# Ve460 Control Systems Analysis and Design Chapter 3 Block Diagram and Signal Flow Graphs

Jun Zhang

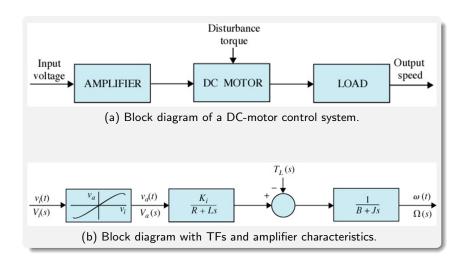
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### 3-1 Block diagrams

- Used to model all types of system: plant and controller;
- Describe cause-and-effect relationship;
- Describe composition and interconnection.







### 3-1-1 Block Diagrams of Control Systems

### Sensing devices: addition & subtraction

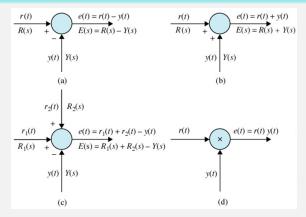


Figure 3.1: Block diagram element of typical sensing devices of control systems. (a) Subtraction; (b) Addition; (c) Addition and Subtraction; (d) Multiplication.



### A feedback control system (an LTI system!)

r: reference input

y: output (controlled variable)

b: feedback signal

u: actuating signal

H(s): feedback TF

G(s)H(s) = L(s): Loop TF

G(s): forward-path TF

M(s) = Y(s)/R(s): closed-loop TF or system TF



# How to calculate M(s)?

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$$Y(s) = G(s)(R(s) - B(s))$$
  
=  $G(s)R(s) - G(s)H(s)Y(s)$ 

$$\Rightarrow (1 + G(s)H(s))Y(s) = G(s)R(s)$$

$$M(s) = \frac{Y(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)}$$



### 3-1-2 Multivariable System

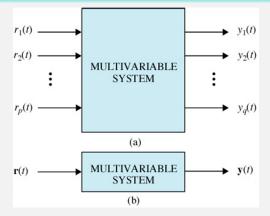


Figure 3.2: Block diagram representations of a multivariable system



### A Multivariable Feedback Control System

$$\begin{array}{c|c}
R(s) & \downarrow s & U(s) \\
r(t) & \downarrow s & U(s)
\end{array}$$

$$\begin{array}{c|c}
G(s) & \downarrow s & Y(s) \\
\hline
Y(s) & = G(s)U(s) \\
U(s) & = R(s) - B(s) \\
B(s) & = H(s)Y(s)
\end{array}$$
Transfer Matrices

Therefore

$$Y(s) = G(s)R(s) - G(s)B(s)$$
  
=  $G(s)R(s) - G(s)H(s)Y(s)$ ,

and

$$(I + G(s)H(s))Y(s) = G(s)R(s)$$
  
 $M(s) = (I + G(s)H(s))^{-1}G(s).$ 



### Example 1

$$G(s) = \begin{bmatrix} \frac{1}{s+1} & -\frac{1}{s} \\ 2 & \frac{1}{s+2} \end{bmatrix}, \qquad H(s) = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix},$$

$$I + G(s)H(s) = \begin{bmatrix} \frac{s+2}{s+1} & -\frac{1}{s} \\ 2 & \frac{s+3}{s+2} \end{bmatrix}.$$

Note that

$$\begin{bmatrix} a & b \\ c & d \end{bmatrix}^{-1} = \frac{1}{ad - bc} \begin{bmatrix} d & -b \\ -c & a \end{bmatrix},$$

then

$$(I+G(s)H(s))^{-1}=rac{1}{\Delta}egin{bmatrix} rac{s+3}{s+2} & rac{1}{s} \ -2 & rac{s+2}{s+1} \end{bmatrix},$$



where

$$\Delta = \frac{s+2}{s+1} \cdot \frac{s+3}{s+2} + \frac{2}{s} = \frac{s^2+5s+2}{s(s+1)}.$$

Therefore

$$M(s) = (I + G(s)H(s))^{-1}G(s)$$

$$= \frac{s(s+1)}{s^2 + 5s + 2} \begin{bmatrix} \frac{s+3}{s+2} & \frac{1}{s} \\ -2 & \frac{s+2}{s+1} \end{bmatrix} \begin{bmatrix} \frac{1}{s+1} & \frac{1}{s} \\ 2 & \frac{2}{s} \end{bmatrix}$$

$$= \frac{s(s+1)}{s^2 + 5s + 2} \begin{bmatrix} \frac{3s^2 + 9s + 4}{s(s+2)(s+1)} & -\frac{1}{s} \\ 2 & \frac{3s+2}{s(s+1)} \end{bmatrix}.$$



### Matlab Control System Toolbox

#### We first use tf to create transfer function models.

tf

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Transfer function model

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#### Description

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Use tf to create real-valued or complex-valued transfer function models, or to convert dynamic system models to transfer function form.

Transfer functions are a frequency-domain representation of linear time-invariant systems. For instance, consider a continuous-time SISO dynamic system represented by the transfer function sys(s) = N(s)/D(s), where s = jw and N(s) and D(s) are called the numerator and denominator polynomials, respectively. The tf model object can represent SISO or MIMO transfer functions in continuous time or discrete time.

You can create a transfer function model object either by specifying its coefficients directly, or by converting a model of another type (such as a state-space model ss) to transfer-function form. For more information, see Transfer Functions.

You can also use  $\mathsf{tf}$  to create generalized state-space (genss) models or uncertain state-space (uss) models.

#### Creation

#### Syntax

```
sys = tf(numerator,denominator)
```

- sys = tf(numerator,denominator,ts)
- sys = tf(numerator,denominator,ltisys)



#### Create a Matlab transfer function model for:

$$G(s) = \frac{s}{s^2 + s + 1}.$$

```
>> num=[1 0];
>> den=[1 1 1];
>> Gs=tf(num, den)
Gs =
  s^2 + s + 1
Continuous-time transfer function.
```

### Alternatively, we can use rational expression

```
>> s = tf('s');
>> Gs1=s/(s^2+s+1)
Gs1 =
  s^2 + s + 1
Continuous-time transfer function.
```

### Consider Example 1 again:

$$G(s) = \begin{bmatrix} rac{1}{s+1} & -rac{1}{s} \\ 2 & rac{1}{s+2} \end{bmatrix}, \qquad H(s) = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}.$$

```
>> Gnum={1 -1; 2 1}; Gden={[1 1], [1 0]; 1 [1 2]};
>> Gs2=tf(Gnum, Gden)
```

Gs2 =

From input 1 to output...

s + 1 2:

From input 2 to output...

2:

s + 2Continuous-time transfer function.



### We now use feedback in Matlab Control System Toolbox.

#### feedback

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Feedback connection of multiple models

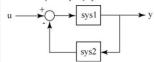
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#### Syntax

```
sys = feedback(sys1,sys2)
sys = feedback(sys1,sys2,feedin,feedout)
sys = feedback(sys1,sys2, 'name')
sys = feedback(____,sign)
```

#### Description

sys = feedback(sys1,sys2) returns a model object sys for the negative feedback interconnection of model objects sys1,sys2.



From the figure, the dosed-loop model sys has u as input vector and y as output vector. Both models, sys1 and sys2, must either be continuous or discrete with identical sample times.



sys =

From input 1 to output...

$$3 s^2 + 9 s + 4$$

$$s^3 + 7 s^2 + 12 s + 4$$

From input 2 to output...  $-s^2 - 3 s - 2$ 

Continuous-time transfer function.

From earlier computation:

$$\frac{s(s+1)}{s^2 + 5s + 2}$$

$$\cdot \begin{bmatrix} \frac{3s^2 + 9s + 4}{s(s+2)(s+1)} & -\frac{1}{s} \\ 2 & \frac{3s+2}{s(s+1)} \end{bmatrix}$$



$$\begin{bmatrix} a & b \\ c & d \end{bmatrix}^{-1} = \frac{1}{ad - bc} \begin{bmatrix} d & -b \\ -c & a \end{bmatrix}$$

In Matlab, we can calculate it as

```
>> syms a b c d
>> inv([a b; c d])
ans =
[ d/(a*d - b*c), -b/(a*d - b*c)]
[-c/(a*d - b*c), a/(a*d - b*c)]
>> pretty(ans)
 ad-bc ad-bc
  ad-bc ad-bc/
```

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$$G(s) = \begin{bmatrix} rac{1}{s+1} & -rac{1}{s} \\ 2 & rac{1}{s+2} \end{bmatrix}, \qquad H(s) = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

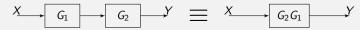
```
>> syms s;
>> Gs=[1/(s+1) -1/s; 2 1/(s+2)];
>> Hs=eye(2);
>> simplify( inv(eye(2)+Gs*Hs)*Gs )
ans =
[(3*s^2 + 9*s + 4)/(s^3 + 7*s^2 + 12*s + 4), -(s + 1)/(s^2 + 5*s + 2)]
              (2*s*(s + 1))/(s^2 + 5*s + 2), (3*s + 2)/(s^2 + 5*s + 2)]
>> pretty(ans)
 3 s + 9 s + 4 s + 1
 3 2 2 2
s + 7 s + 12 s + 4 s + 5 s + 2
      2 s (s + 1) 3 s + 2
     -----, 2 2 1
s + 5 s + 2 s + 5 s + 2 /
```



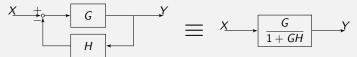
### Simplification of Block Diagram

### Simple cases

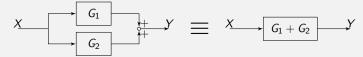
a) Combining blocks in series (Matlab command series)



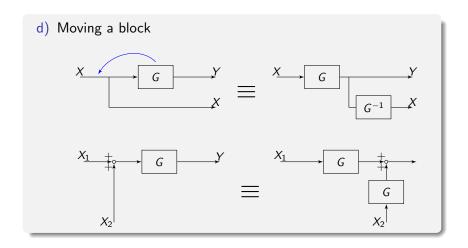
b) Elimination of feedback loop (Matlab command feedback)



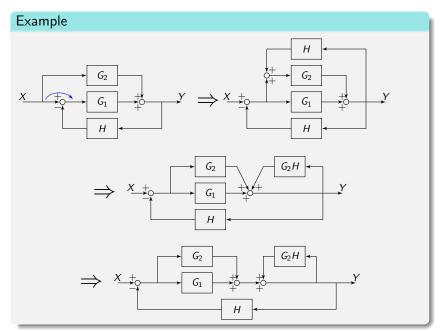
c) Combining blocks in parallel (Matlab command parallel)













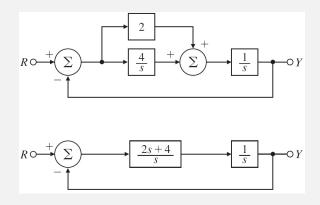
$$M(s) = rac{(G_1 + G_2) \cdot rac{1}{1 - G_2 H}}{1 + (G_1 + G_2) \cdot rac{1}{1 - G_2 H} \cdot H}$$

$$= \frac{G_1 + G_2}{1 - G_2 H + (G_1 + G_2)H}$$
$$= \frac{G_1 + G_2}{1 + G_1 H}.$$



## Block diagram simplification using Matlab

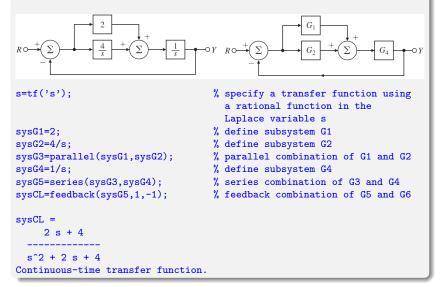
Consider the following example:



$$G(s) = \frac{\frac{2s+4}{s}\frac{1}{s}}{1+\frac{2s+4}{s}\frac{1}{s}} = \frac{2s+4}{s^2+2s+4}.$$



### To use Matlab, label the blocks as follows:





# 3-2 Signal Flow Graphs (SFGs)

# Simplified version of block diagram

Cause-and-effect relation:

$$y_2=a_{12}y_1$$

$$\bigcirc$$
 $\begin{array}{c}
a_{12} \\
y_1
\end{array}$ 
 $\begin{array}{c}
y_2
\end{array}$ 

### SFG basic properties

- Applies only to linear systems;
- Cause-and-effect algebraic equations.
- Node: variable;
- Signal flows along the direction of arrow;
- $y_k \rightarrow y_i$ : arrow represents dependence;
- a<sub>ki</sub>: gain.



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**1** Value of node  $= \sum$  all signals entering the node:

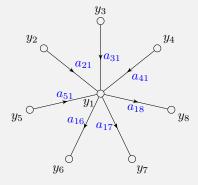
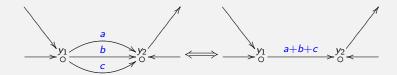


Figure 3.3:  $y_1 = a_{21}y_2 + a_{31}y_3 + a_{41}y_4 + a_{51}y_5$ .

2 Value of node is transmitted through all branches leaving the node, e.g.,  $y_6 = a_{16}y_1$ ,  $y_7 = a_{17}y_1$ ,  $y_8 = a_{18}y_1$ .



3 Parallel branches in the same direction.



4 Series connection of unidirectional branches

$$\overset{y_1}{\circ} \xrightarrow{a_{12}} \overset{a_{23}}{\circ} \xrightarrow{a_{23}} \overset{y_3}{\circ} \xrightarrow{a_{34}} \overset{y_4}{\circ} \qquad \Longleftrightarrow \qquad \overset{y_1}{\circ} \xrightarrow{a_{34}} \overset{a_{23}}{\circ} \overset{a_{12}}{\circ} \xrightarrow{y_4} \overset{y_4}{\circ}$$



### Example: Feedback Control System

$$R(s) \xrightarrow{1} E(s) \xrightarrow{G(s)} Y(s) \xrightarrow{1} Y(s) \xrightarrow{1} Y(s)$$

$$Y(s) = G(s) \cdot E(s) = G(s)(R(s) - H(s) \cdot Y(s))$$
  
 
$$\therefore (1 + G(s)H(s))Y(s) = G(s) \cdot R(s)$$

$$\therefore \frac{Y(s)}{R(s)} = \frac{G(s)}{1 + G(s)H(s)}$$



### Example: Constructing SFG from algebraic equations

$$\begin{cases} y_2 = a_{12}y_1 + a_{32}y_3 \\ y_3 = a_{23}y_2 + a_{43}y_4 \\ y_4 = a_{24}y_2 + a_{34}y_3 + a_{44}y_4 \\ y_5 = a_{25}y_2 + a_{45}y_4 \end{cases}$$

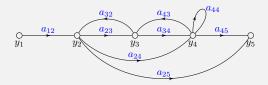
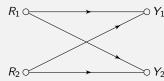


Figure 3.4: Signal Flow Graph example



# Multivariable systems



$$Y_1(s) = G_{11}(s)R_1(s) + G_{21}(s)R_2(s)$$

$$Y_2 = Y_2(s) = G_{12}(s)R_1(s) + G_{22}(s)R_2(s)$$

Matrix representation:

$$\begin{bmatrix} Y_1(s) \\ Y_2(s) \end{bmatrix} = \begin{bmatrix} G_{11}(s) & G_{21}(s) \\ G_{12}(s) & G_{22}(s) \end{bmatrix} \begin{bmatrix} R_1(s) \\ R_2(s) \end{bmatrix}$$



In general, we have

$$G_{ij}(s) = \frac{Y_j(s)}{R_i(s)}, \qquad Y(s) = G(s)R(s),$$

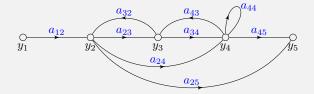
$$R(s) = \begin{bmatrix} R_1(s) \\ \vdots \\ R_p(s) \end{bmatrix}, \quad Y(s) = \begin{bmatrix} Y_1(s) \\ \vdots \\ Y_q(s) \end{bmatrix},$$

$$G(s) = egin{bmatrix} G_{11}(s) & G_{21}(s) & \cdots & G_{p1}(s) \ G_{12}(s) & G_{22}(s) & \cdots & G_{p2}(s) \ dots & dots & dots & dots \ G_{1q}(s) & G_{2q}(s) & \cdots & G_{pq}(s) \end{bmatrix}.$$

Note that the indices here are different from a usual matrix.



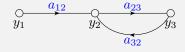
# Definition of SFG Terms



- **1** Input node: has only outgoing branch,  $y_1$ ;
- 2 Output node: has only incoming branch, y<sub>5</sub>.

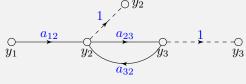


Some SFG does not have output node, e.g.,



(a) Original signal flow graph

Can add a new node to make an output node:



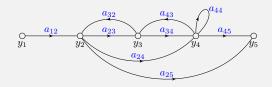
(b) Modified signal flow graph

### More definitions of SFG Terms

- 3 Path: continuous succession of branches traversed in the same direction;
- 4 Path gain: product of branch gains along a path;
- Forward path: a path from an input node to an output node, and no node traversed more than once;
- 6 Forward path gain: path gain of a forward path.



### We have three forward paths between $y_1$ and $y_5$



$$M_1: {\stackrel{y_1}{\circ}} \xrightarrow{a_{12}} {\stackrel{a_{22}}{\circ}} \xrightarrow{a_{23}} {\stackrel{y_3}{\circ}} \xrightarrow{a_{34}} {\stackrel{y_4}{\circ}} \xrightarrow{a_{45}} {\stackrel{y_5}{\circ}}$$

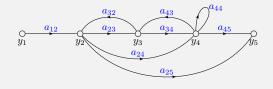
$$M_2: {\stackrel{y_1}{\circ}} \xrightarrow{a_{12}} {\stackrel{y_2}{\circ}} {\stackrel{y_5}{\circ}}$$

$$M_3: {\stackrel{g_1}{\circ}} \xrightarrow{\stackrel{a_{12}}{\rightarrow}} {\stackrel{y_2}{\circ}} \xrightarrow{\stackrel{y_4}{\circ}} \xrightarrow{\stackrel{a_{45}}{\circ}} {\stackrel{y_5}{\circ}}$$

$$M_1 = a_{45}a_{34}a_{23}a_{12}, \quad M_2 = a_{25}a_{12}, \quad M_3 = a_{45}a_{24}a_{12}.$$



8 Loop gain: path gain of a loop



$$L_{1}: \bigcirc \bigvee_{y_{2}} \bigcirc \bigvee_{a_{23}} \bigcirc \bigvee_{y_{3}} \qquad \qquad L_{2}: \bigcirc \bigvee_{y_{3}} \bigcirc \bigvee_{a_{34}} \bigcirc \bigvee_{y_{4}} \bigvee_{a_{34}} \bigvee_{y_{4}} \bigcup_{a_{34}} \bigcup_{a_{34}} \bigvee_{y_{4}} \bigcup_{a_{34}} \bigcup_{$$

$$L_1 = a_{23}a_{32}, L_2 = a_{34}a_{43}, L_3 = a_{44}, L_4 = a_{24}a_{43}a_{32}.$$



1 Non-touching Loops: loops do not share a common node, e.g.,

$$0 \stackrel{a_{32}}{\longleftarrow} 0$$

$$y_2 \stackrel{y_3}{\longrightarrow} y_3$$

and



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#### 3-2-6 Gain Formula for SFG (Mason's formula)

SFG with N forward paths and K loops:

$$M = \frac{y_{\text{out}}}{y_{\text{in}}} = \frac{1}{\Delta} \sum_{k=1}^{N} M_k \Delta_k$$
: gain between  $y_{\text{in}}$  and  $y_{\text{out}}$ ,

where

 $\Delta$ : 1–(sum of the gains of all individual loops)

+(sum of products of gains of all possible combinations of TWO non-touching loops)

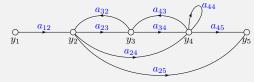
-(sum of products of gains of all possible combinations of THREE non-touching loops) · · ·

 $M_k$ : gain of the k-th forward path between  $y_{in}$  and  $y_{out}$ ,

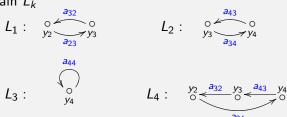
 $\Delta_k$ : the  $\Delta$  for that part of the SFG that is non-touching with kth forward path.



### Example



**1** Loop gain  $L_k$ 



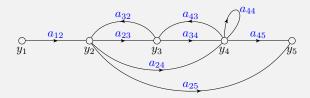
$$L_1 = a_{23}a_{32}, \ L_2 = a_{34}a_{43}, \ L_3 = a_{44}, \ L_4 = a_{24}a_{43}a_{32}.$$

- 2 Two non-touching loops:  $L_1$  and  $L_3$ ,  $(a_{23}a_{32})(a_{44})$ .
- 3 No three non-touching loop.

$$\Delta = 1 - (L_1 + L_2 + L_3 + L_4) + (L_1 L_3).$$

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# Forward path gain $M_k$ and $\Delta_k$

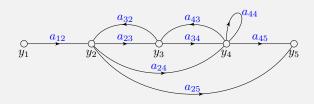


**1**  $M_1$  and  $\Delta_1$ :

$$M_1 = a_{12}a_{23}a_{34}a_{45};$$

Non-touching part: none  $\Longrightarrow \Delta_1 = 1$ .





**2**  $M_2$  and  $\Delta_2$ :

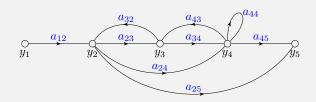
$$y_1$$
  $\xrightarrow{a_{12}}$   $y_2$   $y_5$   $y_5$ 

$$M_2 = a_{12}a_{25};$$

Non-touching part:  $\Longrightarrow \Delta_2 = 1 - (a_{43}a_{34} + a_{44}).$ 







**3**  $M_3$  and  $\Delta_3$ :



$$M_3 = a_{45}a_{24}a_{12}$$
;

Non-touching part:  $\Longrightarrow \Delta_3 = 1$ .







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Putting everything together, we have

$$\begin{split} \Delta &= 1 - \left( L_1 + L_2 + L_3 + L_4 \right) + L_1 L_3 \\ &= 1 - \left( a_{23} a_{32} + a_{34} a_{43} + a_{44} + a_{24} a_{43} a_{32} \right) + a_{23} a_{32} a_{44}, \end{split}$$

and

$$M_1 = a_{12}a_{23}a_{34}a_{45}, \quad \Delta_1 = 1;$$
  
 $M_2 = a_{12}a_{25}, \quad \Delta_2 = 1 - (a_{34}a_{43} + a_{44});$   
 $M_3 = a_{45}a_{24}a_{12}, \quad \Delta_3 = 1;$ 

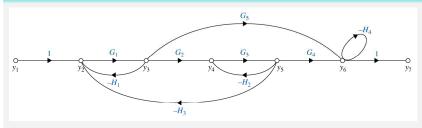
Finally, we obtain

$$\therefore \frac{y_5}{y_1} = \frac{M_1 \Delta_1 + M_2 \Delta_2 + M_3 \Delta_3}{\Delta}$$

$$= \frac{a_{12} a_{23} a_{34} a_{45} + a_{12} a_{25} (1 - a_{34} a_{43} - a_{44}) + a_{12} a_{24} a_{45}}{\Delta}$$



#### 3-2-7 Gain formula between output nodes & non-input nodes



What if we want  $\frac{y_7}{y_2}$ , where  $y_2$  is not an input?

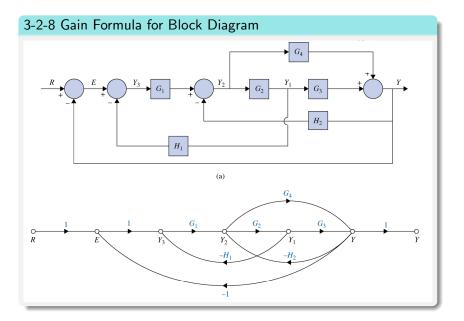
$$\frac{y_7}{y_2} = \frac{y_7/y_1}{y_2/y_1} = \frac{\sum M_k \Delta_k|_{\text{from } y_1 \text{ to } y_7}/\Delta}{\sum M_k \Delta_k|_{\text{from } y_1 \text{ to } y_2}/\Delta},$$

 $\Delta$  is independent of the inputs and outputs

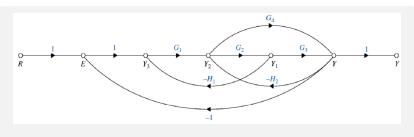
$$\therefore \frac{y_7}{y_2} = \frac{y_7/y_1}{y_2/y_1} = \frac{\sum M_k \Delta_k|_{\text{from } y_1 \text{ to } y_7}}{\sum M_k \Delta_k|_{\text{from } y_1 \text{ to } y_2}}$$



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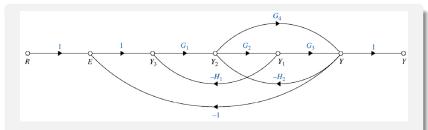
$$M_1: R \rightarrow E \rightarrow Y_3 \rightarrow Y_2 \rightarrow Y_1 \rightarrow Y \quad M_1 = G_1G_2G_3$$
  
 $M_2: R \rightarrow E \rightarrow Y_3 \rightarrow Y_2 \rightarrow Y \quad M_2 = G_1G_4$ 

Note that neither path has non-touching loops.

$$\Delta = 1 - \underbrace{(G_{1}G_{2}(-H_{1}) + G_{4}(-H_{2}) + (G_{2}G_{3})(-H_{2})}_{Y_{2}-Y_{1}-Y_{2}} + \underbrace{(G_{2}G_{3})(-H_{2}) + (G_{2}G_{3})(-H_{2})}_{Y_{2}-Y_{1}-Y-E} + \underbrace{(G_{2}G_{3})(-H_{2}) + (G_{2}G_{3})(-H_{2})}_{E-Y_{3}-Y_{2}-Y-E}$$

$$= 1 + G_{1}G_{2}H_{1} + G_{4}H_{2} + G_{2}H_{3}H_{2} + G_{1}G_{2}G_{3} + G_{1}G_{4}$$





$$\therefore \frac{Y(s)}{R(s)} = \frac{M_1 + M_2}{\Delta} = \frac{G_1 G_2 G_3 + G_1 G_4}{\Delta}$$

$$\frac{E(s)}{R(s)} = \frac{1 - (G_1 G_2 (-H_1) + G_4 (-H_2) + (G_2 G_3) (-H_2))}{\Delta}$$

$$= \frac{1 + G_1 G_2 H_1 + G_4 H_2 + G_2 G_3 H_2}{\Delta}$$

$$\therefore \frac{Y(s)}{E(s)} = \frac{G_1 G_2 G_3 + G_1 G_4}{1 + G_1 G_2 H_1 + G_4 H_2 + G_2 G_3 H_2}$$



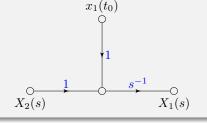
## 3-3 State Diagram

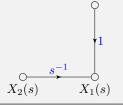
Use SFG to describe state equation and differential equation. Consider

$$\frac{dx_1(t)}{dt} = x_2(t),$$

$$sX_1(s) - x_1(t_0) = X_2(s),$$

$$\therefore X_1(s) = \frac{X_2(s)}{s} + \frac{x_1(t_0)}{s}.$$





 $x_1(t_0)$ 



## Now consider a differential equation

$$\frac{d^{n}y(t)}{dt^{n}} + a_{n}\frac{d^{n-1}y(t)}{dt^{n-1}} + \dots + a_{2}\frac{dy(t)}{dt} + a_{1}y(t) = r(t)$$

$$\Rightarrow \frac{d^{n}y(t)}{dt^{n}} = -a_{n}\frac{d^{n-1}y(t)}{dt^{n-1}} - \dots - a_{2}\frac{dy(t)}{dt} - a_{1}y(t) + r(t)$$

