

Electrical Engineering
IIT Madras



EE 3101 Digital Signal Processing

BS Electronic Systems programme

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Session # 12

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EE3101 Digital Signal ProcessingEE3101

Session 12

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Session 12

Outline

Last session

- DTFT Symmetry Props
- DFT Theorems

Week 5-6

O&S ch2 Section 7-9

Discrete-Time Fourier Transform (DTFT): Complex exponentials as eigensignals of LTI systems—DTFT definition—inversion formula—properties—relationship to continuous-time Fourier series (CTFS)

✓ O&S ch2 Sec 7 Representation of Sequences by Fourier Transforms

✓ ch2 Sec 8 Symmetry Properties of Fourier Transform

✓ ch2 Sec 9 Fourier Transform Theorems

Properties
&
ApplicationsReading Assignment

O&S ch2 Sec 7, Sec 8 and Sec 9

Discrete Time Fourier Transform (DTFT)

$$\begin{aligned} \text{DT seq } x[n] &\xrightarrow[\text{DTFT}]{\mathcal{F}} X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \\ x[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega \xleftarrow[\text{Inv DTFT}]{\mathcal{F}^{-1}} X(e^{j\omega}) \end{aligned}$$

Condition for existence of DTFT

$$\text{If } \sum_{n=-\infty}^{\infty} |x[n]| < \infty \Rightarrow |X(e^{j\omega})| < \infty \quad \forall \omega$$

Absolute Summability

DTFT exists & Convergence is UNIFORM

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$X_M(e^{j\omega}) = \sum_{n=-M}^M x[n] e^{-j\omega n}$$

$$\text{uniform convergence } \lim_{M \rightarrow \infty} X_M(e^{j\omega}) = X(e^{j\omega})$$

Special cases

1. Sequences not absolutely summable, but square summable

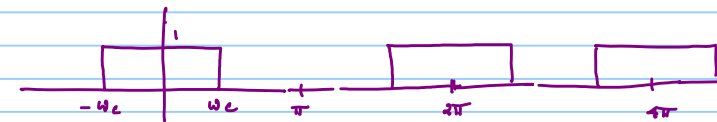
$$\sum_{n=-\infty}^{\infty} |x[n]|^2 < \infty \quad \sum_{n=-\infty}^{\infty} |x[n]| \rightarrow \infty \quad \text{DTFT converges in mean square sense}$$

Mean Square Convergence $X_M(e^{j\omega}) = \sum_{n=-M}^M x[n] e^{-j\omega n}$

$$\lim_{M \rightarrow \infty} \int_{-\pi}^{\pi} |X(e^{j\omega}) - X_M(e^{j\omega})|^2 d\omega \rightarrow 0$$

$$X(e^{j\omega}) = \text{rect}\left(\frac{\omega}{2\omega_c}\right) = \begin{cases} 1 & |\omega| \leq \omega_c \\ 0 & \omega_c < |\omega| \leq \pi \end{cases}$$

Ideal LPF



$$\frac{\omega_c}{\pi} \text{sinc}\left(\frac{\omega_c n}{\pi}\right) \longleftrightarrow \sum_{m=-\infty}^{\infty} \text{rect}\left(\frac{\omega - 2\pi m}{2\omega_c}\right)$$

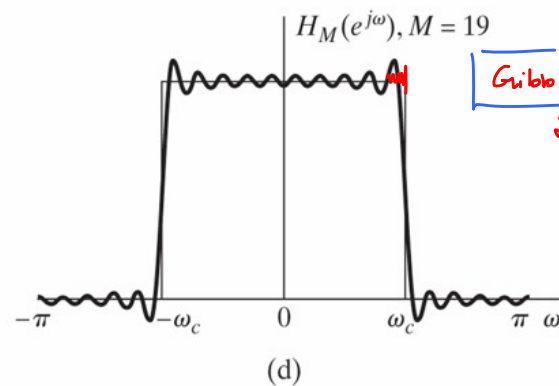
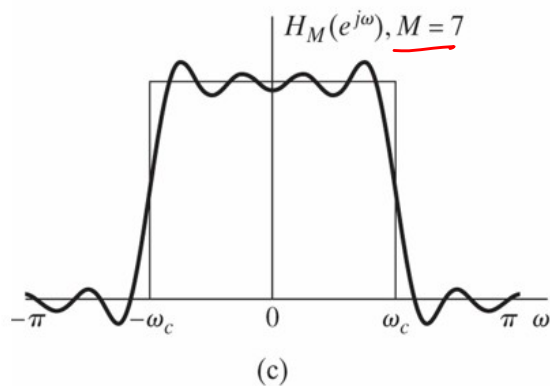
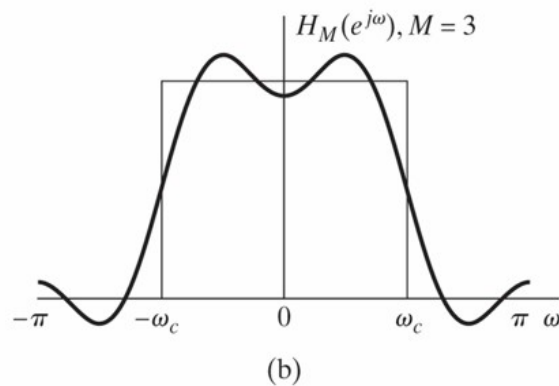
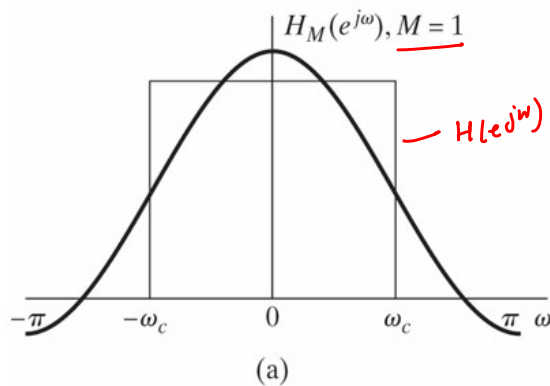
2. DTFT can exist for some sequences that are not absolutely summable and not square summable

Examples

$$\begin{cases} x[n] = 1 \quad \forall n & \longleftrightarrow & X(e^{j\omega}) = \sum_{k=-\infty}^{\infty} 2\pi \delta(\omega + 2\pi k) \\ x[n] = e^{j\omega_0 n} & \longleftrightarrow & X(e^{j\omega}) = \sum_{k=-\infty}^{\infty} 2\pi \delta(\omega - \omega_0 + 2\pi k) \end{cases}$$

D&S

Fig 2.1



Gibbs phenomenon

$\lim_{M \rightarrow \infty}$

Convergence in Mean-Square Sense

- ① Abrupt transition (discontinuity) cannot be suppressed via uniform convergence
- ② Convergence in Mean Square

$$H_M(e^{j\omega}) = \sum_{n=-M}^M h[n] e^{-j\omega n}$$

$$h[n] = \begin{cases} \frac{\omega_c}{\pi} \text{sinc}\left(\frac{\omega_c n}{\pi}\right) & n \neq 0 \\ \frac{\omega_c}{\pi} & n = 0 \end{cases}$$

DTFT

TABLE 2.1 SYMMETRY PROPERTIES OF THE FOURIER TRANSFORM

Sequence $x[n]$	Fourier Transform $X(e^{j\omega})$
1. $x^*[n]$	$X^*(e^{-j\omega})$
2. $x^*[-n]$	$X^*(e^{j\omega})$
3. $\text{Re}\{x[n]\}$	$X_e(e^{j\omega})$ (conjugate-symmetric part of $X(e^{j\omega})$)
4. $j\text{Im}\{x[n]\}$	$X_o(e^{j\omega})$ (conjugate-antisymmetric part of $X(e^{j\omega})$)
5. $x_e[n]$ (conjugate-symmetric part of $x[n]$)	$X_R(e^{j\omega}) = \text{Re}\{X(e^{j\omega})\}$
6. $x_o[n]$ (conjugate-antisymmetric part of $x[n]$)	$jX_I(e^{j\omega}) = j\text{Im}\{X(e^{j\omega})\}$
The following properties apply only when $x[n]$ is real:	
7. Any real $x[n]$	$X(e^{j\omega}) = X^*(e^{-j\omega})$ (Fourier transform is conjugate symmetric)
8. Any real $x[n]$	$X_R(e^{j\omega}) = X_R(e^{-j\omega})$ (real part is even)
9. Any real $x[n]$	$X_I(e^{j\omega}) = -X_I(e^{-j\omega})$ (imaginary part is odd)
10. Any real $x[n]$	$ X(e^{j\omega}) = X(e^{-j\omega}) $ (magnitude is even)
11. Any real $x[n]$	$\angle X(e^{j\omega}) = -\angle X(e^{-j\omega})$ (phase is odd)
12. $x_e[n]$ (even part of $x[n]$)	$X_R(e^{j\omega})$
13. $x_o[n]$ (odd part of $x[n]$)	$jX_I(e^{j\omega})$

$$x[n] \xleftrightarrow{\text{DTFT}} X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

① Conjugation $X^*(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x^*[n] e^{j\omega n}$

② Replace $\omega \leftarrow -\omega$ $X^*(e^{-j\omega}) = \sum_{n=-\infty}^{\infty} x^*[n] e^{-j\omega n}$

③ Replace $m = -n$ $X^*(e^{j\omega}) = \sum_{m=-\infty}^{\infty} x^*[-m] e^{-j\omega m}$

Result $x[n] \longleftrightarrow X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$



TABLE 2.2 FOURIER TRANSFORM THEOREMS

Sequence $x[n]$ $y[n]$	Fourier Transform $X(e^{j\omega})$ $Y(e^{j\omega})$
1. $ax[n] + by[n]$	$aX(e^{j\omega}) + bY(e^{j\omega})$
2. $x[n - n_d]$ (n_d an integer)	$e^{-j\omega n_d} X(e^{j\omega})$
3. $e^{j\omega_0 n} x[n]$	$X(e^{j(\omega - \omega_0)})$
4. $x[-n]$	$X(e^{-j\omega})$ $X^*(e^{j\omega})$ if $x[n]$ real.
5. $nx[n]$	$j \frac{dX(e^{j\omega})}{d\omega}$
6. $x[n] * y[n]$	$X(e^{j\omega})Y(e^{j\omega})$
7. $x[n]y[n]$	$\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta})Y(e^{j(\omega - \theta)})d\theta$
Parseval's theorem:	
8. $\sum_{n=-\infty}^{\infty} x[n] ^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) ^2 d\omega$	
9. $\sum_{n=-\infty}^{\infty} x[n]y^*[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})Y^*(e^{j\omega})d\omega$	

Time shift

Freq shift

Time reversal

Differentiation

Convolution in Time

Multiplication in Time

periodic convolution

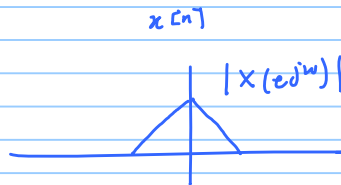
Integrating over one period

Convolution of periodic functions

#3 Freq Shifting

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$e^{j\omega_0 n} x[n] \longleftrightarrow X(e^{j(\omega - \omega_0)})$$



$$x_1[n] = e^{j\omega_0 n} x[n]$$

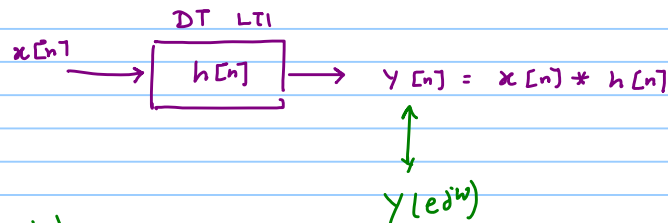


#6 Convolution Theorem

$$x[n] \longleftrightarrow X(e^{j\omega})$$

$$h[n] \longleftrightarrow H(e^{j\omega})$$

$$x[n] * h[n] \longleftrightarrow X(e^{j\omega}) H(e^{j\omega})$$



#8,9 Parseval Theorem

$$E_x = \sum_{n=-\infty}^{\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

$$\sum_{n=-\infty}^{\infty} x[n] y^*[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y^*(e^{j\omega}) d\omega \quad (\text{Generalized Parseval's Theorem})$$

#5 Differentiation

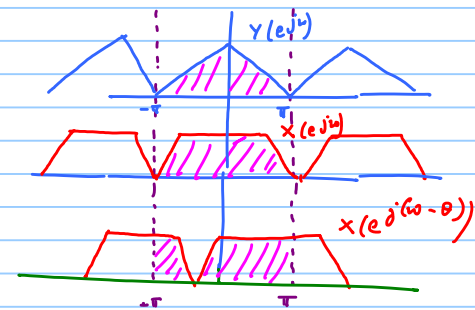
$$x[n] \longleftrightarrow X(e^{j\omega})$$

$$nx[n] \longleftrightarrow j \frac{d}{d\omega} X(e^{j\omega})$$

#7 Multiplication $x[n] \longleftrightarrow X(e^{j\omega})$ $y[n] \longleftrightarrow Y(e^{j\omega})$

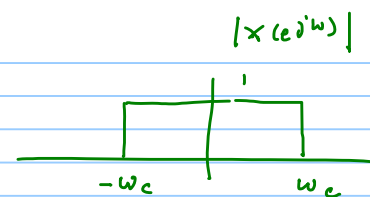
$$w[n] = x[n]y[n] \longleftrightarrow W(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} Y(e^{j\theta}) X(e^{j(\omega-\theta)}) d\theta$$

Periodic Convolution



Ex

$$x[n] = \frac{w_c}{\pi} \operatorname{sinc}\left(\frac{w_c n}{\pi}\right) \longleftrightarrow \sum_{m=-\infty}^{\infty} \operatorname{rect}\left(\frac{\omega - 2\pi m}{2w_c}\right)$$



$$E_x = \sum_{n=-\infty}^{\infty} |x[n]|^2$$

$x[n]$

Parseval's Thm

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

$$E_x = \sum_{n=-\infty}^{\infty} |x[n]|^2 = \frac{w_c}{\pi}$$

$$= \frac{1}{2\pi} \int_{-w_c}^{w_c} 1 \cdot d\omega = \frac{w_c}{\pi}$$

Ex. 1

$$x[n] = 1 \quad \forall n \quad \longleftrightarrow \quad 2\pi \sum_{m=-\infty}^{\infty} \delta(\omega - 2\pi m)$$

Ex 2.

Given $a^n u[n] \longleftrightarrow \frac{1}{1 - ae^{-j\omega}}$

$$|a| < 1$$

$$(-1)^n a^n u[n]$$

$$e^{j\omega_0 n} x[n] \longleftrightarrow X(e^{j(\omega - \omega_0)})$$

Method 2 Freq. shift property

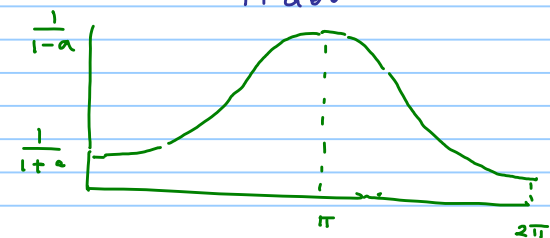
$$(-1)^n = e^{+j\pi n}$$

$$(-1)^n a^n u[n] = \underbrace{e^{j\pi n}}_{(-1)^n} a^n u[n]$$

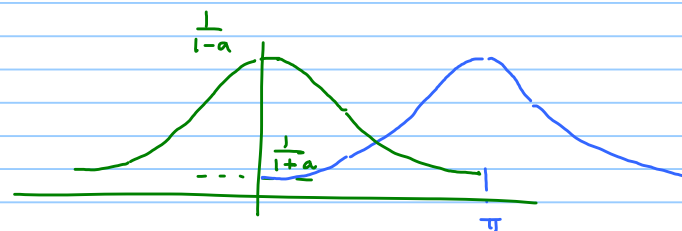
$$e^{j\pi n} a^n u[n] \longleftrightarrow \frac{1}{1 + ae^{-j\omega}}$$

Method 1

$$x_1[n] = (-a)^n u[n] \longleftrightarrow X_1(e^{j\omega}) = \frac{1}{1 + ae^{-j\omega}}$$



High Pass Filter



Ex 3

$$x[n] \longleftrightarrow X(e^{j\omega})$$

$$nx[n] \longleftrightarrow j \frac{d}{d\omega} X(e^{j\omega})$$

$$a^n u[n] \longleftrightarrow \frac{1}{1 - ae^{-j\omega}}$$

$$na^n u[n] \longleftrightarrow j \frac{d}{d\omega} \left(\frac{1}{1 - ae^{-j\omega}} \right) = \frac{ae^{-j\omega}}{(1 - ae^{-j\omega})^2} \quad (1)$$

(3L)

$$x_1[n] \longleftrightarrow \frac{1}{(1 - ae^{-j\omega})^2}$$

Calculation of Inverse DTFT

$$ax_1[n] \longleftrightarrow \frac{a}{(1 - ae^{-j\omega})^2}$$

$$ax_1[n-1] \longleftrightarrow \frac{ae^{-j\omega}}{(1 - ae^{-j\omega})^2} \quad (2)$$

$$(1) = (2)$$

$$\text{LHS } ax_1[n-1] = na^{n-1}u[n]$$

Replace $n \leftarrow n+1$

$$x_1[n] = (n+1)a^n u[n+1]$$

$$(n+1)a^n u[n+1] \longleftrightarrow \frac{1}{(1 - ae^{-j\omega})^2}$$

Ex 4. Property of convolution

$$x_1[n] = x_2[n] = \{1 \quad 1 \quad 1\}$$

$$X_1(e^{j\omega}) = X_2(e^{j\omega}) = e^{j\omega} + 1 + e^{-j\omega}$$

$$x_1[n] * x_2[n] = \{1 \quad 2 \quad 3 \quad 2 \quad 1\}$$

$$x_1[n] * x_2[n] = X_1(e^{j\omega}) X_2(e^{j\omega})$$

$$= (e^{j\omega} + 1 + e^{-j\omega})(e^{j\omega} + 1 + e^{-j\omega})$$

$$= e^{j2\omega} + e^{j\omega} + 1$$

$$e^{j\omega} + 1 + e^{-j\omega}$$

$$+ 1 + e^{-j\omega} + e^{-j2\omega}$$

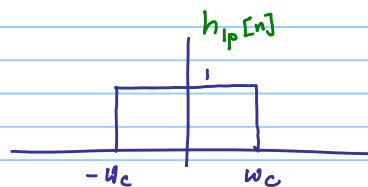
$$\hline e^{j2\omega} + 2e^{j\omega} + 3 + 2e^{-j\omega} + e^{-j2\omega}$$

$$x_1[n] * x_2[n] = \{1, 2, 3, 2, 1\}$$

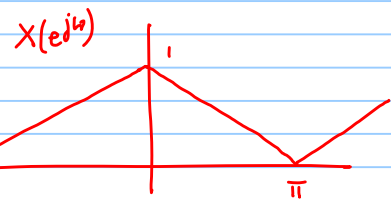
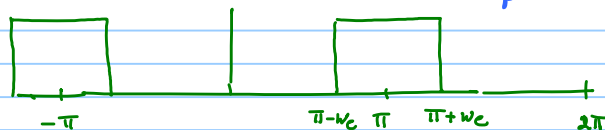
Property of Convolution

Ideal Filters

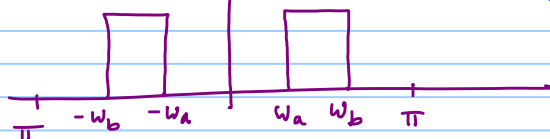
Ideal LPF



Ideal HPF $\leftrightarrow h_{hp}[n]$



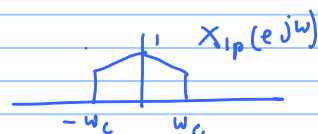
Bandpass filter $\leftrightarrow h_{bp}[n]$



Bandstop filter $\leftrightarrow h_{bs}[n]$



$$x[n] * h_{lp}[n] \leftrightarrow$$



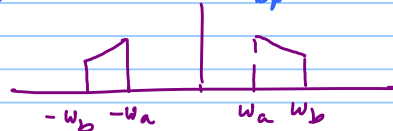
$$x[n] * h_{hp}[n] \leftrightarrow$$

$X_{hp}(e^{j\omega})$



$$x[n] * h_{bp}[n] \leftrightarrow$$

$X_{bp}(e^{j\omega})$

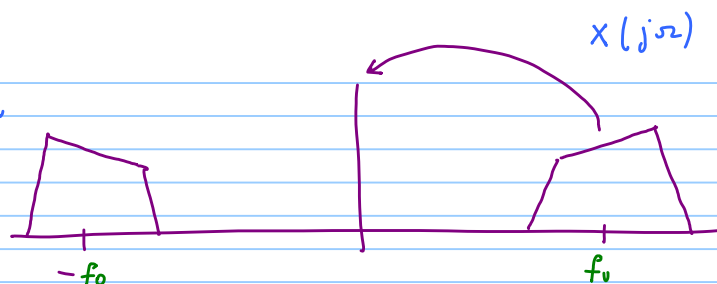


$$x[n] * h_{bs}[n] \leftrightarrow$$



Ex 6

Freq. down conversion



$$\begin{aligned} X_{lp}(e^{j\omega}) &= X(e^{j(\omega + \omega_s)}) \\ &= X(e^{j(\omega + \frac{\pi}{2})}) \end{aligned}$$

$$x_{lp}[n] = \underline{e^{-j\frac{\pi}{2}n}} \uparrow x[n]$$

n	0	1	2	3	4	5	6	7
$e^{-j\frac{\pi}{2}n}$	1	-j	-1	j	1	-j	-1	j

Sample signal

$$\Omega_s = 2\Omega_N \times 2$$

$$T_s = \frac{1}{4f_0}$$

Ex 7 MA filter

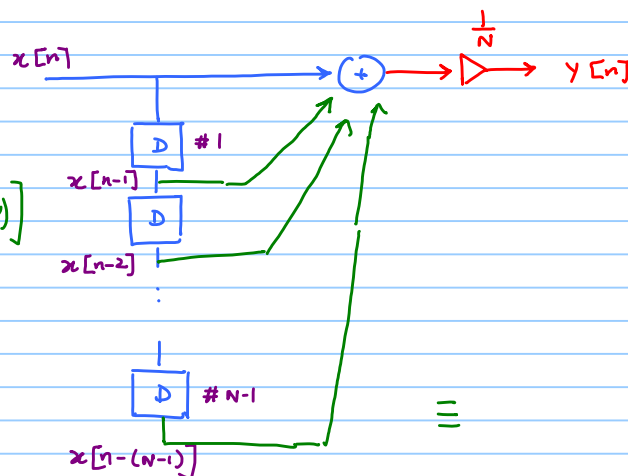
$$y[n] = \frac{1}{N} \sum_{k=0}^{N-1} x[n-k]$$

Apply DTFT

$$Y(e^{j\omega}) = \frac{1}{N} [X(e^{j\omega}) + e^{-j\omega} X(e^{j\omega}) + \dots + e^{-j\omega(N-1)} X(e^{j\omega})]$$

$$= \frac{X(e^{j\omega})}{N} [1 + e^{-j\omega} + \dots + e^{-j\omega(N-1)}]$$

$$Y(e^{j\omega}) = \frac{X(e^{j\omega})}{N} \left(\frac{1 - e^{-j\omega N}}{1 - e^{-j\omega}} \right)$$

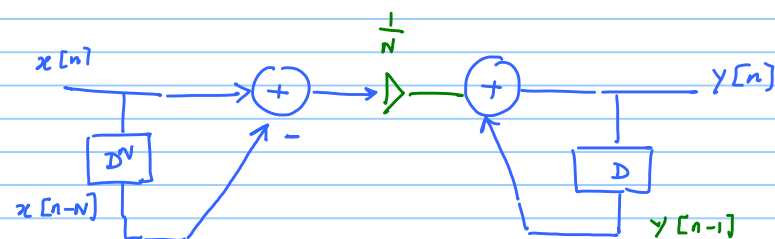


LCCDE

$$Y(e^{j\omega}) (1 - e^{-j\omega}) = \frac{1}{N} X(e^{j\omega}) (1 - e^{-j\omega N})$$

Inv. DTFT

$$y[n] - y[n-1] = \frac{1}{N} (x[n] - x[n-N])$$



Canonical form (# computation)

Ex 7b

LTI system represented by the LCCDE

$$\rightarrow y[n] - \frac{1}{2}y[n-1] = x[n] - \frac{1}{4}x[n-1]$$

Apply DTFT

$$Y(e^{j\omega}) - \frac{1}{2}e^{-j\omega}Y(e^{j\omega}) = X(e^{j\omega}) - \frac{1}{4}e^{-j\omega}X(e^{j\omega})$$

$$Y(e^{j\omega})\left(1 - \frac{1}{2}e^{-j\omega}\right) = X(e^{j\omega})\left(1 - \frac{1}{4}e^{-j\omega}\right)$$

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})} = \frac{1 - \frac{1}{4}e^{-j\omega}}{1 - \frac{1}{2}e^{-j\omega}} = \frac{1}{1 - \frac{1}{2}e^{-j\omega}} - \frac{1}{4} \frac{e^{-j\omega}}{1 - \frac{1}{2}e^{-j\omega}} = H(e^{j\omega})$$

$$h[n] = \left(\frac{1}{2}\right)^n u[n] - \frac{1}{4} \left(\frac{1}{2}\right)^{n-1} u[n-1]$$

Task.

Sketch the circuit for LCCDE

$$y[n] = h[n] * x[n]$$

$$\text{DTFT} \Rightarrow Y(e^{j\omega}) = H(e^{j\omega}) X(e^{j\omega})$$

$$H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})}$$

TABLE 3-3 FOURIER TRANSFORM PAIRS

Sequence	Fourier Transform
✓ 1. $\delta[n]$	1
✓ 2. $\delta[n - n_0]$	$e^{-j\omega n_0}$
✓ 3. 1 $(-\infty < n < \infty)$	$\sum_{k=-\infty}^{\infty} 2\pi \delta(\omega + 2\pi k)$
✓ 4. $a^n u[n] \quad (a < 1)$	$\frac{1}{1 - ae^{-j\omega}}$
5. $u[n]$	$\frac{1}{1 - e^{-j\omega}} + \sum_{k=-\infty}^{\infty} \pi \delta(\omega + 2\pi k)$
→ ✓ 6. $(n+1)a^n u[n] \quad (a < 1)$	$\frac{1}{(1 - ae^{-j\omega})^2}$
7. $\frac{r^n \sin \omega_p (n+1)}{\sin \omega_p} u[n] \quad (r < 1)$	$\frac{1}{1 - 2r \cos \omega_p e^{-j\omega} + r^2 e^{-j2\omega}}$
✓ 8. $\frac{\sin \omega_c n}{\pi n}$	$X(e^{j\omega}) = \begin{cases} 1, & \omega < \omega_c, \\ 0, & \omega_c < \omega \leq \pi \end{cases}$ Ideal LPP
✓ 9. $x[n] = \begin{cases} 1, & 0 \leq n \leq M \\ 0, & \text{otherwise} \end{cases}$	$\frac{\sin[\omega(M+1)/2]}{\sin(\omega/2)} e^{-j\omega M/2}$
✓ 10. $e^{j\omega_0 n}$	$\sum_{k=-\infty}^{\infty} 2\pi \delta(\omega - \omega_0 + 2\pi k)$
✓ 11. $\cos(\omega_0 n + \phi)$	$\sum_{k=-\infty}^{\infty} [\pi e^{j\phi} \delta(\omega - \omega_0 + 2\pi k) + \pi e^{-j\phi} \delta(\omega + \omega_0 + 2\pi k)]$

$$\frac{1}{2\pi} \longleftrightarrow \sum_{k=-\infty}^{\infty} \delta(\omega + 2\pi k)$$

$$\begin{aligned} \# 11 \quad \cos(\omega_0 n + \phi) &= \frac{1}{2} [e^{j(\omega_0 n + \phi)} + e^{-j(\omega_0 n + \phi)}] \\ &= \frac{e^{j\phi}}{2} e^{j\omega_0 n} + \frac{e^{-j\phi}}{2} e^{-j\omega_0 n} \end{aligned}$$

Apply # 10

$$\begin{aligned} \cos(\omega_0 n + \phi) &\longleftrightarrow \pi e^{j\phi} \sum_{k=-\infty}^{\infty} \delta(\omega - \omega_0 + 2\pi k) \\ &\quad + \pi e^{-j\phi} \sum_{k=-\infty}^{\infty} \delta(\omega + \omega_0 + 2\pi k) \end{aligned}$$