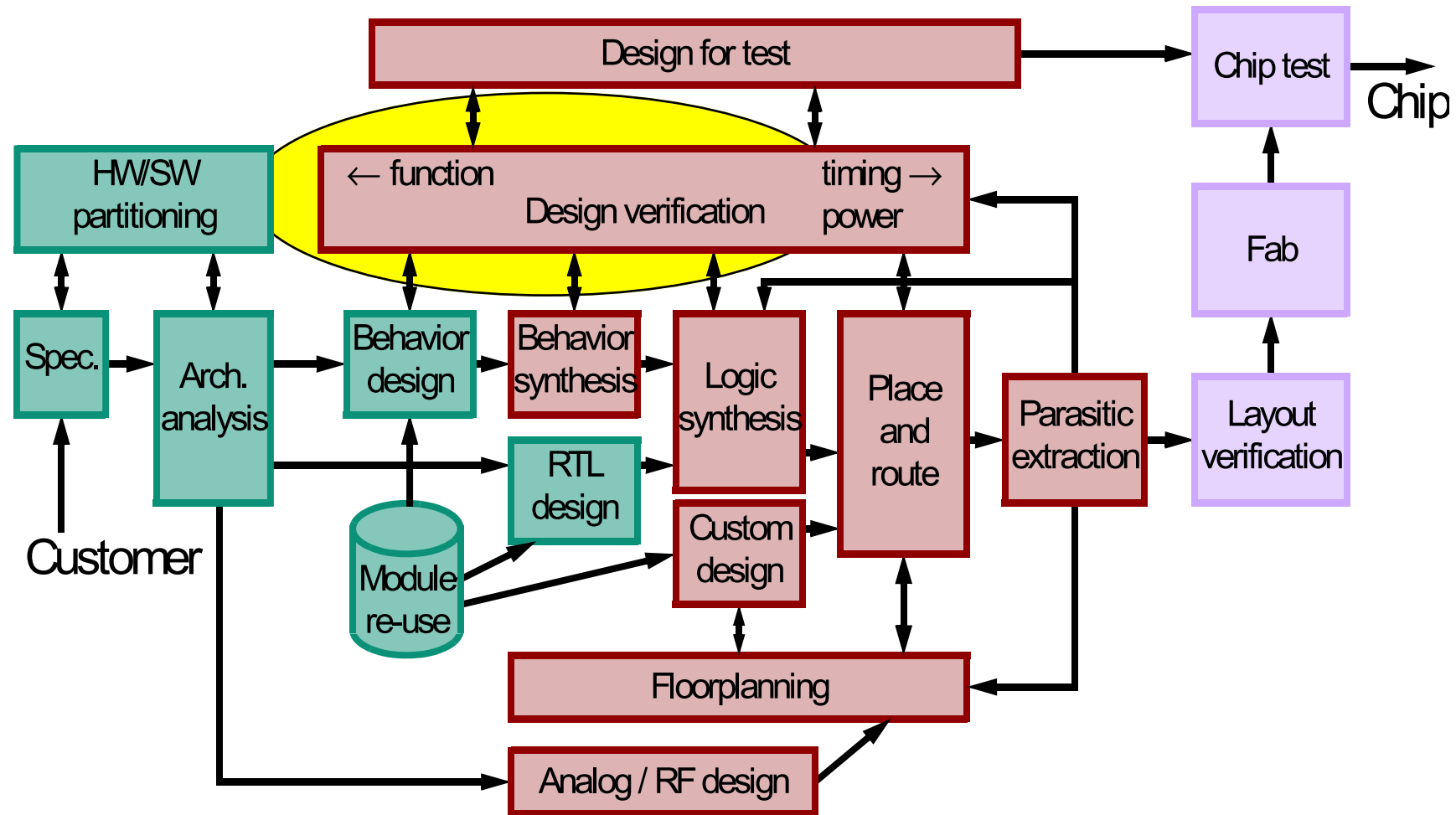


Functional Verification

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Functional Verification



Some Types of Verification

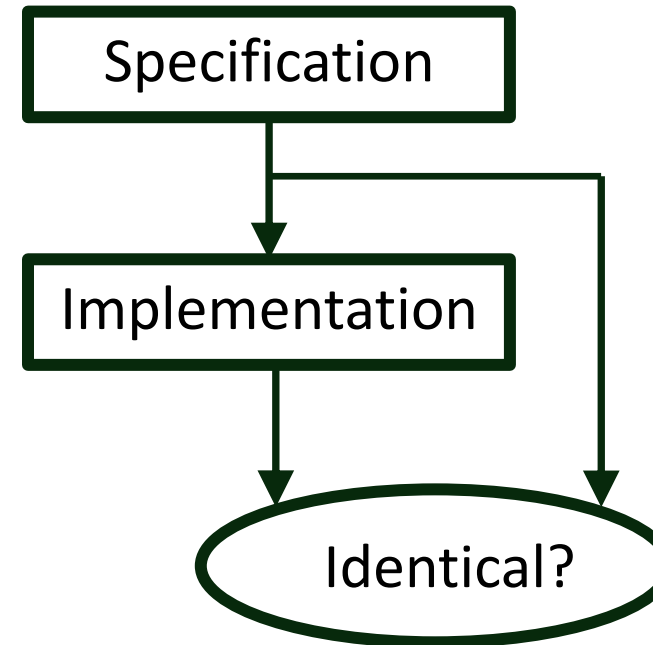
- Prototyping: PCB-based test (bread-boarding).
- Equivalence checking: check consistency specification → implementation.
- Correctness-by-construction: prove refinement preserves behavior.
- Simulation at different levels.
- Timing analysis (static) for delay verification.
- Layout verification (DRC, ERC, Extraction, LVS ...).

Verification Options for HDL Code

- Logic (digital) simulation.
 - Slow (but not compared to circuit simulation).
 - Not exhaustive (depends on test vector set).
- Formal verification.
 - Large state space - practical only for some problems.
 - Non-intuitive - demands expert user.
- 'Linters' (from lint).
 - A heuristic static parsing strategy for inspection of code.
- Acceleration and emulation using special purpose hardware.

Context of Functional Verification

1. Determine intent - *specification*.
2. Determine actual function - *implementation*.
3. Compare intended function and actual function.
4. Consider the confidence of comparison results.



Challenges with Specification Definition

- Even for derivative designs (re-used), application use-cases are often used to determine the intended function - the specification.
 - With system complexity and radically new consumer patterns, calling for immature applications, how to establish use-cases?
- Specifications are often first expressed in natural languages.
 - A natural language specification is ambiguous, has gaps and may be full of mistakes/misunderstandings.
 - Specifications in an executable form offer both a conciseness and allow for deliberations on HW/SW partitionings.

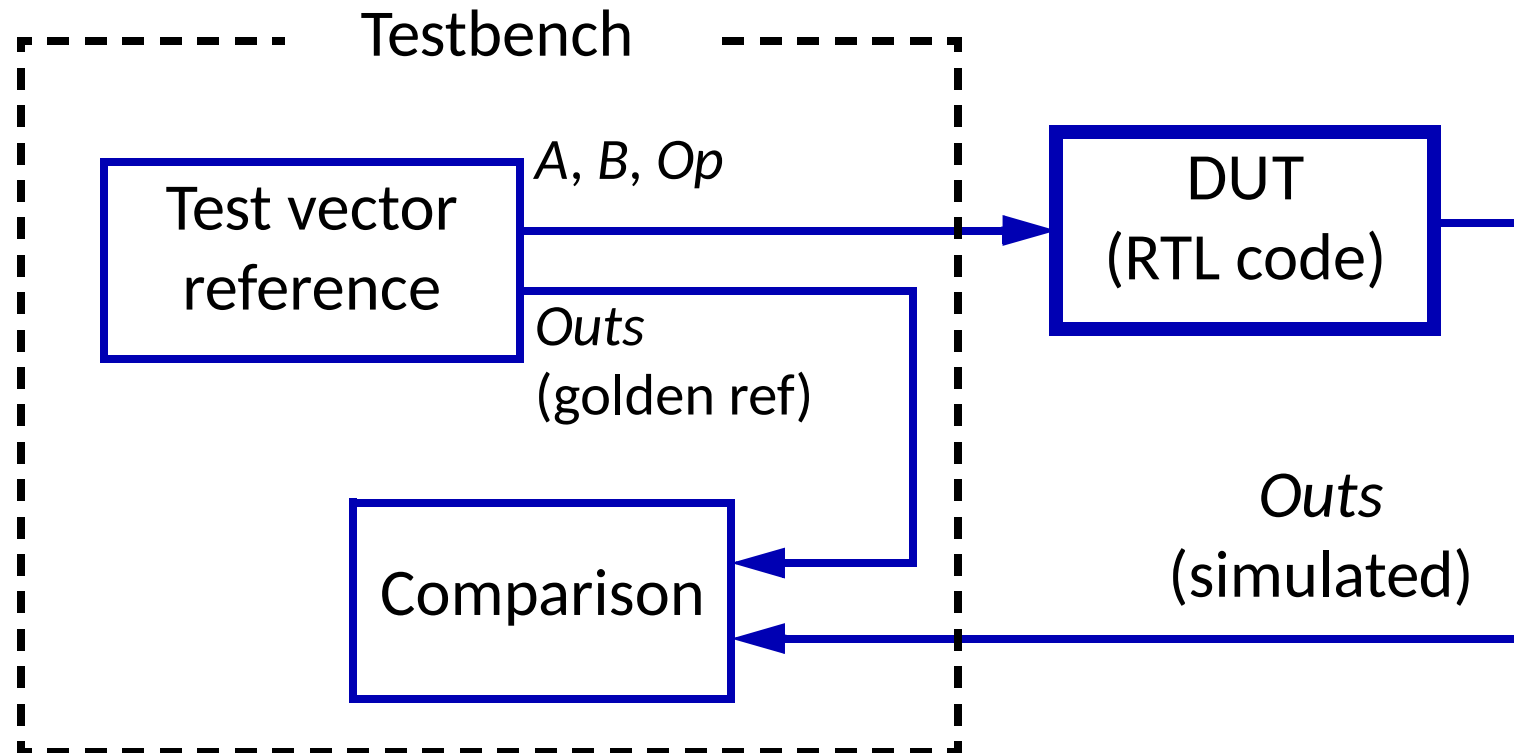
Logic Simulation

1. Given a piece of HDL code and a digital stimuli, the test vectors, ...
 2. evaluate the HDL code using a testbench ...
 3. by, for each stimulus, checking the HDL code output result against “golden” reference.
- Purpose:
 - Find design implementation errors early, before circuit design.
 - Verify that implementation matches specification.

Dedicated Verification Teams?

- How do we create the golden reference?
 - Since the verification reference implementation itself is not targeting a product, it may become a “second-class citizen”.
- Use separate design and verification teams!
 - Only possible for larger corporations.
- Determine intent - specification.
- Determine actual function - *design implementation*.
- Determine a *verification implementation*.

Logic Simulation Context



Logic Simulation Strategies 1(3)

- Consider the use of assertions inside the testbench.
 - ALU opcode should not be outside the valid range.
 - A half adder should never deliver $\text{Sum} = 1$ and $\text{Carry} = 1$.
 - Assertions become an 'executable databook' for blocks.
- Stimulate HDL code using test vectors for directed tests.
 - Instruction Set Simulator (ISS) simulation using benchmarks.
 - Test vectors that target blocks that are known to be prone to errors.
- Stimulate HDL code using random test vectors.
 - Pseudo-randomization can uncover errors in unexpected places.

Logic Simulation Strategies 2(3)

- Evaluate the simulation output either by ...
 - waveform eyeballing (use only for debugging !).
 - comparison against a “golden” reference file:
An expected results file.
 - comparison against a “golden” reference design:
A behavioral model that we know is correct or a previous design with the same functionality and cycle behavior.
 - checking the output log of assertions.

Logic Simulation Strategies 3(3)

- Since we can never exhaustively simulate real systems, we need to decide on an acceptable confidence level.
- The metric of coverage - percentage of items verified out of all possible items:
 - code coverage (syntax): % of RTL code lines simulated
(use metric sensibly... 100% code coverage on flawed design?)
 - functional coverage (semantics):
% of functional features simulated.
 - parameter coverage (semantics):
% of operational ranges simulated, for example,
is the range of depth of a FIFO fully simulated?

Logic Simulation Flow

- Apply random test vectors.
 - Check code coverage [[Sec. 9.3 of Ch9_FunctionalVerification](#)].
 - Block coverage: Proportion of blocks (collection of lines after control statements) visited.
 - Expression coverage: Proportion of minterms [[Lecture 4](#)] visited.
 - Toggle coverage: Proportion of toggling signals in instance/module.
- Add directed test vectors for corner cases that happened not to be covered.

Bugs

- Handle bugs!
- The life of a bug ...
 - opened:
when detected (by design or verification eng), the bug is logged.
 - verified:
when the designer confirms it is indeed a bug.
 - fixed:
when the designer has removed the bug.
 - closed:
when the surrounding code works fine.

Bottom-Up Verification

- When designing a complex digital module, it is good practice to conceptualize it top down, but in terms of verification, the module and its sub-blocks should be verified bottom up.
- The order in which sub-blocks should be implemented and verified should be clear.
- When functionality of individual sub-blocks has been verified, verify functionality of the combination of sub-blocks.
- Finally, you will reach the top design.

Build Testbenches

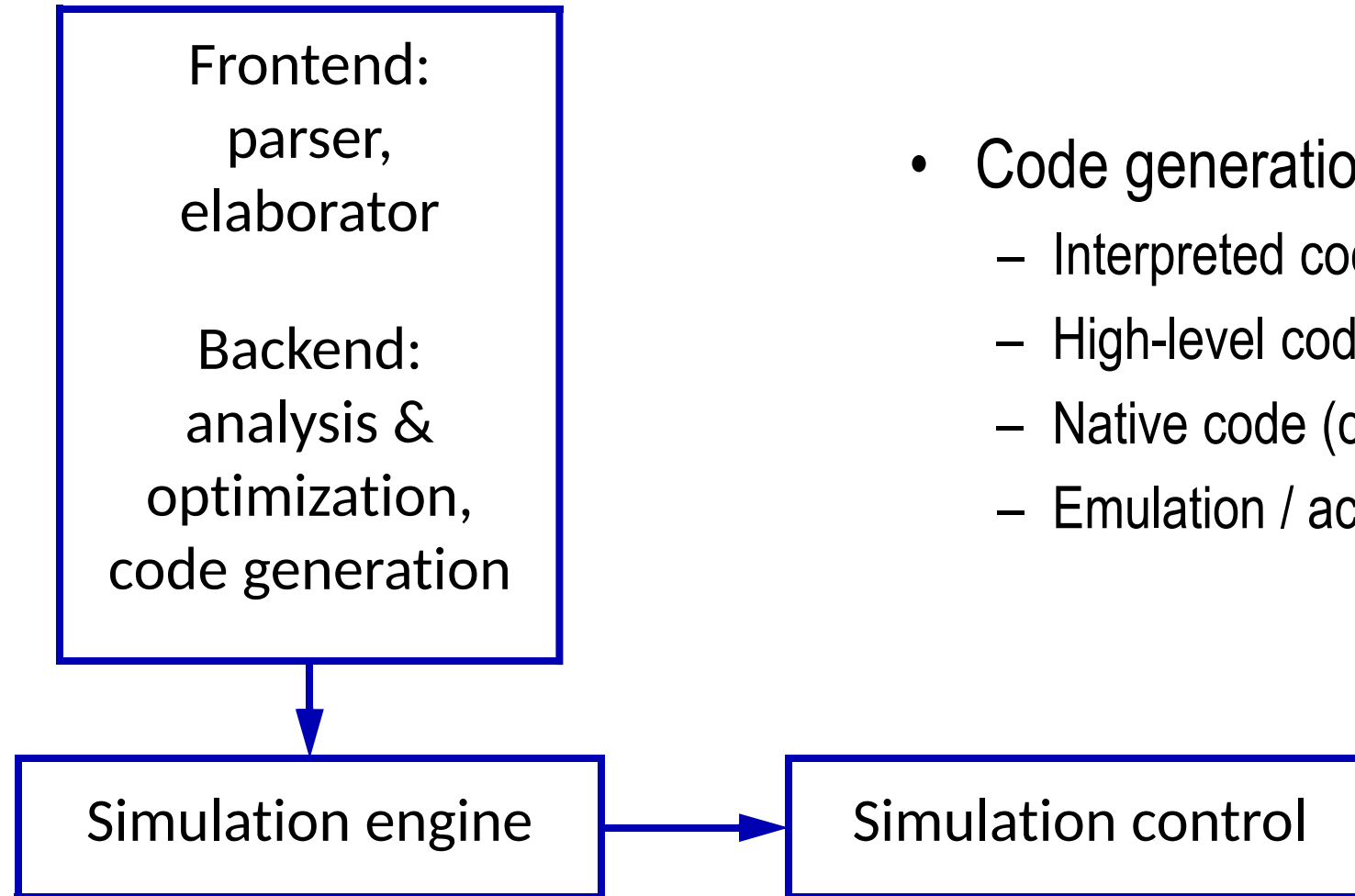
- A testbench for HDL code consists of non-synthesizable code that can interface to input test vectors and expected output vectors.
- Several testbenches, for the same HDL design, can exist:
 - The chance to spot errors (even in the testbench!) increases with the use of alternate testbenches.
Conformity does not always work in your favor!
 - Verification of functionality may warrant a different testbench than one which is used to establish maximal speed (cf RCA vs Sklansky adder).
 - An ALU, for example, can be verified stand-alone or in the context of preceding and/or succeeding hardware.

Testbench Example

```
entity testbench is
end testbench;

architecture behavioral of testbench is
    component XYZ is                                -- Declare DUT component
        port()
    end component;
    constant CLK_PERIOD      : time := 1 ns;        -- Define constants
    signal clk                : STD_LOGIC:= '0';     -- Declare internal signals
begin
    u0: XYZ port map ();                            -- Instantiate DUT
    `generate clock (and reset)'                    -- One or two processes
    `read in test vectors and store in arrays'
    `loop for CYCLES:                                -- Testbench main process
        `send test vector to DUT'
        `read DUT output and compare (assert) to reference'
        `write errors to file'
        wait
    end loop;
end behavioral;
```

Logic Simulators



- Code generation:
 - Interpreted code.
 - High-level code (C etc.).
 - Native code (common).
 - Emulation / acceleration.

(In-Circuit) Emulation

- Palladium Z1 (Cadence).
- Veloce2 (Siemens Mentor).
 - ZeBu (Synopsys, previously EVE, France).
- Features:
 - Acceleration of simulation.
 - Emulation of hardware (FPGA for ASIC).
 - May allow for early SW development (see prototyping systems).



source: Cadence



source: Siemens Mentor

NVIDIA Indus – based on Palladium



source: NVIDIA

Logic Simulation Engines

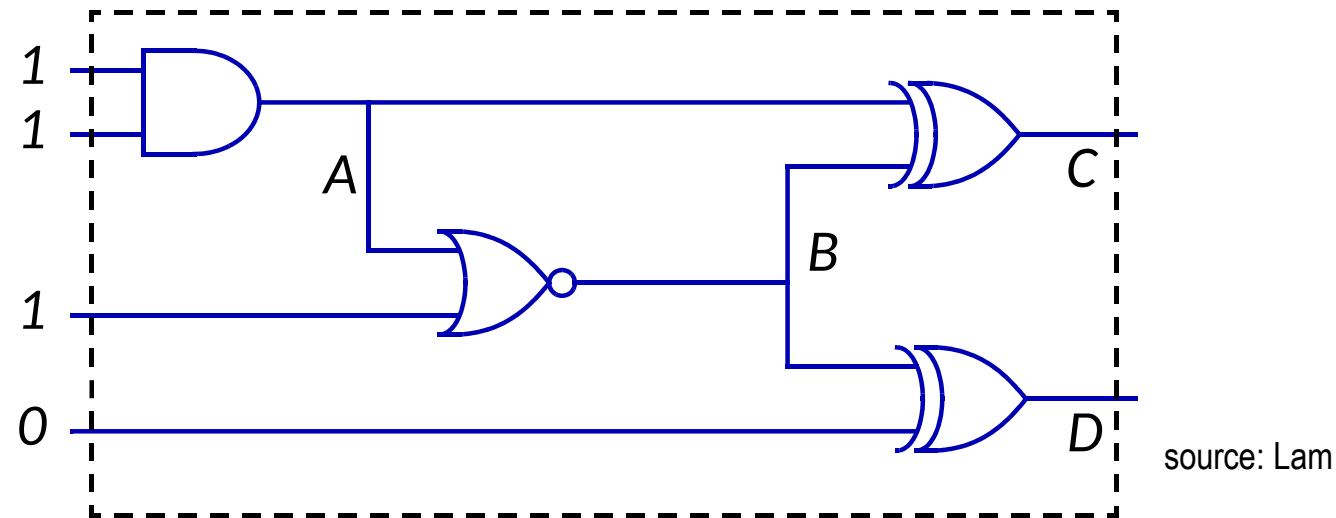
- Cycle-based (time-driven) approaches.
 - Evaluate logic states on *all nodes* between registers
 - Pursue the steady-state (per cycle) values.
 - New data evaluation every clock cycle.
- Event-driven approaches.
 - Evaluate *only* those nodes that *change* state.
 - Can find timing information, if delay models are accurate.
- Cycle-based approaches are faster for signal switching activities above 1%.

Logic Evaluation and Representation

- Unidirectional signal flow (*cf* circuit simulators.)
- When the basic element is a gate, we call it gate-level simulation.
- Two-state representation: 0 and 1.
- Four-state representation: 0, 1, X and Z.
 - X represents either an uninitialized state or that several drivers enforce conflicting values onto the node.
 - Z represents a high-impedance (floating) node.
- Four-state representation is used during circuit power-up; then the simulator switches to two states to increase simulation speed.

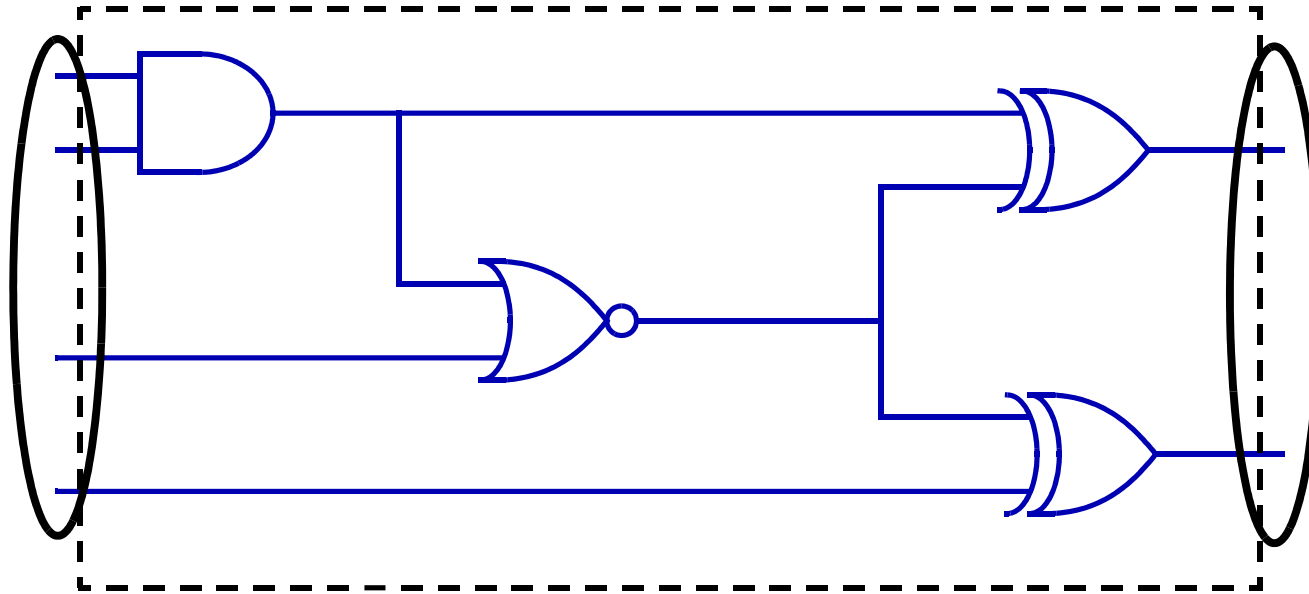
Cycle-Based Logic Simulation

Evaluate all logic between register boundaries



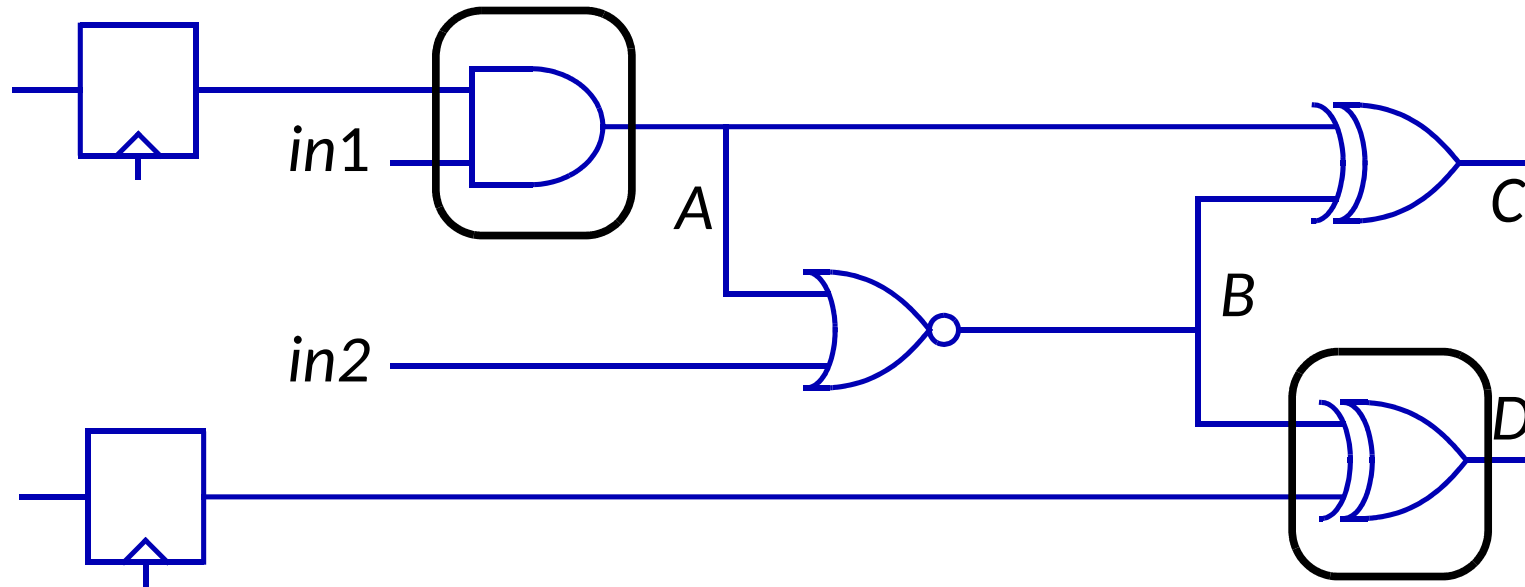
1. $A = 1 \text{ AND } 1 = 1$
2. $B = A \text{ NOR } 1 = 1 \text{ NOR } 1 = 0$
3. $C = A \text{ XOR } B = 1 \text{ XOR } 0 = 1$
4. $D = B \text{ XOR } 0 = 0 \text{ XOR } 0 = 0$

Primary Inputs and Outputs



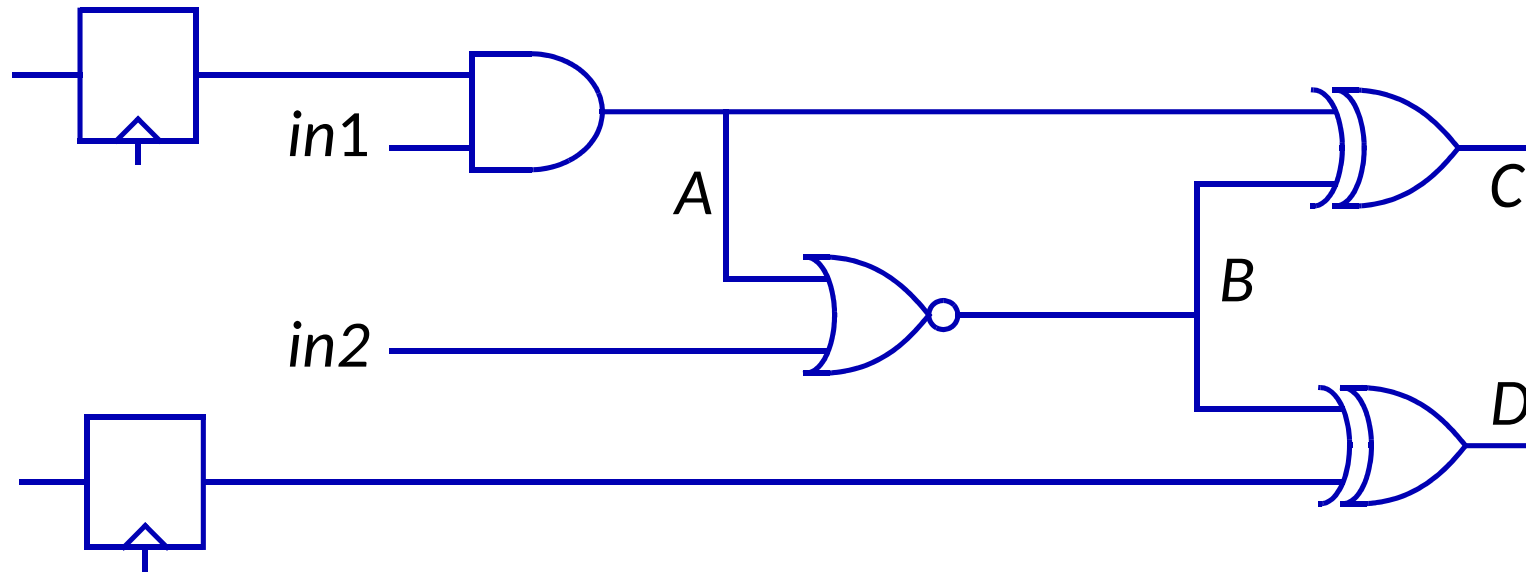
Generally the main inputs and outputs,
those on the boundary of a block,
are called primary inputs and outputs

Order of Evaluation Matters



FF outputs are the first event in an event-driven approach \Rightarrow
gates with inputs from FFs would evaluate first \Rightarrow
an outdated value of *B* would define *D*

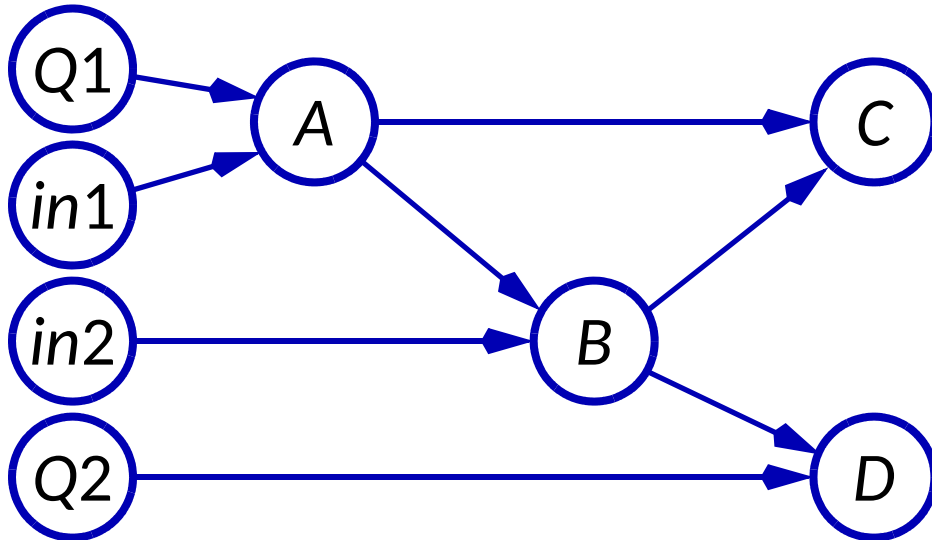
Levelization



In cycle-based simulation approaches, levelization ensures that a gate evaluates its output only after its inputs have been updated.

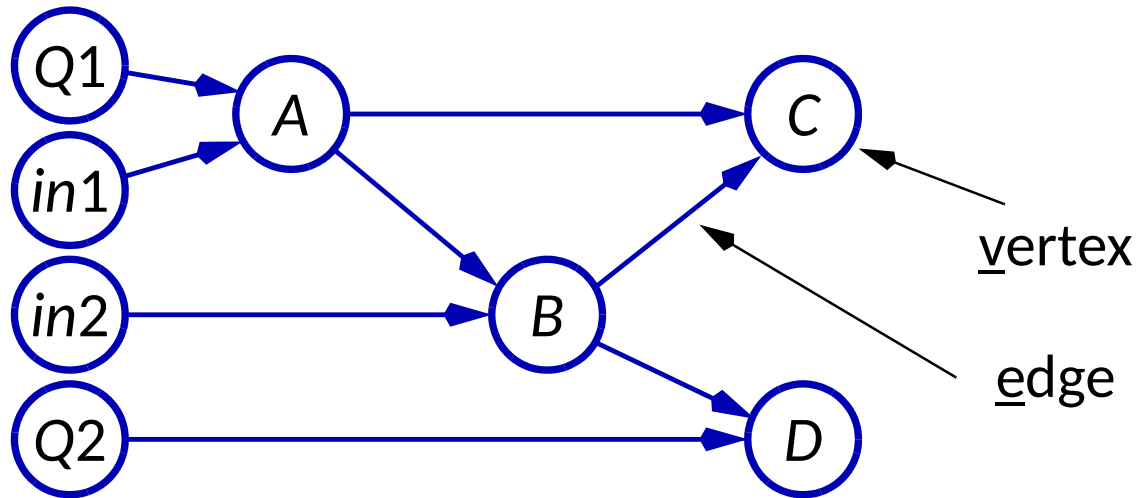
This is important since each gate only is evaluated once!

Algorithm for Levelization 1(5)



- To establish in which order to evaluate the gates, we have to use a systematic approach - an *algorithm*:
- Graph theory is a branch of discrete mathematics.

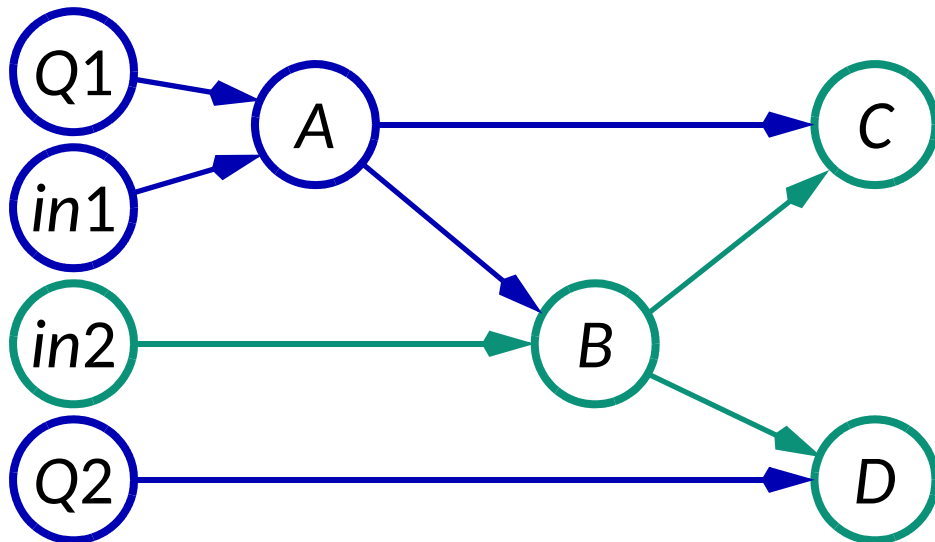
Algorithm for Levelization 2(5)



- Use so-called depth-first search, starting at any primary input:

“Trace through graph until no more outgoing edge exists or until the next vertex has already been visited, backtrack and insert the visited vertex in an ordered evaluation list.”

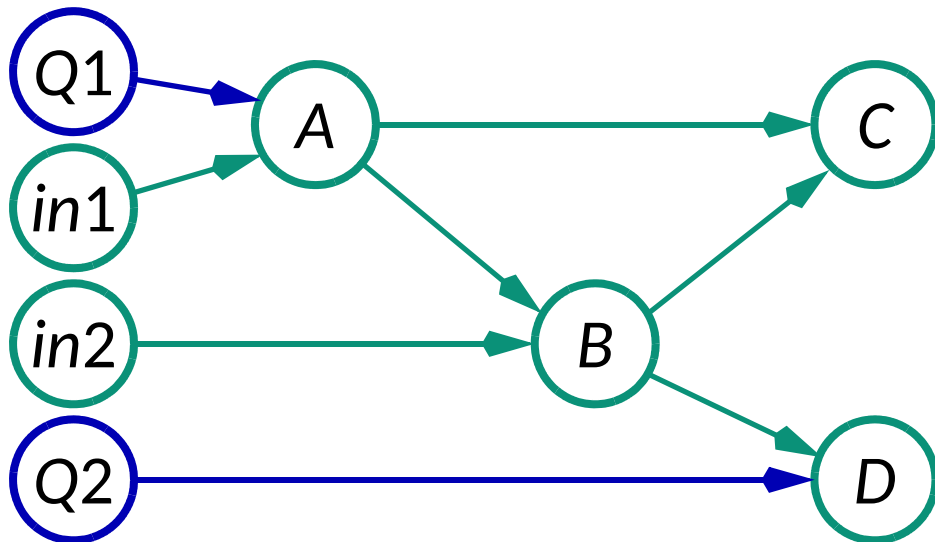
Algorithm for Levelization 3(5)



1. Start for example at *in2* and follow the edge to *B*.
2. Continue to *C* or *D*; here we (arbitrarily) choose *D*.
3. No edge: return and save *D*.
4. Go to *C*.
5. No edge: return and save *C*.
6. Return and save *B*.
7. Back at *in2*, we are done.
Thus, save *in2*.

Evaluation list so far: ***in2, B, C, D***

Algorithm for Levelization 4(5)

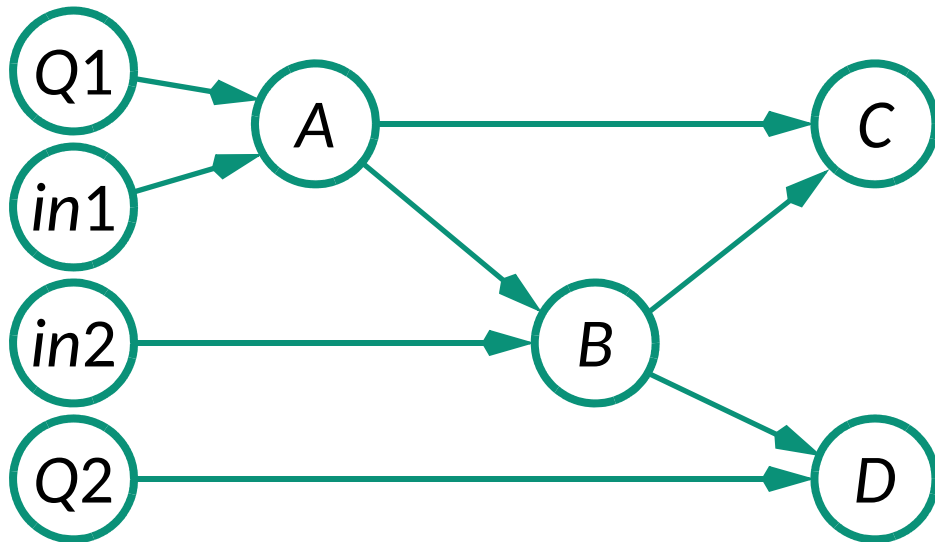


1. Now start with *in1* and follow the edge to *A*.
2. In trying to continue to *B* or *C*, we discover these vertices have already been visited. Thus, save *A*.
3. Return and save *in1*.

Evaluation list so far:

in1, A, in2, B, C, D

Algorithm for Levelization 5(5)



1. Now start with Q1.
2. In trying to continue to A, we discover this vertex has been visited. Thus, save Q1.
3. After trying Q2, we end up saving Q2.

Final evaluation list:

Q2, Q1, in1, A, in2, B, C, D

A logic-gate evaluation that follows this sequence guarantees that gate inputs always are *evaluated before* the gate's output is evaluated.

The Language of Algorithms

“Trace through graph until no more outgoing edge exists or until the next vertex has already been visited, backtrack and insert the visited vertex in an ordered evaluation list.”

Input: $G(V, E)$, Output: List of ordered nodes

TopologicalSort (G)

{while (node v in V is not marked visited) VISIT(v) }

VISIT(v) {

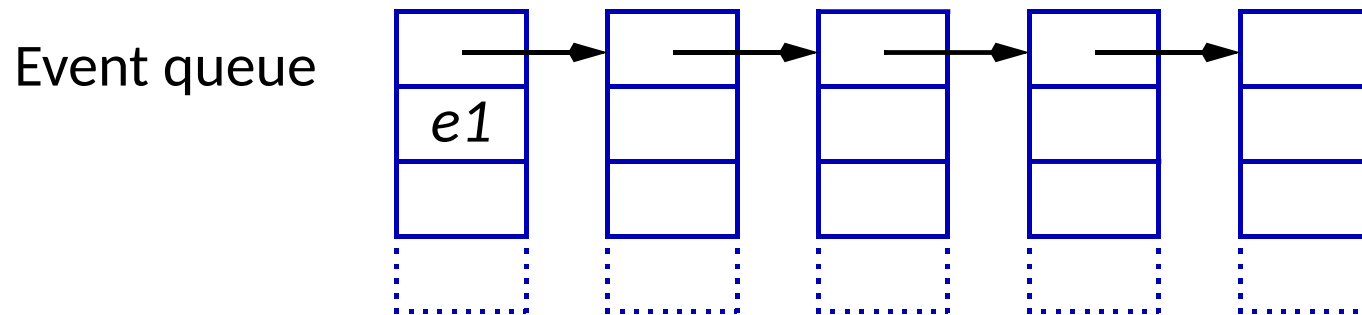
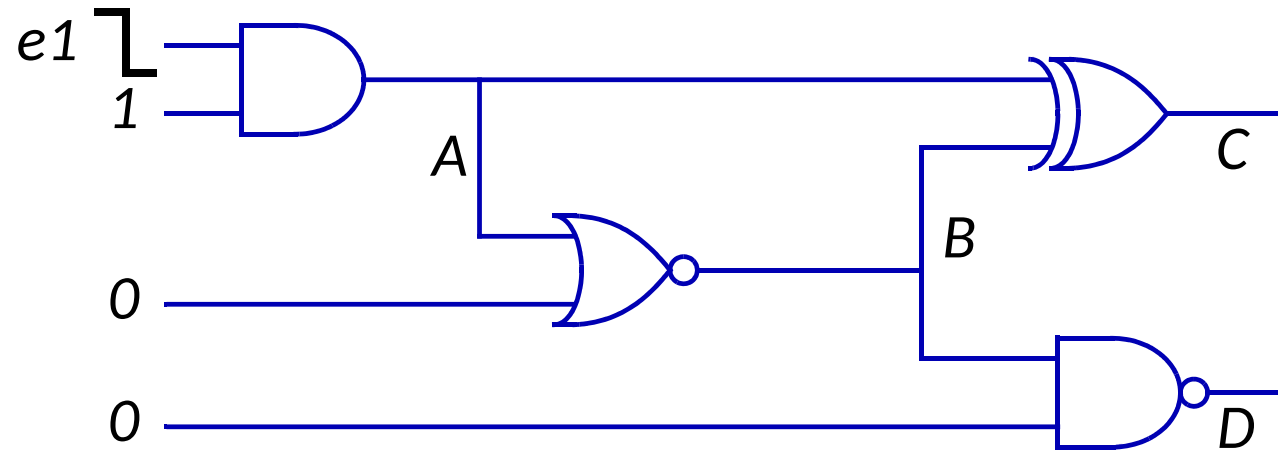
mark v visited;

for each (u taken from the fanout of v)

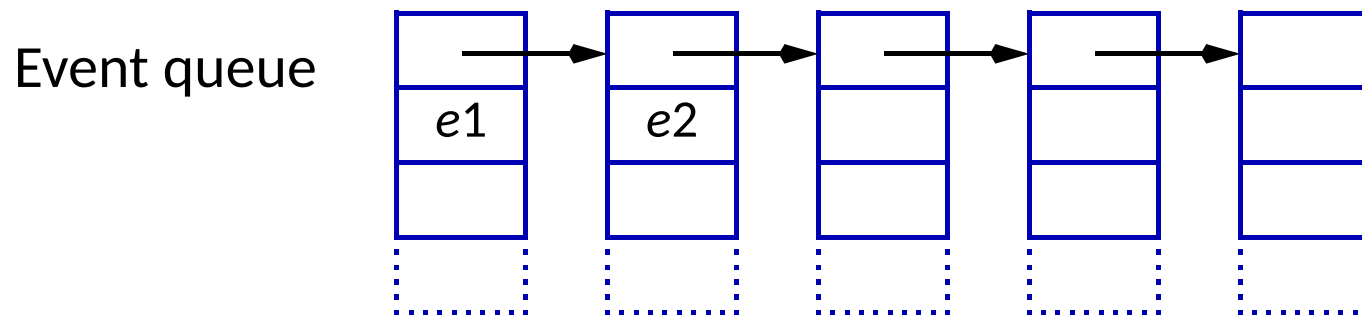
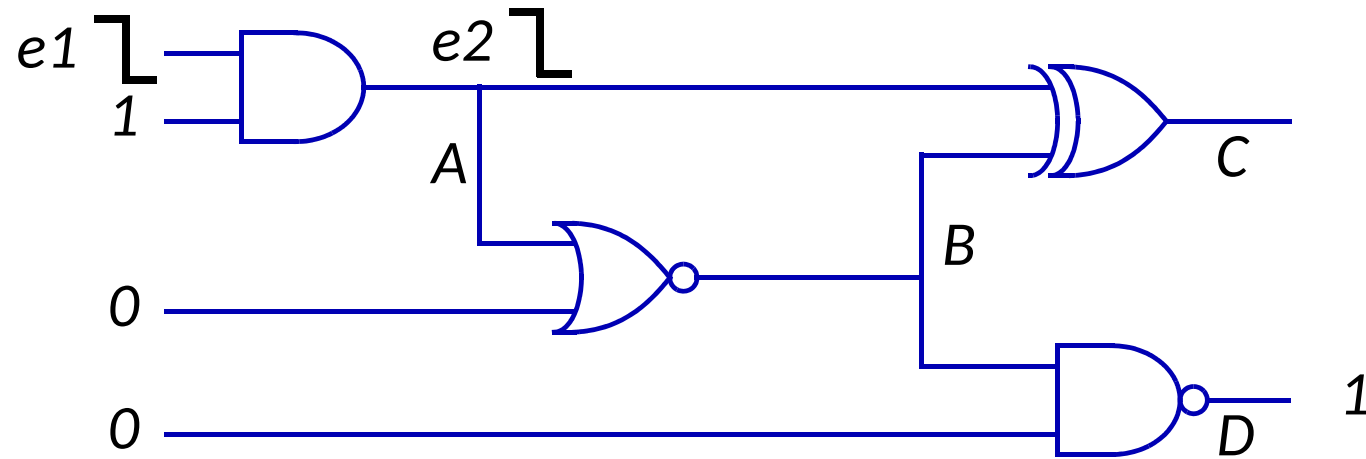
 if (u is not marked visited) VISIT(u);

insert u in front of List; }

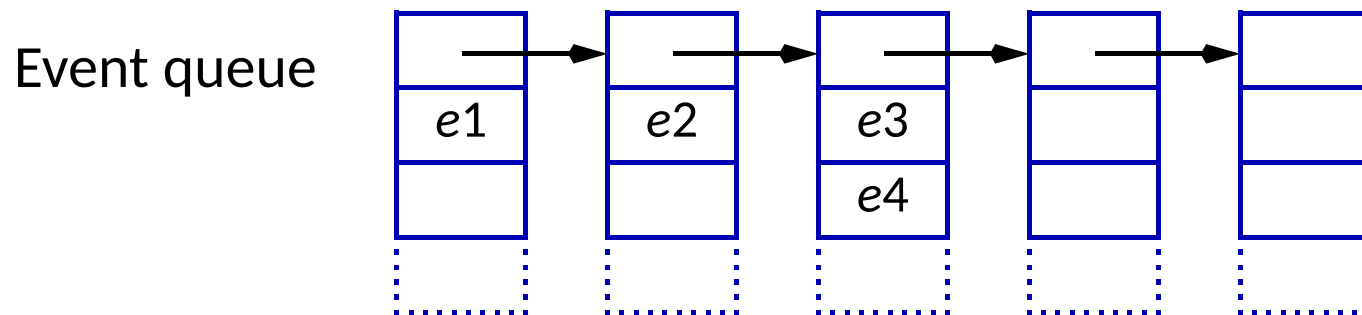
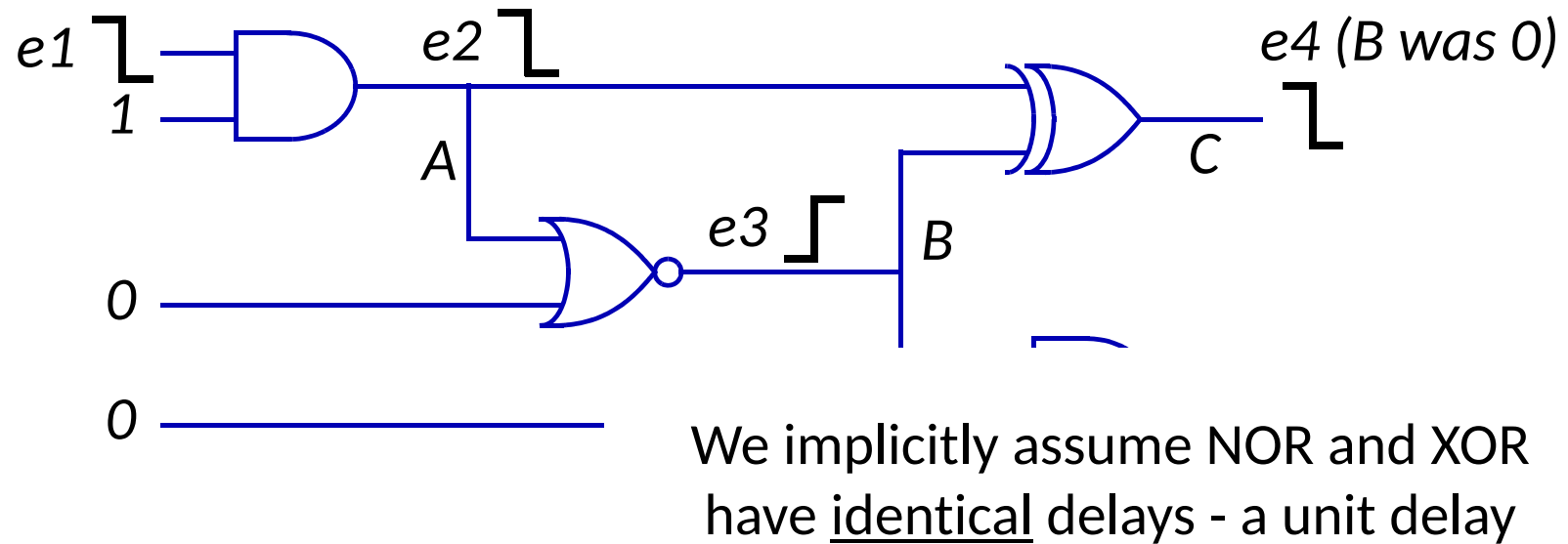
Event-Driven Simulation 1(5)



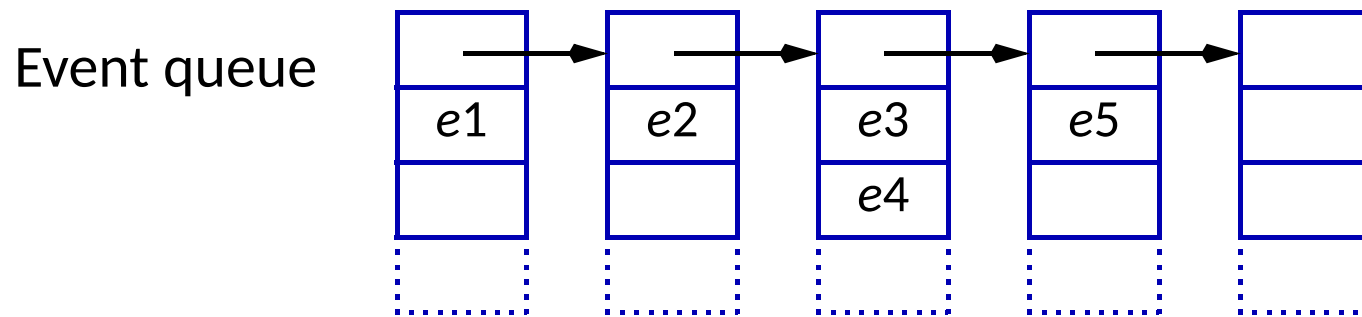
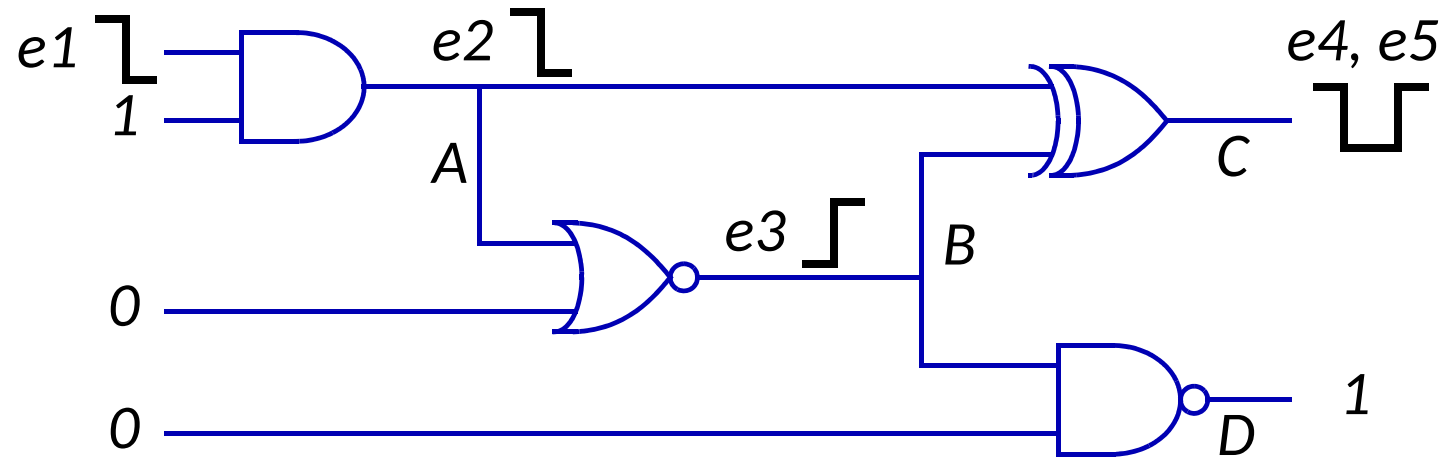
Event-Driven Simulation 2(5)



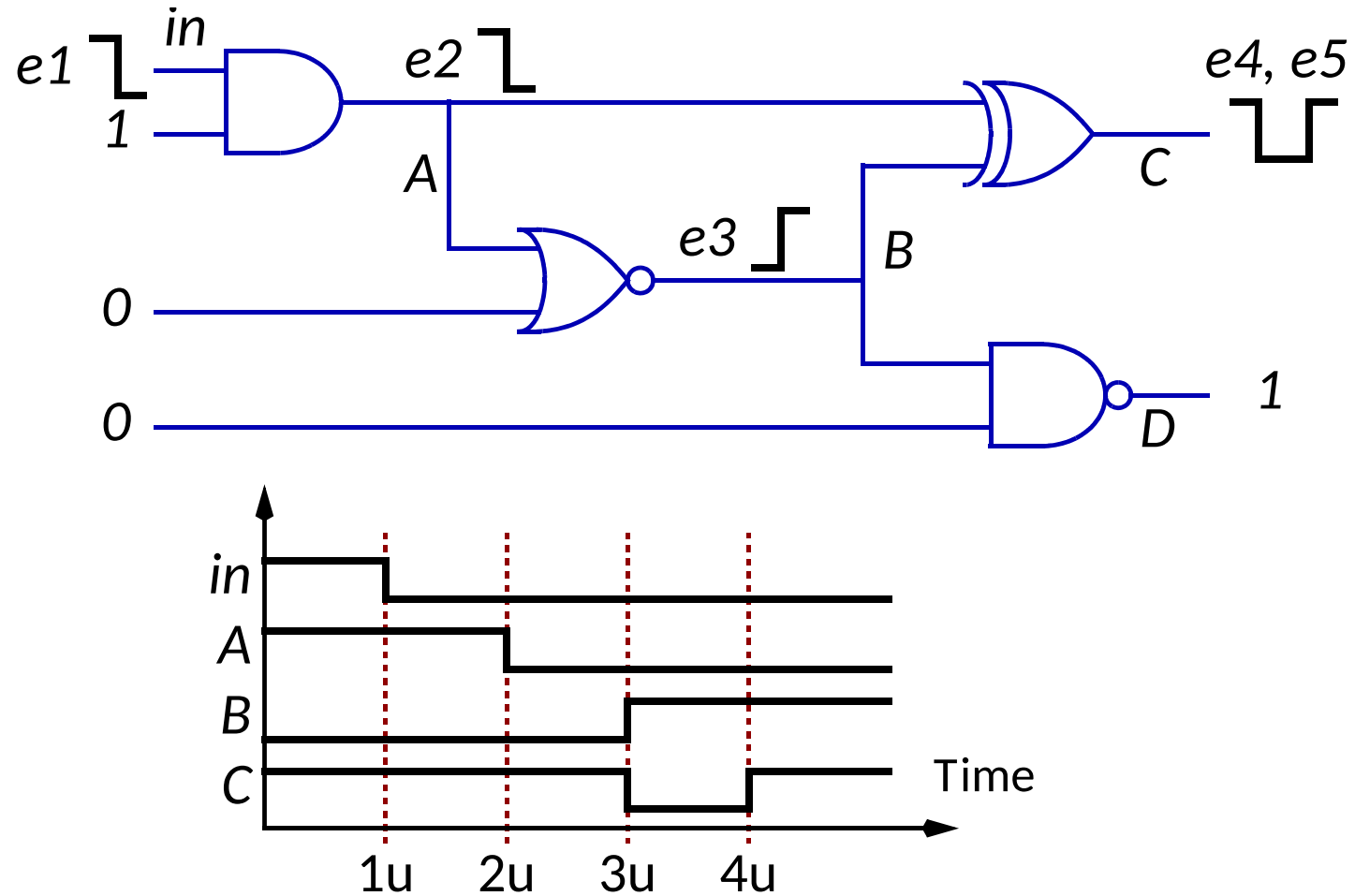
Event-Driven Simulation 3(5)



Event-Driven Simulation 4(5)



Event-Driven Simulation 5(5)



Functional vs RTL Signoff

- Logic simulation can be employed for either HDL code or for netlists of gates (we'll do both in the labs).
- *Functional signoff* is not hardware centric (VHDL code in lab 1).
- *RTL signoff* is hardware centric (synthesized netlist in lab 2-3) and allows verification of
 - power up.
 - shut down.
 - configuration modes.
 - reset.
 - power dissipation.

Functional Verification: Conclusion

- Electronic systems grow more complex \Rightarrow take functional verification very seriously, use designated verification engineers.
- Problem of university education: Limited complexity of engineering challenges \Rightarrow verification is relatively easy.
- Knowledge of simulation principles helps handle the inevitable EDA tools.
- Make use of testbenches.
- *Read more in [Vol 1/Ch 16 Digital simulation](#).*