

$$y(t) = 3x(t)e^t + \text{I.E. (initial energy)}$$

- Causal
- Non-causal : offline processing
- Linear
- Superposition

Input

$$x_1 = u$$

$$x_2 = 3u$$

$$x_3 = x_1 + x_2 = 4u$$

Output

$$y_1$$

$$y_2$$

$$y_3 = y_1 + y_2$$

- Time-invariance

Input

$$x_1 = x(t)$$

$$x_2 = x(t-2)$$

Output

$$y_1 = 3x(t)e^x$$

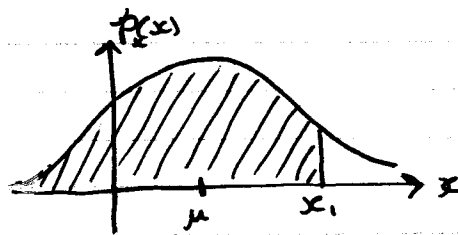
$$y_2 = 3x(t-2)e^x$$

✗

$$y_3(t-2) = 3x(t-2)e^{t-2}$$

### (2.3) Review of Probability Concepts

pdf - Prob. distribution Function

 $p_x(x)$ 

Prob.

$$P_x(x \leq x_1) = \int_{-\infty}^{x_1} p_x(x) dx, \quad \text{Gaussian pdf}$$

### 2) Statistical moments

- 1st order moment

$$\mu = \int_{-\infty}^{\infty} x p_x(x) dx$$

$$= E\{x\}$$

↖ expectation

- 2nd-order moment

$$\text{Var}(x) = \sigma^2 = \int_{-\infty}^{\infty} (x-\mu)^2 p_x(x) dx$$

Variance

$$\text{Standard dev. } \sigma = E\{(x-\mu)^2\}$$

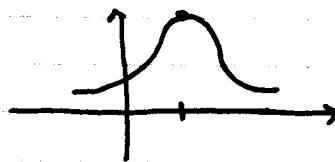
$$\begin{aligned}\sigma^2 &= E\{x^2 - 2x\mu + \mu^2\} \\ &= E\{x^2\} - 2\mu E\{x\} + \mu^2 \\ &= E\{x^2\} - 2\mu^2 + \mu^2\end{aligned}$$

- 3<sup>rd</sup> moment

$$\mu_3 = E\{(x-\mu)^3\}$$

$$SK = \frac{\mu_3}{\sigma^3}; \text{ skewness}$$

(unitless)

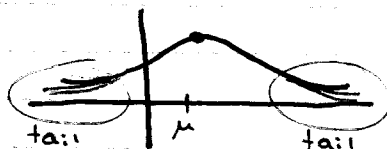


- 4<sup>th</sup> moment

$$\mu_4 = E\{(x-\mu)^4\}$$

Kurtosis

$$KU = \frac{\mu_4}{\sigma^4}$$



Two Variables

$$\text{COV}(x_1, x_2)$$

Coefficient

$$\rho_{12} = \frac{\text{COV}(x_1, x_2)}{\sqrt{\text{Var}(x_1)\text{Var}(x_2)}}$$

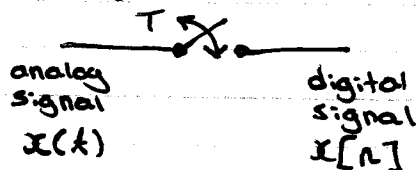
## (2.4) Sampling + Aliasing

Collect data, analog signal

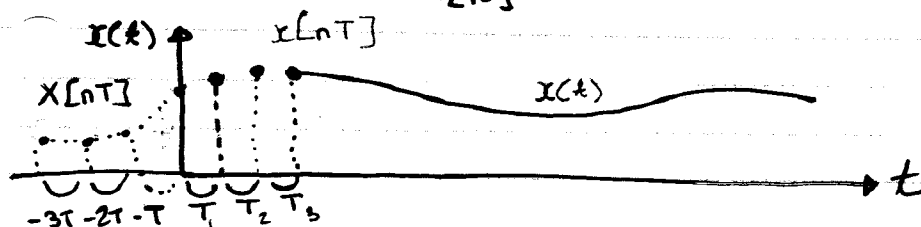
Computer → digital signal

- Sampling

digital switch



round brackets for analog signal  
square brackets for digital signal



$T$  = time interval (sec)

Analog to Digital converter  
(ADC)

← (DAC) inverse  
- control motor, etc.

$$x[nT] = x(t)|_{t=nT}$$

for  $n = 0, 1, 2, \dots$

(in theory,  $n = -2, -1, 0, 1, \dots$ )

Sampling Frequency,

$$f_s = \frac{1}{T} \text{ (Hz)}$$

$$f_s = 20 \text{ Hz}, T = \frac{1}{f_s} = \frac{1}{20} = 0.05 \text{ sec}$$

$$f_s = 15000 \text{ Hz}, T = \frac{1}{f_s} \dots$$

$x[n]$

$n$  = discrete time value

$$f_s \uparrow \uparrow, T = \frac{1}{f_s} \downarrow \downarrow$$

data size  $\uparrow \uparrow$ , processing speed  $\downarrow \downarrow$

$$T \geq 0.003 \text{ sec}$$

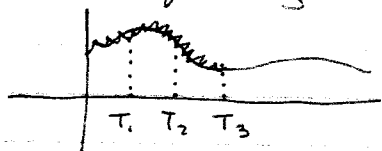
online control, real time

machine learning

(diagnosis  
prognosis)

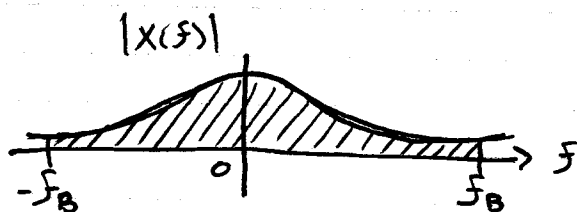
$$\text{If } f_s \downarrow \downarrow, T = \frac{1}{f_s} \uparrow \uparrow$$

high frequency components would be lost

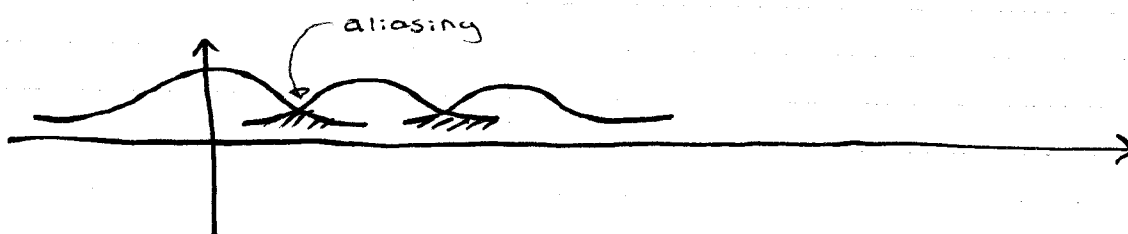


- Actual signals are time-limited

$$-f_B \leq f \leq f_B$$



Sampling ( $2f_B$ )

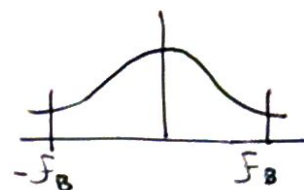


2) Aliasing

All the signals are time limited

$$f_s = 15,000 \text{ Hz}$$

Frequency components would be non-limited

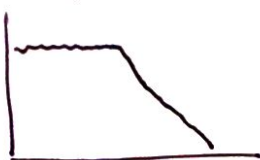


$$x[n] = x[nT] = x(t) |_{t=nT}$$

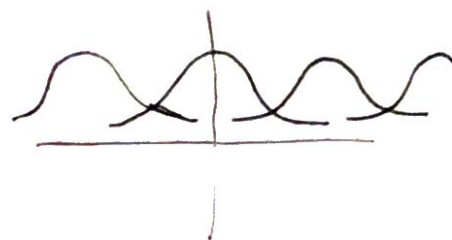
$$n = \dots -2, -1, 0, 1, 2 \dots$$

$$f_{\max} = f_B \text{ Hz}$$

FFT



Sampled continuously:



Most important Freq. Components  $(-f_B, f_B)$

$$f_s = 2f_B$$

$$[0, f_B] \text{ or } [-f_B, f_B]$$

Contains extra Frequency components overlapped  
From high Freq. region to the low Freq. region.  
 $(f_B, +\infty)$

~ aliasing

$$f_s = 40 \text{ shots/sec}$$

$$= 20 \sim 25 \text{ pictures/sec}$$

Before doing ADC

anti-aliasing Filter to remove Freq. (Low pass Filter) components higher than  $f_B$

Nyquist Freq

$$f_n = 2f_B$$

Sampling Freq

$$f_s \geq f_n = 2f_B$$

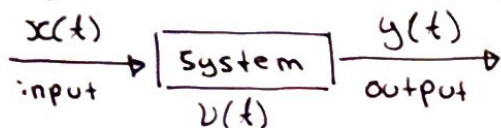
↖ analog filter

CD

20 kHz

 $f_s = 44.1 \text{ kHz}$ 

(2.4) - Convolution representation



$$y(t) = x(t) \otimes v(t)$$

1) Convolution for continuous signals

$$x(t) \otimes v(t) = \int_{-\infty}^{\infty} x(\tau) v(t-\tau) d\tau$$

$$\begin{cases} x(t) = 0, & \text{if } t < 0 \\ v(t) = 0, & \text{if } t < 0 \end{cases} \quad \begin{cases} \text{if } \tau < 0 \\ x(\tau) = 0 \end{cases}$$

$$v(t-\tau) = 0, \text{ if } t-\tau < 0 \quad \begin{cases} t-\tau < 0 \\ t < \tau \end{cases}$$

$$x(t) \otimes v(t) = \int_0^t x(\tau) v(t-\tau) d\tau$$

$$\int_{-\infty}^{\infty} |x(t)| dt < \infty$$

$$\int_{-\infty}^{\infty} |v(t)| dt < \infty$$

2) Conv. for discrete signals

$$x[n], v[n]$$

$$\dots, -2, -1, 0 < n < 1, 2, \dots$$

$$x[n] \otimes v[n] = \sum_{i=-\infty}^{\infty} x[i] v[n-i]$$

$$\text{if } x[n] = 0; n < 0$$

$$\text{if } v[n] = 0; n < 0$$

$$x[i] = 0 \text{ if } i < 0$$

$$v[n-i] = 0 \text{ if } n-i < 0, i > n$$

$$x[n] \otimes v[n] = \begin{cases} \sum_{i=0}^n x[i] v[n-i] & ; i = 0, 1, 2, \dots, n \\ 0 & ; \text{if } i < 0, i > n \end{cases}$$

## Convolution computation procedures :

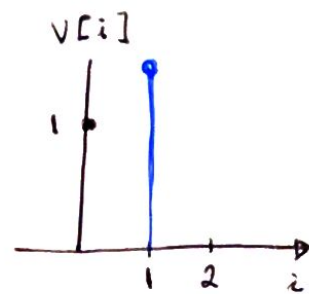
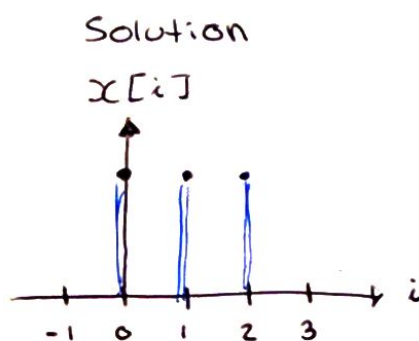
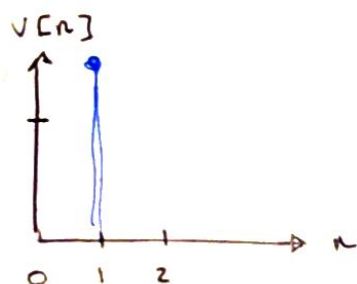
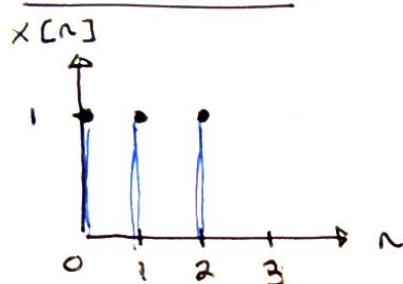
- 1) Changing the discrete time index  $n$  to  $i$  in signals  $x[n]$  and  $v[n]$ . The resulting signals  $x[i]$  and  $v[i]$  are then functions of the discrete-time index  $i$ .
- 2) Determining  $v[n-i]$ . The signal  $v[n-i]$  is a folded and shifted version of the signal  $v[i]$ . More precisely,  $v[-i]$  is  $v[i]$  folded about the vertical axis, and  $v[n-i]$  is  $v[-i]$  shifted by  $n$  steps. If  $n > 0$ ,  $v[n-i]$  is an  $n$ -step right shift of  $v[-i]$ . In contrast, if  $n < 0$ ,  $v[n-i]$  is an  $n$ -step left shift of  $v[-i]$ .
- 3) Computing the convolution

$V[n] \rightarrow V[i]$ , changed index

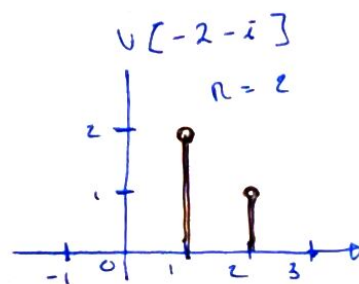
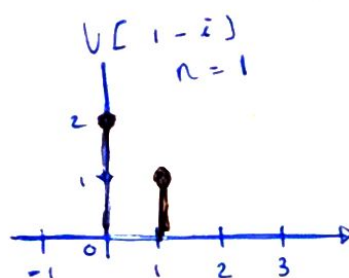
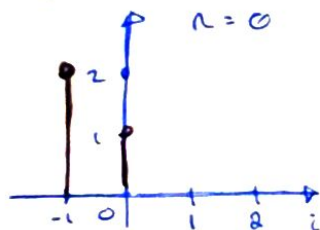
$\rightarrow V[-i]$  Folding

$\rightarrow V[n-i]$  Shifting

### Example 1



$V[-i] = V[0-i]$   
 $n=0$



... etc.

### 3) Conv. Properties

- Associativity,

$x[n], v[n], w[n]$

$$x[n] \otimes (v[n] \otimes w[n])$$

$$= (x[n] \otimes v[n]) \otimes w[n]$$

- Commutativity,

$$x[n] \otimes v[n] = v[n] \otimes x[n]$$

$$\sum_{i=-\infty}^{\infty} x[i]v[n-i] = \sum_{i=-\infty}^{\infty} v[i]x[n-i]$$

- Distributivity with Addition

$$x[n] \otimes (v[n] + w[n]) = x[n] \otimes v[n] + x[n] \otimes w[n] \quad (5)$$