

 \Rightarrow

$$\mathbf{\underline{w}}_{mse} = \mathbf{R}_{x}^{-1} \cdot \mathbf{\underline{r}}_{xe^{*}} = \left(\mathbf{\underline{a}} \cdot \mathbf{\underline{a}}^{h} \cdot \sigma_{s}^{2} + \sigma_{n}^{2} \cdot \mathbf{I}\right)^{-1} \cdot \mathbf{\underline{a}} \cdot \sigma_{s}^{2}$$