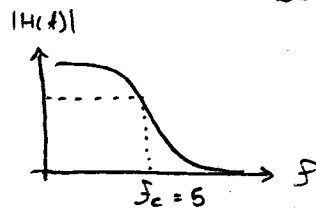


- Corresponding LPF (BW, CV-1 order) - $H(s)$
- $\omega_c = 1 \text{ rad/s}$
- Get $H(s)$
- Transform $H(s) \rightarrow$ desired Filter
Frequency transform

Example 4.2

Design a three-pole Butterworth low-pass Filter with a bandwidth of 5 Hz.

Solution :

Given :

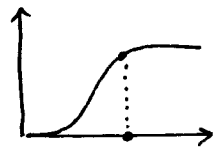
$$H(s) = \frac{(s - z_1)}{(s - p_1)(s - p_2)(s - p_3)}$$

$$H(s) = \frac{2s+1}{s^3 + 2s^2 + s + 4}$$

• • • MATLAB

Example 4.3

Design a 3-pole high-pass Filter with cutoff Frequency $\omega = 4 \text{ Hz}$

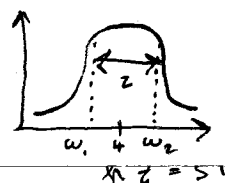
Solution :

$$\omega_c = 4 \text{ Hz}$$

$$\omega_c = 2\pi \times 4 \text{ (rad/s)}$$

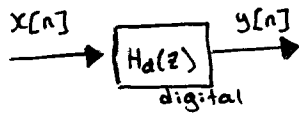
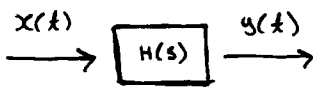
Example 4.4

In example 4.2, a three-pole Butterworth lowpass Filter was transformed to a bandpass Filter with the passband centered at $\omega = 4 \text{ Hz}$. The bandwidth is equal to 2 Hz.

Solution :

4.3 - Design of Digital Filters

1) Digital Filter



- DTFT. (discrete time Fourier transform)

$$X(\Omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\Omega n} \quad ; \quad \text{where} \quad -\pi \leq \Omega \leq \pi$$

- DFT (discrete Fourier transform)

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi k n / N)} \quad ; \quad \text{where} \quad \Omega = \frac{2\pi k}{N} \sim 2\pi f \left(\frac{1}{N}\right) \sim \omega$$

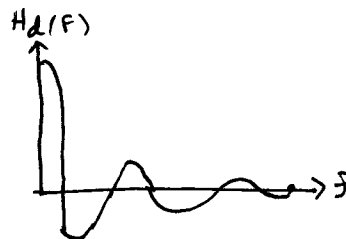
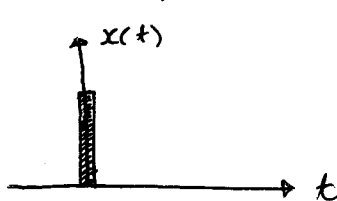
$$\Omega = \omega T$$

- ZT (z-transform)

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n} \quad ; \quad \text{where} \quad z = e^{j\Omega} = e^{j\omega T} = e^{sT}$$

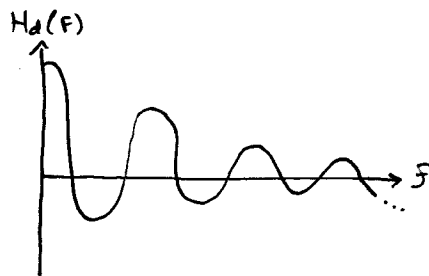
$$z = e^{sT}$$

- Input impulse



Finite number of steps

FIR: Finite impulse response filter



infinite number of steps

IIR: infinite impulse response

2) Design of IIR Filters

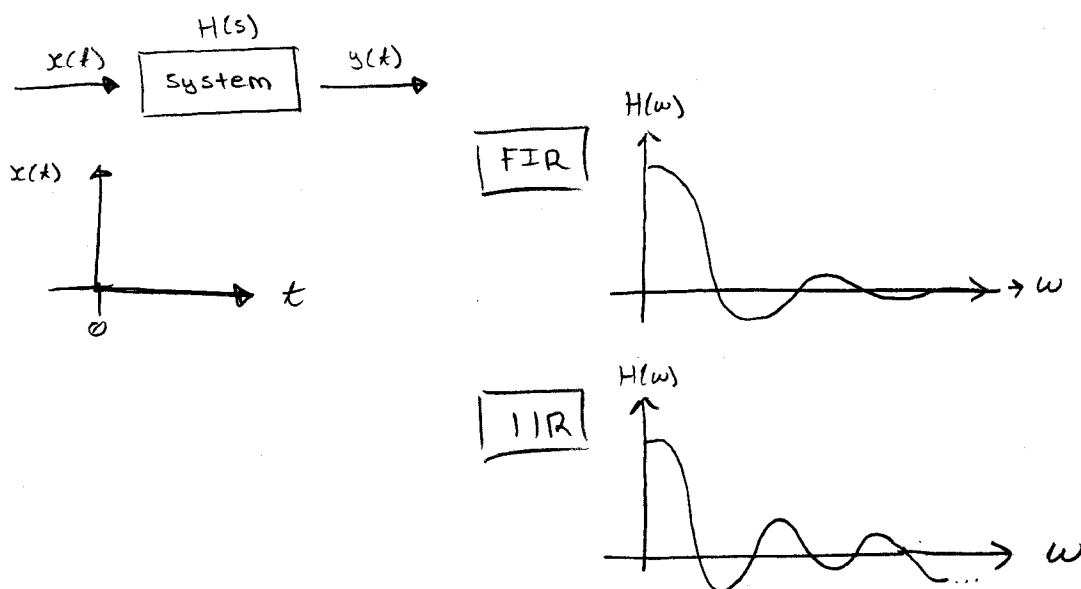
- analog filter

- digitization, $H_d(z)$

$$z = e^{sT}$$

$$\ln z = sT$$

$$s = \frac{1}{T} \ln(z)$$



Design of IIR Filters

- analog filter prototype

$$H(s), H(\omega)$$

- transform analog prototype

→ digital filter: $H_d(z)$

$$s = \frac{1}{T} \ln(z)$$

T = time data sample interval $1/f_s$

Taylor : $\ln z = z - z^2/2 + z^3/3 - z^4/4 + \dots$

Bilinear transformation

$$s = \frac{1}{T} \ln(z) \approx \left(\frac{2}{T}\right) \left(\frac{z-1}{z+1}\right)$$

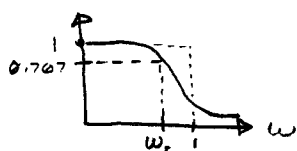
$$H_d(z) = H(z) = H\left(\frac{2}{T} \frac{z-1}{z+1}\right)$$

$$H(s) \sim h(x) \text{ (approximate)}$$

3) Warping Errors

$$H_d(z)$$

LPF with cutoff freq. ω_c



Digital Filter

$$\Omega_c = \omega_c T$$

$H_d(z)$ has cutoff freq.

$$\Omega_c = 2 \tan^{-1}(\omega_c T / 2) \neq \omega_c T$$

~ warping error

approx. bilinear transformation

gay

pre-warping

$$\Omega_c = 2 \tan^{-1}(\omega_c T / 2)$$

$$\Omega_c / 2 = \tan^{-1}(\omega_c T / 2)$$

$$\tan(\Omega_c / 2) = (\omega_c T / 2)$$

$$\omega_c = (2/T) \tan(\Omega_c / 2)$$

Cut-off Freq. of analog prototype

$$\omega_p = (2/T) \tan(\Omega_c / 2) = (2/T) \tan(\omega_c T / 2)$$

to replace $\omega_c \rightarrow \Omega_c = \omega_c T$

Example 4.8 Consider the two-pole Butterworth

Filter with transfer Function :

$$H(s) = \frac{\omega_c^2}{s^2 + \sqrt{2}\omega_c s + \omega_c^2}$$

Filter with $\omega_c = 2$, and $T = 0.2$

$$\rightarrow f_s = 1/T = 5 \text{ Hz}$$

digital Filter

$$H_d(z) = H(s) \Big|_{s=(2/T)(z-1)/(z+1)}$$

$$\begin{aligned} &= \frac{\omega_c^2}{\left(\frac{2}{T}(z-1)/(z+1)\right)^2 + \sqrt{2}\omega_c \left(\frac{2}{T}(z-1)/(z+1)\right) + \omega_c^2} \\ &= \frac{0.0308 z^2 + 0.0605 z + 0.0308}{z^2 - 1.4514 z + 0.5724} \end{aligned}$$

$$\Omega_c = (2/T) \tan^{-1}(\omega_c T / 2) = 0.3948$$

$$\text{Desired : } \Omega_c = \omega_c T = 2 \times 0.2 = 0.4$$

$$\begin{aligned} \omega_p &= (2/T) \tan(\Omega_c / 2) = (2/0.2) \tan(0.4/2) \\ &= 2.027 \text{ rad/s} \end{aligned}$$

$$\omega_p \rightarrow \omega_c \quad (\text{Prewarping})$$

$$H_d = \frac{0.0309 (z^2 + 2z + 1)}{z^2 - 1.444 z + 0.5682}$$

$$\Omega_c = \omega_c T = 0.4$$

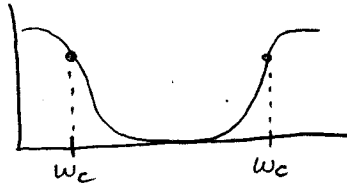
Solution

$$H_d(z) = -2$$

$$(-\pi \sim \pi)$$

Design OF IIR Filters in MATLAB

- bilinear



- butter (prototype, Freq. transformation, pre-warping)

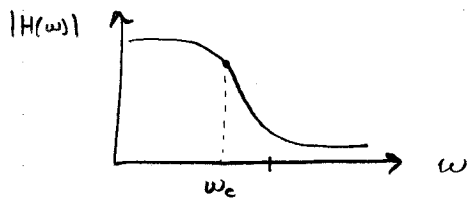
- cheby ($\omega_c = \omega_c T / \pi$)

- Filter

Solution

$$y(t) = 1 + \cos t + \cos(5t)$$

$\omega = 1$ $\omega = 5$



$$1 < \omega_c < 5$$