# Signals and Systems

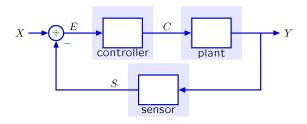
Lecture 12: CT Feedback and Control

Instructor: Prof. Yunlong Cai Zhejiang University

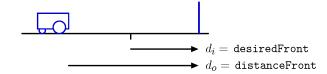
04/03/2025
Partly adapted from the materials provided on the MIT OpenCourseWare

#### Feedback and Control

Feedback: simple, elegant, and robust framework for control.



Last time: robotic driving.



#### Feedback and Control

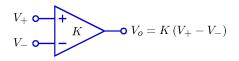
This week: using feedback to enhance performance.

#### Examples:

- increasing speed and bandwidth
- controlling position instead of speed
- reducing sensitivity to parameter variation
- reducing distortion
- stabilizing unstable systems
  - magnetic levitation
  - inverted pendulum

### Op-amps

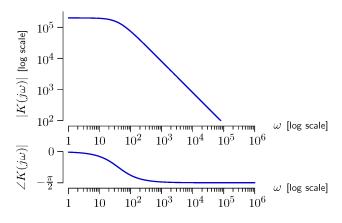
An "ideal" op-amp has many desireable characteristics.



- high speed
- large bandwidth
- high input impedance
- low output impedance
- ...

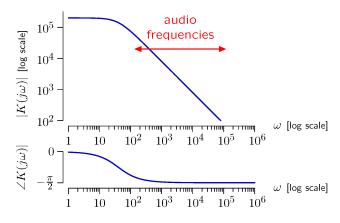
It is difficult to build a circuit with all of these features.

The gain of an op-amp depends on frequency.



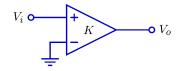
Frequency dependence of LM741 op-amp.

Low-gain at high frequencies limits applications.

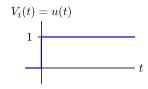


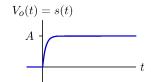
Unacceptable frequency response for an audio amplifier.

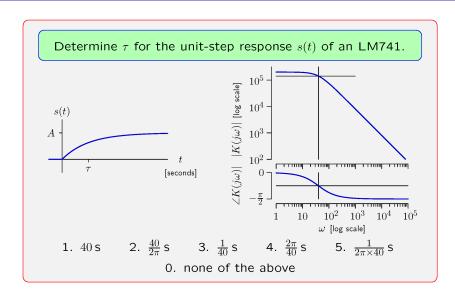
An ideal op-amp has fast time response.



Step response:







Determine the step response of an LM741.

System function:

$$K(s) = \frac{\alpha K_0}{s + \alpha}$$

Impulse response:

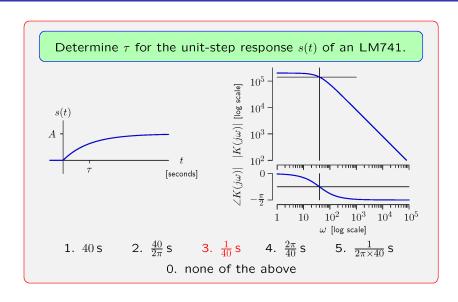
$$h(t) = \alpha K_0 e^{-\alpha t} u(t)$$

Step response:

$$s(t) = \int_{-\infty}^{t} h(\tau)d\tau = \int_{0}^{t} \alpha K_0 e^{-\alpha \tau} d\tau = \frac{\alpha K_0 e^{-\alpha \tau}}{-\alpha} \Big|_{0}^{t} = K_0 (1 - e^{-\alpha t}) u(t)$$

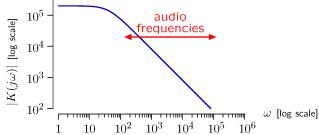
Parameters:

$$A=K_0=2\times 10^5$$
 
$$\tau=\frac{1}{\alpha}=\frac{1}{40}\,\mathrm{s}$$

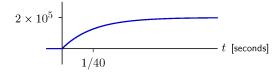


Performance parameters for real op-amps fall short of the ideal.

Frequency Response: high gain but only at low frequencies.

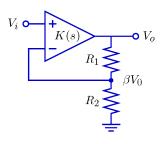


Step Response: slow by electronic standards.



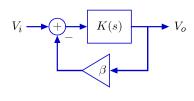
We can use feedback to improve performance of op-amps.





$$V_{-} = \beta V_o = \left(\frac{R_2}{R_1 + R_2}\right) V_o$$

#### 6.003 model

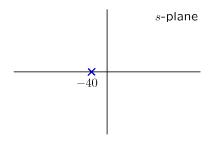


$$\frac{V_o}{V_i} = \frac{K(s)}{1 + \beta K(s)}$$

#### **Dominant Pole**

Op-amps are designed to have a dominant pole at low frequencies:

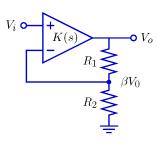
 $\rightarrow$  simplifies the application of feedback.



$$lpha=40\,\mathrm{rad/s}=rac{40\,\mathrm{rad/s}}{2\pi\,\mathrm{rad/cycle}}pprox 6.4\,\mathrm{Hz}$$

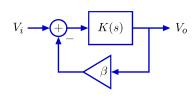
Using feedback to improve performance parameters.





$$V_- = \beta V_o = \left(\frac{R_2}{R_1 + R_2}\right) V_o$$

#### 6.003 model

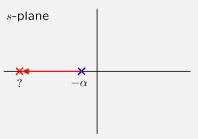


$$\frac{V_o}{V_i} = \frac{K(s)}{1 + \beta K(s)}$$

$$= \frac{\frac{\alpha K_0}{s + \alpha}}{1 + \beta \frac{\alpha K_0}{s + \alpha}}$$

$$= \frac{\alpha K_0}{s + \alpha + \alpha \beta K_0}$$

What is the most negative value of the closed-loop pole that can be achieved with feedback?



1. 
$$-\alpha(1+\beta)$$

1. 
$$-\alpha(1+\beta)$$
 2.  $-\alpha(1+\beta K_0)$ 

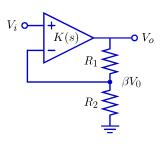
3. 
$$-\alpha(1+K_0)$$
 4.  $-\infty$ 

4. 
$$-\infty$$

5. none of the above

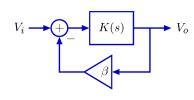
Using feedback to improve performance parameters.





$$V_- = \beta V_o = \left(\frac{R_2}{R_1 + R_2}\right) V_o$$

#### 6.003 model



$$\begin{split} \frac{V_o}{V_i} &= \frac{K(s)}{1 + \beta K(s)} \\ &= \frac{\frac{\alpha K_0}{s + \alpha}}{1 + \beta \frac{\alpha K_0}{s + \alpha}} \\ &= \frac{\alpha K_0}{s + \alpha + \alpha \beta K_0} \end{split}$$

What is the most negative value of the closed-loop pole that can be achieved with feedback?

Open loop system function:  $\frac{\alpha K_0}{s+\alpha}$ 

 $\rightarrow$  pole:  $s = -\alpha$ .

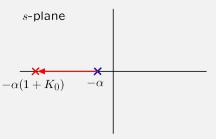
Closed-loop system function:  $\frac{\alpha K_0}{s + \alpha + \alpha \beta K_0}$ 

 $\rightarrow$  pole:  $s = -\alpha(1 + \beta K_0)$ .

The feedback constant is  $0 \le \beta \le 1$ .

ightarrow most negative value of the closed-loop pole is  $s=-lpha(1+K_0).$ 

What is the most negative value of the closed-loop pole that can be achieved with feedback? 3



1. 
$$-\alpha(1+\beta)$$

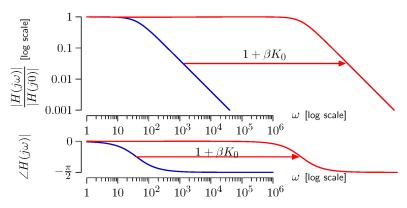
1. 
$$-\alpha(1+\beta)$$
 2.  $-\alpha(1+\beta K_0)$ 

3. 
$$-\alpha(1+K_0)$$
 4.  $-\infty$ 

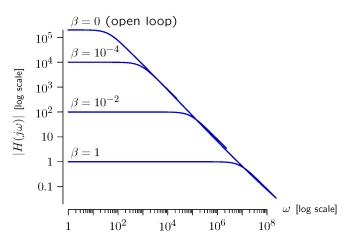
4. 
$$-\infty$$

5. none of the above

Feedback extends frequency response by a factor of  $1+\beta K_0$  ( $K_0=2\times 10^5$ ).



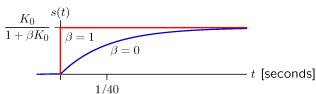
Feedback produces higher bandwidths by **reducing** the gain at low frequencies. It trades gain for bandwidth.



Feedback makes the time response faster by a factor of  $1+\beta K_0$  ( $K_0=2\times 10^5$ ).

Step response

$$s(t) = \frac{K_0}{1 + \beta K_0} (1 - e^{-\alpha(1 + \beta K_0)t}) u(t)$$



Feedback produces faster responses by **reducing** the final value of the step response. It trades gain for speed.

Step response

$$s(t) = \frac{K_0}{1+\beta K_0} (1-e^{-\alpha(1+\beta K_0)t}) u(t)$$
 
$$2\times 10^5 \begin{array}{c} s(t) & \beta \\ 0 & 0.5\times 10^{-5} \\ 1.5\times 10^{-5} & t \text{ [seconds]} \end{array}$$

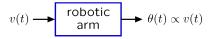
The maximum rate of voltage change  $\frac{ds(t)}{dt}\Big|_{t=0,1}$  is not increased.

Feedback improves performance parameters of op-amp circuits.

- can extend frequency response
- can increase speed

Performance enhancements are achieved through a reduction of gain.

We wish to build a robot arm (actually its elbow). The input should be voltage v(t), and the output should be the elbow angle  $\theta(t)$ .

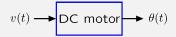


We wish to build the robot arm with a DC motor.

$$v(t) \longrightarrow \mathrm{DC} \ \mathrm{motor} \longrightarrow \theta(t)$$

This problem is similar to the head-turning servo in 6.01!

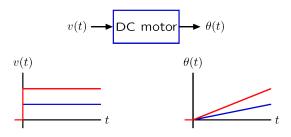
What is the relation between v(t) and  $\theta(t)$  for a DC motor?



- 1.  $\theta(t) \propto v(t)$
- 2.  $\cos \theta(t) \propto v(t)$
- 3.  $\theta(t) \propto \dot{v}(t)$
- 4.  $\cos \theta(t) \propto \dot{v}(t)$
- 5. none of the above

What is the relation between v(t) and  $\theta(t)$  for a DC motor?

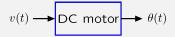
To first order, the rotational speed  $\dot{\theta}(t)$  of a DC motor is proportional to the input voltage v(t).



First-order model: integrator

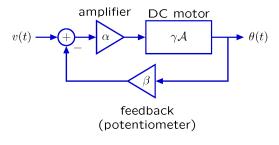


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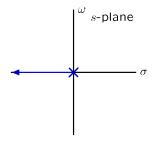
Use proportional feedback to control the angle of the motor's shaft.



$$\frac{\Theta}{V} = \frac{\alpha \gamma \mathcal{A}}{1 + \alpha \beta \gamma \mathcal{A}} = \frac{\alpha \gamma \frac{1}{s}}{1 + \alpha \beta \gamma \frac{1}{s}} = \frac{\alpha \gamma}{s + \alpha \beta \gamma}$$

The closed loop system has a single pole at  $s=-\alpha\beta\gamma$ .

$$\frac{\Theta}{V} = \frac{\alpha \gamma}{s + \alpha \beta \gamma}$$



As  $\alpha$  increases, the closed-loop pole becomes increasingly negative.

Find the impulse and step response.

The system function is

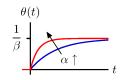
$$\frac{\Theta}{V} = \frac{\alpha \gamma}{s + \alpha \beta \gamma} \,.$$

The impulse response is

$$h(t) = \alpha \gamma e^{-\alpha \beta \gamma t} u(t)$$

and the step response is therefore

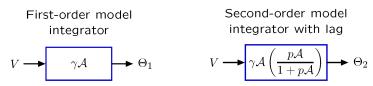
$$s(t) = \frac{1}{\beta} \left( 1 - e^{-\alpha\beta\gamma t} \right) u(t) \,.$$



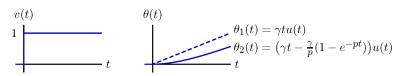
The response is faster for larger values of  $\alpha$ .

Try it: Demo.

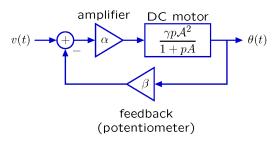
The speed of a DC motor does not change instantly if the voltage is stepped. There is lag due to rotational inertia.



Step response of the models:

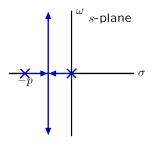


Analyze second-order model.



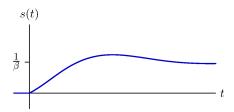
$$\frac{\Theta}{V} = \frac{\frac{\alpha\gamma p\mathcal{A}^2}{1+pA}}{1+\frac{\alpha\beta\gamma p\mathcal{A}^2}{1+pA}} = \frac{\alpha\gamma p\mathcal{A}^2}{1+pA+\alpha\beta\gamma p\mathcal{A}^2} = \frac{\alpha\gamma p}{s^2+ps+\alpha\beta\gamma p}$$
$$s = -\frac{p}{2} \pm \sqrt{\left(\frac{p}{2}\right)^2 - \alpha\beta\gamma p}$$

For second-order model, increasing  $\alpha$  causes the poles at 0 and -p to approach each other, collide at s=-p/2, then split into two poles with imaginary parts.

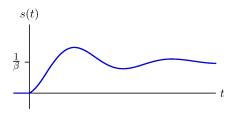


Increasing the gain  $\alpha$  does not increase speed of convergence.

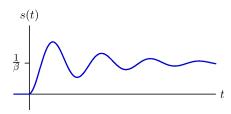
Step response.



Step response.

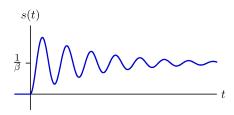


Step response.



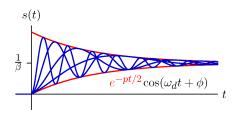
## Motor Controller

Step response.



## Motor Controller

Step response.



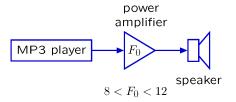
Using feedback to enhance performance.

#### Examples:

- improve performance of an op amp circuit.
- control position of a motor.
- reduce sensitivity to unwanted parameter variation.
- reduce distortions.
- stabilize unstable systems
  - magnetic levitation
  - inverted pendulum

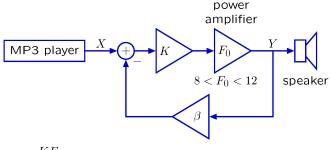
Reducing sensitivity to unwanted parameter variation.

Example: power amplifier



Changes in  $F_0$  (due to changes in temperature, for example) lead to undesired changes in sound level.

Feedback can be used to compensate for parameter variation.



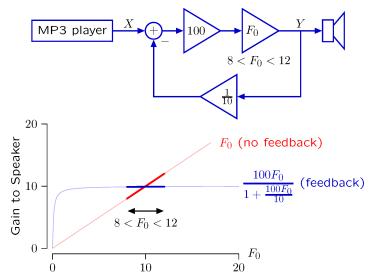
$$H(s) = \frac{KF_0}{1 + \beta KF_0}$$

If K is made large, so that  $\beta KF_0 \gg 1$ , then

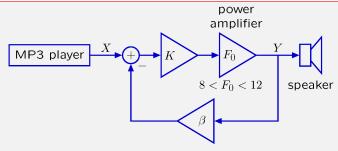
$$H(s) \approx \frac{1}{\beta}$$

independent of K or  $F_0!$ 

Feedback reduces the change in gain due to change in  $F_0$ .



## Check Yourself



Feedback greatly reduces sensitivity to variations in K or  $F_0$ .

$$\lim_{K \to \infty} H(s) = \frac{KF_0}{1 + \beta KF_0} \to \frac{1}{\beta}$$

What about variations in  $\beta$ ? Aren't those important?

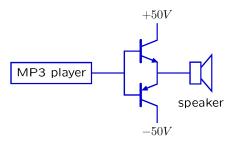
## Check Yourself

What about variations in  $\beta$ ? Aren't those important?

The value of  $\beta$  is typically determined with resistors, whose values are quite stable (compared to semiconductor devices).

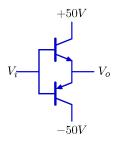
Feedback can compensate for parameter variation even when the variation occurs rapidly.

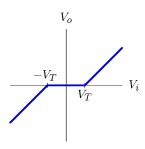
Example: using transistors to amplify power.



This circuit introduces "crossover distortion."

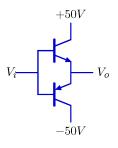
For the upper transistor to conduct,  $V_i - V_o > V_T$ . For the lower transistor to conduct,  $V_i - V_o < -V_T$ .

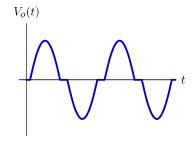




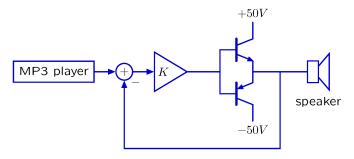
Crossover distortion can have dramatic effects.

Example: crossover distortion when the input is  $V_i(t) = B\sin(\omega_0 t)$ .

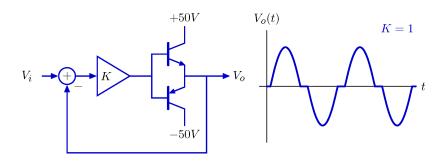




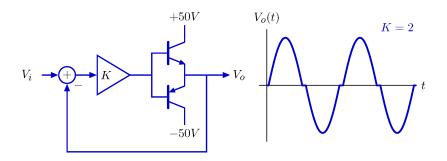
Feedback can reduce the effects of crossover distortion.



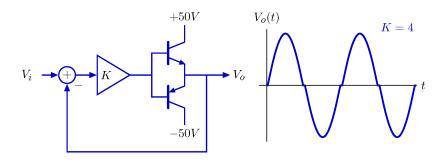
When K is small, feedback has little effect on crossover distortion.



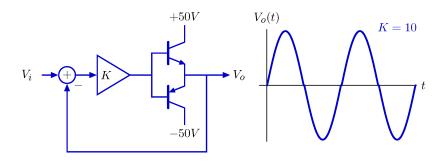
As K increases, feedback reduces crossover distortion.



As K increases, feedback reduces crossover distortion.

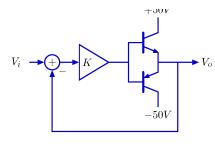


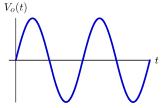
As K increases, feedback reduces crossover distortion.



### Demo

- original
- no feedback
- K = 2
- $\bullet$  K=4
- K = 8
- K = 16
- original





J.S. Bach, Sonata No. 1 in G minor Mvmt. IV. Presto Nathan Milstein violin

Using feedback to enhance performance.

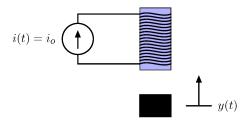
### Examples:

- improve performance of an op amp circuit.
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## Control of Unstable Systems

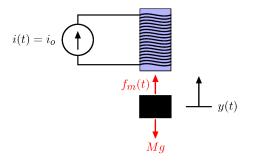
Feedback is useful for controlling **unstable** systems.

Example: Magnetic levitation.



# Control of Unstable Systems

Magnetic levitation is unstable.



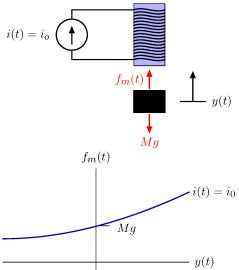
Equilibrium (y = 0): magnetic force  $f_m(t)$  is equal to the weight Mg.

Increase  $y \rightarrow$  increased force  $\rightarrow$  further increases y.

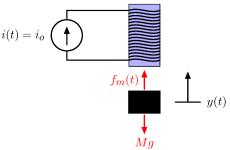
Decrease  $y \rightarrow$  decreased force  $\rightarrow$  further decreases y.

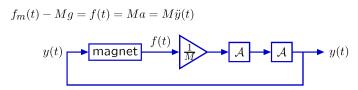
Positive feedback!

The magnet generates a force that depends on the distance y(t).

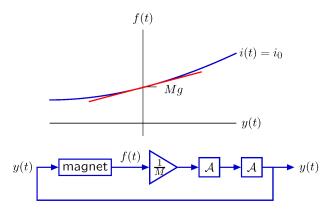


The net force accelerates the mass.





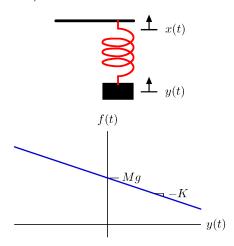
Over small distances, magnetic force grows  $\approx$  linearly with distance.



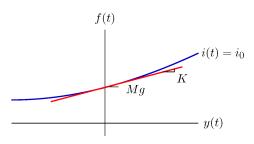
# Levitation with a Spring

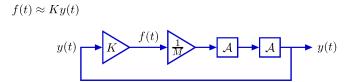
Relation between force and distance for a spring is opposite in sign.

$$F = K(x(t) - y(t)) = M\ddot{y}(t)$$



Over small distances, magnetic force nearly proportional to distance.





# Block Diagrams

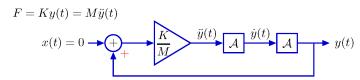
Block diagrams for magnetic levitation and spring/mass are similar.

### Spring and mass

$$F = K\left(x(t) - y(t)\right) = M\ddot{y}(t)$$

$$x(t) \xrightarrow{K} \ddot{y}(t) \xrightarrow{K} y(t)$$

### Magnetic levitation



## Check Yourself

How do the poles of these two systems differ?

### Spring and mass

$$F = K\left(x(t) - y(t)\right) = M\ddot{y}(t)$$

$$x(t) \xrightarrow{\qquad \qquad } K \xrightarrow{\qquad \qquad } \ddot{y}(t) \xrightarrow{\qquad \qquad } y(t)$$

### Magnetic levitation

$$F = Ky(t) = M\ddot{y}(t)$$

$$x(t) = 0 \longrightarrow + \longrightarrow K \qquad \ddot{y}(t) \longrightarrow A \qquad \dot{y}(t) \longrightarrow y(t)$$

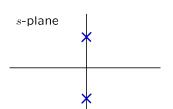
## Check Yourself

How do the poles of the two systems differ?

Spring and mass

$$F = K(x(t) - y(t)) = M\ddot{y}(t)$$

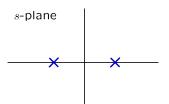
$$\frac{Y}{X} = \frac{\frac{K}{M}}{s^2 + \frac{K}{M}} \ \rightarrow \ s = \pm j \sqrt{\frac{K}{M}}$$



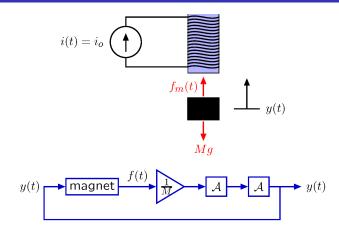
Magnetic levitation

$$F = Ky(t) = M\ddot{y}(t)$$

$$s^2 = \frac{K}{M} \rightarrow s = \pm \sqrt{\frac{K}{M}}$$

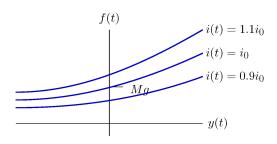


# Magnetic Levitation is Unstable



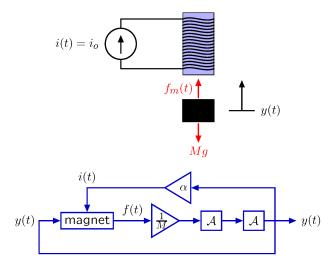
# Magnetic Levitation

We can stabilize this system by adding an additional feedback loop to control i(t).



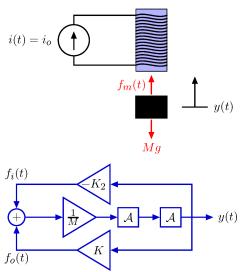
## Stabilizing Magnetic Levitation

Stabilize magnetic levitation by controlling the magnet current.



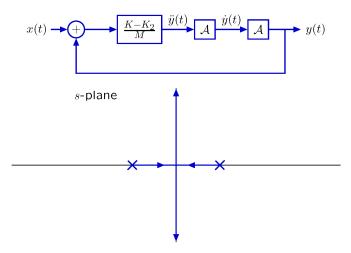
# Stabilizing Magnetic Levitation

Stabilize magnetic levitation by controlling the magnet current.



# Magnetic Levitation

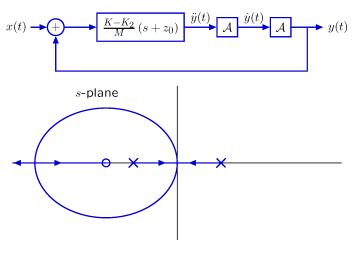
Increasing  $K_2$  moves poles toward the origin and then onto  $j\omega$  axis.



But the poles are still marginally stable.

# Magnetic Levitation

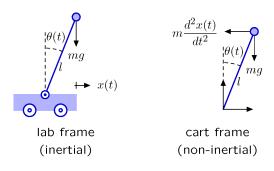
Adding a zero makes the poles stable for sufficiently large  $K_2$ .



Try it: Demo [designed by Prof. James Roberge].

## **Inverted Pendulum**

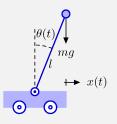
As a final example of stabilizing an unstable system, consider an inverted pendulum.

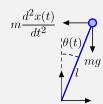


$$\underbrace{ml^2}_{I} \frac{d^2\theta(t)}{dt^2} = \underbrace{mg}_{\text{force}} \underbrace{l\sin\theta(t)}_{\text{distance}} - \underbrace{m\frac{d^2x(t)}{dt^2}}_{\text{force}} \underbrace{l\cos\theta(t)}_{\text{distance}}$$

## Check Yourself: Inverted Pendulum

### Where are the poles of this system?

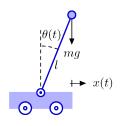




$$ml^{2}\frac{d^{2}\theta(t)}{dt^{2}} = mgl\sin\theta(t) - m\frac{d^{2}x(t)}{dt^{2}}l\cos\theta(t)$$

## Check Yourself: Inverted Pendulum

Where are the poles of this system?



$$m\frac{d^2x(t)}{dt^2} + \frac{1}{\theta(t)} mg$$

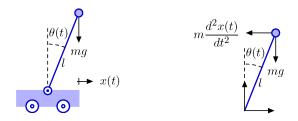
$$ml^{2} \frac{d^{2}\theta(t)}{dt^{2}} = mgl\sin\theta(t) - m\frac{d^{2}x(t)}{dt^{2}}l\cos\theta(t)$$
$$ml^{2} \frac{d^{2}\theta(t)}{dt^{2}} - mgl\theta(t) = -ml\frac{d^{2}x(t)}{dt^{2}}$$

$$H(s) = \frac{\Theta}{X} = \frac{-mls^2}{ml^2s^2 - mal} = \frac{-s^2/l}{s^2 - a/l}$$
 poles at  $s = \pm \sqrt{\frac{g}{l}}$ 

poles at 
$$s = \pm \sqrt{\frac{g}{l}}$$

## **Inverted Pendulum**

This unstable system can be stablized with feedback.



Try it. Demo. [originally designed by Marcel Gaudreau]

Using feedback to enhance performance.

#### Examples:

- improve performance of an op amp circuit.
- control position of a motor.
- reduce sensitivity to unwanted parameter variation.
- reduce distortions.
- stabilize unstable systems
  - magnetic levitation
  - inverted pendulum

# Assignments

- Reading Assignment: Chap. 11.0-11.2, Review Chap. 9.7-9.8, 10.7-10.8
- Homework 6