

# Signals and Systems

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## Lecture 13: Continuous-time Fourier Series

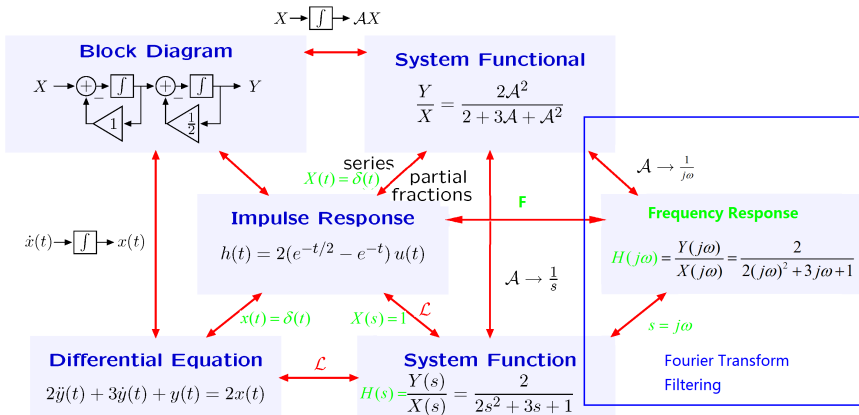
Instructor: Prof. Yunlong Cai  
Zhejiang University

04/08/2025

Adapted from the materials provided on  
the MIT OpenCourseWare

# Review: System Representations

Relations among representations.



# Fourier Representations

Fourier series represent **signals** in terms of **sinusoids**.

→ leads to a new representation for **systems** as **filters**.

# Harmonics

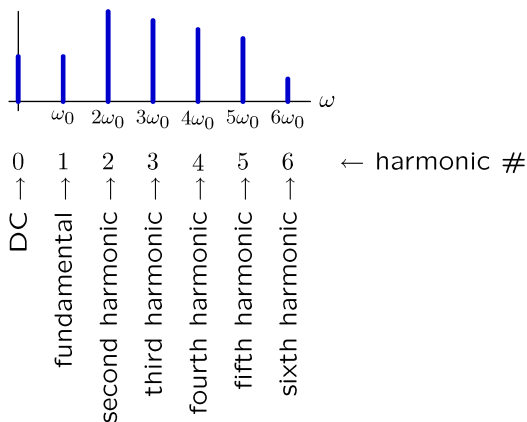
A **harmonically related** set of complex exponentials is a set of periodic exponentials with fundamental frequencies that are all multiples of a single positive frequency  $\omega_0$ :

$$\phi_k(t) = e^{jk\omega_0 t}, \quad k = 0, \pm 1, \pm 2, \dots$$

- $k = 0$ , DC component;
- $k = \pm 1$ , the first harmonic components;
- $k = \pm 2$ , the second harmonic components;
- ...

# Fourier Series

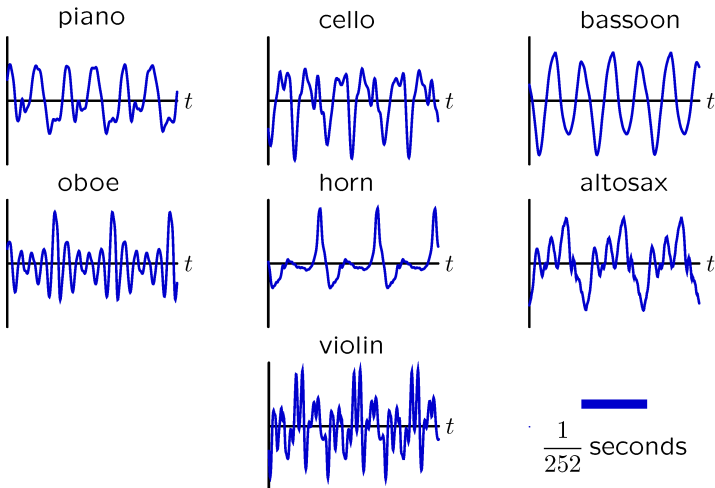
Representing signals by their harmonic components.



# Musical Instruments

Harmonic content is natural way to describe some kinds of signals.

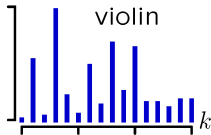
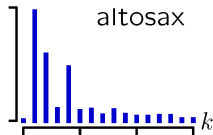
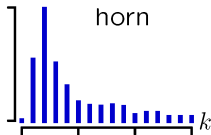
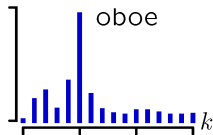
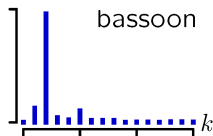
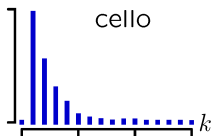
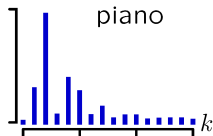
Ex: musical instruments (<http://theremin.music.uiowa.edu/MIS>)



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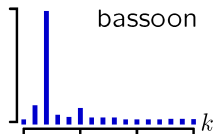
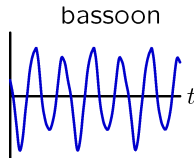
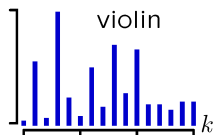
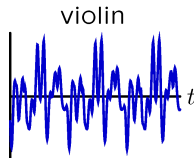
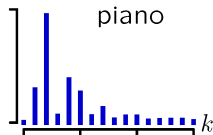
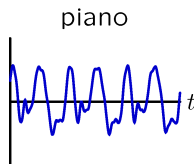
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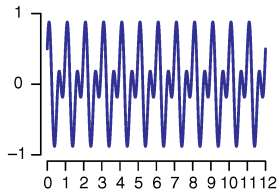




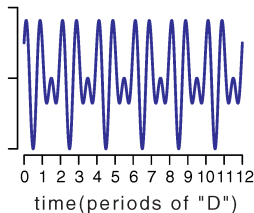
# Harmonics

Harmonic structure determines consonance and dissonance.

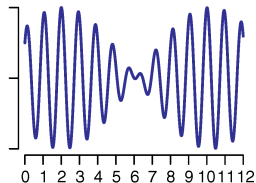
octave (D+D')



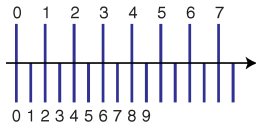
fifth (D+A)



D+E $\flat$

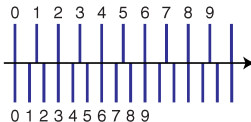


D'



D

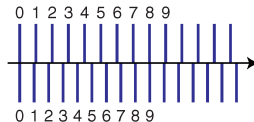
A



D

harmonics

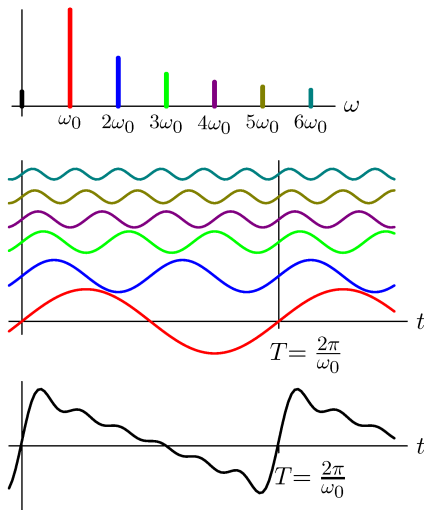
E $\flat$



D

# Harmonic Representations

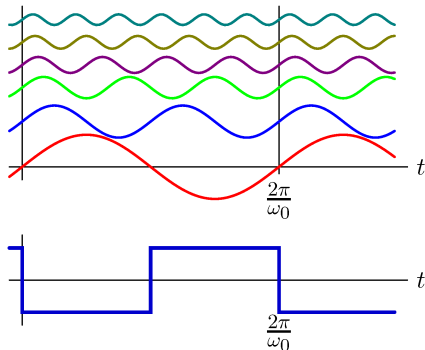
What signals can be represented by sums of harmonic components?



Only periodic signals: all harmonics of  $\omega_0$  are periodic in  $T = 2\pi/\omega_0$ .

# Harmonic Representations

Is it possible to represent ALL periodic signals with harmonics?  
What about discontinuous signals?



Fourier claimed YES — even though all harmonics are continuous!  
Lagrange ridiculed the idea that a discontinuous signal could be written as a sum of continuous signals.

We will assume the answer is YES and see if the answer makes sense.

# Separating harmonic components

Underlying properties.

1. Multiplying two harmonics produces a new harmonic with the same fundamental frequency:

$$e^{jk\omega_0 t} \times e^{jl\omega_0 t} = e^{j(k+l)\omega_0 t}.$$

2. The integral of a harmonic over any time interval with length equal to a period  $T$  is zero unless the harmonic is at DC:

$$\begin{aligned} \int_{t_0}^{t_0+T} e^{jk\omega_0 t} dt &\equiv \int_T e^{jk\omega_0 t} dt = \begin{cases} 0, & k \neq 0 \\ T, & k = 0 \end{cases} \\ &= T\delta[k] \end{aligned}$$

# Separating harmonic components

Assume that  $x(t)$  is periodic in  $T$  and is composed of a weighted sum of harmonics of  $\omega_0 = 2\pi/T$ .

$$x(t) = x(t + T) = \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 kt}$$

Then

$$\begin{aligned} \int_T x(t) e^{-jl\omega_0 t} dt &= \int_T \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 kt} e^{-j\omega_0 lt} dt \\ &= \sum_{k=-\infty}^{\infty} a_k \int_T e^{j\omega_0 (k-l)t} dt \\ &= \sum_{k=-\infty}^{\infty} a_k T \delta[k-l] = Ta_l \end{aligned}$$

Therefore

$$a_k = \frac{1}{T} \int_T x(t) e^{-j\omega_0 kt} dt = \frac{1}{T} \int_T x(t) e^{-j\frac{2\pi}{T} kt} dt$$

# Fourier Series

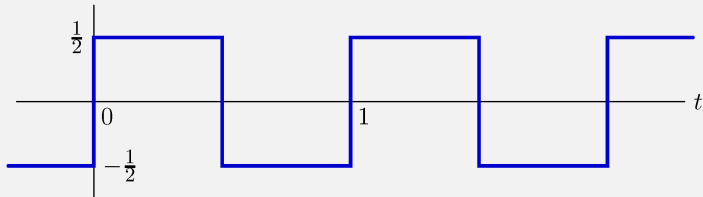
Determining harmonic components of a periodic signal.

$$a_k = \frac{1}{T} \int_T x(t) e^{-j \frac{2\pi}{T} kt} dt \quad (\text{"analysis" equation})$$

$$x(t) = x(t + T) = \sum_{k=-\infty}^{\infty} a_k e^{j \frac{2\pi}{T} kt} \quad (\text{"synthesis" equation})$$

# Check Yourself

Let  $a_k$  represent the Fourier series coefficients of the following square wave.

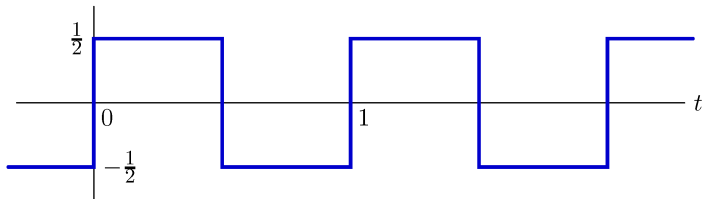


How many of the following statements are true?

1.  $a_k = 0$  if  $k$  is even
2.  $a_k$  is real-valued
3.  $|a_k|$  decreases with  $k^2$
4. there are an infinite number of non-zero  $a_k$
5. all of the above

# Check Yourself

Let  $a_k$  represent the Fourier series coefficients of the following square wave.



$$\begin{aligned} a_k &= \int_T x(t) e^{-j\frac{2\pi}{T}kt} dt = -\frac{1}{2} \int_{-\frac{1}{2}}^0 e^{-j2\pi kt} dt + \frac{1}{2} \int_0^1 e^{-j2\pi kt} dt \\ &= \frac{1}{j4\pi k} (2 - e^{j\pi k} - e^{-j\pi k}) \\ &= \begin{cases} \frac{1}{j\pi k} ; & \text{if } k \text{ is odd} \\ 0 ; & \text{otherwise} \end{cases} \end{aligned}$$



# Check Yourself

Let  $a_k$  represent the Fourier series coefficients of the following square wave.

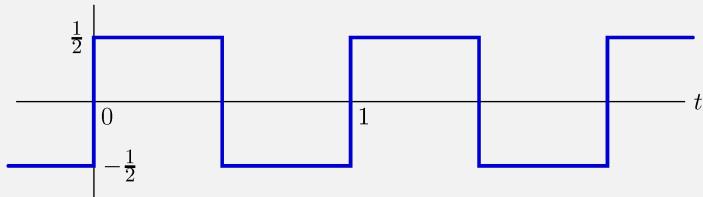
$$a_k = \begin{cases} \frac{1}{j\pi k} & \text{if } k \text{ is odd} \\ 0 & \text{otherwise} \end{cases}$$

How many of the following statements are true?

1.  $a_k = 0$  if  $k$  is even ✓
2.  $a_k$  is real-valued ✗
3.  $|a_k|$  decreases with  $k^2$  ✗
4. there are an infinite number of non-zero  $a_k$  ✓
5. all of the above ✗

# Check Yourself

Let  $a_k$  represent the Fourier series coefficients of the following square wave.



How many of the following statements are true? 2

1.  $a_k = 0$  if  $k$  is even ✓
2.  $a_k$  is real-valued ✗
3.  $|a_k|$  decreases with  $k^2$  ✗
4. there are an infinite number of non-zero  $a_k$  ✓
5. all of the above ✗

# Check Yourself

Determine the Fourier series representation for

$$x(t) = \cos 4\pi t + 2 \sin 8\pi t$$

# Check Yourself

Determine the Fourier series representation for

$$x(t) = \cos 4\pi t + 2 \sin 8\pi t$$

$$\begin{aligned} \text{Euler's relation} &= \frac{1}{2} [e^{j4\pi t} + e^{-j4\pi t}] + \frac{2}{2j} [e^{j8\pi t} - e^{-j8\pi t}] \\ \text{(memorize!)} & \end{aligned}$$
$$\omega_0 = 4\pi \quad T = \frac{2\pi}{\omega_0} = \frac{2\pi}{4\pi} = \frac{1}{2}$$

$$\begin{aligned} a_0 &= 0 \\ a_1 &= \frac{1}{2} \\ a_{-1} &= \frac{1}{2} \\ a_2 &= \frac{1}{j} \\ a_{-2} &= -\frac{1}{j} \\ a_3 &= 0 \\ a_{-3} &= 0 \end{aligned}$$

# Check Yourself

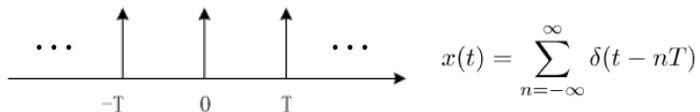
Determine the Fourier series representation for



$$x(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

# Check Yourself

Determine the Fourier series representation for


$$x(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

$$\begin{aligned} a_k &= \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt = \frac{1}{T} \int_{-T/2}^{T/2} \delta(t) e^{-jk\omega_0 t} dt \\ &= \frac{1}{T} \quad \text{for all } k ! \end{aligned}$$

$$\Downarrow$$
$$x(t) = \frac{1}{T} \sum_{k=-\infty}^{\infty} e^{jk\omega_0 t}$$

— All components have:

- (1) the same amplitude,  
&
- (2) the same phase.

# Fourier Series Properties

If a signal is differentiated in time, its Fourier coefficients are multiplied by  $j\frac{2\pi}{T}k$ .

Proof: Let

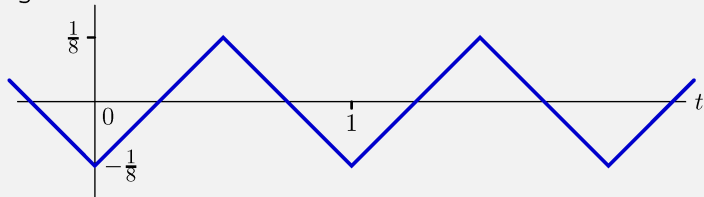
$$x(t) = x(t + T) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt}$$

then

$$\dot{x}(t) = \dot{x}(t + T) = \sum_{k=-\infty}^{\infty} \left( j\frac{2\pi}{T}k a_k \right) e^{j\frac{2\pi}{T}kt}$$

# Check Yourself

Let  $b_k$  represent the Fourier series coefficients of the following triangle wave.



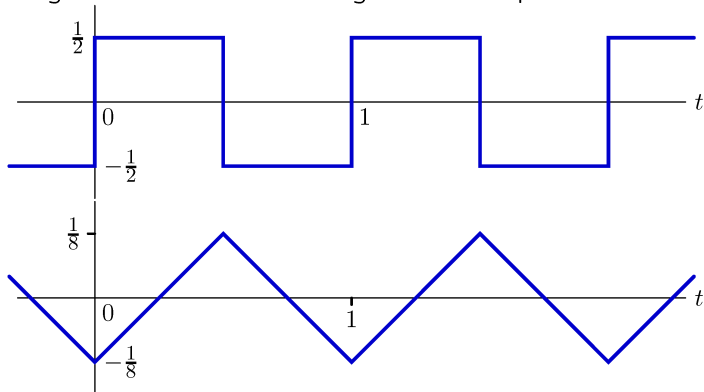
How many of the following statements are true?

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2.  $b_k$  is real-valued
3.  $|b_k|$  decreases with  $k^2$
4. there are an infinite number of non-zero  $b_k$
5. all of the above



# Check Yourself

The triangle waveform is the integral of the square wave.



Therefore the Fourier coefficients of the triangle waveform are  $\frac{1}{j2\pi k}$  times those of the square wave.

$$b_k = \frac{1}{jk\pi} \times \frac{1}{j2\pi k} = \frac{-1}{2k^2\pi^2} ; k \text{ odd}$$

# Check Yourself

Let  $b_k$  represent the Fourier series coefficients of the following triangle wave.

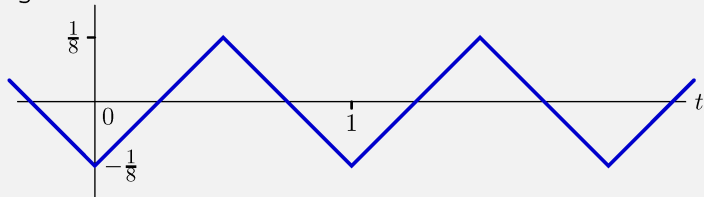
$$b_k = \frac{-1}{2k^2\pi^2} ; k \text{ odd}$$

How many of the following statements are true?

1.  $b_k = 0$  if  $k$  is even ✓
2.  $b_k$  is real-valued ✓
3.  $|b_k|$  decreases with  $k^2$  ✓
4. there are an infinite number of non-zero  $b_k$  ✓
5. all of the above ✓

# Check Yourself

Let  $b_k$  represent the Fourier series coefficients of the following triangle wave.



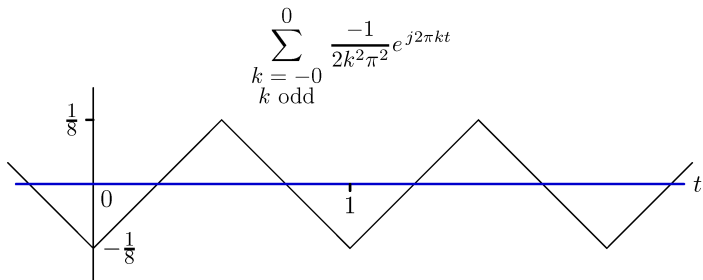
How many of the following statements are true? 5

1.  $b_k = 0$  if  $k$  is even ✓
2.  $b_k$  is real-valued ✓
3.  $|b_k|$  decreases with  $k^2$  ✓
4. there are an infinite number of non-zero  $b_k$  ✓
5. all of the above ✓

# Fourier Series

One can visualize convergence of the Fourier Series by incrementally adding terms.

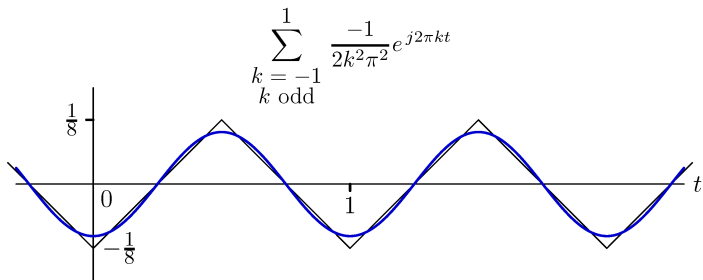
Example: triangle waveform



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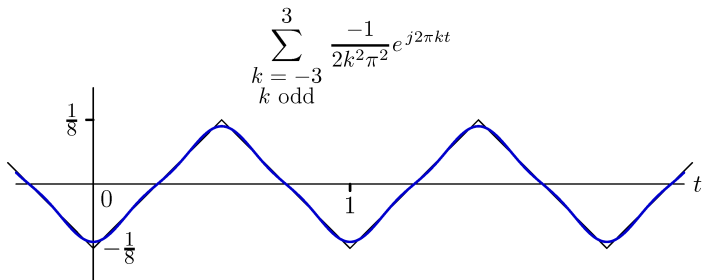
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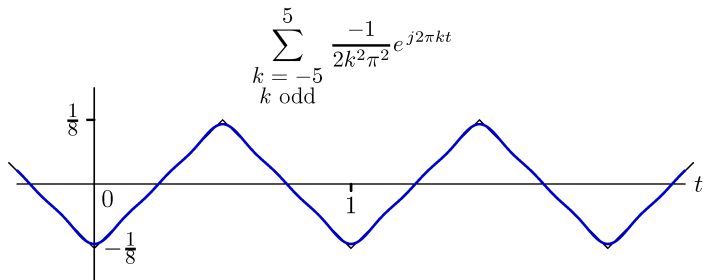
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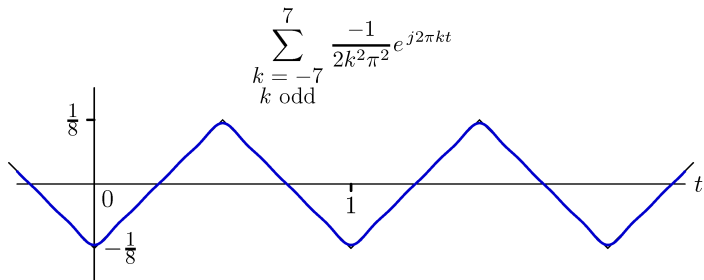
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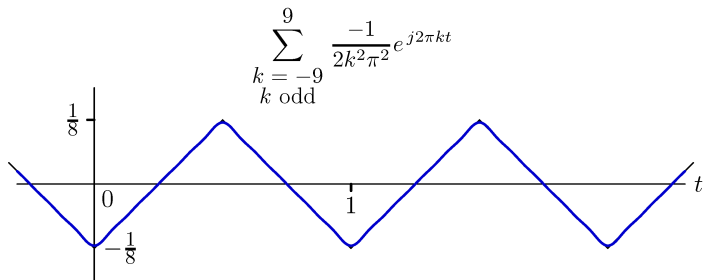




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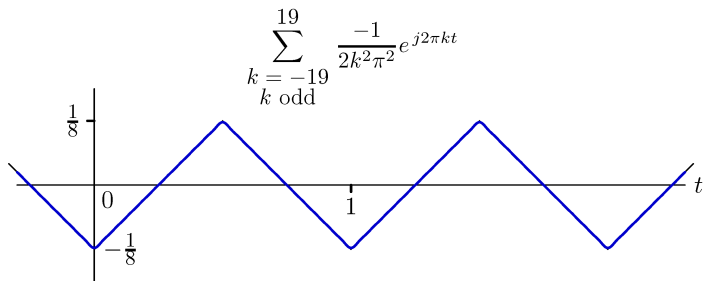
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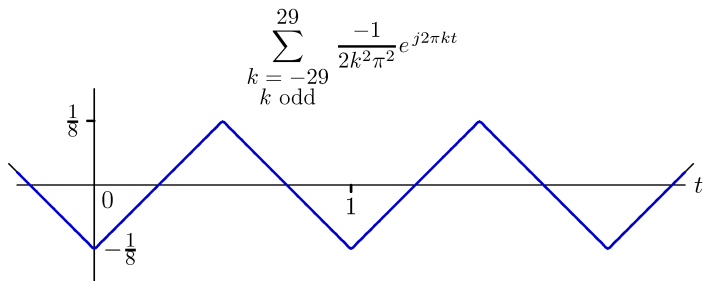
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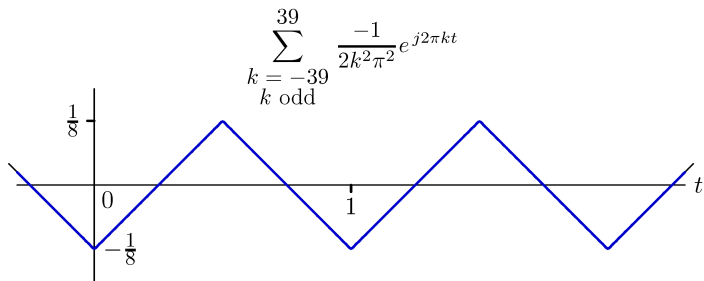
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Example: triangle waveform

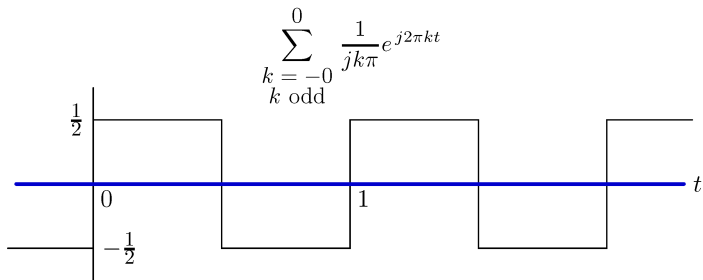


Fourier series representations of functions with discontinuous slopes converge toward functions with discontinuous slopes.

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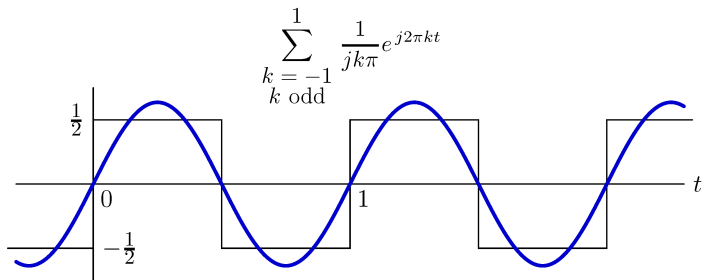
Example: square wave



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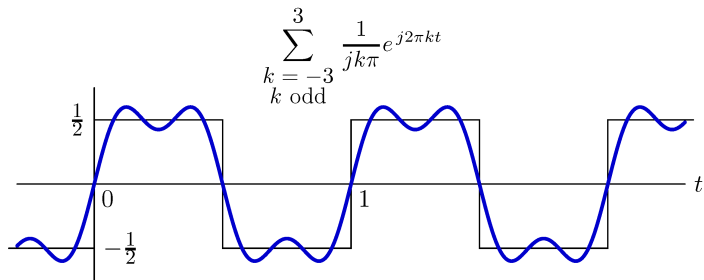
Example: square wave



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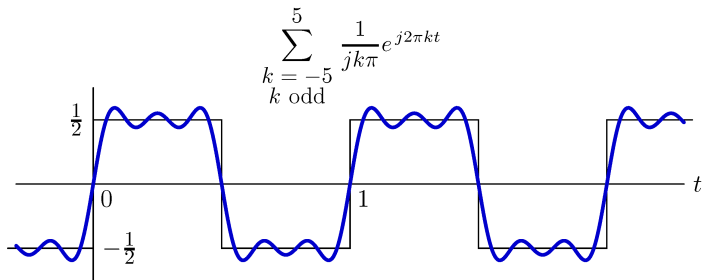
Example: square wave



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One can visualize convergence of the Fourier Series by incrementally adding terms.

Example: square wave

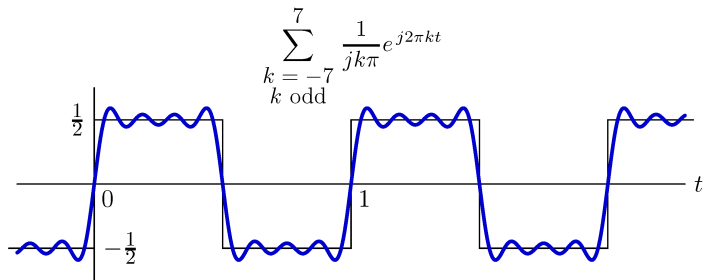




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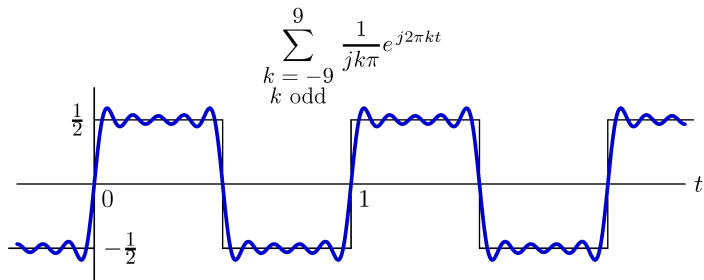
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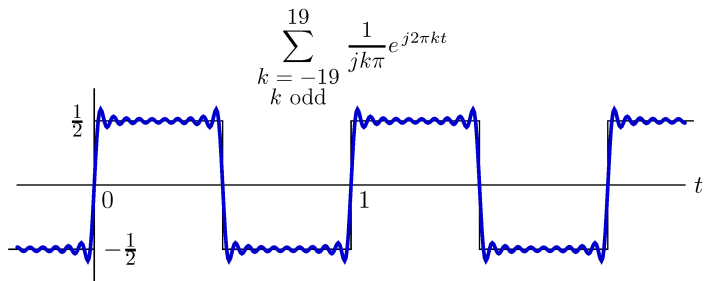
Example: square wave



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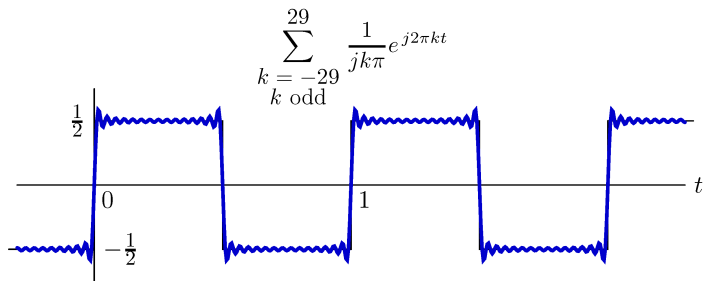
Example: square wave



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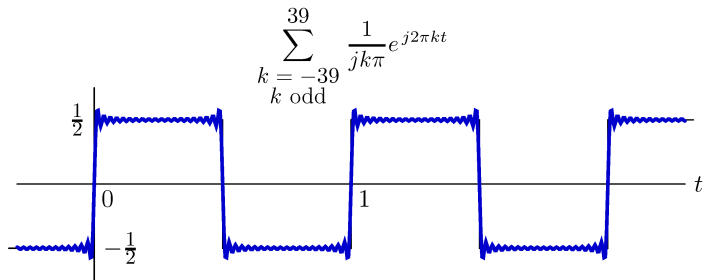
Example: square wave



# Fourier Series

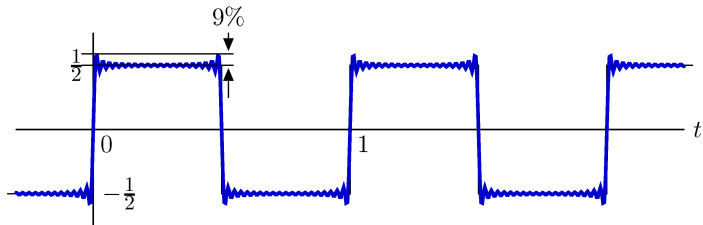
One can visualize convergence of the Fourier Series by incrementally adding terms.

Example: square wave



# Fourier Series

Partial sums of Fourier series of discontinuous functions “ring” near discontinuities: Gibb’s phenomenon.



This ringing results because the magnitude of the Fourier coefficients is only decreasing as  $\frac{1}{k}$  (while they decreased as  $\frac{1}{k^2}$  for the triangle).

You can decrease (and even eliminate the ringing) by decreasing the magnitudes of the Fourier coefficients at higher frequencies.

# Gibbs Phenomenon

$$x_N(t) = \sum_{k=-N}^N a_k e^{jk\omega_0 t}$$

- As  $N \rightarrow \infty$ ,  $x_N(t)$  exhibits Gibb's phenomenon at points of discontinuity;
- As  $N$  increases, the ripples in the partial sums become compressed toward the discontinuity;
- For any finite value of  $N$ , the peak amplitude of the ripples remains constant.

# Fourier Series: Convergence

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t} \quad \leftarrow \quad x_N(t) = \sum_{k=-N}^N a_k e^{jk\omega_0 t}$$

$$\rightarrow e_N(t) = x(t) - x_N(t) = x(t) - \sum_{k=-N}^N a_k e^{jk\omega_0 t}$$

$$\rightarrow E_N = \int_T |e_N(t)|^2 dt$$

$$\rightarrow \left\{ \begin{array}{l} E_N \rightarrow 0, \quad N \rightarrow \infty \\ a_k = \frac{1}{T} \int_T x(t) e^{-jk\omega_0 t} dt \\ \int_T |x(t)|^2 dt < \infty \end{array} \right.$$



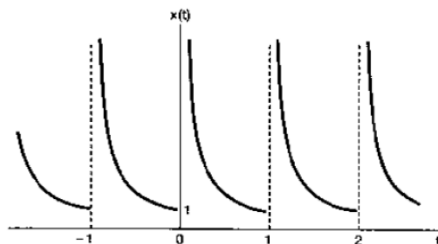
# Fourier Series: The Dirichlet Conditions

- **Condition 1:**

Over any period,  $x(t)$  must be absolutely integrable

$$\int_T |x(t)| dt < \infty$$

## Counter example



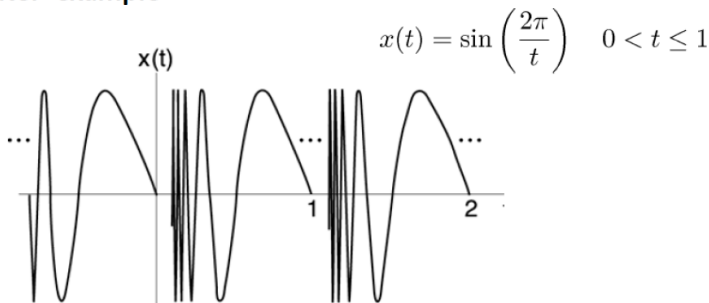
$$x(t) = \frac{1}{t}, \quad 0 < t \leq 1$$

# Fourier Series: The Dirichlet Conditions

- **Condition 2:**

In any finite interval of time,  $x(t)$  is of bounded variation; i.e. there are no more than a finite number of maxima and minima during any single period of the signal.

## Counter example

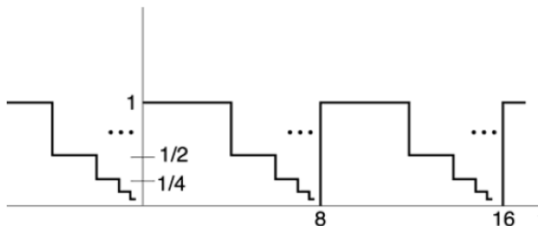


# Fourier Series: The Dirichlet Conditions

- **Condition 3:**

In any finite interval of time, there are only a finite number of discontinuities. Each of these discontinuities is finite.

## Counter example



# Fourier Series: Summary

Fourier series represent periodic signals as sums of sinusoids.

- valid for an extremely large class of periodic signals
- valid even for discontinuous signals such as square wave

However, convergence as  $\#$  harmonics increases can be complicated.

# Filtering

The output of an LTI system is a “filtered” version of the input.

Input: Fourier series  $\rightarrow$  sum of complex exponentials.

$$x(t) = x(t + T) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt}$$

Complex exponentials: eigenfunctions of LTI systems.

$$e^{j\frac{2\pi}{T}kt} \rightarrow H(j\frac{2\pi}{T}k)e^{j\frac{2\pi}{T}kt}$$

Output: same eigenfunctions, amplitudes/phases set by system.

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt} \rightarrow y(t) = \sum_{k=-\infty}^{\infty} a_k H(j\frac{2\pi}{T}k) e^{j\frac{2\pi}{T}kt}$$

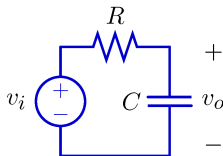
# Filtering

Notion of a filter.

LTI systems

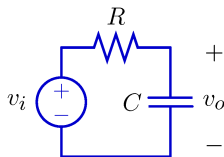
- cannot create new frequencies.
- can scale magnitudes and shift phases of existing components.

Example: Low-Pass Filtering with an RC circuit



# Lowpass Filter

Calculate the frequency response of an RC circuit.



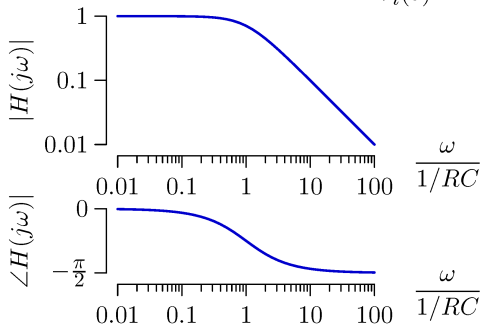
$$\text{KVL: } v_i(t) = Ri(t) + v_o(t)$$

$$\text{C: } i(t) = C\dot{v}_o(t)$$

$$\text{Solving: } v_i(t) = RC\dot{v}_o(t) + v_o(t)$$

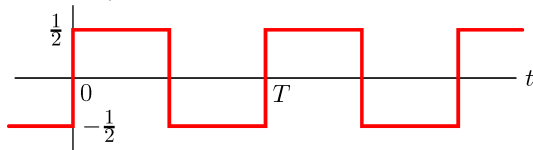
$$V_i(s) = (1 + sRC)V_o(s)$$

$$H(s) = \frac{V_o(s)}{V_i(s)} = \frac{1}{1 + sRC}$$

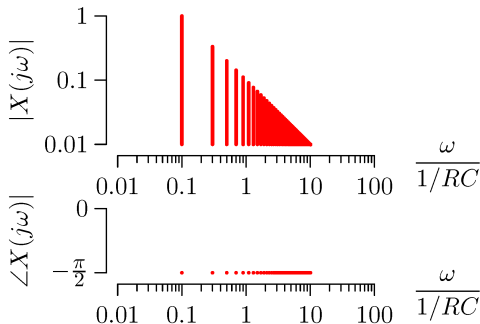


# Lowpass Filtering

Let the input be a square wave.



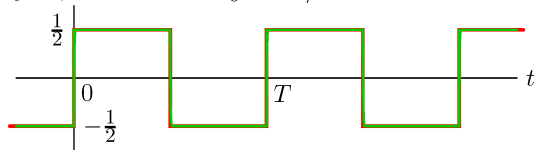
$$x(t) = \sum_{k \text{ odd}} \frac{1}{j\pi k} e^{j\omega_0 k t} ; \quad \omega_0 = \frac{2\pi}{T}$$



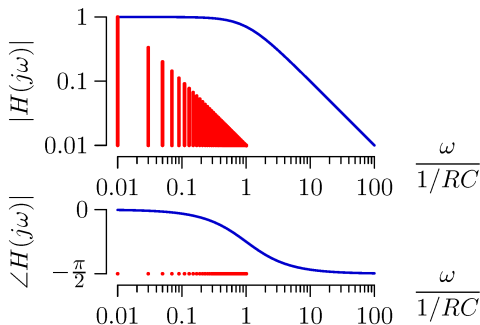


# Lowpass Filtering

Low frequency square wave:  $\omega_0 \ll 1/RC$ .

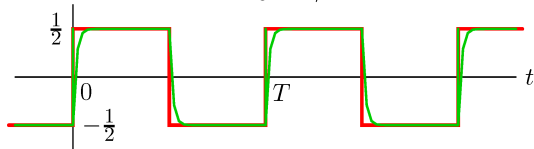


$$x(t) = \sum_{k \text{ odd}} \frac{1}{j\pi k} e^{j\omega_0 k t} ; \quad \omega_0 = \frac{2\pi}{T}$$

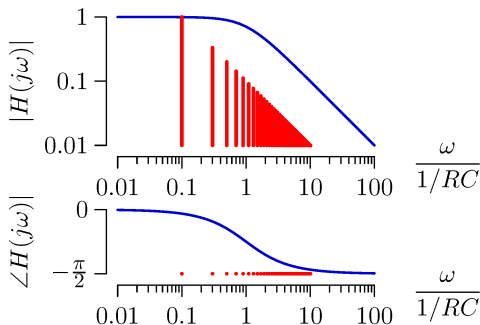


# Lowpass Filtering

Higher frequency square wave:  $\omega_0 < 1/RC$ .

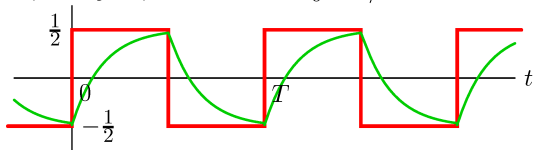


$$x(t) = \sum_{k \text{ odd}} \frac{1}{j\pi k} e^{j\omega_0 k t} ; \quad \omega_0 = \frac{2\pi}{T}$$

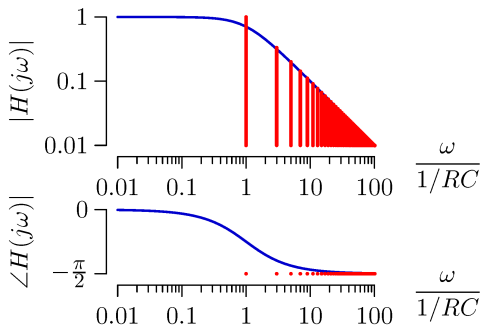


# Lowpass Filtering

Still higher frequency square wave:  $\omega_0 = 1/RC$ .

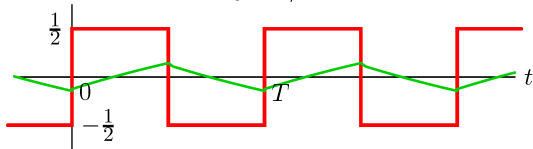


$$x(t) = \sum_{k \text{ odd}} \frac{1}{j\pi k} e^{j\omega_0 k t} ; \quad \omega_0 = \frac{2\pi}{T}$$

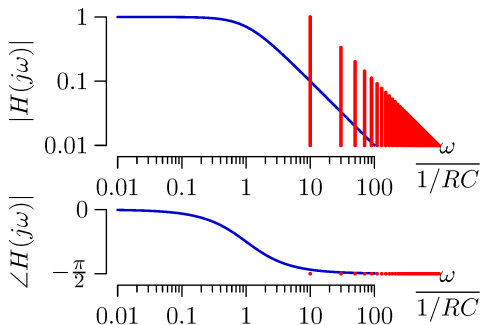


# Lowpass Filtering

High frequency square wave:  $\omega_0 > 1/RC$ .



$$x(t) = \sum_{k \text{ odd}} \frac{1}{j\pi k} e^{j\omega_0 k t} ; \quad \omega_0 = \frac{2\pi}{T}$$



# Fourier Series: Summary

Fourier series represent signals by their frequency content.

Representing a signal by its frequency content is useful for many signals, e.g., music.

Fourier series motivate a new representation of a system as a filter.

# Assignments

- Reading Assignment: Ch. 1.3, Ch. 3.0 - 3.5, Ch. 3.9
- Homework 7