# EE285 Electronics I

Amplifiers

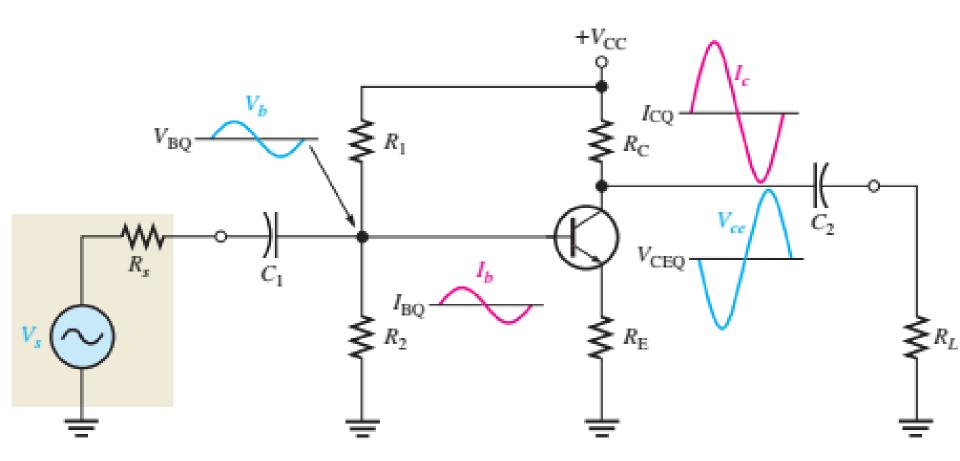
Dr. Lasith Yasakethu

#### Notations

- DC quantities are identified by nonitalic uppercase (capital) subscripts such as  $I_C$ ,  $I_E$ ,  $V_C$ , and  $V_{CE}$
- Lowercase italic subscripts are used to indicate ac quantities: for example,  $I_c$ ,  $I_e$ ,  $I_b$ ,  $V_c$ , and  $V_{ce}$  (rms values are assumed unless otherwise stated)
- Instantaneous quantities are represented by both lowercase letters and subscripts such as  $i_c$ ,  $i_e$ ,  $i_b$ , and  $v_{ce}$
- Lowercase subscripts are used to identify ac resistance values. For example,  $R_c$  is the ac collector resistance, and  $R_C$  is the dc collector resistance
- Resistance values internal to the transistor use a lowercase r to show it is an ac resistance. An example is the internal ac emitter resistance,  $r_e$ .

#### ▶ The Linear Amplifier

- A linear amplifier provides amplification of a signal without any distortion so that the output signal is an exact amplified replica of the input signal
- A voltage-divider biased transistor with a sinusoidal ac source capacitively coupled to the base through  $C_1$  and a load capacitively coupled to the collector through  $C_2$  is shown in Figure
- The capacitors ideally appear as shorts to the signal voltage
- The sinusoidal source voltage causes the base voltage to vary sinusoidally above and below its dc bias level,  $V_{\rm BQ}$ . The resulting variation in base current produces a larger variation in collector current because of the current gain of the transistor



- As the sinusoidal collector current increases, the collector voltage decreases.
- The collector current varies above and below its Q-point value,  $I_{CQ}$ , in phase with the base current.
- The sinusoidal collector-to-emitter voltage varies above and below its Q-point value,  $V_{\text{CEO}}$ , out of phase with the base voltage
- A transistor always produces a **phase inversion** between the base voltage and the collector voltage

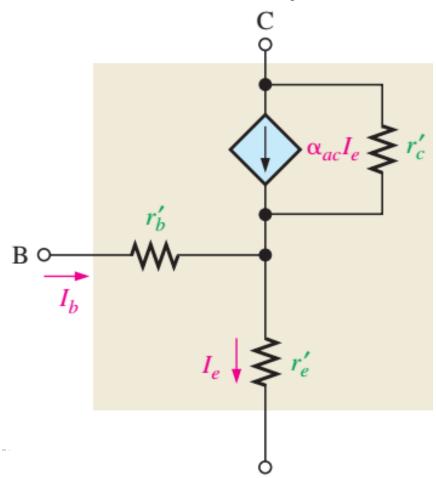
#### > r Parameters

The five *r* parameters commonly used for BJTs are given in Table. The italic lowercase letter *r* with a prime denotes resistances internal to the transistor.

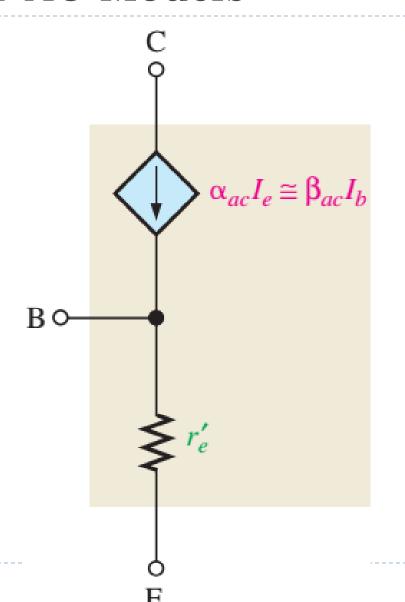
r PARAMETER	DESCRIPTION
$\alpha_{ac}$	ac alpha $(I_c/I_e)$
$\beta_{ac}$	ac beta $(I_c/I_b)$
$r'_e$	ac emitter resistance
$r_b'$	ac base resistance
$r_c'$	ac collector resistance

#### ▶ r-Parameter Transistor Model

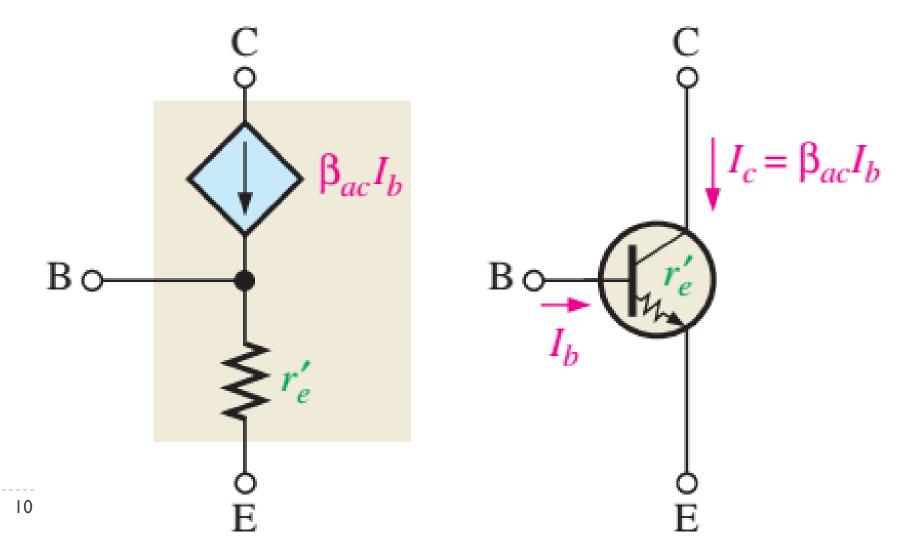
An *r*-parameter model for a BJT is shown in Figure



- For most general analysis work, it can be simplified as follows:
  - The effect of the ac base resistance (r<sub>b</sub>') is usually small enough to neglect, so it can be replaced by a short
  - The ac collector resistance  $(r_c)$  is usually several hundred kilohms and can be replaced by an open
  - The resulting simplified *r*-parameter equivalent circuit is shown in Figure



Relation of transistor symbol to r-parameter model



### • Determining $r_e$ by a Formula

- For amplifier analysis, the ac emitter resistance,  $r_e$  is the most important of the r parameters
- To calculate the approximate value of r<sub>e</sub>' you can use the below equation

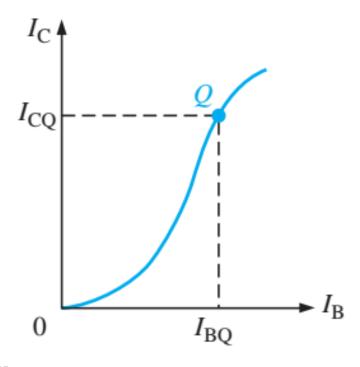
$$r_e' \cong \frac{25 \,\mathrm{mV}}{I_\mathrm{E}}$$

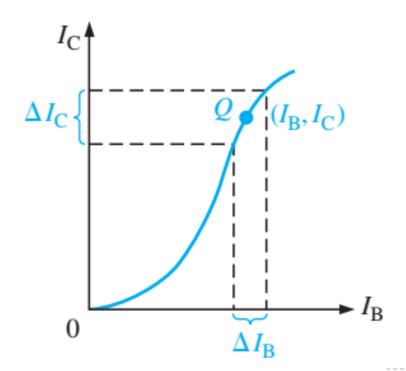
This is temperature dependent and above formula is based on an ambient temperature of 20°C

Determine the  $r'_e$  of a transistor that is operating with a dc emitter current of 2 mA.

$$r'_e \cong \frac{25 \text{ mV}}{I_E} = \frac{25 \text{ mV}}{2 \text{ mA}} = 12.5 \Omega$$

- Comparison of the AC Beta ( $\beta_{ac}$ ) to the DC Beta ( $\beta_{DC}$ )
  - For a typical transistor, a graph of  $I_C$  versus  $I_B$  is nonlinear, as shown in Figure





(a)  $\beta_{DC} = I_C/I_B$  at Q-point

(b)  $\beta_{ac} = \Delta I_{\rm C} / \Delta I_{\rm B}$ 

- If you pick a Q-point on the curve and cause the base current to vary an amount  $\Delta l_B$  then the collector current will vary an amount  $\Delta l_C$  as shown in part (b)
- At different points on the nonlinear curve, the ratio  $\Delta I_{\rm C}/\Delta I_{\rm B}$  will be different, and it may also differ from the  $I_{\rm C}/I_{\rm B}$  ratio at the Q-point
- Since  $\beta_{DC} = I_C/I_B$  and  $\beta_{AC} = \Delta I_C/\Delta I_B$  the values of these two quantities can differ slightly

#### h Parameters

- A manufacturer's datasheet typically specifies h (hybrid) parameters  $(h_i, h_r, h_f, \text{ and } h_o)$  because they are relatively easy to measure
- The four basic ac h parameters and their descriptions are given in Table

h PARAMETER	DESCRIPTION	CONDITION
$h_i$	Input impedance (resistance)	Output shorted
$h_r$	Voltage feedback ratio	Input open
$h_f$	Forward current gain	Output shorted
$h_o$	Output admittance (conductance)	Input open

- Each of the four h parameters carries a second subscript letter to designate the common-emitter (e), common-base (b), or common-collector (c) amplifier configuration, as listed in Table
- The term *common* refers to one of the three terminals (E, B, or C) that is referenced to ac ground for both input and output signals.

CONFIGURATION	h PARAMETERS
Common-Emitter	$h_{ie},h_{re},h_{fe},h_{oe}$
Common-Base	$h_{ib},h_{rb},h_{fb},h_{ob}$
Common-Collector	$h_{ic},h_{rc},h_{fc},h_{oc}$

#### Relationships of h Parameters and r Parameters

The ac current ratios,  $\alpha_{ac}$  and  $\beta_{ac}$  convert directly from h parameters as follows:

$$\alpha_{ac} = h_{fb}$$

$$\beta_{ac} = h_{fe}$$

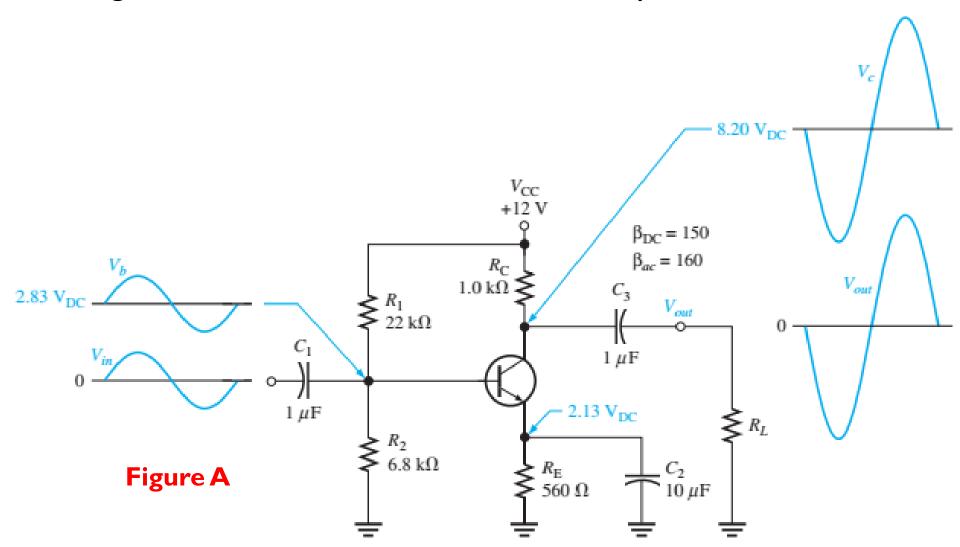
 $\triangleright$  The following formulas show how to convert to r parameters

$$r'_{e} = \frac{h_{re}}{h_{oe}}$$

$$r'_{c} = \frac{h_{re} + 1}{h_{oe}}$$

$$r'_b = h_{ie} - \frac{h_{re}}{h_{oe}} (1 + h_{fe})$$

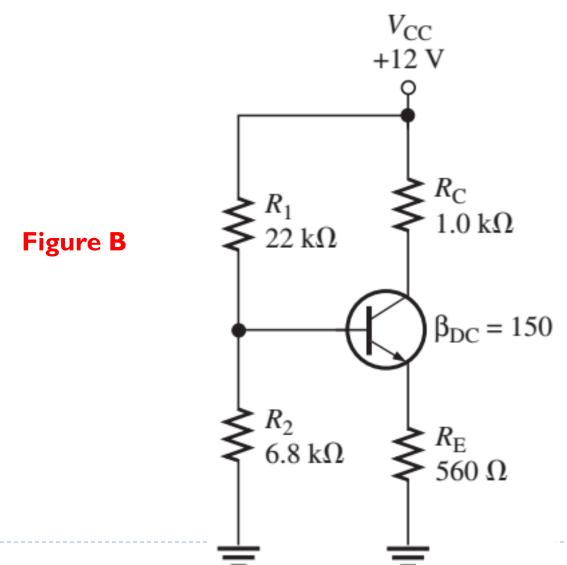
▶ Figure shows a **common-emitter** amplifier



- The input signal,  $V_{in}$ , is capacitively coupled to the base terminal, the output signal,  $V_{out}$ , is capacitively coupled from the collector to the load
- Because the ac signal is applied to the base terminal as the input and taken from the collector terminal as the output, the emitter is common to both the input and output signals
- There is no signal at the emitter because the bypass capacitor effectively shorts the emitter to ground at the signal frequency
- The output signal is 180°out of phase with the input signal

#### DC Analysis

- To analyze the amplifier in Figure A, the dc bias values must first be determined
- To do this, a dc equivalent circuit is developed by removing the coupling and bypass capacitors because they appear open as far as the dc bias is concerned
- This also removes the load resistor and signal source.
- ▶ The dc equivalent circuit is shown in Figure B



Theveninizing the bias circuit and applying Kirchhoff's voltage law to the base-emitter circuit,

$$R_{\text{TH}} = \frac{R_1 R_2}{R_1 + R_2} = \frac{(6.8 \text{ k}\Omega)(22 \text{ k}\Omega)}{6.8 \text{ k}\Omega + 22 \text{ k}\Omega} = 5.19 \text{ k}\Omega$$

$$V_{\text{TH}} = \left(\frac{R_2}{R_1 + R_2}\right) V_{\text{CC}} = \left(\frac{6.8 \text{ k}\Omega}{6.8 \text{ k}\Omega + 22 \text{ k}\Omega}\right) 12 \text{ V} = 2.83 \text{ V}$$

$$V_{TH} - I_B R_{TH} - V_{BE} - I_E R_E = 0$$

$$I_E = \frac{V_{\text{TH}} - V_{\text{BE}}}{R_E + R_{\text{TH}}/\beta_{\text{DC}}} = \frac{2.83 \text{ V} - 0.7 \text{ V}}{560 \Omega + 34.6 \Omega} = 3.58 \text{ mA}$$

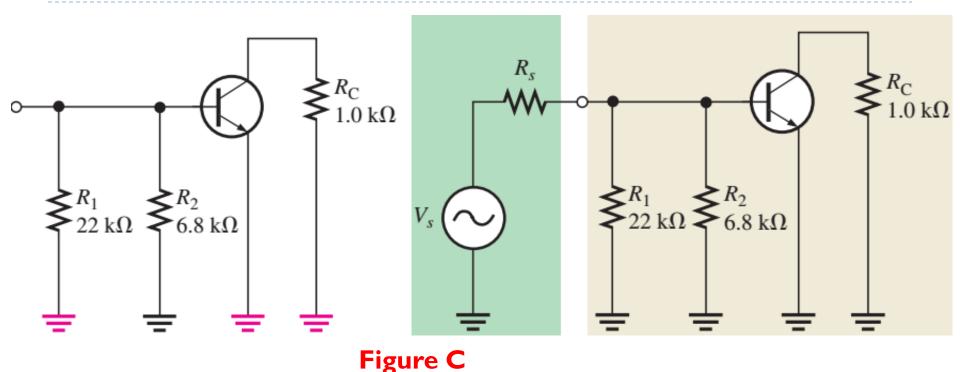
$$I_C \cong I_E = 3.58 \text{ mA}$$

$$V_E = I_E R_E = (3.58 \text{ mA})(560 \Omega) = 2 \text{ V}$$

$$V_{\rm B} = V_{\rm E} + 0.7 \,\text{V} = 2.7 \,\text{V}$$
  
 $V_{\rm C} = V_{\rm CC} - I_{\rm C}R_{\rm C} = 12 \,\text{V} - (3.58 \,\text{mA})(1.0 \,\text{k}\Omega) = 8.42 \,\text{V}$   
 $V_{\rm CE} = V_{\rm C} - V_{\rm E} = 8.42 \,\text{V} - 2 \,\text{V} = 6.42 \,\text{V}$ 

### AC Analysis

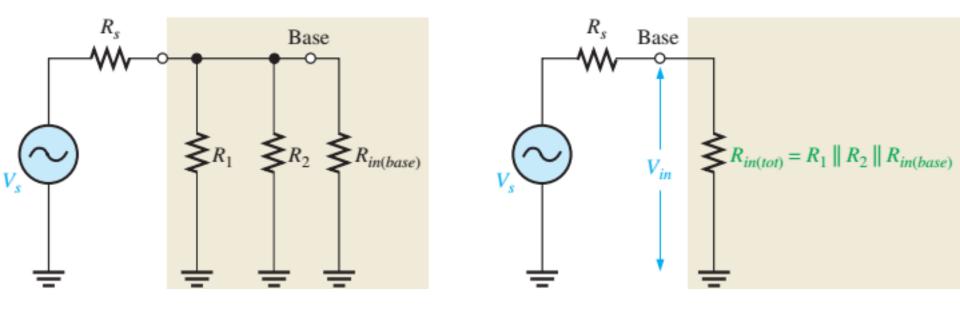
- To analyze the ac signal operation of an amplifier, an ac equivalent circuit is developed as follows:
  - The capacitors  $C_1$ ,  $C_2$ , and  $C_3$  are replaced by effective shorts because their values are selected so that  $X_C$  is negligible at the signal frequency and can be considered to be  $0\Omega$
  - The dc source is replaced by ground
- The ac equivalent circuit for the common-emitter amplifier in Figure A is shown in Figure C
- Notice that both  $R_C$  and  $R_I$  have one end connected to ac ground (red) because, in the actual circuit, they are connected to  $V_{CC}$
- <sub>24</sub> which is, in effect, ac ground



• In ac analysis, the ac ground and the actual ground are treated as the same point electrically. The amplifier in Figure A is called a common-emitter amplifier because the bypass capacitor  $C_2$  keeps the emitter at ac ground. Ground is the common point in the circuit.

#### Signal (AC) Voltage at the Base

- An ac voltage source,  $V_s$ , is shown connected to the input in Figure C
- If the internal resistance of the ac source is  $0\Omega$  then all of the source voltage appears at the base terminal
- If, however, the ac source has a nonzero internal resistance, then three factors must be taken into account in determining the actual signal voltage at the base. These are
  - $\triangleright$  the source resistance  $(R_s)$ ,
  - $\blacktriangleright$  the bias resistance  $(R_1 || R_2)$  and
  - the ac input resistance at the base of the transistor  $(R_{in(base)})$



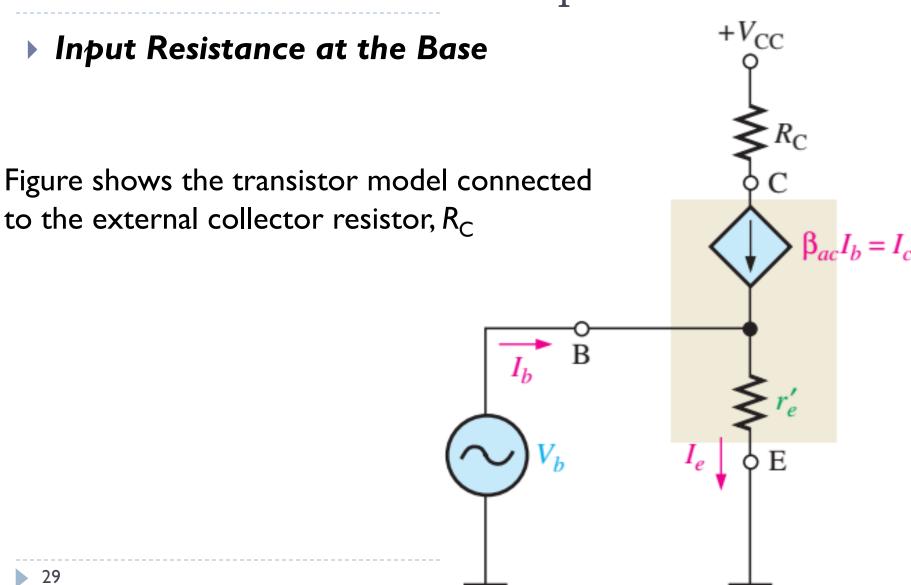
•  $R_1, R_2$ , and  $R_{in(base)}$  are combined in parallel to get the total **input** resistance,  $R_{in(tot)}$ , which is the resistance "seen" by an ac source connected to the input

$$R_{in(tot)} = R_1 \| R_2 \| R_{in(base)}$$

Signal voltage at the base of the transistor is found by the voltagedivider formula as follows:

$$V_b = \left(\frac{R_{in(tot)}}{R_s + R_{in(tot)}}\right) V_s$$

If  $R_s \ll R_{in(tot)}$  then  $V_b \approx V_s$  where  $V_b$  is the input voltage,  $V_{in}$ , to the amplifier



The input resistance looking in at the base is

$$R_{in(base)} = \frac{V_{in}}{I_{in}} = \frac{V_b}{I_b}$$

- The base voltage is:  $V_b = I_e r_e'$
- ▶ and since  $I_e \approx I_c$ :  $I_b \approx {^{I_e}}/{\beta_{ac}}$
- Substituting for  $V_b$  and  $I_b$

$$R_{in(base)} = \frac{V_b}{I_b} = \frac{I_e r_e'}{I_{e/\beta_{ac}}}$$

$$R_{in(base)} = \beta_{ac} r'_e$$

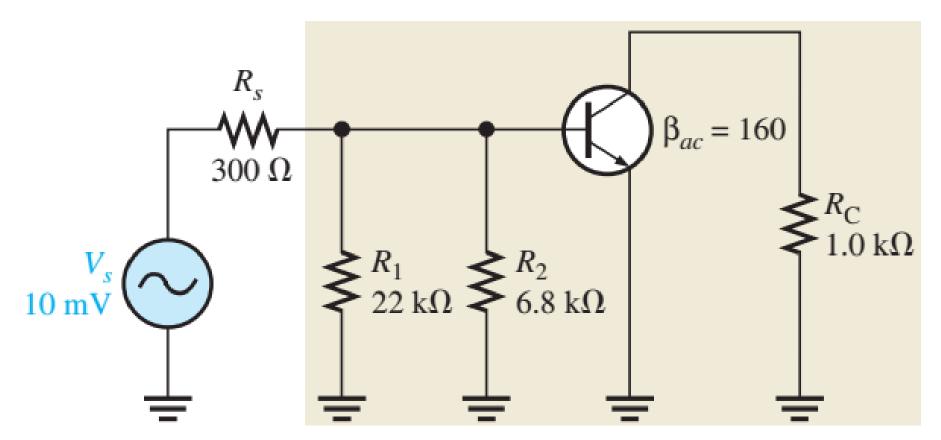
#### Output Resistance

The **output resistance** of the common-emitter amplifier is the resistance looking in at the collector and is approximately equal to the collector resistor

$$R_{out} \cong R_{\rm C}$$

Actually,  $R_{out} = R_C || r'_c$  but since the internal ac collector resistance of the transistor  $r'_c$ , is typically much larger than  $R_C$ , the approximation is usually valid

Determine the signal voltage at the base of the transistor in Figure. This circuit is the ac equivalent of the amplifier in Figure A with a 10 mV rms, 300  $\Omega$  signal source.  $I_E$  was previously found to be 3.80 mA.



First, determine the ac emitter resistance.

$$r'_e \cong \frac{25 \text{ mV}}{I_E} = \frac{25 \text{ mV}}{3.80 \text{ mA}} = 6.58 \Omega$$

Then,

$$R_{in(base)} = \beta_{ac} r'_{e} = 160(6.58 \,\Omega) = 1.05 \,\mathrm{k}\Omega$$

Next, determine the total input resistance viewed from the source.

$$R_{in(tot)} = R_1 \| R_2 \| R_{in(base)} = \frac{1}{\frac{1}{22 \text{ k}\Omega} + \frac{1}{6.8 \text{ k}\Omega} + \frac{1}{1.05 \text{ k}\Omega}} = 873 \Omega$$

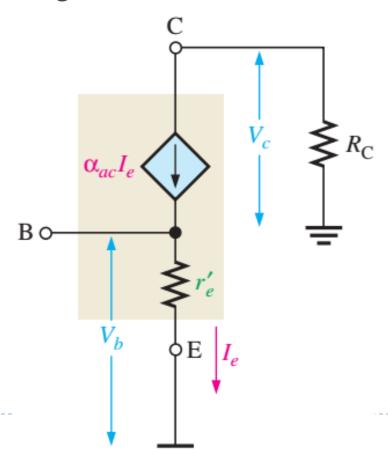
The source voltage is divided down by  $R_s$  and  $R_{in(tot)}$ , so the signal voltage at the base is the voltage across  $R_{in(tot)}$ .

$$V_b = \left(\frac{R_{in(tot)}}{R_s + R_{in(tot)}}\right) V_s = \left(\frac{873 \Omega}{1173 \Omega}\right) 10 \text{ mV} = 7.44 \text{ mV}$$

As you can see, there is significant attenuation (reduction) of the source voltage due to the source resistance and amplifier's input resistance combining to act as a voltage divider.

#### Voltage Gain

The ac voltage gain expression for the common-emitter amplifier is developed using the below model circuit



The gain is the ratio of ac output voltage at the collector  $(V_c)$  to ac input voltage at the base  $(V_b)$ 

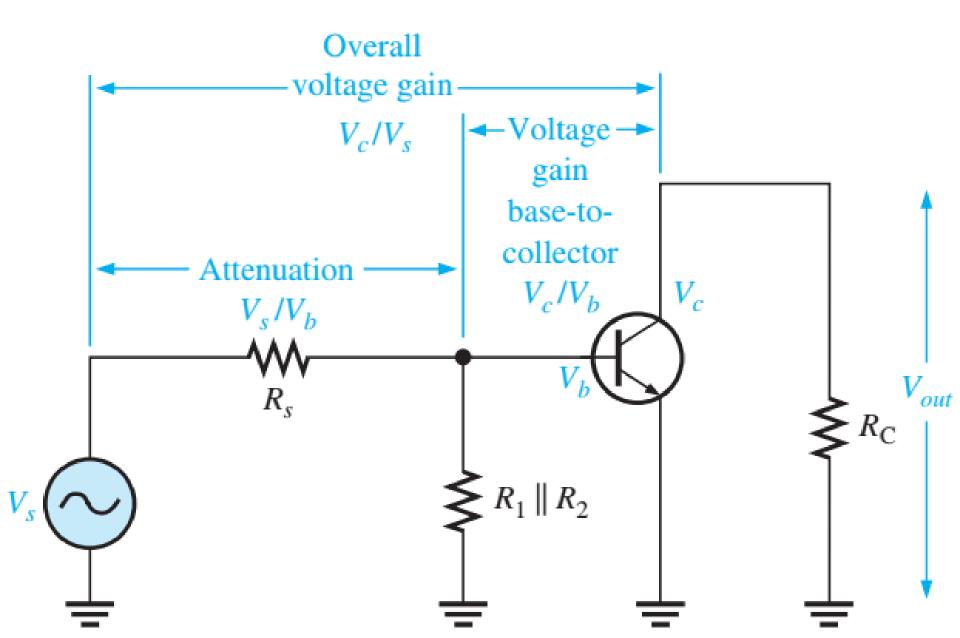
$$A_V = \frac{V_{out}}{V_{in}} = \frac{V_c}{V_b}$$

- Notice in the figure that:  $V_c = \alpha_{ac} I_e R_C \approx I_e R_C$  and  $V_b = I_e r_e'$
- Therefore,  $A_V = \frac{I_e R_C}{I_e r'_e}$
- $A_V = \frac{R_C}{r'_e}$
- Equation gives the voltage gain from base to collector. To get the overall gain of the amplifier from the source voltage to collector, the attenuation of the input circuit must be included

#### Attenuation

- Attenuation is the reduction in signal voltage as it passes through a circuit and corresponds to a gain of less than 1.
- For example, if the signal amplitude is reduced by half, the attenuation is 2, which can be expressed as a gain of 0.5 because gain is the reciprocal of attenuation.
- Assume that the amplifier in Figure has a voltage gain from base to collector of  $A_v$  and the attenuation from the source to the base is  $V_s/V_b$ . This attenuation is produced by the source resistance and total input resistance of the amplifier acting as a voltage divider and can be expressed as

Attenuation = 
$$\frac{V_s}{V_b} = \frac{R_s + R_{in(tot)}}{R_{in(tot)}}$$



The overall voltage gain of the amplifier, is the voltage gain from base to collector,  $V_c/V_b$ , times the reciprocal of the attenuation,  $V_b/V_s$ 

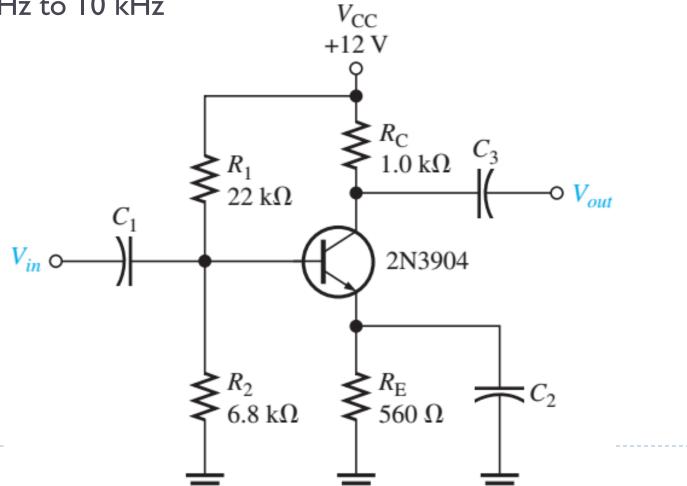
$$A_{v}' = \left(\frac{V_{c}}{V_{b}}\right) \left(\frac{V_{b}}{V_{s}}\right) = \frac{V_{c}}{V_{s}}$$

### ▶ Effect of the Emitter Bypass Capacitor

- The emitter bypass capacitor, which is  $C_2$  in Figure A, provides an effective short to the ac signal around the emitter resistor, thus keeping the emitter at ac ground. With the bypass capacitor, the gain of a given amplifier is maximum and equal to  $A_V = \frac{R_C}{r_e'}$
- The value of the bypass capacitor must be large enough so that its reactance over the frequency range of the amplifier is very small (ideally 0  $\Omega$ ) compared to  $R_{\rm E}$ . A good rule-of-thumb is that the capacitive reactance,  $X_{\rm C}$ , of the bypass capacitor should be at least 10 times smaller than  $R_{\rm E}$  at the minimum frequency for which the amplifier must operate

$$X_C \leq \frac{R_E}{10}$$

Select a minimum value for the emitter bypass capacitor,  $C_2$ , in Figure if the amplifier must operate over a frequency range from 200 Hz to 10 kHz  $V_{--}$ 



The  $X_C$  of the bypass capacitor,  $C_2$ , should be at least ten times less than  $R_E$ .

$$X_{C2} = \frac{R_{\rm E}}{10} = \frac{560 \,\Omega}{10} = 56 \,\Omega$$

Determine the capacitance value at the minimum frequency of 200 Hz as follows:

$$C_2 = \frac{1}{2\pi f X_{C2}} = \frac{1}{2\pi (200 \text{ Hz})(56 \Omega)} = 14.2 \ \mu\text{F}$$

This is the minimum value for the bypass capacitor for this circuit. You can always use a larger value, although cost and physical size may impose limitations.

### Voltage Gain Without the Emitter Bypass Capacitor

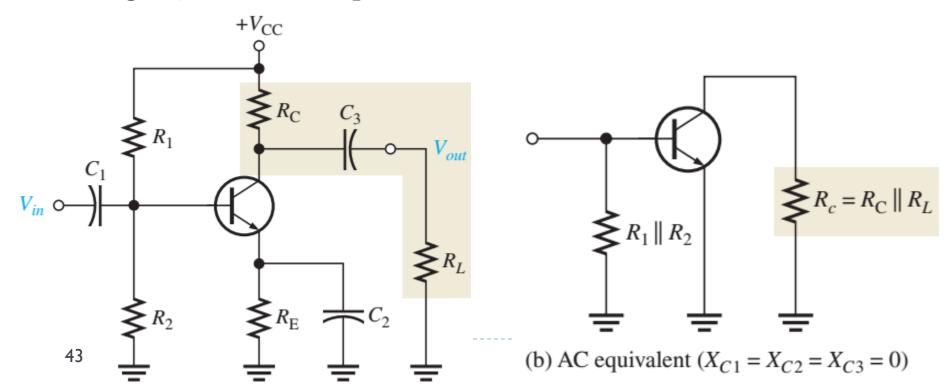
Without the bypass capacitor, the emitter is no longer at ac ground. Instead,  $R_{\rm E}$  is seen by the ac signal between the emitter and ground and effectively adds to in the voltage gain formula (refer model circuit used to derive volt. gain with bypass capacitor)

$$A_{v} = \frac{R_{\rm C}}{r'_{e} + R_{\rm E}}$$

The effect of  $R_E$  is to decrease the ac voltage gain

### Effect of a Load on the Voltage Gain

When a resistor,  $R_L$ , is connected to the output through the coupling capacitor  $C_3$ , as shown in Figure, it creates a load on the circuit. The collector resistance at the signal frequency is effectively  $R_C$  in parallel with  $R_L$ .



▶ The total ac collector resistance is

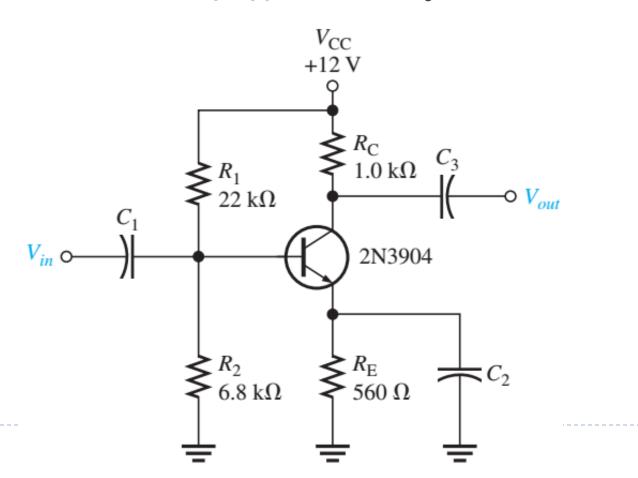
$$R_c = \frac{R_{\rm C}R_L}{R_{\rm C} + R_L}$$

Replacing  $R_C$  with  $R_c$  in the voltage gain expression gives

$$A_{v} = \frac{R_{c}}{r'_{e}}$$

When  $R_c < R_C$  because of  $R_L$ , the voltage gain is reduced. However, if  $R_L >> R_C$  then  $R_c \approx R_C$  and the load has very little effect on the gain

Calculate the base-to-collector voltage gain of the amplifier in Figure when a load resistance of 5 k $\Omega$  is connected to the output. The emitter is effectively bypassed and  $r'_e$ = 6.58  $\Omega$ 



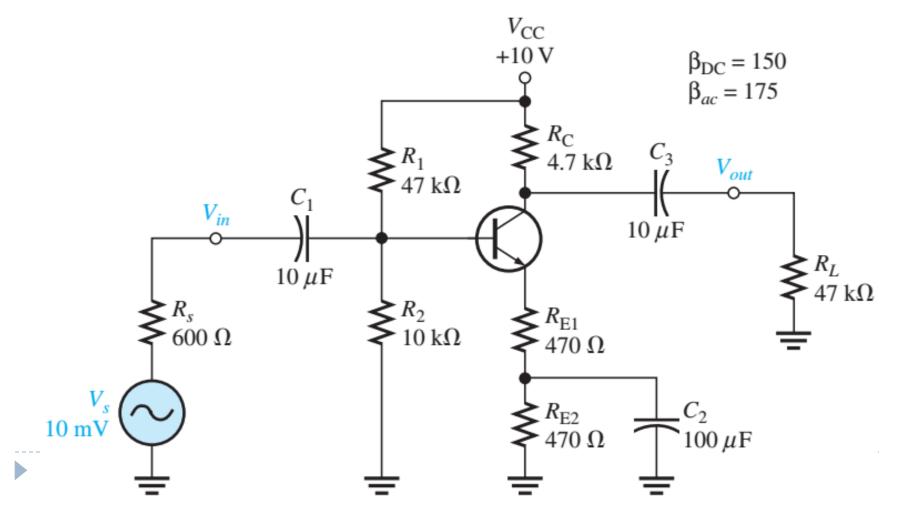
The ac collector resistance is

$$R_c = \frac{R_{\rm C}R_L}{R_{\rm C} + R_L} = \frac{(1.0\,{\rm k}\Omega)(5\,{\rm k}\Omega)}{6\,{\rm k}\Omega} = 833\,\Omega$$

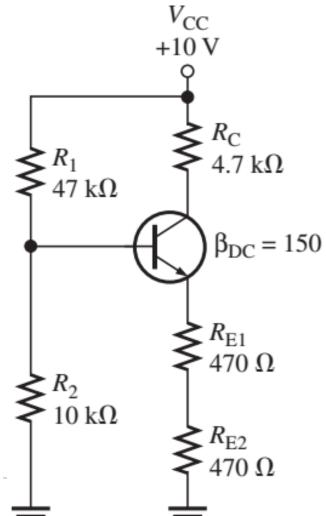
Therefore,

$$A_v = \frac{R_c}{r'_e} = \frac{833 \ \Omega}{6.58 \ \Omega} = 127$$

For the amplifier in Figure a) Determine the dc collector voltageb) Determine the ac collector voltage



Determine the dc bias values using the dc equivalent circuit in below Figure  $_{
m extsf{v}}$ 



Apply Thevenin's theorem and Kirchhoff's voltage law to the base-emitter circuit in Figure

$$R_{\text{TH}} = \frac{R_1 R_2}{R_1 + R_2} = \frac{(47 \text{ k}\Omega)(10 \text{ k}\Omega)}{47 \text{ k}\Omega + 10 \text{ k}\Omega} = 8.25 \text{ k}\Omega$$

$$V_{\text{TH}} = \left(\frac{R_2}{R_1 + R_2}\right) V_{\text{CC}} = \left(\frac{10 \text{ k}\Omega}{47 \text{ k}\Omega + 10 \text{ k}\Omega}\right) 10 \text{ V} = 1.75 \text{ V}$$

$$I_{\text{E}} = \frac{V_{\text{TH}} - V_{\text{BE}}}{R_{\text{E}} + R_{\text{TH}}/\beta_{\text{DC}}} = \frac{1.75 \text{ V} - 0.7 \text{ V}}{940 \Omega + 55 \Omega} = 1.06 \text{ mA}$$

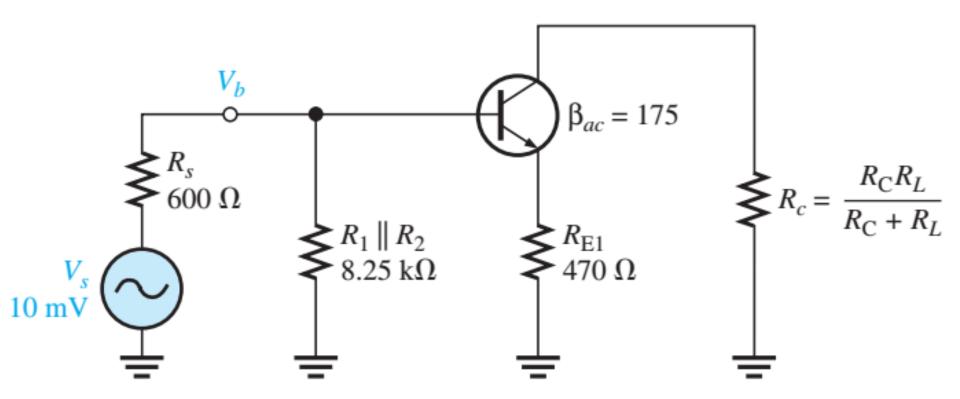
$$I_{\text{C}} \approx I_{\text{E}} = 1.06 \text{ mA}$$

$$V_{\text{E}} = I_{\text{E}}(R_{\text{E1}} + R_{\text{E2}}) = (1.06 \text{ mA})(940 \Omega) = 1 \text{ V}$$

$$V_{\text{B}} = V_{\text{E}} + 0.7 \text{ V} = 1 \text{ V} - 0.7 \text{ V} = 0.3 \text{ V}$$

$$V_{\text{C}} = V_{\text{CC}} - I_{\text{C}}R_{\text{C}} = 10 \text{ V} - (1.06 \text{ mA})(4.7 \text{ k}\Omega) = 5.02 \text{ V}$$

The ac analysis is based on the ac equivalent circuit in below Figure



The first thing to do in the ac analysis is calculate  $r'_e$ .

$$r'_e \cong \frac{25 \,\text{mV}}{I_E} = \frac{25 \,\text{mV}}{1.06 \,\text{mA}} = 23.6 \,\Omega$$

Next, determine the attenuation in the base circuit. Looking from the 600  $\Omega$  source, the total  $R_{in}$  is

$$R_{in(tot)} = R_1 \| R_2 \| R_{in(base)}$$
  
 $R_{in(base)} = \beta_{ac}(r'_e + R_{E1}) = 175(494 \Omega) = 86.5 \text{ k}\Omega$ 

Therefore,

$$R_{in(tot)} = 47 \,\mathrm{k}\Omega \,\|\, 10 \,\mathrm{k}\Omega \,\|\, 86.5 \,\mathrm{k}\Omega = 7.53 \,\mathrm{k}\Omega$$

The attenuation from source to base is

Attenuation 
$$= \frac{V_s}{V_b} = \frac{R_s + R_{in(tot)}}{R_{in(tot)}} = \frac{600 \Omega + 7.53 \text{ k}\Omega}{7.53 \text{ k}\Omega} = 1.08$$

Before  $A_v$  can be determined, you must know the ac collector resistance  $R_c$ .

$$R_c = \frac{R_C R_L}{R_C + R_I} = \frac{(4.7 \,\mathrm{k}\Omega)(47 \,\mathrm{k}\Omega)}{4.7 \,\mathrm{k}\Omega + 47 \,\mathrm{k}\Omega} = 4.27 \,\mathrm{k}\Omega$$

The voltage gain from base to collector is

$$A_v \cong \frac{R_c}{R_{\rm E1}} = \frac{4.27 \,\mathrm{k}\Omega}{470 \,\Omega} = 9.09$$

The overall voltage gain is the reciprocal of the attenuation times the amplifier voltage gain.

$$A_v' = \left(\frac{V_b}{V_s}\right) A_v = (0.93)(9.09) = 8.45$$

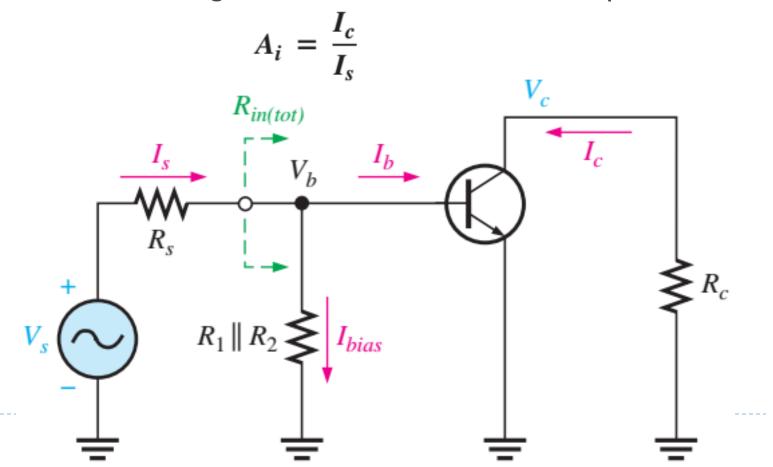
The source produces 10 mV rms, so the rms voltage at the collector is

$$V_c = A'_v V_s = (8.45)(10 \,\text{mV}) = 84.5 \,\text{mV}$$

#### Current Gain

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The current gain from base to collector is  $I_c/I_b$  or  $\beta_{ac}$ . However, the overall current gain of the common-emitter amplifier is



▶  $I_s$  is the total signal input current produced by the source, part of which  $(I_b)$  is base current and part of which  $(I_{bias})$  goes through the bias circuit  $(R_1 || R_2)$  as shown in Figure. The source "sees" a total resistance of  $R_s + R_{in(tot)}$ . The total current produced by the source is

$$I_s = \frac{V_s}{R_s + R_{in(tot)}}$$

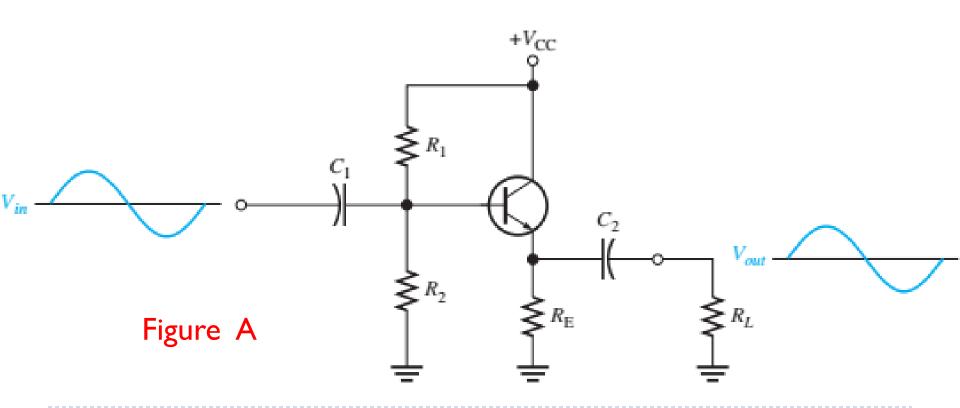
#### Power Gain

The overall power gain is the product of the overall voltage gain  $(A'_{\nu})$  and the overall current gain  $(A_{i})$ .

$$A_p = A_v' A_i$$

• where  $A'_v = V_c/V_s$ 

The **common-collector** (CC) amplifier is usually referred to as an emitter-follower (EF). An **emitter-follower** circuit with voltage-divider bias is shown in Figure.

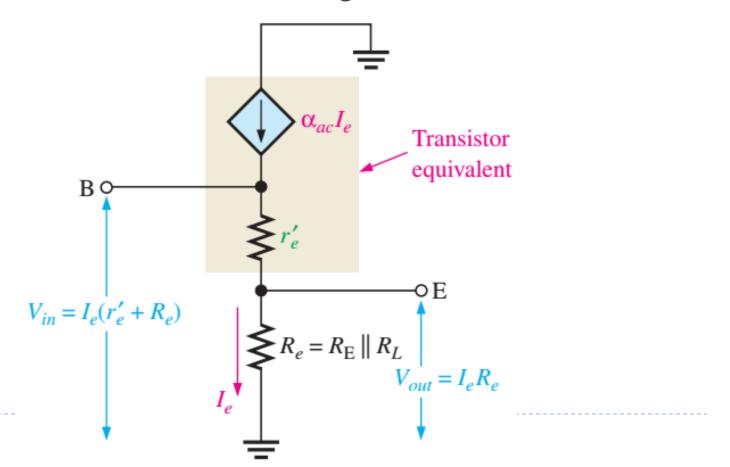


- Notice that the input signal is capacitively coupled to the base, the output signal is capacitively coupled from the emitter, and the collector is at ac ground.
- There is no phase inversion, and the output is approximately the same amplitude as the input.

### Voltage Gain

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As in all amplifiers, the voltage gain is  $A_v = V_{out}/V_{in}$ . The ac model for the emitter follower is shown in Figure



$$V_{out} = I_e R_e$$

$$V_{in} = I_e(r_e' + R_e)$$

Therefore, the voltage gain is

$$A_v = \frac{I_e R_e}{I_e (r_e' + R_e)}$$

$$A_v = \frac{R_e}{r_e' + R_e}$$

• where  $R_e$  is the parallel combination of  $R_E$  and  $R_L$ 

$$A_v = \frac{R_e}{r_e' + R_e}$$

- If there is no load, then  $R_e = R_E$ .
- Notice that the gain is always less than 1. If  $R_E >> r'_e$  then a good approximation is

$$A_{\nu} \cong 1$$

Since the output voltage is at the emitter, it is in phase with the base voltage, so there is no inversion from input to output. Because there is no inversion and because the voltage gain is approximately I, the output voltage closely follows the input voltage in both phase and amplitude; thus the term emitter-follower

### ▶ Input Resistance

- The emitter-follower is characterized by a high input resistance; this is what makes it a useful circuit. Because of the high input resistance, it can be used as a buffer to minimize loading effects when a circuit is driving a low-resistance load
- The derivation of the input resistance, looking in at the base of the common-collector amplifier, is similar to that for the common-emitter amplifier. In a common-collector circuit, however, the emitter resistor is *never* bypassed because the output is taken across  $R_e$ , which is  $R_E$  in parallel with  $R_L$

$$R_{in(base)} = \frac{V_{in}}{I_{in}} = \frac{V_b}{I_b} = \frac{I_e(r'_e + R_e)}{I_b}$$

Since 
$$I_e \cong I_c = \beta_{ac}I_b$$
,

$$R_{in(base)} \cong \frac{\beta_{ac}I_b(r'_e + R_e)}{I_b}$$

$$R_{in(base)} \cong \beta_{ac}(r'_e + R_e)$$

If  $R_e \gg r'_e$ , then the input resistance at the base is simplified to

$$R_{in(base)} \cong \beta_{ac}R_e$$

The bias resistors in Figure A appear in parallel with  $R_{in(base)}$ , looking from the input source; and just as in the common-emitter circuit, the total input resistance is

$$R_{in(tot)} = R_1 \| R_2 \| R_{in(base)}$$

#### Output Resistance

With the load removed, the output resistance, looking into the emitter of the emitter-follower is approximated as follows:

$$R_{out} \cong \left(\frac{R_s}{\beta_{ac}}\right) \| R_{\rm E}$$

 $\triangleright$   $R_s$  is the resistance of the input source

#### Current Gain

The current gain for the emitter-follower in Figure A is

$$A_i = \frac{I_e}{I_{in}}$$

• where  $I_{in} = V_{in}/R_{in(tot)}$ 

#### Power Gain

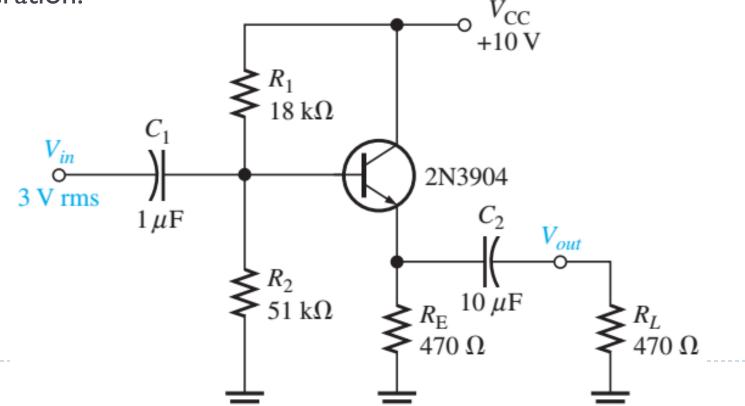
The common-collector power gain is the product of the voltage gain and the current gain. For the emitter-follower, the power gain is approximately equal to the current gain because the voltage gain is approximately I

$$A_p = A_v A_i$$

Since  $A_v \cong 1$ , the power gain is

$$A_p \cong A_i$$

Determine the total input resistance of the emitter-follower in Figure. Also find the voltage gain, current gain, and power gain in terms of power delivered to the load,  $R_L$ . Assume  $\beta_{ac} = 175$  and that the capacitive reactances are negligible at the frequency of operation.



The ac emitter resistance external to the transistor is

$$R_e = R_E \| R_L = 470 \Omega \| 470 \Omega = 235 \Omega$$

The approximate resistance, looking in at the base, is

$$R_{in(base)} \cong \beta_{ac}R_e = (175)(235 \Omega) = 41.1 \text{ k}\Omega$$

The total input resistance is

$$R_{in(tot)} = R_1 \| R_2 \| R_{in(base)} = 18 \text{ k}\Omega \| 51 \text{ k}\Omega \| 41.1 \text{ k}\Omega = 10.1 \text{ k}\Omega$$

The voltage gain is  $A_v \cong 1$ . By using  $r'_e$ , you can determine a more precise value of  $A_v$  if necessary.

$$V_{\rm E} = \left(\frac{R_2}{R_1 + R_2}\right) V_{\rm CC} - V_{\rm BE} = \left(\frac{51 \,\mathrm{k}\Omega}{18 \,\mathrm{k}\Omega + 51 \,\mathrm{k}\Omega}\right) 10 \,\mathrm{V} - 0.7 \,\mathrm{V}$$
$$= (0.739)(10 \,\mathrm{V}) - 0.7 \,\mathrm{V} = 6.69 \,\mathrm{V}$$

Therefore,

$$I_{\rm E} = \frac{V_{\rm E}}{R_{\rm E}} = \frac{6.69 \text{ V}}{470 \Omega} = 14.2 \text{ mA}$$

and

$$r'_e \cong \frac{25 \text{ mV}}{I_E} = \frac{25 \text{ mV}}{14.2 \text{ mA}} = 1.76 \Omega$$

So,

$$A_v = \frac{R_e}{r'_e + R_e} = \frac{235 \ \Omega}{237 \ \Omega} = \mathbf{0.992}$$

The small difference in  $A_v$  as a result of considering  $r'_e$  is insignificant in most cases.

The current gain is  $A_i = I_e/I_{in}$ . The calculations are as follows:

$$I_e = \frac{V_e}{R_e} = \frac{A_v V_b}{R_e} \cong \frac{(0.992)(3 \text{ V})}{235 \Omega} = \frac{2.98 \text{ V}}{235 \Omega} = 12.7 \text{ mA}$$

$$I_{in} = \frac{V_{in}}{R_{in(tot)}} = \frac{3 \text{ V}}{10.1 \text{ k}\Omega} = 297 \,\mu\text{A}$$

$$A_i = \frac{I_e}{I_{in}} = \frac{12.7 \text{ mA}}{297 \,\mu\text{A}} = 42.8$$

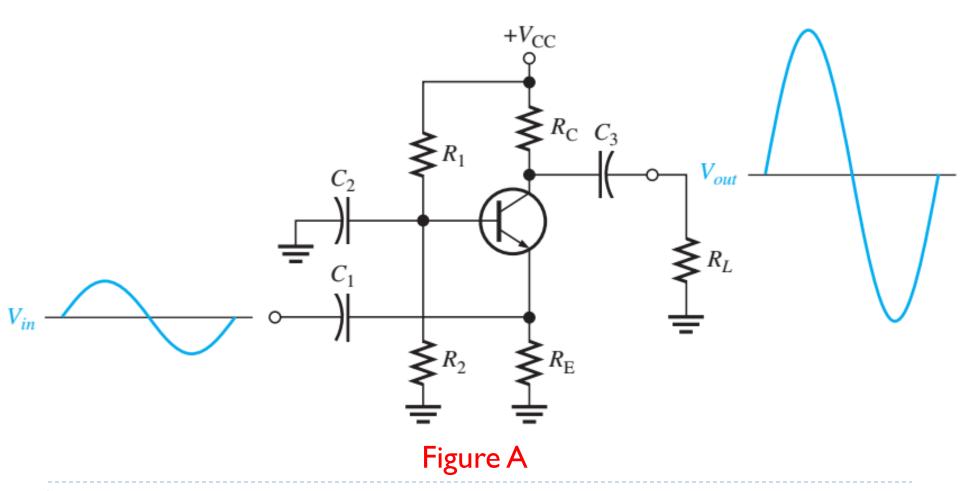
The power gain is

$$A_p \cong A_i = 42.8$$

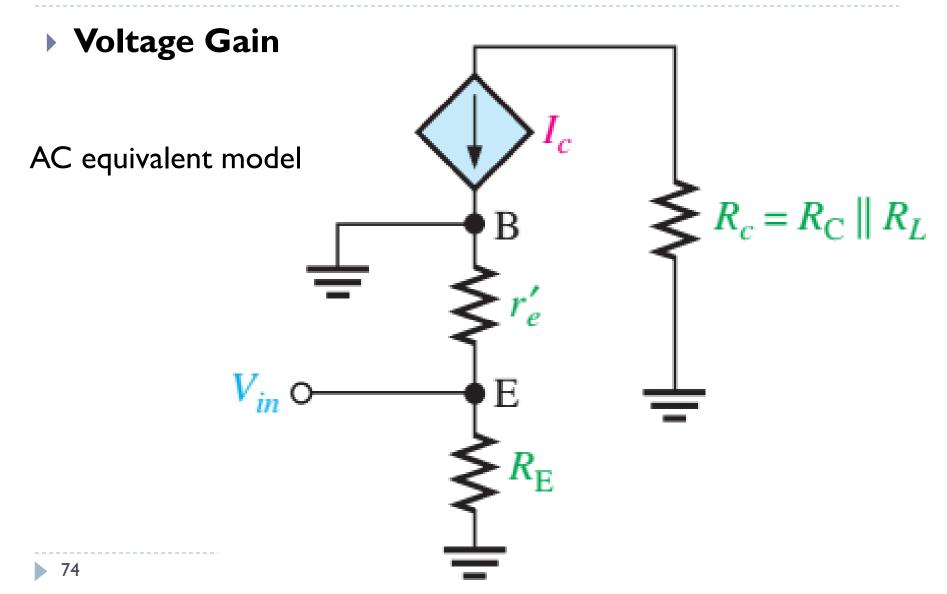
Since  $R_L = R_E$ , one-half of the power is dissipated in  $R_E$  and one-half in  $R_L$ . Therefore, in terms of power to the load, the power gain is

$$A_{p(load)} = \frac{A_p}{2} = \frac{42.8}{2} = 21.4$$

A typical common-base amplifier is shown in Figure A



- The base is the common terminal and is at ac ground because of capacitor  $C_2$ .
- The input signal is capacitively coupled to the emitter
- The output is capacitively coupled from the collector to a load resistor



The voltage gain from emitter to collector is developed as follows  $(V_{in} = V_e, V_{out} = V_c)$ 

$$A_{v} = \frac{V_{out}}{V_{in}} = \frac{V_{c}}{V_{e}} = \frac{I_{c}R_{c}}{I_{e}(r'_{e} \parallel R_{E})} \cong \frac{I_{e}R_{c}}{I_{e}(r'_{e} \parallel R_{E})}$$

If  $R_{\rm E} \gg r_e'$ , then

$$A_v \cong \frac{R_c}{r'_e}$$

where  $R_c = R_C \parallel R_L$ . Notice that the gain expression is the same as for the common-emitter amplifier. However, there is no phase inversion from emitter to collector

### Input Resistance

The resistance, looking in at the emitter, is

$$R_{in(emitter)} = \frac{V_{in}}{I_{in}} = \frac{V_e}{I_e} = \frac{I_e(r'_e \parallel R_E)}{I_e}$$

If 
$$R_{\rm E} \gg r_e'$$
, then

$$R_{in(emitter)} \cong r'_e$$

▶  $R_{\rm E}$  is typically much greater than  $r'_{\rm e}$  so the assumption that  $r'_{\rm e} \mid\mid R_{\rm E} \approx r'_{\rm e}$  is usually valid

### Output Resistance

Looking into the collector, the ac collector resistance,  $r'_c$  appears in parallel with  $R_C$ . As you have previously seen in connection with the CE amplifier,  $r'_c$ , is typically much larger than  $R_C$ , so a good approximation for the output resistance is

$$R_{out} \cong R_{\rm C}$$

#### Current Gain

The current gain is the output current divided by the input current.  $I_c$  is the ac output current, and  $I_e$  is the ac input current. Since  $I_c \approx I_e$  the current gain is approximately I

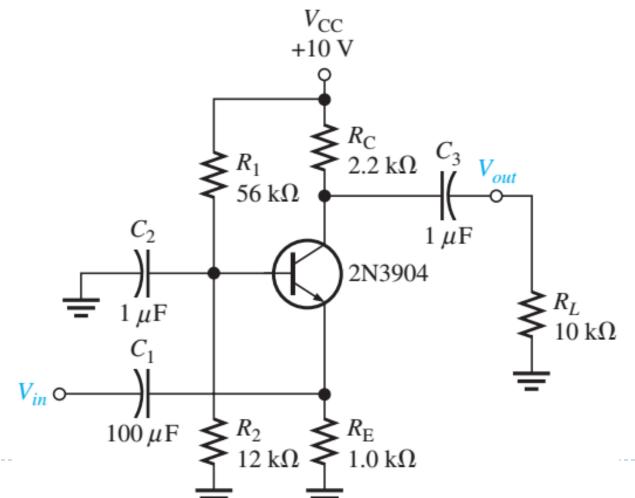
$$A_i \cong 1$$

#### Power Gain

Since the current gain is approximately I for the common-base amplifier and  $A_p = A_v A_i$  the power gain is approximately equal to the voltage gain.

$$A_P \cong A_v$$

Find the input resistance, voltage gain, current gain, and power gain for the amplifier in Figure.  $\beta_{DC} = 250$ .



First, find  $I_E$  so that you can determine  $r'_e$ . Then  $R_{in} \cong r'_e$ .

$$R_{\text{TH}} = \frac{R_1 R_2}{R_1 + R_2} = \frac{(56 \,\mathrm{k}\Omega)(12 \,\mathrm{k}\Omega)}{56 \,\mathrm{k}\Omega + 12 \,\mathrm{k}\Omega} = 9.88 \,\mathrm{k}\Omega$$

$$V_{\text{TH}} = \left(\frac{R_2}{R_2}\right) V_{\text{CO}} = \left(\frac{12 \,\mathrm{k}\Omega}{R_2}\right) V_{\text{CO}} =$$

$$V_{\text{TH}} = \left(\frac{R_2}{R_1 + R_2}\right) V_{\text{CC}} = \left(\frac{12 \,\text{k}\Omega}{56 \,\text{k}\Omega + 12 \,\text{k}\Omega}\right) 10 \,\text{V} = 1.76 \,\text{V}$$

$$I_{\text{E}} = \frac{V_{\text{TH}} - V_{\text{BE}}}{R_{\text{E}} + R_{\text{TH}}/\beta_{\text{DC}}} = \frac{1.76 \,\text{V} - 0.7 \,\text{V}}{1.0 \,\text{k}\Omega + 39.5 \,\Omega} = 1.02 \,\text{mA}$$

Therefore,

$$R_{in} \cong r'_e = \frac{25 \text{ mV}}{I_E} = \frac{25 \text{ mV}}{1.02 \text{ mA}} = 24.5 \Omega$$

Calculate the voltage gain as follows:

$$R_c = R_C \| R_L = 2.2 \text{ k}\Omega \| 10 \text{ k}\Omega = 1.8 \text{ k}\Omega$$

$$A_v = \frac{R_c}{r_c'} = \frac{1.8 \text{ k}\Omega}{24.5 \Omega} = 73.5$$

Also,  $A_i \cong 1$  and  $A_p \cong A_v = 73.5$