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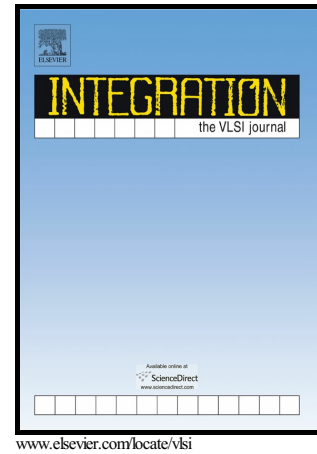


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Low Cost Optimized Trojan Secured Schedule at Behavioral Level for Single & Nested Loop Control Data Flow Graphs (Invited Paper)

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Abstract

Internet of Things (IoT) powered by high level synthesis (HLS) provides huge opportunity of progress in the area of hardware design. However, the present era of hardware design involves globalization which poses security threat to the design integrators that rely on third party intellectual property (IP) cores for increasing design productivity at reduced design time. This paper presents for the first time in the literature, a low cost optimized hardware Trojan secured HLS approach for hardware (or application specific core) designs that is based on single or nested loop control data flow graphs (CDFG) applications. The paper presents multiple novel vendor allocation schemes (each with its own attribute of delay and area) as security constraint that yield Trojan secured schedules at behavioral level. Demonstration of the proposed approach on a nested loop case study asserts our proposed theory. Results on standard benchmarks indicate significant reduction in final solution cost compared to a similar approach.

Keywords—IP; HLS; low cost; loop, CDFGs; Trojan; Security

I. INTRODUCTION

IoT is a system of interconnected electronic/computing devices such as application specific processor, microcontrollers, macro IPs, sensors etc. Many of these applications are algorithms which are data/control intensive driven with requirements of low latency, low power, high throughput etc. which mandates hardware realization. IoT thus presents opportunity for growth in the area of hardware design enabled by high level synthesis. However in high level synthesis, design library comprises of lot of third party IP (3PIP) cores as modules. This is because in order to develop complex designs, third party IPs are used to increase the design productivity. But there are serious security concerns for design integrators due to involvement of third party IPs owing to globalization involved [1,2,3]. In a 3PIP design, an adversary (an untrustworthy vendor) can deliberately infuse a Trojan logic resulting in erroneous computation of a digital design. Such Trojan logic may corrupt the IP by either external activation (such as antennas or sensors) or internal activation (such as internal states of finite state machine (FSM) or counters), resulting in malfunctioning. Therefore, to have a Trojan secured design it should be ensured during HLS [3-7] that any possible infection of 3PIP is detectable. Thus a paradigm shift in 'high level synthesis of IoT' is needed to incorporate trust/security objective besides power, area, energy etc. The designs generated should be equipped with low cost detection capability to fight against Trojan threats that targets alteration of computational value [3, 8, 9]. Typically, the hardware design with Trojan detection (security) capability generated through HLS methodology contains additional register transfer level (RTL) blocks as detection logic resulting from imposed security constraints. Trojan security constraints in the form of dual modular redundancy include duplication of functional modules which incur additional RTL logic, resulting in both delay and area overhead [10, 11]. As providing Trojan detection capability to HLS designs will in most cases require additional digital circuitry hence exploring a low cost solution is crucial besides providing Trojan security.

The novel contributions of this paper are as follows: (*Note: seminal version of this work is presented in iNIS 2015 [11]*)

- 1) Presents a novel low cost security aware HLS for optimized Trojan secured schedule within user constraints.
- 2) Presents a novel encoding to handle single and nested loop CDFGs using Particle Swarm Optimization (PSO) that explores an optimal vendor allocation procedure besides schedule configuration.
- 3) Provides lower cost Trojan security aware solutions that are of better quality than solutions obtained through existing approach [10].

Threat Models

By default a Trojan in a circuit is ineffective as its trigger is not enabled, which indicates there is no payload effect. Under certain rare specific criteria or events the Trojan gets triggered and then its payload makes the circuit output erroneous. Trojan triggering process can be classified into three categories [12]: 1) rare value triggered, 2) time-triggered and 3) both time and value triggered. Payload effect on a triggered Trojan depends on the application [13]. It may disable the system or leak the vital information or change the output value of the circuit.

Class of Trojan which only changes the computational output value of an IP, is targeted in this paper. Following threat model is considered in this paper: An IP which is malicious in nature designed either by an adversary/rogue in the third party house may be present within the library of a HLS tool. This IP behaves normally until triggered by a rogue or adversary and therefore detection of such Trojan becomes difficult during normal validation and manufacturing test.

There could also be other Trojans in 3PIPs which produce other effects such as Trojan that disable functionality i.e. does not allow generating any output, Trojan that leak information such as encrypted data/secret key etc and Trojan that inserts unwanted processing delay/creates reliability problem into the IP. However, these ones are not handled in this current paper.

II. RELATED PRIOR APPROACH

No effort has been done on generating an optimized Trojan security aware schedule for single loop and nested loops CDFGs (based on user constraints) during system level which is capable of providing runtime detection. There are different types of Trojan detection mechanisms available in the literature. Trojans may be inserted in a foundry, therefore different types of detection methods exist: (a) side channel analysis method [14] [15] [16], (b) logic testing method [10] [17] [18] [19] [22] [23]. Moreover, Trojans can also be inserted in an IP, 3PIP or in a design for which detection methods exist such as [20] [21], [10], [18]. Authors in [10], [18] duplicate the IPs in order to detect the presence of a Trojan, however none of the aforesaid approaches handle single and nested CDFG during Trojan detection.

In [14], authors presented a way to construct the fingerprint for all delay paths to detect Trojan circuits. However authors in [15], used the chip characteristics like path delay and power consumption (of the manufactured chips) to detect Trojan within. Authors in [16], proposed a side channel analysis approach to detect the Trojans. The approach is capable of detecting malicious hardware modifications where the Trojan detection is done based on path delay and power of the manufactured chips. Moreover, the approach mentioned in [17], generated a multiple supply transient current integration methodology to detect Trojan within the circuits. In [19] authors worked on code-coverage analysis where list of suspicious signals is identified to detect Trojans. The signals which stay stable during the coverage analysis are considered suspicious since Trojans do not change the state. The analysis includes line, statement, finite state machine (FSM) and path coverage at the RTL. In this approach, multiple test vectors are added to activate the uncovered parts of the design, but this requires a huge verification time. Moreover, redundant circuits are removed from the list of suspicious signals and components since these remain at the same logic level during verification. A sequential automatic test pattern generation (ATPG) is used to generate special patterns to change the signal values during simulation. An equivalence analysis is on the suspicious signal list in order to reduce the number of signals but the runtime overhead incurred is very high. Also, in this approach the quantity of suspicious signals keeps increasing with the increase in the complexity of Trojan inserted. Moreover, it is also difficult to achieve 100% code-coverage for all IPs. Also Trojan detection cannot be accomplished by tests used for detecting manufacturing faults, by analysis of parametric signals and reverse engineering. Moreover, test vectors generation [12] specific to Trojan detection also exists which depend on logic and fault simulator as well as on ATPG tool; however it incurs high runtime overhead/complexity for large number of circuit inputs during testing phase. Further, multiple excitation of rare occurrences (MERO) [22] based compact test pattern generation technique also exists for Trojan detection which starts with a golden netlist, random patterns, rare nodes and number of times to activate a node to rare value. However this technique is sensitive to Trojan sample size and number of times a rare condition is satisfied, in order to create a balance between estimate coverage and simulation time as well as coverage and test vector set size respectively. Without carefully selecting the sensitivity parameters, the results may be imprecise. Further, in case of a 3PIP in HLS, no golden model exists. Additionally in [23], procedure to identify circuit sites where Trojan may be inserted in untrusted foundry is also proposed, however it is not beneficial for detection of Trojan inserted in 3PIP module of HLS library at architectural level. None of the approaches mentioned above considers Trojan security of single and nested loop CDFGs during HLS.

Furthermore, authors in [10, 18] adopted a concurrent error detection (CED) approach for Trojan detection of DFGs. The approaches use a diverse set of 3PIP vendors for Trojan identification. However, the approaches do not have mechanism to explore efficient vendor allocation procedure for Trojan secured hardware. Efficient vendor allocation procedure for secured hardware is critical as it affects the final area-latency of the solution. Not considering exploration of an efficient vendor allocation procedure results in inferior quality solutions (as discussed later in results section). Further in [10, 18], the authors have only targeted DFGs and no efforts have been done to handle single and nested loop based CDFGs. In [20], a simulation based Trojan detection approach is proposed where the statistical correlation of Trojan logic compared to rest of the circuit is weak. A weighted circuit graph is generated based on the input and output of the circuit and their interconnection. The reachability plots identify the weak relation, thus identify the Trojan. In [21] the Trojan detection process is divided into tracer and checker. The tracer identifies those signals which contain un-activated entries. Out of these signals, the checker analyzes to determine the signals which contain redundant inputs and hence potentially affected by Trojan. None of the above techniques focus on solving the design space exploration (DSE) problem of generating an optimized Trojan secured schedule for single and nested loop CDFGs during HLS based on area-delay constraints capable of providing run-time detection. This paper aims to provide a low cost Trojan security aware HLS solution for single loop and nested loop CDFGs that affects computational output based on user specified constraints.

Fig.1. Trojan logic inserted by an adversary in a BCD Adder IP (inputs A & B); Note: When Enable (En) = 0 (rare event), then Trojan blocks get activated and produces incorrect computational output ($O_3O_2O_1O_0$)

III. MOTIVATIONAL EXAMPLE: TROJAN IN 3PIP MODULE

Let us assume a case during HLS which explains the Trojans in 3PIP module and their effect. Fig. 1 shows an IP of BCD adder designed by an untrusted third party. This IP is included in a HLS tool library. In the given figure, the area marked in red is a possible location where insertion of Trojan is made by malevolent alteration of the IP by an untrusted vendor. Under normal conditions, the module behaves like a functionally correct IP i.e. the upper mux passes '1' and lower mux passes '0' to finally form bits '0110' (decimal of '6'). This value is added to correct any invalid BCD summation produced and yield a result between valid decimal values 0 to 9. However if an adversary triggers the select (S) of the circuit through external or internal activation the Trojan logic will impact the system and change the computational output value. This is the only impact that this class of Trojan makes, as it does not attempt to steal/leak secret information (or cause any other harm). *Note: The existence of Trojan in a 3PIP in a HLS library typically cannot be prevented as it depends on the presence of a rogue element in the third party design house, which a design integrator is unaware of. The only way to minimize the possibility of a Trojan existence during high level synthesis is using IPs from trusted third party sources in tool library. As there is no guarantee of a Trojan free 3PIP in HLS library, therefore low cost Trojan security methodology is critical. Another alternative to prevent Trojan is designing all IPs within in-house design team (i.e. without outsourcing to third party vendors). However, this drastically increases design cycle time and affects design productivity.*

To exploit Trojan for affecting computational output value, an attacker does not need to know the architecture of the system. The triggering of the Trojan can be achieved through numerous possibilities such as activation through Trojan time bomb (FSM counter), sensing a specific design signal, external antenna etc. Activation of the Trojan logic, would directly impact the computational output value of the entire system. Lower level techniques like side channel analysis and RTL simulation cannot detect such Trojans because Trojan logics can only be triggered through specific predefined conditions like FSM counter value, sensing a specific design signal, external antenna etc. Also, Trojans are only activated by a rogue when deployed in real time situation for change in computational output value. Therefore, even after a functional verification check (RTL simulation), the Trojan within IP might get undetected since the IP will perform normally/ correctly due to it being dormant at that stage. This is because of two reasons: (a) the IP cores are generally considered pre-verified functional blocks thus many design integrators do not rigorously perform verification. This provides opportunity to escape any Trojan detection; (b) During functional verification/RTL simulation/other lower level tests (during pre-integration phase), the Trojan logic does not get triggered easily as it gets activated only under rare condition. It normally remains dormant and escapes the detection process performed by in-house design integrator team. For example, consider a scenario for Fig.1 where the BCD adder IP black box provides three input signals Enable (En), A (4-bit) and B (4-bit) and output signal $O_3O_2O_1O_0$. During functional verification/RTL simulation test, En is always be kept 'on' (by feeding logic '1' in test bench) by the design verification team as they falsely believe it to be an IP enable signal (without being aware that it is designed in disguise to act as a trigger signal for Trojan). Under En = 1, the BCD adder IP would behave normally, thus passing RTL simulation/functional verification test. This is because Trojan would have only got activated under rare event (enable reset i.e. En = 0 which normally the design integrator would never make during verification test). Thus after integration, whenever the IP 'En' signal remains 'off', it actually continues to work maliciously (due to Trojan logic activation) thereby impacting the remaining system through wrong output value. Moreover, detection of Trojan in an IP during HLS is very difficult since there is no golden IPs available.

TABLE 1 PROPOSED NOMENCLATURE

p	Population size	V_j	Type of vendor
X_i	Resource set of a design solution	C_T^{DMR}	Total CS required to execute the loop of CDFG DMR completely
D	Total available resource types	C_{body}^{DMR}	Number of CS required to execute loop body of CDFG DMR once
$N(R_D)$	Number of instances of a resource type 'D'	I	Maximum number of iteration (loop count)
U	Candidate unrolling factor value	C_{first}^{DMR}	Number of CS required to execute first iteration of the $CDFG^{DMR}$
I_1, I_2	Corresponding iteration counts of loops	Δ	Delay of one CS in nanoseconds
A_v	Vendor allocation Procedure	$D(op_i^{V_j})$	Delay of operation i , assigned to vendor V_j
A_T^{DMR}	Total area of a dual modular redundant (DMR) design	$c.s$	Control steps
A_{cons}^{DMR}	User specified area constraint	T_{max}^{DMR}	DMR solution with maximum time in the design space
A_{max}^{DMR}	DMR solution with maximum area in the design space	$C_j(X_i)$	Cost of solution with resource set X_i
T_E^{DMR}	Total execution time of a dual modular redundant (DMR) design	\vec{R}_n	Resource array
T_{cons}	User specified execution time constraint	$R_{d_i}^+$	New resource value or 'U' value of particle X_i in d^{th} dimension
$R_i^{V_j}$	Number of instances utilized from vendor V_j for a resource type R_i	$V_{d_i}^+$	New velocity of i^{th} particle in d^{th} dimension
n	Maximum number of instances of resource type R_i for vendor V_i	k and ' k '	Maximum value of node
$A(R_i^{V_j})$	Area of a resource type (R_i) corresponding to vendor (V_j)	$D(op_i^{V_j})$	Delay of operation i , assigned to vendor V_j
$R_{d_{lbi}}$	Resource value of X_{lbi} in d^{th} dimension	R_{d_i}	Resource value or 'U' value of particle X_i in d^{th} dimension
ω	Inertia weight	$R_{d_{gb}}$	Resource value of X_{gb} in d^{th} dimension

b_1, b_2	Acceleration coefficients which balances the effect of cognitive and social factor during exploration	r_1, r_2	Random numbers providing stochasticity
X_{lbi}^{SL}	Local best particle of Single loop and Nested loop	X_{gb}^{SL}	Global best particle of Single loop and Nested loop
X_{lbi}^{NL}		X_{gb}^{NL}	

IV. SECURITY AWARE HLS: FORMULATION AND MODELS

The aim of proposed approach is to explore the design space, comprising of candidate solutions for scheduling and vendor assignment, for single and nested loop CDFGs (as discussed in the upcoming sections). The output of the exploration results in a user constraint optimized Trojan detectable schedule for loop based CDFGs. To instantiate the proposed approach, example of area and delay constraints have been considered in the current paper.

A. Problem Definition

1) Encoding for Handling Single Loop Based CDFGs:

Determine an optimal solution,

$$X_i = \{N(R_1), N(R_2), \dots, N(R_D), U, A_v\},$$

while exploring the design space of a given single loop based CDFG and satisfying conflicting user constraints and minimizing the overall cost.

2) Encoding for Handling Nested Loop Based CDFGs:

Determine an optimal solution,

$$X_i = \{N(R_1), N(R_2), \dots, N(R_D), I_1, I_2, A_v\},$$

while exploring the design space of a given nested loop based CDFG and satisfying conflicting user constraints and minimizing the overall cost.

The problem can be formulated as:

$$\begin{aligned} &\text{Find: Optimal } (X_i), \\ &\text{with minimum hybrid Cost } t(A_T^{DMR}, T_E^{DMR}), \\ &\text{subjected to: } A_T^{DMR} \leq A_{cons} \text{ and } T_E^{DMR} \leq T_{cons} \text{ and Trojan security,} \end{aligned}$$

The variables have been defined in Table 1. The details and impact of vendor allocation procedure are explained in later sections.

B. Evaluation Models

In the proposed work, each particle position represents a hybrid encoding of candidate solutions for scheduling resource configurations (X_i), loop unrolling factor (U) and vendor assignment procedure information (A_v) in the design space [11].

Fig.2. Proposed Security aware PSO-DSE methodology

1) Area model: Total area consumed (A_T^{DMR}) by a resource set is given by:

$$A_T^{DMR} = \sum_{j=1}^2 \sum_{i=1}^n (A(R_i^{V_j}) * R_i^{V_j}) \quad (1)$$

The variables have been defined in Table 1. Note: The area component includes area due to functional resources, interconnect units (mux and demux), comparator (for error detection) as well as overhead incurred from internal buffering (during temporary storage of operation output in DMR scheduling).

2) Execution Time (Delay) Model:

a) For Handling Single Loop Based CDFGs

The execution delay is evaluated after derivation of the delay model using the following two cases [11]:

Case 1: When ' U ' is equal to one (indicates no unrolling) then:

Total # of control steps (CS) = # of CS required executing loop body once * number of duplicate iterations of loop body:

$$C_T^{DMR} = (C_{body}^{DMR} * \alpha) = \left(C_{body}^{DMR} * \left\lceil \frac{I}{U} \right\rceil \right) \quad (2)$$

Where $\alpha = \lceil I/U \rceil$

$$\text{Since } U=I, \quad C_T^{DMR} = (C_{body}^{DMR} * I) \quad (3)$$

The variables have been defined in Table 1. *Note: Eqn. (2) is also valid when 'U' evenly divides the loop count (I) (i.e. $I \% U == 0$)*
Case 2: When 'U' unevenly divides I: In such a case, $I \bmod U$ iterations will be executed sequentially, therefore, the total no. of CSs is:

$$C_T^{DMR} = \left(C_{body}^{DMR} * \left\lfloor \frac{I}{U} \right\rfloor \right) + (I \bmod U) * C_{first}^{DMR} \quad (4)$$

{CSs for unrolled loop} {CSs for sequential loop}

The variables have been defined in Table 1. Hence the execution time for the system is calculated as:

$$T_E^{DMR} = \Delta * C_T^{DMR} \quad (5)$$

b) For Handling Nested Loop Based CDFGs:

The delay for nested loop based CDFG is evaluated by fully unrolling the CDFG on the basis of iteration counts (I_1, I_2). Further, for given 'D' functional resources the delay is:

$$T_E^{DMR} = \sum_{c,s=1}^{c,s(\max)} \sum_{j=1}^2 \text{Max}(D(op_i^{Vj}), \dots, D(op_k^{Vj}), D(op_i^{Vj}), \dots, D(op_k^{Vj})) \quad (6)$$

Where, $1 \leq i \leq k$ and ' $i \leq i \leq k$ '. (Here, operations in original and duplicate is labeled as i and ' i ' respectively). The variables have been defined in Table 1.

3) *Fitness function:* The fitness function (considering execution time and area consumption of a solution) is inspired from [24] and is formulated as:

$$C_f(X_i) = W_1 \frac{A_T^{DMR} - A_{cons}}{A_{max}^{DMR}} + W_2 \frac{T_E^{DMR} - T_{cons}}{T_{max}^{DMR}} \quad (7)$$

The variables have been defined in Table 1. W_1 and W_2 are the user defined weights both kept at $\frac{1}{2}$ during exploration to provide equal preference.

V. PROPOSED SECURITY AWARE PSO-DSE METHODOLOGY

The proposed DSE methodology is driven through PSO process for generation of optimal Trojan secured schedule based on area-delay constraint (shown in Fig. 2). A Trojan secured schedule is obtained as follows: a) duplicate all the operations of original CDFG to form a duplicate one, which together is called '*dual modular redundant system*' b) Perform scheduling of operations of DMR system to obtained scheduled DMR c) Finally, perform hardware allocation of operations of DMR schedule based on Trojan security allocation rules (i.e. assigning hardware from distinct/multiple vendors to similar operations of original and duplicate unit). The proposed methodology presents a novel encoding scheme for the particle which comprises of resource configuration (\vec{R}_n), unrolling factor (U) in case of single loop applications, iteration counts (I_1, I_2) in case of nested loop applications and vendor allocation procedure type (A_v). Therefore, a particle position (candidate design solution) is labeled as X_i for :

$$X_i = (\vec{R}_n, U, A_v) \quad (8)$$

in case of single loop based CDFGs, while

$$X_i = (\vec{R}_n, I_1, I_2, A_v) \quad (9)$$

for nested loop based CDFGs.

Where, \vec{R}_n indicates the resource array (resource configuration e.g. number of adders, multipliers etc.). The reason behind incorporating the last dimension with vendor allocation procedure type ' A_v ' is elaborated in section V.E.

A. Motivation of proposed work

The proposed work is motivated from the following three uncovered aspects of the literature: **(a)** during detection of Trojans, exploration to optimize the Trojan secured schedule based on user constraint is not performed. **(b)** recent approaches [10] have only attempted to keep distinct hardware (from different vendor) assignment to similar operations in original and duplicate units of a dual modular redundant system to achieve detection but have not probed into a detailed vendor allocation procedure of hardware units inside a DMR schedule (as there can be more than one ways to implement distinct hardware assignment condition). **(c)** Exploration of simultaneous Trojan secured schedule and unrolling factor for single loop based control data flow graphs have not been addressed. Further, integrating the exploration of an efficient vendor allocation procedure concurrently with the aforesaid task has not been performed.

B. Pre-processing of unrolling factor candidates

A pre-processing algorithm for unrolling factors has been deployed to screen the large unfit unrolling factor (which are potential sources for greater delay due to multiple sequential loops involved) candidates before exploration initiates.

1) For Single Loop Based CDFGs

The position of particle as defined earlier in eqn. (8) & (9) can be expanded as follows. The particle position 'X_i' is given as:

$$X_i = (N(R_1), (N(R_2), \dots, (N(R_d), (N(R_D), U, A_v)) \quad (10)$$

The particles are uniformly distributed over the design space and are initialized as:

$$X_1 = (\min(R_1), \min(R_2), \dots, \min(R_D), \min(U), A_v) \quad (11)$$

$$X_2 = (\max(R_1), \max(R_2), \dots, \max(R_D), \max(U), A_v) \quad (12)$$

$$X_3 = ((\min(R_1) + \max(R_1))/2, \dots, ((\min(R_D) + \max(R_D))/2, (\min(U) + \max(U))/2, A_v) \quad (13)$$

While rest of the particles (X₄...X_n) is initialized by:-

$$X_n = ((\min(R_1) + \max(R_1))/2 \pm \alpha, (\min(R_2) + \max(R_2))/2 \pm \alpha, \dots, (\min(R_d) + \max(R_d))/2 \pm \alpha, \dots, (\min(U) + \max(U))/2 \pm \alpha, \text{random}(A_v)) \quad (14)$$

Where, α is a random integer between *min* and *max* of particular resource type or unrolling factor.

2) For Nested Loop Based CDFGs

The position of particle as defined earlier in eqn. (8) & (9) can be expanded as follows. Particle position 'X_i' is given as:

$$X_i = (N(R_1), (N(R_2), \dots, (N(R_d), (N(R_D), I_1, I_2, A_v)) \quad (15)$$

The particles are uniformly distributed over the design space and are initialized as:

$$X_1 = (\min(R_1), \min(R_2), \dots, \min(R_D), I_1, I_2, A_v) \quad (16)$$

$$X_2 = (\max(R_1), \max(R_2), \dots, \max(R_D), \max(U), A_v) \quad (17)$$

$$X_3 = ((\min(R_1) + \max(R_1))/2, \dots, ((\min(R_D) + \max(R_D))/2, (\min(U) + \max(U))/2, A_v) \quad (18)$$

While rest of the particles (X₄...X_n) is initialized by:-

$$X_n = ((\min(R_1) + \max(R_1))/2 \pm \alpha, (\min(R_2) + \max(R_2))/2 \pm \alpha, \dots, (\min(R_d) + \max(R_d))/2 \pm \alpha, \dots, (\min(U) + \max(U))/2 \pm \alpha, \text{random}(A_v)) \quad (19)$$

Where, α is a random integer between *min* and *max* of particular resource type or unrolling factor.

Note: The initialization process selects the best positions through all combinations of A_v value. Further, these selected positions are used as initial solutions to begin the exploration.

D. Particle Movement using Velocity

In PSO-DSE, each dimension (*d*) of a particle position X_i (except the last dimension) is updated using the following function [24]:

$$R_{d_i}^+ = R_{d_i} + V_{d_i}^+ \quad (20)$$

The variable V_{d_i}⁺ is updated by eqn. (21) as follows:

$$V_{d_i}^+ = \omega V_{d_i} + b_1 r_1 [R_{d_{lbi}} - R_{d_i}] + b_2 r_2 [R_{d_{gb}} - R_{d_i}] \quad (21)$$

The local best (X_{lbi}^{SL} and X_{lbi}^{NL} for single and nested loop) and global best (X_{gb}^{SL} and X_{gb}^{NL} for single and nested loop) particle positions are updated using eqn. (22) & (23) for single loop based applications while using eqn. (24) & (25) for nested loop based applications:

$$X_{lbi}^{SL} = \{R_{lbi}, \dots, R_{D-1_{lbi}}, U\} \quad (22)$$

$$X_{gb}^{SL} = \{R_{lgb}, \dots, R_{D-1_{gb}}, U\} \quad (23)$$

$$X_{lbi}^{NL} = \{R_{lbi}, \dots, R_{D-1_{lbi}}, I_1, I_2\} \quad (24)$$

$$X_{gb}^{NL} = \{R_{lgb}, \dots, R_{D-1_{gb}}, I_1, I_2\} \quad (25)$$

The variables have been defined in Table 1. The particle positions are used to generate a DMR schedule (SDFG^{DMR}) with distinct vendor assignment rule to detect the presence of Trojan.

E. Motivation of using Vendor Type 'A_v' in Particle Encoding during DSE

The proposed approach uses two distinct/multiple vendors for assigning to similar operations of original and duplicate units. The goal is to perform the similar operations of DMR not only through distinct hardware but also distinct hardware from distinct vendors. This enables original and duplicate version to produce computationally different output in case hardware of one vendor is Trojan infected. However, distinct vendor assignment rule for Trojan detection can be implemented in more than one ways; therefore the most efficient allocation procedure of hardware units from distinct vendor to similar operations (in original and duplicate) is explored through proposed scheme. The schedule with allocation generated indicates a Trojan secured system. The obtained Trojan secured DMR schedule is evaluated on metrics of area and delay. The cost of the Trojan secured scheduling solution explored includes area component due to functional resources, interconnect units (mux and demux), comparator as well as overhead incurred from internal buffering (temporary storage of operation output). This internal buffering contributes to the total area since in a DMR similar operation is being done on the two copies (original and duplicate) at different times. The system needs to keep the outputs from both units stored in some internal buffer to compare only when both outputs are ready. This

process of evaluating design solutions (particle positions) evolves through the proposed DSE using PSO to generate an optimal Trojan fault secured DMR system that satisfies A_{cons} , T_{cons} as well as minimizes hybrid cost.

Fig. 3: Scheduling and Binding of FIR for $X_i = 2(+), 2(*), 2(<)$, $U=2$, $I=4$ based on Vendor Allocation Mode $A_v = 00$; $T_E^{DMR} = 45080$ ns and $A_T^{DMR} = 13064$ au

Fig. 4: Scheduling and Binding of FIR for $X_i = 2(+), 2(*), 2(<)$, $U=2$, $I=4$ based on Vendor Allocation Mode $A_v = 01$; $T_E^{DMR} = 43080$ ns and $A_T^{DMR} = 17996$ au

Fig. 5: Scheduling and Binding of FIR for $X_i = 2(+), 2(*), 2(<)$, $U=2$, $I=4$ based on Vendor Allocation Mode $A_v = 10$; $T_E^{DMR} = 45080$ ns and $A_T^{DMR} = 13064$ au

Fig. 6: Scheduling and Binding of FIR for $X_i = 2(+), 2(*), 2(<)$, $U=2$, $I=4$ based on Vendor Allocation Mode $A_v = 11$; $T_E^{DMR} = 45070$ ns and $A_T^{DMR} = 15096$ au

Fig. 7(a) 'C' code for original loop of Autocorrelation benchmark (adopted from [30])

Fig 7(b): Scheduling and Binding of Auto Correlation Filter for $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, based on vendor allocation mode $A_v = 00$. $T_E^{DMR} = 45350.0$ ns and $A_T^{DMR} = 17996.0$ au

In order to detect Trojans a minimum of two vendors is always needed to provide distinctness (Note: cases where Trojans lead to disabling hardware units are ignored in this paper, as it falls outside the scope of the proposed work). However, technique of usage of the two vendors during allocation inside the DMR scheduling (i.e. assignment process of each vendor IPs inside the system during allocation) dictates the final latency and area of entire system. This is because same resource type/IP from two different vendors have different area and delay. Note: Further it is assumed that the IP characteristics from vendors (V_1 and V_2) are as follows: Multiplier and adder provided by vendor V_1 has area = '2468au' & '2034au', latency = '10000ns' & '265ns' (where values have been extracted from [24]; assuming 1 au = 1 transistor) while multiplier and adder provided by vendor V_2 has area = '2464au' & '2032au', latency = '11000ns' & '270ns' respectively. The values of area and delay of modules assumed is with respect to 90nm technology scale adopted from [25], [26], [27]. Hence, merely using distinctive vendor assignment for detection without probing into the procedure of allocation (assignment) of vendor type in DMR system may lead to skipping of an alternate better solution in context of DSE of Trojan secured schedule. Therefore, exploration of an additional dimension ' A_v ' (indicating allocation procedure of IP's from different vendor type) which can either be in one of the following distinct vendor allocation modes, mode 1: '00' or mode 2: '01' or Type 3: '10' or Type 4: '11' is incorporated in the bacterium encoding along with resource array. The value of ' A_v ' as '00' or '01' or '10' or '11' is interpreted as follows:

Rule 1: Vendor allocation procedure (Type 1): $A_v = V_x V_y = 00$

- Alternate vendor assignment to operations in control step of a unit. (Example in Fig. 3, operation 1 & 3 assigned alternatively to ' V_1 ' and ' V_2 '. Next multiplication if any would have been assigned to ' V_1 ' alternately).
- Similar operations of both U^{OG} and U^{DP} being assigned to different vendors.

Rule 2: Vendor allocation procedure (Type 2): $A_v = V_x V_y = 01$

- All operations of a specific unit being strictly assigned to resources of same vendor type (say: all operations of original unit strictly assigned to vendor ' V_1 ' and all operations of duplication to vendor ' V_2 ').
- Similar operations of both original unit (U^{OG}) and duplicate unit (U^{DP}) being assigned to different vendors.

Rule 3: Vendor allocation procedure (Type 3): $A_v = V_x V_y = 10$

- All operations within critical path (p^{cri}) of a specific unit being strictly assigned to a vendor type while all operations of non critical path through **alternate** vendor type (say all operations of original unit belonging to p^{cri} strictly assigned to vendor ' V_1 ' while all operations of non p^{cri} to alternate vendor ' V_2 ').
- Operations of critical path of U^{OG} and U^{DP} are assigned to distinct vendors.
- Similar operations of non critical path in both U^{OG} and U^{DP} being assigned to different vendors.

Rule 4: Vendor allocation procedure (Type 4): $A_v = V_x V_y = 11$

- **Alternate** vendor assignment to operations belonging to subsequent unrolled nested loop iterations within a unit.
- Similar operations of unrolled nested loop iteration in both U^{OG} and U^{DP} assigned to different vendors.

Fig 8: Scheduling and Binding of Auto Correlation Filter for $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, based on vendor allocation mode $A_v = 01$. $T_E^{DMR} = 43350.0\text{ns}$ and $A_T^{DMR} = 31060.0\text{au}$

Fig 9: Scheduling and Binding of Auto Correlation Filter for $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, based on vendor allocation mode $A_v = 10$; $T_E^{DMR} = 43340.0\text{ns}$ and $A_T^{DMR} = 31060.0\text{au}$

F. Handling of Single Loop CDFGs: An Example

As described before in section V.E, consider cases which use a list scheduled CDFG^{DMR} for FIR benchmark unrolled twice with: (i) resource constraint of $X_i = 2(+), 2(*), 1(<)$; $U = 2$; iteration count (I) = 4 $A_v = 00$ (illustrated in Fig. 3). The corresponding $T_E^{DMR} = 45080\text{ns}$ and $A_T^{DMR} = 13064\text{au}$ (ii) $X_i = 2(+), 2(*), 1(<)$; $U = 2$; iteration count (I) = 4 $A_v = 01$ (illustrated in Fig. 4). The corresponding $T_E^{DMR} = 43080\text{ns}$ and $A_T^{DMR} = 17996\text{au}$ (iii) $X_i = 2(+), 2(*), 1(<)$; $U = 2$; iteration count (I) = 4, $A_v = 10$ (illustrated in Fig. 5). The corresponding $T_E^{DMR} = 45080\text{ns}$ and $A_T^{DMR} = 13064\text{au}$ (iv) $X_i = 2(+), 2(*), 1(<)$; $U = 2$; iteration count (I) = 4, $A_v = 11$ (illustrated in Fig. 6). The corresponding $T_E^{DMR} = 45070\text{ns}$ and $A_T^{DMR} = 15096\text{au}$.

Clearly a difference is observed in the delay and area of the generated scheduling solutions where all abide the rule of hardware from distinct vendor assignment to similar operations for detectability. The schedules with allocation generated in Fig 3, 4, 5 and 6 are Trojan secured (with two distinct vendors needed), however one is better than the other in different parameter. Only using distinct vendor assignment for hardware without probing into the procedure of allocation of vendor type in DMR system may lead to missing of better solutions in context of DSE. Therefore in context of DSE, it is worth to explore the additional dimension incorporated in the proposed particle encoding in eqn. (8) and (9).

Fig 10: Scheduling and Binding of Auto Correlation Filter for $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, based on vendor allocation mode $A_v = 11$; $T_E^{DMR} = 45350.0\text{ns}$ and $A_T^{DMR} = 26128.0\text{au}$

G. Handling of Nested Loop Based CDFGs : An Example

On the basis of discussion in section IV.E, consider the four cases as illustrated in Fig. 7(b), 8, 9 and 10; which show the list scheduled nested loop based CDFG for Auto Correlation Filter shown in Fig. 7(a) with $I_1=3$, $I_2=2$ iteration counts respectively with:

- Resource constraint of $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, $A_v = 00$. The corresponding $T_E^{DMR} = 45350.0\text{ns}$ and $A_T^{DMR} = 17996.0\text{au}$. As seen from Fig. 7(b) operations of U^{OG} are assigned through alternate vendor assignment in a control step, for example, operation 1 and 5 are assigned alternatively to 'V₁' and 'V₂'. For Trojan detection distinct vendor assignment is given to U^{DP} based on rule 1.
- Resource constraint of $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, $A_v = 01$. The corresponding $T_E^{DMR} = 43350.0\text{ns}$ and $A_T^{DMR} = 31060.0\text{au}$. For example, in Fig. 8 all operations of U^{OG} are assigned to vendor type 'V₁'. For Trojan detection distinct vendor assignment is given to U^{DP} based on rule 2.
- Resource constraint of $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, $A_v = 10$. The corresponding $T_E^{DMR} = 43340.0\text{ns}$ and $A_T^{DMR} = 31060.0\text{au}$. From Fig. 9 it can be seen that, all operations belonging to $p^{cri}(1, 2, 4, 5, 6, 8, 9, 10, 12)$ of U^{OG} are assigned to vendor type V₁, while all operations belonging to $p^{non-cri}(3, 7, 11, 15, 16, 13, 14)$ are assigned through alternate vendor type. For Trojan detection distinct vendor assignment is given to U^{DP} based on rule 3.
- Resource constraint of $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, $A_v = 11$. The corresponding $T_E^{DMR} = 45350.0\text{ns}$ and $A_T^{DMR} = 26128.0\text{au}$. G_n is the unrolled nested loop iterations; where, G_1 , G_2 , G_3 are unrolled nested loop with respect to first, second and third outer loop iteration respectively. In Fig. 10, operations belonging to subsequent unrolled nested loop iterations are assigned through alternate vendor procedure. For example, operations 1, 2, 3 and 4 belonging to G_1 are assigned to V₁, while operations 5, 6, 7 and 8 belonging to G_2 are assigned to V₂ respectively. For Trojan detection distinct vendor assignment is given to U^{DP} based on rule 4.

It can be seen from that there is a difference in the area and delay values of the generated Trojan secured scheduling solutions, though all have distinct vendor assignment to similar operations of original and duplicate unit for detectability. The CDFG^{DMR} schedules with vendor allocation details generated in Fig. 7(b), 8, 9 and 10 are Trojan secured (with two vendors required in all cases) with duplication of all the operations of the original. For example, suppose we have a candidate design solution $X_i = 2(+), 3(*), 2(<)$, $I_1 = 3$, $I_2 = 2$, A_v for building a Trojan secured DMR schedule. This indicates that the nested CDFG has two loops (I_1

and I_2 , with I_1 as outer loop and I_2 as the inner loop, see Fig. 7(a)), where 2 iterations of I_2 are carried out with 3 iterations of I_1 and duplication is done for all the operations corresponding to the iteration count to create a complete unrolled CDFG^{DMR}. Once the CDFG^{DMR} is created both the original and duplicate unit will be scheduled based on the resource constraint information available in X_i : 2(+), 3(*), 2(<). However, it can be seen that one schedule is better than its counterpart in either area or execution delay depending on the mode A_v , irrespective of the fact that the schedule resources and iteration counts are same. Therefore in context of low cost scheduling, exploration of A_v (which can either be Mode 1: '00' or Mode 2: '01' or Mode 3: '10' or Mode 4: '11') dimension along with resource array is important to design a Trojan secured schedule for nested loop applications.

Once the schedules and the corresponding T_E^{DMR} and A_T^{DMR} are obtained, the cost of obtained Trojan secured DMR schedule is further evaluated using eqn. (7). This process of evaluating design solutions (particle positions) evolves through the proposed DSE using PSO to generate an optimal Trojan fault secured DMR system that satisfies A_{cons} , T_{cons} as well as minimizes hybrid cost.

TABLE 2 RESULTS OF PROPOSED APPROACH

Benchmark [29]	T_{cons} (us)	T_E^{DMR} (us)	A_{cons} (au)	A_T^{DMR} (au)
AUTOCORRELATION FILTER ($I_1=2, I_2=3$)	50.04	34.62	19000	17996
AUTOCORRELATION FILTER ($I_1=3, I_2=3$)	65.00	35.43	23000	22928
AUTOCORRELATION FILTER ($I_1=4, I_2=2$)	55.00	24.43	28000	27860
DHMC ($I_1=2, I_2=3$)	420.00	262.33	35000	33524
DHMC ($I_1=3, I_2=3$)	620.00	507.35	35000	26994
DHMC ($I_1=4, I_2=2$)	550.00	327.80	36000	35992
DIFFERENTIAL EQUATION (U = 4)	182.21	120.81	43301	26994
FFT (U = 2)	200.00	156.85	41000	26128
FIR (U = 6)	40.00	23.62	27000	22928
TEST CASE (U = 6)	130.00	99.00	29000	22062

Note: 1a.u. = 1 Transistor; us = microseconds

H. Velocity clamping , Resource Clamping and Stopping Criterion

The velocity clamping is necessary to control unwarranted drift control. Additionally, resource clamping is performed to control 'swarm explosion'. Both clampings have been adopted from [24]. The proposed methodology terminates when a) the maximum number of iterations exceeds 100 (S_1), or when no improvement is visible in X_{gb} over ' ϵ ' number of iteration. ($\epsilon=10$) (S_2).

VI. EXPERIMENTAL RESULTS

This section shows the results of the proposed methodology, divided into following sub-sections: (a) Experimental setup and benchmarks which discusses the machine setup, control parameters and benchmarks used (b) Analysis of results which presents the proposed results for single and nested loop based CDFGs based on user constraints and (c) Comparison with related prior research in terms of implementation cost and final architectural solution explored.

TABLE 3 RESULTS OF APPROACH [10]

Benchmark [29]	T_{cons} (us)	T_E^{DMR} (us)	A_{cons} (au)	A_T^{DMR} (au)
AUTOCORRELATION FILTER ($I_1=2, I_2=3$)	50.04	33.81	19000	30396
AUTOCORRELATION FILTER ($I_1=3, I_2=3$)	65.00	45.08	23000	31278
AUTOCORRELATION FILTER ($I_1=4, I_2=2$)	55.00	45.08	28000	31404
DHMC ($I_1=2, I_2=3$)	420.00	286.35	35000	52824
DHMC ($I_1=3, I_2=3$)	620.00	440.89	35000	66054
DHMC ($I_1=4, I_2=2$)	550.00	327.80	32000	35992
DIFFERENTIAL EQUATION ($U=4$)	182.21	131.27	43301	29640
FFT ($U=4$)	200.00	179.24	41000	33974
FIR ($U=6$)	40.00	34.08	27000	24692
TEST CASE ($U=6$)	130.00	99.27	29000	23826

Note: 1a.u. = 1 Transistor; us = microseconds

A. Experimental Setup and Benchmarks

The proposed approach as well as [10] both have been implemented in java and run on Intel Core-i5-3210M CPU with 3MB L3 cache memory, 4GB DDR3 primary memory and processor frequency of 2.5 GHz. During experiment, 15 runs were executed for proposed PSO-DSE. During the experiments conducted, it was found that the proposed approach is scalable and is able to handle problems of medium/large size. Further during the experiments, the following optimal settings from [24] for PSO framework were fixed: ω (inertia weight) = linearly decreasing between 0.9 to 0.1; $b1$ & $b2$ (acceleration coefficient) = 2; $r1$ & $r2$ (random numbers) = 1; stopping criterion = S_1 or S_2 ; p (swarm size) = 3 or 5 or 7. The benchmarks used during experiments comprises of loop based control intensive applications derived from multimedia and signal processing domain. More explicitly, single and nested loop based control data flow graphs were used as standard benchmarks from [23, 25]. These benchmarks are typically data and control intensive applications which consume huge power and processing delay. The benchmark application is made up of operations that utilize third party micro IPs such as adder, subtractor, comparator etc. present in the tool library. For evaluating the Trojan detection capability (i.e. testing purpose) of the proposed approach, Trojan logic was inserted into these micro 3PIPs by altering the digital logic of the circuit through a combination of additional gates, multiplexers, flip-flops, control signals etc. A demonstrative example which shows the procedure for Trojan insertion in BCD adder present as 3PIP module in HLS tool library is shown in Fig. 1 (section III). This type of modelling is realistic as mostly in real life cases the Trojan gets triggered by activation of a control line (such as 'En' in Fig.1).

TABLE 4 COMPARISON OF PROPOSED DSE APPROACH WITH [10]

Benchmark [29] [30]	Final design solution for Trojan Secured schedule (Proposed)	Final design solution for Trojan Secured schedule [10]	Cost of final solution (proposed)	Cost of final solution [10]
AUTOCORRELATION FILTER ($I_1=2, I_2=3$)	2(+), 4(*), 2(<), $I_1=2, I_2=3, A_v=00$	3(+), 5(*), 2(<), $I_1=2, I_2=3, A_v=01$	-0.134	0.096
AUTOCORRELATION FILTER ($I_1=3, I_2=3$)	2(+), 8(*), 2(<), $I_1=3, I_2=3, A_v=00$	3(+), 5(*), 2(<), $I_1=3, I_2=3, A_v=01$	-0.149	0.025

AUTOCORRELATION FILTER ($I_1=4, I_2=2$)	2(+), 8(*), 2(<), $I_1=4, I_2=2, A_v=00$	3(+), 5(*), 2(<), $I_1=4, I_2=2, A_v=01$	-0.173	-0.017
DHMC ($I_1=2, I_2=3$)	4(+), 6(*), 2(<), $I_1=2, I_2=3, A_v=00$	3(+), 5(*), 2(<), $I_1=2, I_2=3, A_v=01$	-0.140	0.057
DHMC ($I_1=3, I_2=2$)	3(+), 4(*), 2(<), $I_1=3, I_2=2, A_v=00$	3(+), 5(*), 2(<), $I_1=3, I_2=2, A_v=01$	-0.131	0.173
DHMC ($I_1=4, I_2=2$)	4(+), 6(*), 2(<), $I_1=4, I_2=2, A_v=00$	3(+), 5(*), 2(<), $I_1=4, I_2=2, A_v=01$	-0.136	-0.103
DIFFERENTIAL EQUATION (U = 4)	2(+), 2(-), 5(*), 2(<), U = 4, $A_v=00$	2(+), 2(-), 5(*), 2(<), U = 4, $A_v=01$	-0.232	-0.193
FFT (U = 4)	4(+), 2(-), 4(*), 1(<), U = 2, $A_v=00$	5(+), 2(-), 5(*), 2(<), U = 4, $A_v=01$	-0.185	-0.088
FIR (U = 6)	4(+), 8(*), 2(<), U = 6, $A_v=00$	3(+), 5(*), 2(<), U = 6, $A_v=01$	-0.175	-0.074
TEST CASE (U = 6)	4(+), 4(*), 2(<), U = 3, $A_v=00$	3(+), 3(*), 2(<), U = 6, $A_v=01$	-0.191	-0.166

B. Analysis of Results

1) For Single LoopBased CDFGs

For single loop based applications, the proposed approach is able to attain final solutions which lie within the user provided constraints of area and execution delay (as well as minimizes the hybrid cost in eqn. (7)), as shown in Table 2. For example, for Differential Equation benchmark, the proposed approach generates the final solution, (X_i): $N(R_{adder})$, $N(R_{multiplier})$, $N(R_{comparator})$, U , $A_v = 3(+)$, $3(-)$, $5(*)$, $2(<)$, $U = 4$, $A_v = 00$, with $T_E^{DMR} = 120.81\mu s$ and $A_T^{DMR} = 29136$ au, which completely satisfies the area-delay constraints ($A_{cons} = 43301$ au and $T_{cons} = 182.21\mu s$). (NOTE: A_{cons} and T_{cons} could be any value between the minimum (A_{min}^{DMR} , T_{min}^{DMR}) and maximum value (A_{max}^{DMR} , T_{max}^{DMR})). Similar behavior is observed for the other benchmarks.

2) For Nested Loop Based CDFGs

In case of nested loop based CDFGs, the proposed approach is able to attain final solutions which lie within the user provided constraints of area and execution delay (as well as minimizes the hybrid cost in eqn. (7)), as shown in Table 2. For example, for Auto correlation Filter benchmark, the proposed approach generates the final solution, (X_i): $N(R_{adder})$, $N(R_{multiplier})$, $N(R_{comparator})$, $I_1, I_2, A_v = 2(+)$, $4(*)$, $2(<)$, $I_1=2, I_2=3, A_v=00$, with $T_E^{DMR} = 34.620\mu s$ and $A_T^{DMR} = 17996.00$ au, which completely satisfies the area-delay constraints ($A_{cons} = 19000.00$ au and $T_{cons} = 50.04\mu s$).

C. Comparison with Related Prior Research

It is important to note that there is no work in the literature that explores the low cost Trojan secured schedule either single-dimensionally (comprising of only scheduling resources) or multi-dimensionally (comprising of scheduling resources ($\overline{R_n}$), ' U ' (for single loop) or I_1, I_2 (for nested loops) with respect to user specified constraints.

TABLE 5 COMPARISON OF EXPLORATION TIME FOR FINDING A LOW COST TROJAN SECURED DATAPATH SOLUTION WITH RESPECT TO SWARM SIZE P FOR PROPOSED APPROACH

Benchmark [29] [30]	Swarm Size, p	Exploration Time (in ms)
AUTOCORRELATION FILTER ($I_1=2, I_2=3$)	3	2144
	5	4305
	7	4893
AUTOCORRELATION FILTER ($I_1=3, I_2=2$)	3	3297
	5	4091
	7	8080
AUTOCORRELATION FILTER ($I_1=4, I_2=2$)	3	4928
	5	5536
	7	7860
DHMC	3	21836

	7	62345
	3	29823
DHMC ($I_1 = 3, I_2 = 3$)	5	49876
	7	68243
	3	29066
DHMC ($I_1 = 4, I_2 = 2$)	5	45327
	7	65827
	3	2084
DIFFERENTIAL EQUATION ($U = 4$)	5	6359
	7	9243
	3	6856
FFT ($U = 4$)	5	9132
	7	9205
	3	638
FIR ($U = 6$)	5	1132
	7	1775
	3	1156
TEST CASE ($U = 6$)	5	3382
	7	4537

Note: ms = milliseconds

The costs function for both [10] and proposed approach, both considers the area of single comparator/error detection block responsible for runtime Trojan detection at the final output. However, the cost function does not have any impact of comparator/error detection block as it is used in both approaches. Table 3 shows the results of the approach [10], while Table 4 shows the comparative results of the proposed approach with [10], in terms of cost and final Trojan secured hardware solution found. Although [10] does not handle single and nested loop CDFGs during Trojan security, however for comparison purpose we have fully re-implemented [10] by feeding the completely unrolled version of each CDFG (both single and nested loop). Complete unrolling of the CDFG (by eliminating the loop converts it into a straightforward DFG) was required because [10] cannot handle loop unrolling factor concurrently with hardware resource configuration during exploration. As seen the proposed approach generates substantially better results in comparison to [10]. This is due to following reasons:

- (a) The proposed approach comprises of a supplementary option in encoding for exploration of an optimal allocation procedure of vendors A_v , which bears an impact on the final optimization quality;
- (b) In case of single loop based applications, the proposed approach comprises of another supplementary option in encoding for exploration of an optimal unrolling factor;
- (c) The proposed approach provides exploration of schedule resources based on user constraints which also bears an effect on the final design quality.

It can be seen from the results, that even though for some benchmarks the number of resources in the final solution for [10] are lower than the proposed solution, it has no optimization of loop unrolling (for single loop CDFGs) and vendor allocation procedure. Since there is no feature for exploring an optimal unrolling factor and distinct vendor allocation in [10] hence it uses the maximum unroll factor value as well as same vendor allocation to each unit rule ($A_v = 01$) during DMR design thereby resulting in higher cost. Therefore, distinct vendor allocation type $A_v = 00$ always yields lower cost final solution than $A_v = 01$.

Further, Table 5 highlights the impact of population size p on the runtime of proposed approach. It can be seen that for all benchmarks, the runtime of the proposed approach to find the solutions escalates with the growth in population size. This is due to the increase in computational complexity per iteration. However, there is no improvement in the quality of the solution found.

D. Detection ability of solutions generated

The proposed approach deals with Trojan in the IP that change its computational value. However, during real time deployment, the Trojan becomes effective when triggered (by an adversary) through a rare event value, thereby remaining inactive before that. Since, triggering through rare event value is performed through intelligent disguise hence it is very difficult to detect Trojan during RTL simulation/other lower level tests by a trustworthy in-house design integrator. Detailed explanation on how a sample Trojan gets triggered only through a rare event value is explained in section III.

Type of Trojan inserted for evaluation: The work presented in this paper targets only a specific class of Trojan. Trojans that on triggering only change the output computational value (and have no other effect) are inserted within the components of HLS library for testing purpose i.e. the IPs were modelled by inserting malevolent logic in the module (such as adder/subtractor etc) of the HLS library. Further for testing purpose during experiments, the inserted Trojan was triggered for evaluating the detection capability of the proposed approach. An example of such a Trojan and its payload is shown in Fig. 1. Additionally, a sequence of test vectors generated through 8th order polynomial Linear-Feedback Shift Register (LFSR) as ATPG are fed into the DMR schedule for comprehensive coverage of Trojan detection in the final explored solution. (*Note: there are no specific test sets that make detection harder or easier*). The results generated indicate that a detection rate of 100% was achieved while handling such Trojans. This is because these Trojans on activation only change the computational value in the output from original and duplicate unit as payload without producing any other effect.

VII. CONCLUSION

This paper presented a novel low cost methodology for optimized Trojan secured scheduling at behavioural level for control data flow graphs (single and nested loop). It is capable of providing secured information processing during high level synthesis. Results indicated improvement in quality of final solution obtained compared to a similar related work.

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Highlights

- Low cost optimized Trojan secured HLS approach
- Handles Trojan security of single or nested loop control data flow graphs (CDFG) applications.
- Multiple novel vendor allocation schemes as security constraint

Figure 1

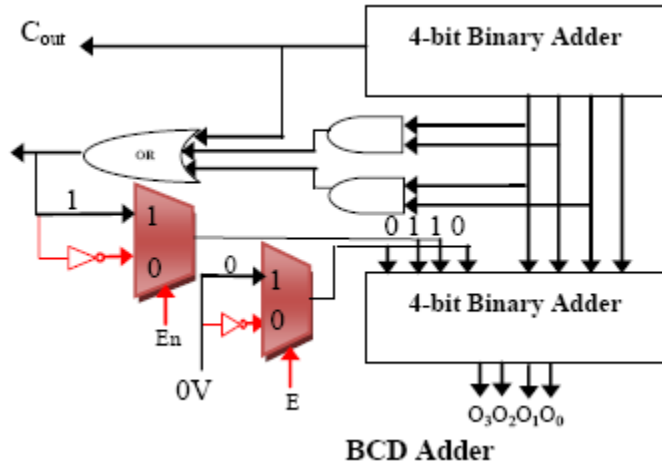


Figure 2

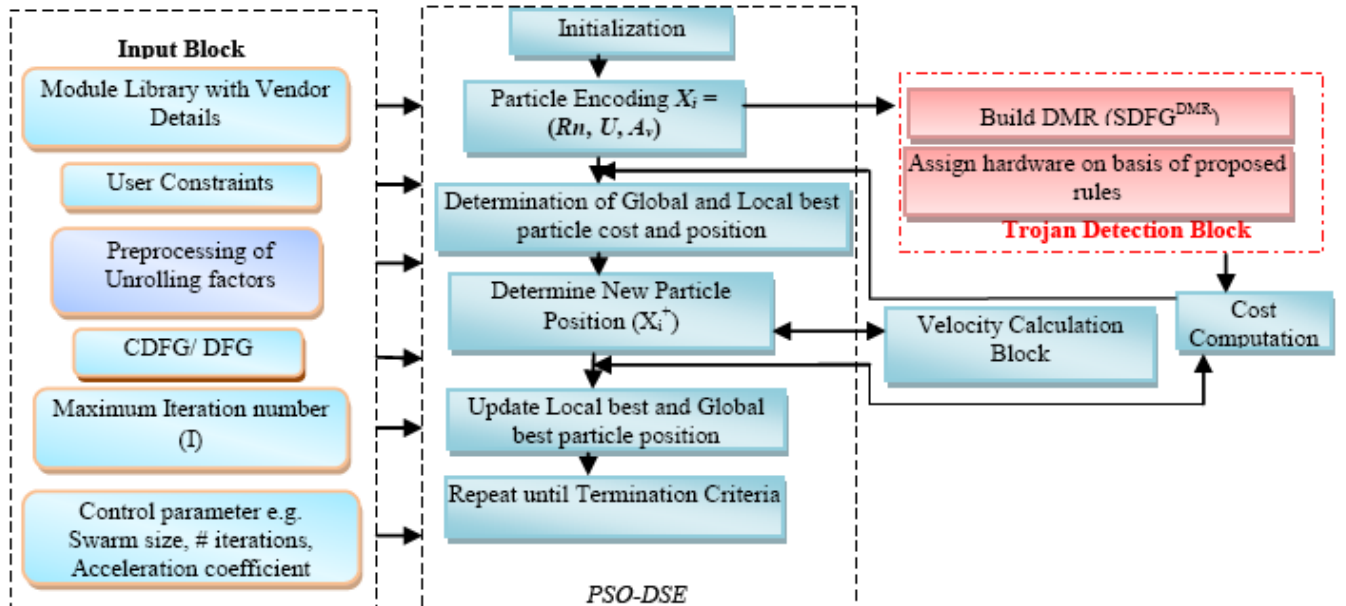


Figure 3

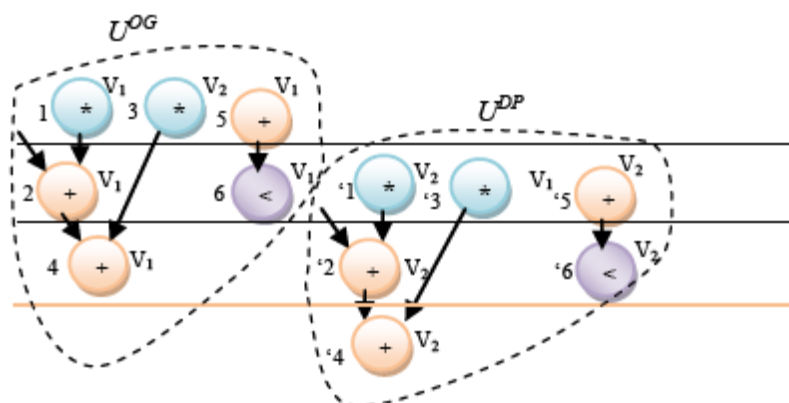


Figure 4

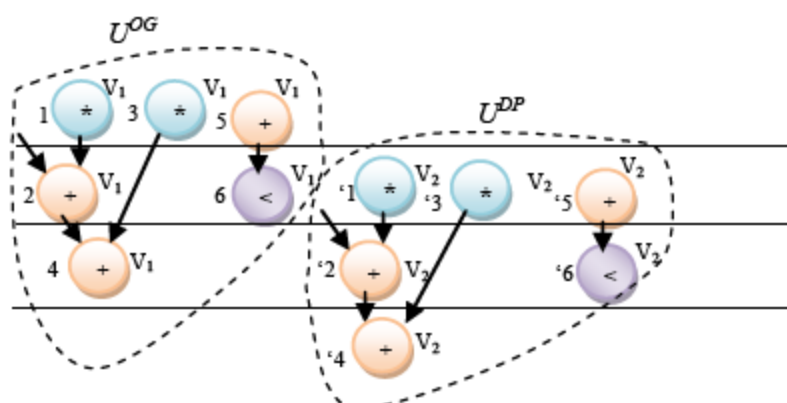


Figure 5

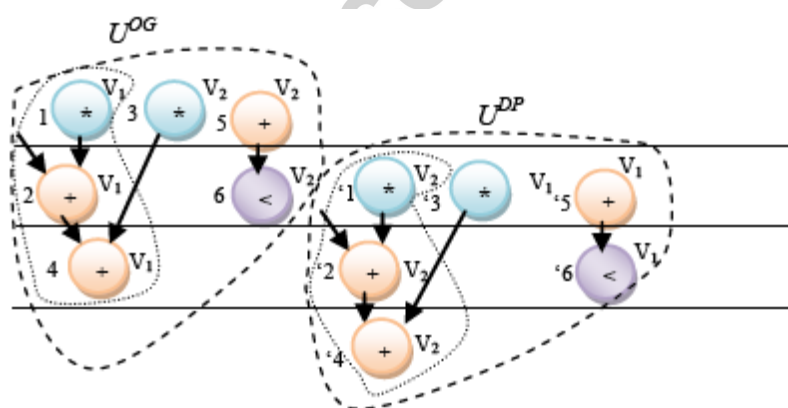


Figure 6

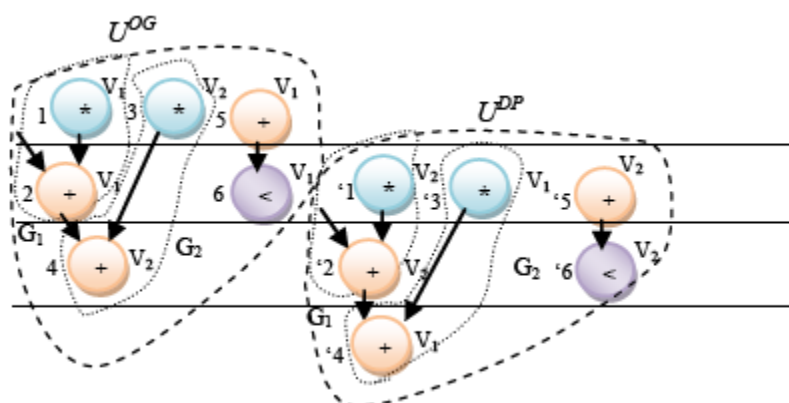


Figure 7(a)

```

For (i=0; i<I1; i++) // I1 = Outer loop
{
    For (j=0; j<I2; j++) I2 = Inner loop
    {
        Ai += Bj * Bj-i;
    }
}

```

Figure 7(b)

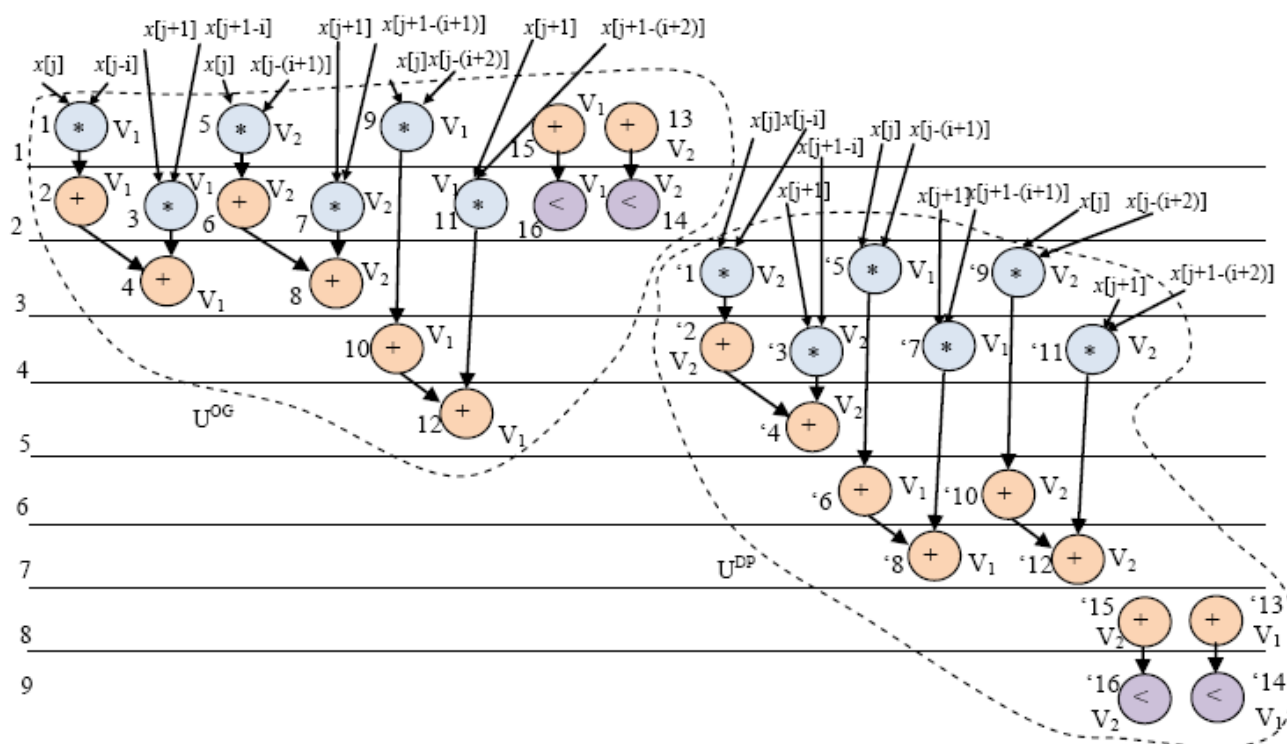


Figure 8

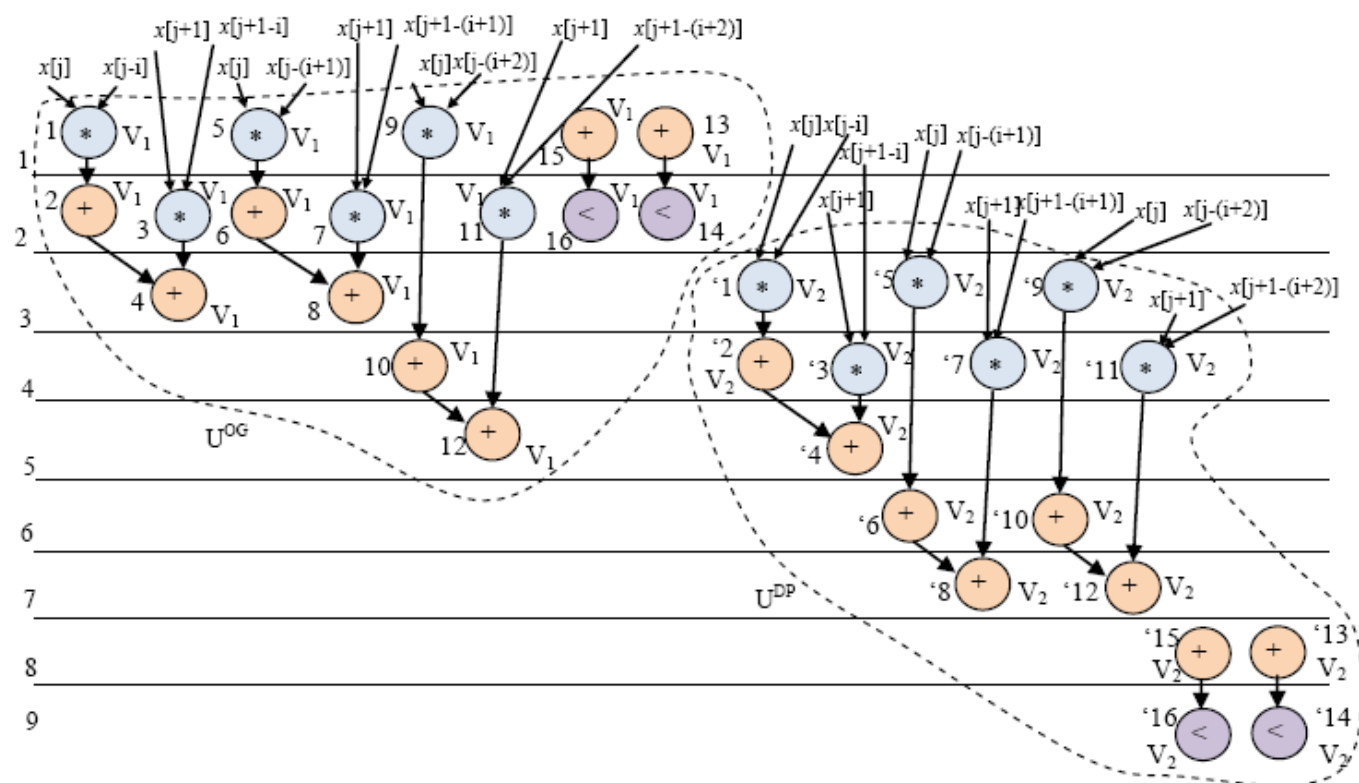


Figure 9

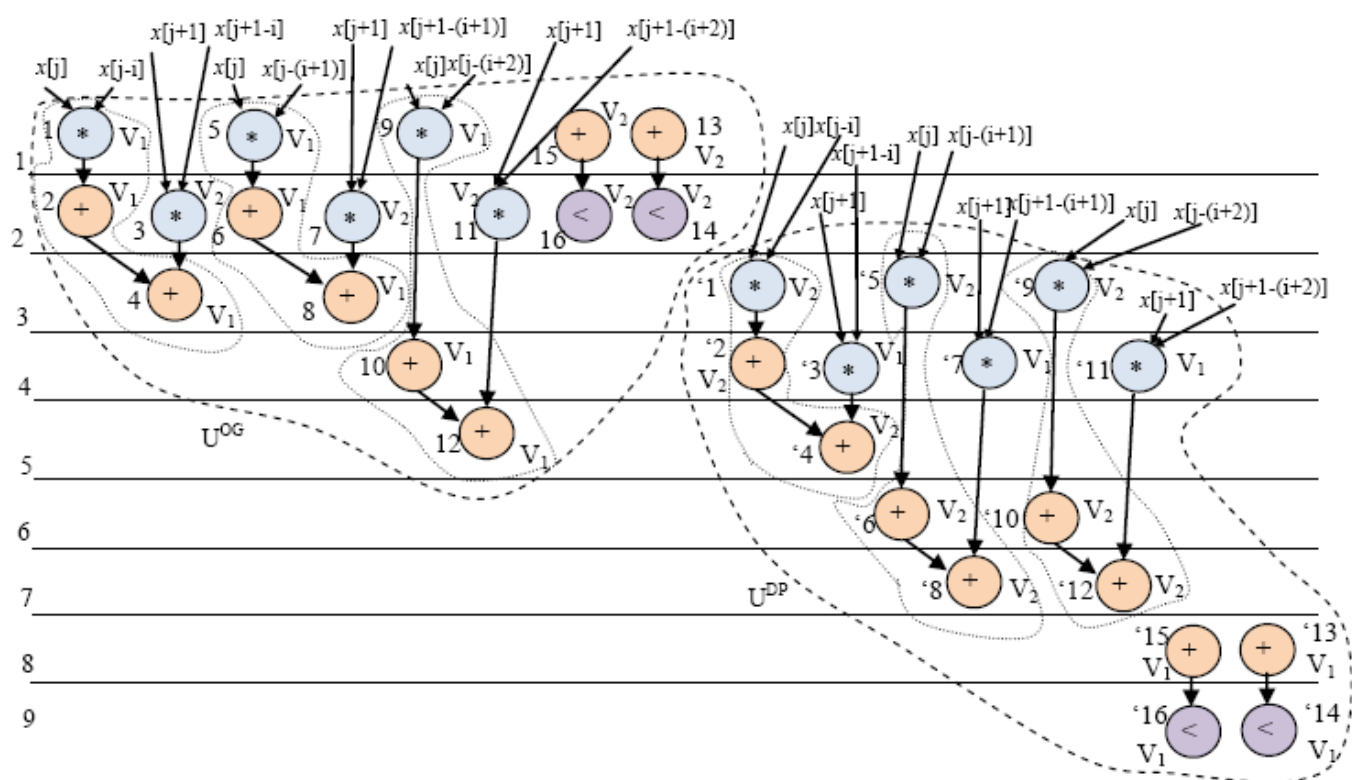


Figure 10

