

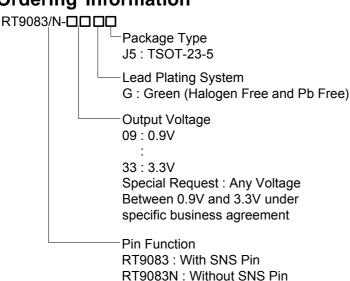
# 30μA I<sub>O</sub>, 250mA Low-Dropout Linear Regulator

## **General Description**

The RT9083 is a low-dropout (LDO) voltage regulators with enable function that operates from 1.2V to 5.5V. It provides up to 250mA of output current and offers low-power operation in miniaturized packaging.

The features of low quiescent current as low as  $30\mu A$  and almost zero disable current is ideal for powering the battery equipment to a longer service life. The RT9083 is stable with the ceramic output capacitor over its wide input range from 1.2V to 5.5V and the entire range of output load current (0mA to 250mA).

### **Ordering Information**



### Note:

Richtek products are:

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

#### **Features**

- 30µA Ground Current
- PSRR = 75dB at 1kHz
- Adjustable Output Voltage Available by Specific Application
- ±2% Output Accuracy
- 250mA (V<sub>IN</sub> ≥ 2.3V) Output Current with EN
- Low (0.1μA) Disable Current
- 1.2V to 5.5V Operating Input Voltage
- Dropout Voltage : 0.45V (typ.) at 250mA when  $V_{OUT} \ge 3V$
- Support Fixed Output Voltage 0.9V, 1.05V, 1.2V, 1.5V, 1.8V, 1.9V, 2.5V, 2.7V, 2.8V, 2.9V, 3V, 3.3V
- Stable with Ceramic or Tantalum Capacitor
- Current Limit Protection
- Over Temperature Protection
- TSOT-23-5 Packages Available

### Applications

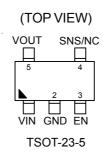
- Portable, Battery Powered Equipment
- Ultra Low Power Microcontrollers
- Notebook Computers

## **Marking Information**

For marking information, contact our sales representative directly or through a Richtek distributor located in your area.



# **Pin Configuration**

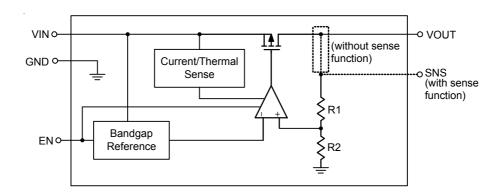


## **Functional Pin Description**

Pin No.	Pin Name	Pin Function			
1	VIN	upply voltage input.			
2	GND	ound.			
3	EN	nable control input.			
SNS Out		Output Voltage Sense.			
4	NC	No internal connection. (RT9083N only)			
5	VOUT	Output of the regulator.			



### **Functional Block Diagram**



## **Operation**

#### **Basic Operation**

The RT9083 is a low quiescent current linear regulator designed especially for low external components system. The input voltage range is from 1.2V to 5.5V.

The minimum required output capacitance for stable operation is 1µF effective capacitance after consideration of the temperature and voltage coefficient of the capacitor.

#### **Output Transistor**

The RT9083 builds in a P-MOSFET output transistor which provides a low switch-on resistance for low dropout voltage applications.

#### **Error Amplifier**

The Error Amplifier compares the internal reference voltage with the output feedback voltage from the internal divider, and controls the Gate voltage of P-MOSFET to support good line regulation and load regulation at output voltage.

#### **Enable**

The RT9083 delivers the output power when it is set to enable state. When it works in disable state, there is no output power and the operation quiescent current is almost zero.

#### **Current Limit Protection**

The RT9083 provides current limit function to prevent the device from damages during over-load or shorted-circuit condition. This current is detected by an internal sensing transistor.

#### **Over Temperature Protection**

The over temperature protection function will turn off the P-MOSFET when the junction temperature exceeds 150°C (typ.),  $V_{IN} \ge 1.5V$  and the output current exceeds 30mA. Once the junction temperature cools down by approximately 20°C, the regulator will automatically resume operation.

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## Absolute Maximum Ratings (Note 1)

<ul> <li>VOUT to VIN6.5V to 0.</li> <li>Power Dissipation, P<sub>D</sub> @ T<sub>A</sub> = 25°C</li> <li>TSOT-23-5 0.43W</li> </ul>	δV
•	3V
TSOT-23-5 0.43W	
1001 200	
Package Thermal Resistance (Note 2)	
TSOT-23-5, $\theta_{JA}$ 230.6°C/W	
• Lead Temperature (Soldering, 10 sec.) 260°C	
• Junction Temperature 150°C	
• Storage Temperature Range –65°C to 15	0°C
ESD Susceptibility (Note 3)	
HBM (Human Body Model) 2kV	
MM (Machine Model) 200V	

## **Recommended Operating Conditions**

• Input Voltage, VIN	- 1.2V to 5.5V
• Junction Temperature Range	40°C to 125°C
• Ambient Temperature Range	40°C to 85°C

### **Electrical Characteristics**

 $(V_{OUT} + 1 < V_{IN} < 5.5V, T_A = 25^{\circ}C$ , unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit		
Output Voltage Range	V <sub>OUT</sub>		0.9		3.3	V		
DC Output Accuracy		I <sub>LOAD</sub> = 1mA	-2		2	%		
	V <sub>DROP</sub>	0.9V ≤ V <sub>OUT</sub> < 1.2V		0.5	0.65			
		1.2V ≤ V <sub>OUT</sub> < 1.5V		0.3	0.4			
Dropout Voltage		1.5V ≤ V <sub>OUT</sub> < 1.8V		0.2	0.24	V		
$(I_{LOAD} = 50mA)$ (Note 5)		1.8V ≤ V <sub>OUT</sub> < 2.5V		0.15	0.18			
		$2.5V \le V_{OUT} < 3V$		0.1	0.15			
		3V ≤ V <sub>OUT</sub>		0.08	0.12			
	VDROP	0.9V ≤ V <sub>OUT</sub> < 1.2V		1.25	1.45			
		1.2V ≤ V <sub>OUT</sub> < 1.5V		1	1.2			
Dropout Voltage		1.5V ≤ V <sub>OUT</sub> < 1.8V		0.81	0.9	V		
$(I_{LOAD} = 250 \text{mA})$ (Note 5)		1.8V ≤ V <sub>OUT</sub> < 2.5V		0.68	0.8			
		$2.5V \le V_{OUT} < 3V$		0.51	0.6			
		3V ≤ V <sub>OUT</sub>		0.45	0.6			
V <sub>CC</sub> Consumption Current	IQ	$I_{LOAD}$ = 20mA, $V_{OUT} \le 5.5V$ $V_{IN} \ge V_{OUT} + V_{DROP}$		30	50	μА		

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Parameter		Symbol	Test Conditions		Min	Тур	Max	Unit	
Shutdown GNE	) Current		V <sub>EN</sub> = 0V			0.1	0.5	μΑ	
Shutdown Leakage Current			V <sub>EN</sub> = 0V, V <sub>OUT</sub> = 0V			0.1	0.5	μΑ	
EN Input Current		I <sub>EN</sub>	V <sub>EN</sub> = 5.5V				0.1	μА	
			10. 1	1.2V	$\leq V_{IN} < 1.5V$			0.6	
				1.5V	$\leq V_{IN} < 1.8V$			0.5	
Line Regulation	1	ΔLINE	$I_{LOAD} = 10mA$	1.8V	$\leq V_{IN} < 2.1V$			0.3	- %
				2.1V	$\leq V_{IN} \leq 5.5V$			0.15	1
Load Regulation	n	ΔLOAD	5mA < I <sub>LOAD</sub> < 250mA				1	%	
Power Supply Rejection Ratio		PSRR	$V_{IN} = 3V$ , $I_{LOAD} = 50$ mA, $C_{OUT} = 1$ µF, $V_{OUT} = 2.5V$ , $f = 1$ kHz			75		dB	
			$ \begin{array}{c} C_{OUT} = 1 \mu F, \\ I_{LOAD} = 30 mA, \\ BW = 10 Hz \ to \\ 100 kHz, \\ V_{IN} = V_{OUT} + 2 V \end{array} \begin{array}{c} V_{OUT} = 0.9 V \\ V_{OUT} = 1.2 V \\ V_{OUT} = 1.8 V \\ V_{OUT} = 3.3 V \end{array} $		V <sub>OUT</sub> = 0.9V		39		μVRMS
					V <sub>OUT</sub> = 1.2V		46		
Output Voltage	Output Voltage Noise				V <sub>OUT</sub> = 1.8V		48		
					V <sub>OUT</sub> = 3.3V		58		1
Output Current Limit		I <sub>LIM</sub>	Peak output current		260	350	500	mA	
Fold-Back Current Limit			V <sub>OUT</sub> = 0.5V x V <sub>OUT(normal)</sub>		150	270	390	mA	
Enable Input Logic-High		VIH	V <sub>IN</sub> = 5V		1.2			\ /	
Voltage	Logic-Low	VIL	V <sub>IN</sub> = 5V					0.4	V
Thermal Shutdown Temperature		T <sub>SD</sub>	I <sub>LOAD</sub> = 30mA, V <sub>IN</sub> ≥ 1.5V			150		°C	
Thermal Shutdown Hysteresis		ΔT <sub>SD</sub>					20		°C

- **Note 1.** Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.  $\theta_{JA}$  is measured at  $T_A$  = 25°C on a high effective thermal conductivity four-layer test board per JEDEC 51-7.
- Note 3. Devices are ESD sensitive. Handling precaution is recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.
- Note 5. The dropout voltage is defined as  $V_{IN} V_{OUT}$ , when  $V_{OUT}$  is 98% of the normal value of  $V_{OUT}$ .



## **Typical Application Circuit**

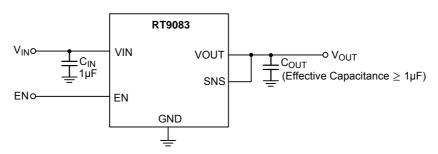


Figure 1. Application with Sense Function

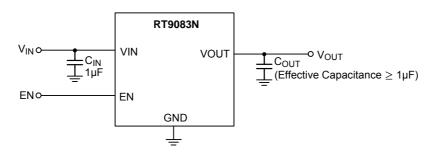


Figure 2. Application without Sense Function

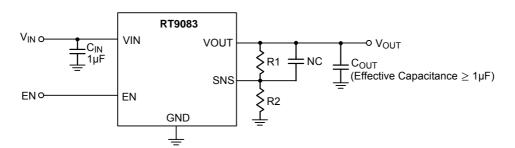
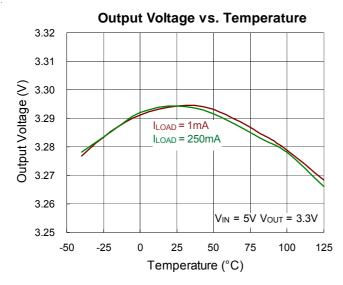
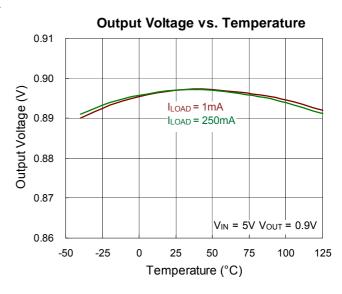


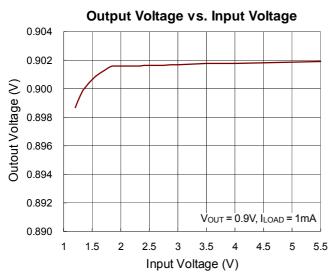
Figure 3. Adjustable Output Voltage Application Circuit

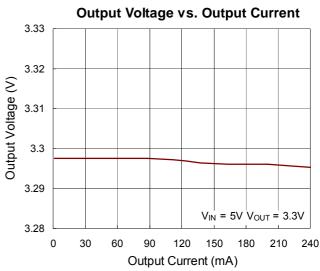


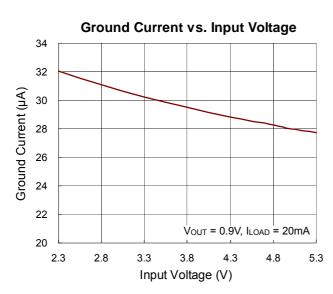
## **Typical Operating Characteristics**

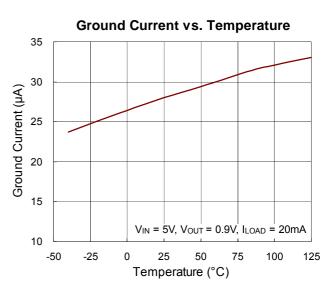






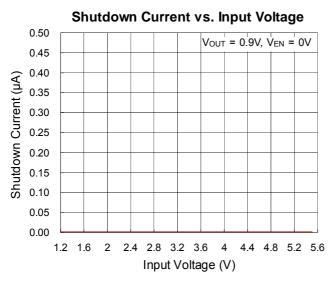


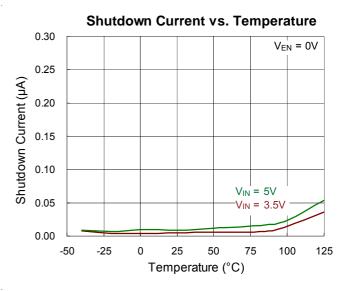


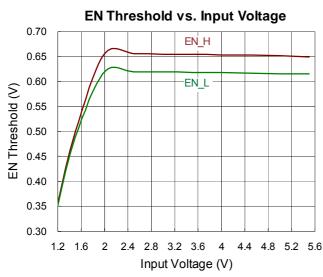


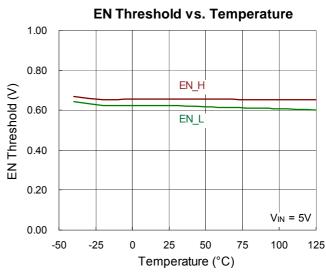
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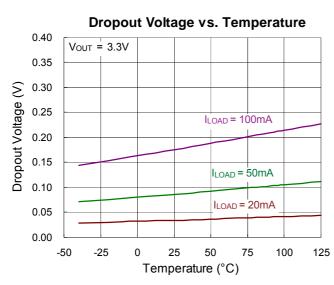


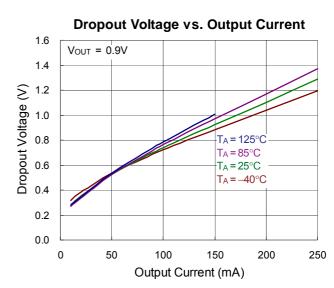






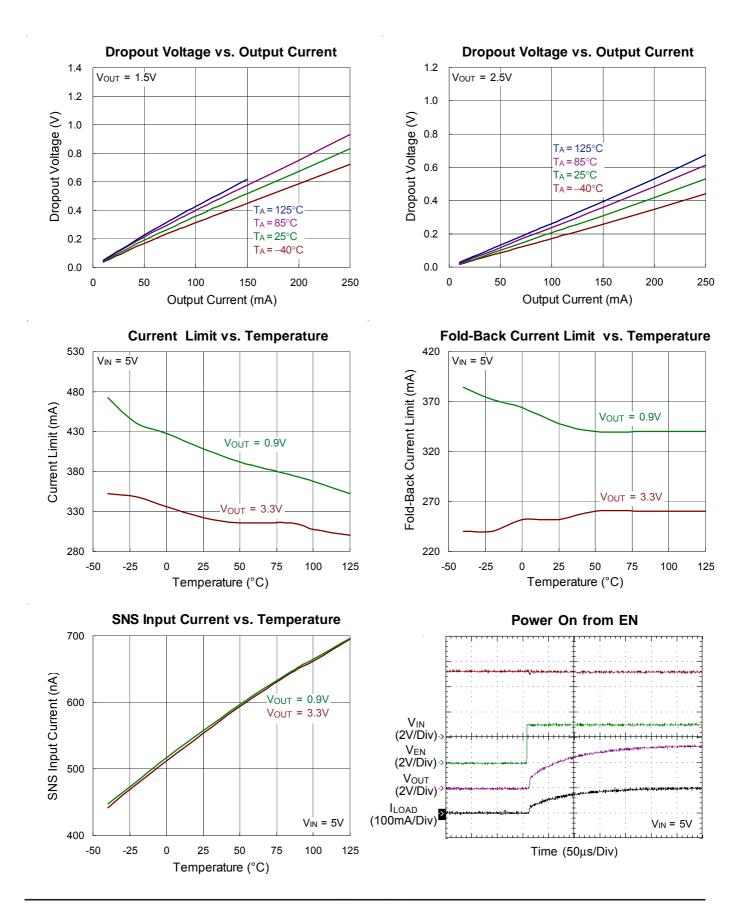






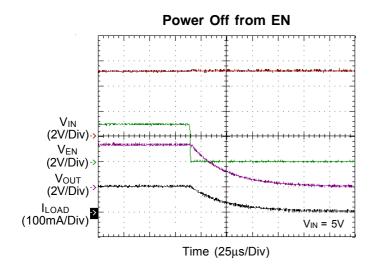
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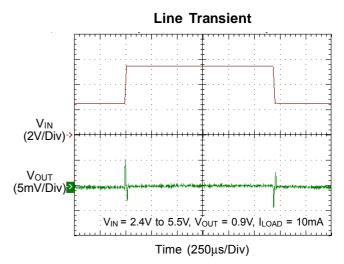


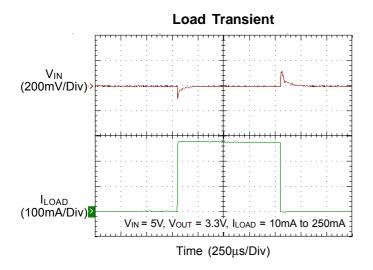


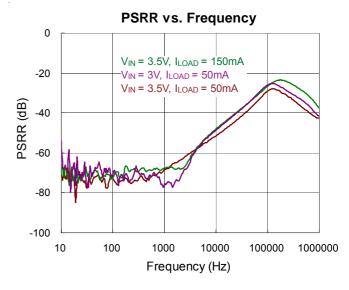
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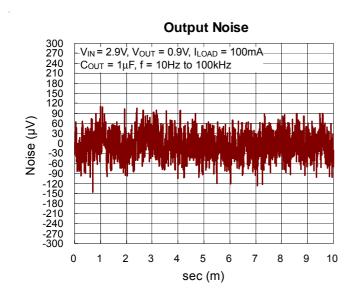


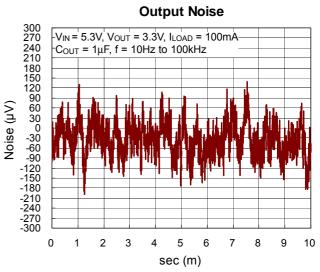












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### **Application Information**

Like any low dropout linear regulator, the RT9083's external input and output capacitors must be properly selected for stability and performance. Use a  $1\mu F$  or larger input capacitor and place it close to the IC's VIN and GND pins. Any output capacitor meeting the minimum  $1m\Omega$ ESR (Equivalent Series Resistance) and effective capacitance larger than 1µF requirement may be used. Place the output capacitor close to the IC's VOUT and GND pins. Increasing capacitance and decreasing ESR can improve the circuit's PSRR and line transient response.

#### **Enable**

The RT9083 has an EN pin to turn on or turn off the regulator, When the EN pin is in logic high, the regulator will be turned on. The shutdown current is almost 0µA typical. The EN pin may be directly tied to V<sub>IN</sub> to keep the part on. The Enable input is CMOS logic and cannot be left floating.

#### **Adjustable Output Voltage Setting**

Because of the small input current at the SNS pin, the RT9083 with SNS pin also can work as an adjustable output voltage LDO. Figure 3 gives the connections for the adjustable output voltage application. The resistor divider from VOUT to SNS sets the output voltage when in regulation.

The voltage on the SNS pin sets the output voltage and is determined by the values of R1 and R2. In order to keep a good temperature coefficient of output voltage, the values of R1 and R2 should be selected carefully to ignore the temperature coefficient of input current at the SNS pin. A current greater than 50µA in the resistor divider is recommended to meet the above requirement. The adjustable output voltage can be calculated using the formula given in equation 1:

$$V_{OUT} = \frac{R1 + R2}{R2} \times V_{SNS} \tag{1}$$

where V<sub>SNS</sub> is determined by the output voltage selections in the ordering information of RT9083.

When we choose  $39k\Omega$  and  $15k\Omega$  as R1 and R2 respectively, and select a 0.9V output at SNS pin, the adjustable output voltage will be set to around 3.24V. Its temperature coefficient in Figure 4 is still perfect in such kind of application.

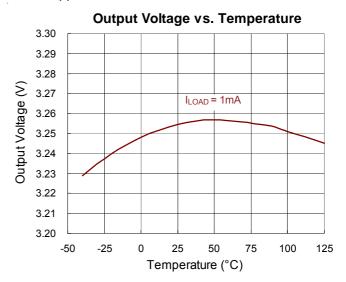


Figure 4. Temperature Coefficient of Adjustable Output Voltage

The minimum recommended 50µA in the resistor divider makes the application no longer an ultra low guiescent LDO. Figure 5 is another fine adjustable output voltage application can keep the LDO still operating in low power consumption. The fine fune range is recommended to be less than 50mV (R1  $\leq$  91k $\Omega$ ) in order to keep a good temperature coefficient of the output voltage.

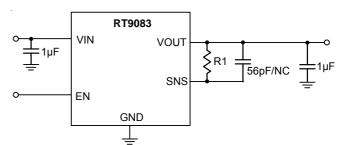


Figure 5. Fine Adjustable Output Voltage Application Circuit

There isn't extra current consumption in the above application. But the temperature coefficient of output voltage will be degraded by the input current at SNS pin. If the tuning range is larger than 50mV, a compensation capacitor (56pF) is required to keep the stability of output voltage. The fine adjustable output voltage is calculated using the formula given in equation2:

$$V_{OUT} = V_{SNS} + I_{SNS} \times R1$$
 (2)

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where I<sub>SNS</sub> is the input Current at SNS pin (typical 550nA at room temperature) and VSNS is determined by the output voltage selections in the ordering information of RT9083.

#### **Current Limit**

The RT9083 contains an independent current limiter, which monitors and controls the pass transistor's gate voltage, limiting the output current to 0.35A (typ.). The current limiting level is reduced to around 250mA named fold-back current limit when the output voltage is further decreased. The output can be shorted to ground indefinitely without damaging the part.

#### **Thermal Considerations**

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications the maximum junction temperature is 125°C and TA is the ambient temperature. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For TSOT-23-5 package, the thermal resistance, θ<sub>JA</sub>, is 230.6°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at  $T_A = 25^{\circ}C$  can be calculated by the following formula:

$$P_{D(MAX)}$$
 = (125°C - 25°C) / (230.6°C/W) = 0.43W for TSOT-23-5 package

The maximum power dissipation depends on the operating ambient temperature for fixed T<sub>J(MAX)</sub> and thermal resistance,  $\theta_{JA}$ . The derating curve in Figure 6 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

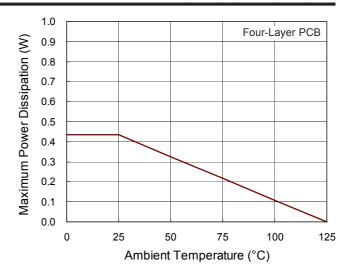
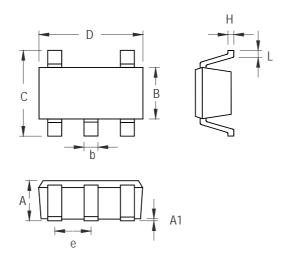


Figure 6. Derating Curve of Maximum Power Dissipation



### **Outline Dimension**



Cumb al	Dimensions	n Millimeters	Dimensions In Inches			
Symbol	Min	Max	Min	Max		
А	0.700	1.000	0.028	0.039		
A1	0.000	0.100	0.000	0.004		
В	1.397	1.803	0.055	0.071		
b	0.300	0.559	0.012	0.022		
С	2.591	3.000	0.102	0.118		
D	2.692	3.099	0.106	0.122		
е	0.838	1.041	0.033	0.041		
Н	0.080	0.254	0.003	0.010		
L	0.300	0.610	0.012	0.024		

**TSOT-23-5 Surface Mount Package** 

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