

Research Paper Presentation

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Title and Author

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Estimation of Observation Error Probability in Wireless Sensor Networks

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Abstract

- Propose for a parallel wireless sensor network (WSN) model
- A decoding technique that well exploits the knowledge of the sensing data to be transmitted from each sensor to the fusion center (FC).
- Algorithm to estimate the observation error probabilities
- The convergence of the algorithm is also evaluated with comparison of bit-error-rate (BER)

System Model

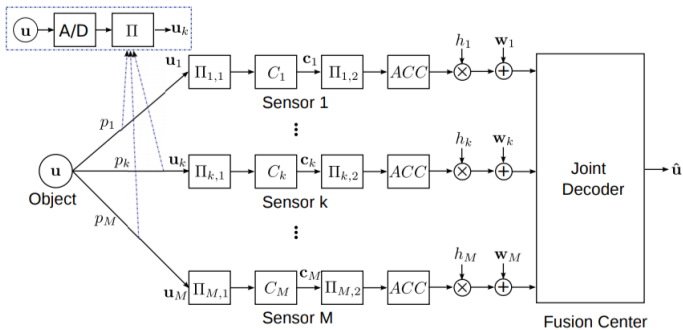


Figure: Structure of proposed system model

System Model

- ① *A/D converter* :
 - ▶ convert analog signal to digital bit sequence
- ② *Interleaver (Π)* :
 - ▶ Intermixes signal thus converting bunch error to random error
- ③ *Channel encoder (C_k)* :
 - ▶ used to detect error and correct some of them
 - ▶ Introduce bit sequence which help in overcoming effect of white noise
- ④ *ACC* :
 - ▶ Doped Accumulator (ACC) takes XOR of signal with some bit sequence
- ⑤ *channel coefficient* :
 - ▶ Scale a signal
- ⑥ *White gaussian noise* :
 - ▶ basic model to mimic effect of random process that occur in nature

System Model

- Error corrupted binary sequence u_k , $k \in \{1, \dots, M\}$ obtained after interleaver Π followed from A/D converter.
- p_k be the bit flipping probability for the sensor k .
- signal u_k is interleaved by $\Pi_{K,1}$
- Then encoded by channel encoder C_k
- encoded bit sequence c_k interleaved by the interleaver $\Pi_{k,2}$ and doped-accumulated by ACC
- then it modulated by BPSK thus gives s_k
- transmitted to the FC over independent static AWGN channels
- finally we get bit sequence y_k

$$y_k = h_k \times s_k + w_k \quad (1)$$

h_k is channel coefficient and w_k is zero mean Gaussian noise

Decoding

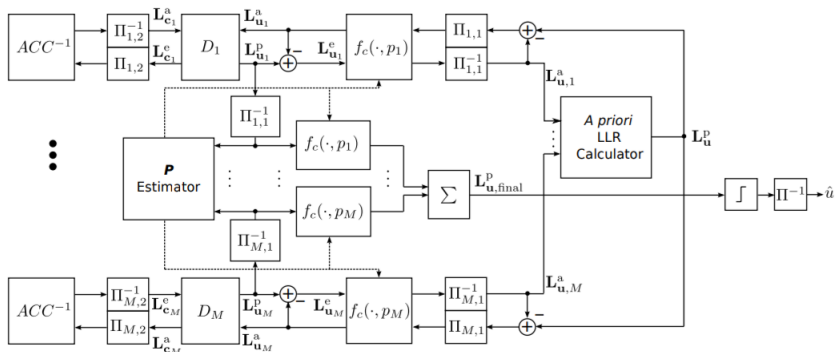


Figure: Proposed decoding strategy for a parallel sensor network

Decoding

Decoding consist of three parts :

① Local iteration

- ▶ In LI extrinsic LLR exchange between decoder D_k and doped accumulator ACC^{-1} takes place
- ▶ takes *priori* LLR L_u^a and gives back *posteriori* LLR L_{uk}^p after performing decoding and ACC^{-1}

② P estimator

- ▶ then a *posteriori* LLR L_{uk}^p containing some error are feed to P estimator to calculate p_k
- ▶ Value of p_k are used by $f_c(\cdot, p_k)$ function

③ Global iteration

- ▶ Extrinsic LLR exchange between Decoders

LI and *GI* are performed until no more relevant gain can be achieved in a final *posteriori* LLR $L_{u,final}^p$

P Estimation

- $j = i + 1$ if $i = 1 \dots M - 1$ and $j = 1$ if $i = M$
- N : number of the LLR pairs with their absolute values larger than a given threshold T

$$\hat{q}_{ij} = \frac{1}{N} \sum_1^N \frac{\exp(L_{ui}^p) + \exp(L_{uj}^p)}{[1 + \exp(L_{ui}^p)] \cdot [1 + \exp(L_{uj}^p)]} \quad (2)$$

Let I be identity matrix of size M and J be defined in (4) of size M

$$\hat{q} = \left[(I + J) - 2 \cdot \text{diag}(\hat{P}) \cdot J \right] \cdot \hat{P} \quad (3)$$

Here $\hat{P} = [\hat{p}_1, \hat{p}_2, \dots, \hat{p}_M]^T$ and $\hat{q} = [\hat{q}_{12}, \hat{q}_{23}, \dots, \hat{q}_{M1}]^T$

The $\text{diag}(\cdot)$ is the operator that forms a diagonal matrix from its argument vector.

P Estimation

$$J = \begin{bmatrix} 0 & 1 & 0 & 0 & \dots 0 \\ 0 & 0 & 1 & 0 & \dots 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ 1 & 0 & 0 & 0 & \dots 0 \end{bmatrix} \quad (4)$$

Objective : to find $\hat{P} \succcurlyeq 0$ which minimize $\|A\hat{P} - \hat{q}\|^2$

where, $A = \left[(I + J) - 2 \cdot \text{diag}(\hat{P}) \cdot J \right]$

We solve this using iterative algorithm summerized in Alogorithm 1. In this algorithm, we use the standard Non negative Least Squares (*lsqnonneg*) algorithm

Algorithm 1: P Estimator

Input: \hat{q} , ϵ , Pre-defined maximum iterations IT_m

Output: $\hat{P} \succcurlyeq 0$ such that $\|A\hat{P} - \hat{q}\|^2$ is minimized

Initialization: $\hat{P}^{(0)} = 0$, Calculate A and $\Delta(0) = \|A\hat{P}^{(0)} - \hat{q}\|^2$;

for $l = 1$ to IT_m **do**

 Calculate $\hat{P}^{(l)}$ by using *lsqnonneg* algorithm;

 Update $A = \left[(I + J) - 2 \cdot \text{diag} \left(\hat{P}^{(l)} \right) \cdot J \right]$;

$\Delta(l) = \|A\hat{P}^{(l)} - \hat{q}\|^2$;

if $\Delta(l) > \Delta(l-1)$ **then**

 Exit For;

end

if $\|\hat{P}^{(l)} - \hat{P}^{(l-1)}\|^2 \leq \epsilon$ **then**

 Exit For;

end

end

$\hat{P} = \hat{P}^{(l)}$;

Significance

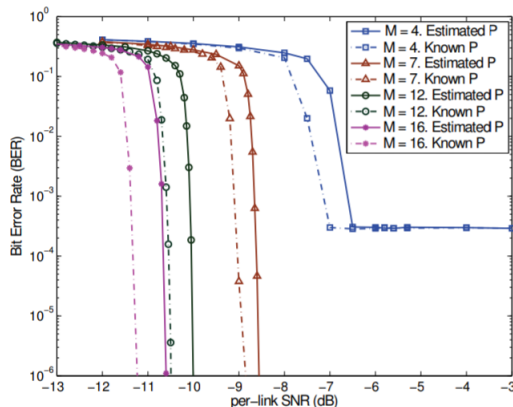


Figure is drawn for

- $T = 2$
- $\epsilon = 10^{-6}$
- Iteration time $IT_M = 20$

using estimated \hat{P} can cause roughly 0.3 to 0.5 dB loss in per-link SNR compared to actual \hat{P}

Our model gives 2 – 3 dB improvement than earlier models