

# Exact surface profile measurement without subtracting dispersion phase through Fourier transform in a white-light scanning interferometer

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A new signal processing is proposed in which the dispersion phase is not subtracted from the detected spectral phase distribution. The linear and bias components in the spectral phase distribution are used to calculate the complex-valued interference signal (CVIS). The simulations verify that the dispersion phase generates an inclination in the measured surface profile along one direction in which the magnitude of the dispersion phase changes linearly. The simulations also show that the position of zero phase nearest the position of amplitude maximum in the CVIS almost does not change due to the bias component, although the random phase noise contained in the interference signal changes the slope of the linear component. Measured surface profiles show that the new signal processing achieves highly accurate measurement by the CVIS. © 2018 Optical Society of America

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## 1. INTRODUCTION

White-light scanning interferometers (WLSIs) have been used to measure three-dimensional shapes of reflecting surfaces since early times. If measurement error in WLSIs is less than several nanometers, a WLSI is a very powerful tool that has a wide measurement region with a measurement accuracy of nanometer order. In order to achieve high measurement accuracy, different techniques to determine a position of a reflecting surface from the interference signal have been proposed. Detected interference signals have two important positions along the scanning position. One is the maximum value of the envelope, which is called envelope peak [1,2], coherence peak [3], or top of envelope [4]. The other is the maximum value of the interference signal, which is called fringe peak [1], phase peak [2], or signal maximum [4]. It has been reported that the fringe peak provides better accuracy in position measurement compared with the envelope peak [1,2,4]. Although the fringe peak is strongly related to the position where the phase of the interference signal is zero, the position of the fringe peak is not exactly the same as the position of zero phase because the interference signal is a product of the coherence function and the cosine function of the phase distribution. Therefore, it is important to obtain exactly the position of zero phase nearest to the envelope peak. The phase-shifting method [3] and the

lock-in evaluation method [5] were used to obtain the position of zero phase nearest to the envelope peak. These two methods focus on the waveform of the real-valued interference signal itself. On the other hand, Fourier transform of the real-valued interference signal or spectral distribution in wavenumber domain contains useful information about the interference signal [2,6–8]. When there is no dispersion effect and no noise effect in the interference signal, a magnitude of nonlinear component in the spectral phase distribution is very small. The position of the reflecting surface has been obtained from the slope of the linear component in the spectral phase distribution [6,7]. In order to utilize information about amplitude and phase, the complex-valued interference signal (CVIS) is calculated by Fourier transform of the positive components of the spectral distribution [8]. The position of zero phase nearest to the position of amplitude maximum can be easily obtained, and its position provides an accurate measurement of surface profile. In Ref. [8] the dispersion phase was exactly eliminated in the spectral phase distribution to get highly accurate measurement.

In this paper, a new signal processing is proposed in which the dispersion phase is not subtracted from the detected spectral phase distribution. A least squares line is calculated in the spectral phase distribution, and the linear and bias components in the spectral phase distribution are used to calculate the CVIS.

How the dispersion phase provides effects on the linear and bias components when a random phase noise exists in the interference signal is examined by simulations. The simulations verify that the dispersion phase generates an inclination in the measured surface profile along one direction in which the magnitude of the dispersion phase changes linearly. The simulations also show the random phase noise changes the slope of the linear component, but the position of zero phase nearest to the position of amplitude maximum in the CVIS almost does not change due to the bias component. These characteristics lead to highly accurate measurement by the new signal processing without using the dispersion phase measured with a spectrally resolved interferometer.

## 2. PRINCIPLE

Figure 1 shows a WLSI with a supercontinuum light source whose spectral intensity is  $I(\sigma)$ , where  $\sigma$  is the wavenumber. The position of an object surface is  $z_o$ , and the position  $z$  of a reference surface is scanned by a piezoelectric transducer (PZT). An interference signal is detected with a camera when the PZT is moving. The interference signal expressed as a function of the scanning position  $z$  is given by

$$S(z) = \int_{-\infty}^{\infty} I(\sigma) \cos[4\pi(z - z_o)\sigma + \varphi_n(z)\sigma + \varphi_d(\sigma)] d\sigma, \quad (1)$$

where phase  $\varphi_d(\sigma)$  is a dispersion phase caused by two sides of unequal length in a cubic beam splitter. The two beams reflected from the object surface and the reference surface have path distances of  $l_1$  and  $l_2$  in the beam splitter, respectively, as shown in Fig. 2. Denoting the distance difference by  $l_e = l_1 - l_2$ , the dispersion phase is given by

$$\varphi_d(\sigma) = 4\pi[n(\sigma) - 1]\sigma l_e, \quad (2)$$

where  $n(\sigma)$  is the refractive index of the beam splitter, and  $l_e$  is a linear function of the position  $x$  where the two beams go out from the beam splitter. At  $x = x_B$ , the value of  $l_e$  is almost equal to zero, and it increases linearly as the position  $x$  approaches  $x_A$ , as shown in Fig. 2. Phase  $\varphi_n(z)$  is generated by an optical path change that is caused by random disturbances such as mechanical vibrations and non-constant speed of the PZT movement. Fourier transform of  $S(z)$  or the

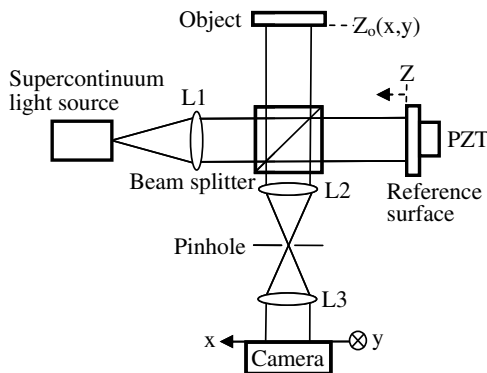


Fig. 1. Schematic of a white-light scanning interferometer.

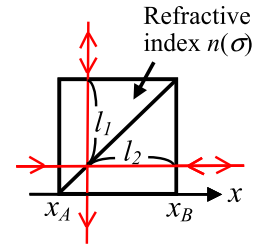


Fig. 2. Generation of dispersion phase by the distance difference of  $l_e = l_1 - l_2$  in the beam splitter.

spectral distribution in the region of positive wavenumbers is expressed as

$$F(\sigma) = I_F(\sigma) e^{-j4\pi z_o \sigma} e^{j\beta(\sigma)}, \quad (\sigma > 0), \quad (3)$$

where  $\beta(\sigma)$  arises from the phases  $\varphi_n(z)\sigma$  and  $\varphi_d(\sigma)$  through Fourier transform. Denoting the least squares line in the phase distribution of Eq. (3) by  $a_0 + (a_1 - 4\pi z_o)\sigma$ , a new spectral distribution having only bias and linear phases is obtained as

$$F_L(\sigma) = I_F(\sigma) e^{j(a_1 - 4\pi z_o)\sigma} e^{ja_0}, \quad (\sigma > 0). \quad (4)$$

Since the phase distribution of Eq. (3) is unwrapped to calculate the least squares line, the value of  $a_0$  is more than  $2\pi$ . Expressing the inverse Fourier transform of  $I_F(\sigma)$  by  $A(2z)e^{j\alpha(2z)}$ , the CVIS derived as inverse Fourier transform of  $F_L(\sigma)$  is given by

$$S_c(z) = A(z - z_s) \exp[j\alpha(z - z_s) + ja_0], \quad (5)$$

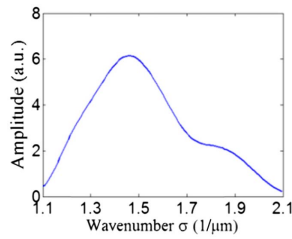
where  $z_s = z_o - (a_1/4\pi)$ ,  $A(z) = A(-z)$ ,  $\alpha(z) = -\alpha(-z)$ , and  $\alpha(0) = 0$ . The period of the phase distribution in the CVIS is  $\lambda_A/2$  around  $z = z_s$ , where  $\lambda_A$  is the weighted average wavelength in the distribution of  $I_F(\sigma)$ . The value of  $a_0$  in Eq. (5) exists in the range from  $-\pi$  to  $\pi$  after the value of  $a_0$  in Eq. (4) is reduced to  $a_0 - 2\pi n$ , where  $n$  is an integer. Equation (5) leads to the following results: a position  $z_a$  of maximum amplitude of  $A(z - z_s)$  is equal to the measurement value  $z_s$  obtained from the slope of the linear component in the spectral phase. The position  $z_p$  of zero phase nearest to the position  $z_a$  is given by

$$z_p = z_s + (a_0 \lambda_A / 4\pi) = z_o - (a_1 / 4\pi) + (a_0 \lambda_A / 4\pi). \quad (6)$$

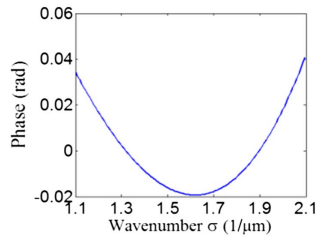
The following characteristics are estimated: the existence of the dispersion phase produces an inclination of a surface profile in the direction of  $x$  axis, because it is expected that the distance difference  $l_e$  in the dispersion phase  $\varphi_d(\sigma)$  changes linearly along the  $x$  axis, and the values of  $a_1$  and  $a_0$  increase linearly with increase in the value of  $l_e$ . The existence of the noise phase  $\varphi_n(z)$  does not generate a large measurement error in the measurement value  $z_p$ , because the value of  $a_0$  caused by  $\varphi_n(z)$  will compensate for a change in  $a_1$  caused by  $\varphi_n(z)$ . These characteristics are verified by simulations and experiments in the following sections.

## 3. SIMULATION

First, it will be explained how the simulation is carried out following the principle. In order to generate the interference signal given by Eq. (1), the spectrum  $I(\sigma)$ , the phase distributions  $\varphi_d(\sigma)$ , and the random phase noise  $\varphi_n(z)$  are necessary. Figure 3 shows the spectrum of the supercontinuum light

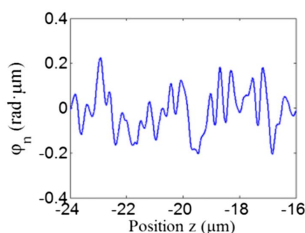


**Fig. 3.** Spectrum  $I(\sigma)$  of the light source.

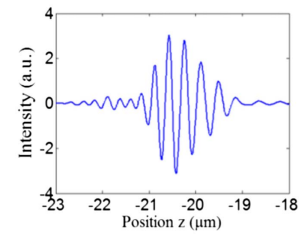


**Fig. 4.** Nonlinear component contained in  $4\pi[n(\sigma) - 1]\sigma$ .

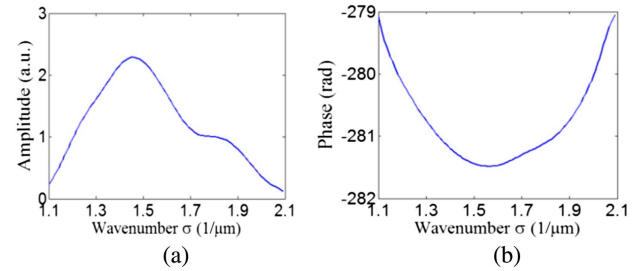
source, which was detected by a spectral analyzer [8]. The wavenumber region of the detected spectrum was from 1.1 to 2.1  $\mu\text{m}^{-1}$ . The difference between the lengths of the two sides of the beam splitter was about 100  $\mu\text{m}$ . The nonlinear component contained in the dispersion phase  $\varphi_d(\sigma)$  at  $l_e = 1 \mu\text{m}$  was calculated with least squares method, as shown in Fig. 4. The dispersion phase  $\varphi_d(\sigma)$  of  $l_e = 40 \mu\text{m}$  around the central position of the beam splitter was used. Figure 5 is the random Gaussian noise  $\varphi_n(z)$  with an average of zero and variance of 0.09  $\text{rad} \cdot \mu\text{m}$ . The interference signal was generated in the region from  $-30 \mu\text{m}$  to  $30 \mu\text{m}$  where the object position  $z_o$  was 1  $\mu\text{m}$ , and the sampling interval  $\Delta z$  and the sampling number  $N$  were 0.0073  $\mu\text{m}$  and 8192, respectively. A part of interference signal whose envelope value was more than one-tenth of its maximum intensity was selected by a rectangular window whose width was from  $-23 \mu\text{m}$  to  $-18 \mu\text{m}$ , as shown in Fig. 6. The other data outside the rectangular window were zero values. Fourier transform was performed on this interference signal of data number  $N$ . The interval of wavenumber was  $\Delta\sigma = 1/(2N\Delta z) = 0.0083 \mu\text{m}^{-1}$ . Figure 7(a) show the amplitude  $I_F(\sigma)$  of Fourier transform  $F(\sigma)$ . The values of  $a_0$  and  $a_1 - 4\pi z_o$  were  $-280.721 \text{ rad}$  and  $257.486 \text{ rad} \cdot \mu\text{m}$ , respectively. The phase distribution of  $F(\sigma)$  contained a large value of  $a_1$  of  $270.053 \text{ rad} \cdot \mu\text{m}$ , which was almost caused by



**Fig. 5.** Gaussian noise with average of zero and variance of 0.09  $\text{rad} \cdot \mu\text{m}$ .



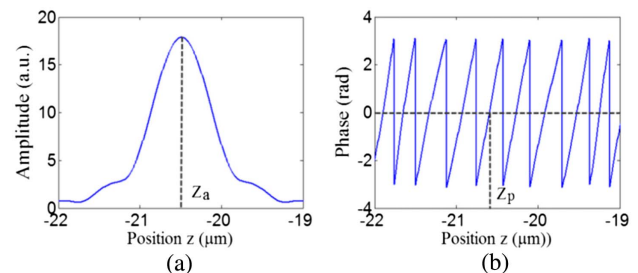
**Fig. 6.** Interference signal generated with the spectrum, the dispersion phase, and the Gaussian noise.



**Fig. 7.** Fourier transform of the interference signal of Fig. 6; (a) amplitude  $I_F(\sigma)$  and (b) phase of  $\beta(\sigma) - a_1\sigma$ .

the dispersion phase at  $l_e = 40 \mu\text{m}$ . In order to show clearly the non-linear component contained in the phase of  $F(\sigma)$ , the phase distribution of  $\beta(\sigma) - a_1\sigma$  is shown in Fig. 7(b), where the  $a_0$  was equal to  $-280.721 \text{ rad} \cdot \mu\text{m}$ . The measurement value  $z_s$  of the object position was obtained from the linear component of the phase distribution as  $z_s = z_o - a_1/4\pi = -20.490 \mu\text{m}$ . By performing inverse Fourier transform on the data of  $F_L(\sigma)$ , the CVIS was obtained as shown in Figs. 8(a) and 8(b). The value of  $z_a$  was  $-20.490 \mu\text{m}$ , which was equal to  $z_s$ . The measurement value  $z_p$  of the object position obtained from this CVIS was  $-20.596 \mu\text{m}$ .

Second, in order to examine the effect of the dispersion phase, the different values of  $l_e$  were used in the simulation when the noise  $\varphi_n$  did not exist. Results are shown in Table 1. The values of  $a_0$  and  $a_1$  change linearly with the increase of  $l_e$ . The interval in  $a_0$  and  $a_1$  between the adjacent values of  $l_e$  were constant by ignoring small fluctuations less than 5 nm and 3  $\text{rad} \cdot \mu\text{m}$ , respectively. The average values of the interval in  $a_0$  and  $a_1$  were about 73.470  $\text{rad}$  and 67.604  $\text{rad} \cdot \mu\text{m}$ , respectively. The values  $z_s - z_o$  and  $z_p - z_o$



**Fig. 8.** Complex-valued interference signal of Fig. 6 obtained by using the least squares line.

**Table 1. Simulation Results at the Different Values of  $l_e$  Without Noise**

$l_e$ ( $\mu\text{m}$ )	0	10	20	30	40
$a_0$ (rad)	12.566	-60.903	-134.371	-207.841	-281.312
$a_1$ (rad $\cdot \mu\text{m}$ )	0.0	67.604	135.206	202.811	270.415
$z_s - z_o$ ( $\mu\text{m}$ )	0.0	-5.380	-10.759	-16.139	-21.519
$z_p - z_o$ ( $\mu\text{m}$ )	0.0	-5.481	-10.633	-16.113	-21.594

were obtained from the CVIS. These two values almost satisfied Eq. (6). The decreasing interval in the values of  $z_s - z_o$  between adjacent values of  $l_e$  was a constant value of 5.380  $\mu\text{m}$  by ignoring fluctuations less than 1 nm. On the other hand, the interval of 5.152  $\mu\text{m}$  in the values of  $z_p - z_o$  between  $l_e = 10 \mu\text{m}$  and  $l_e = 20 \mu\text{m}$  is not equal to the constant interval of 5.481  $\mu\text{m}$  between the other adjacent values of  $l_e$ . After adding a value of  $\lambda_A/2 = 0.329 \mu\text{m}$  to the values of  $z_p - z_o$  at  $l_e = 20 \mu\text{m}$ , 30  $\mu\text{m}$ , and 40  $\mu\text{m}$ , all the intervals in  $z_p - z_o$  between the adjacent values became the constant value of 5.481  $\mu\text{m}$ . It is made clear from the results in Table 1 that the dispersion phase brings only an inclination along the  $x$  axis to a measured surface profile. Therefore, the dispersion phase has no influence on the measurement accuracy.

Third, the simulation was carried out when the dispersion and noise existed together. The simulation results are shown at Table 2. Although all the values are different from the results in Table 1, the values of  $z_s - z_o$  and  $z_p - z_o$  also satisfy Eq. (6). Due to the effect of noise, the interval in the values of  $z_s - z_o$  between the adjacent values of  $l_e$  was not constant, and the deviations from the average value of 5.369  $\mu\text{m}$  were less than 60 nm. The interval of 5.151  $\mu\text{m}$  in the values of  $z_p - z_o$  between  $l_e = 10 \mu\text{m}$  and  $l_e = 20 \mu\text{m}$  was not equal to the constant interval of 5.482  $\mu\text{m}$  between the other adjacent values of  $l_e$ . After adding a value of  $\lambda_A/2 = 0.329 \mu\text{m}$  to the values of  $z_p - z_o$  at  $l_e = 20 \mu\text{m}$ , 30  $\mu\text{m}$ , and 40  $\mu\text{m}$ , all the intervals in  $z_p - z_o$  became almost constant, and the deviation from the average value of 5.482  $\mu\text{m}$  was less than 4 nm. Therefore, the noise has a large influence on the linear change in the values of  $z_s - z_o$ , compared with the values of  $z_p - z_o$ .

In order to show degree of noise resistance, the values of  $z_s - z_o$  and  $z_p - z_o$  were compared between Tables 1 and 2. The differences of the values are denoted by  $\Delta(z_s - z_o)$  and  $\Delta(z_p - z_o)$  in Table 3. The  $\Delta(z_p - z_o)$  and  $\Delta(z_s - z_o)$  are less than 3 nm and 39 nm, respectively. These results make clear that the value  $z_p$  has a higher noise resistance and provides higher accuracy in surface profile measurement.

Finally, the simulation was carried out for different noises at  $l_e = 40 \mu\text{m}$ , as shown in Table 4. Noise I was used in Table 2. The averages and variances of 3 noises were the same. These

**Table 2. Simulation Results at the Different Values of  $l_e$  with Noise**

$l_e$ ( $\mu\text{m}$ )	0	10	20	30	40
$a_0$ (rad)	12.242	-60.381	-134.747	-208.636	-280.721
$a_1$ (rad $\cdot \mu\text{m}$ )	0.185	67.237	135.419	203.301	270.052
$z_s - z_o$ ( $\mu\text{m}$ )	-0.015	-5.351	-10.776	-16.178	-21.490
$z_p - z_o$ ( $\mu\text{m}$ )	0.002	-5.479	-10.630	-16.110	-21.596

**Table 3. Differences in  $z_s - z_o$  and  $z_p - z_o$  between Tables 1 and 2**

$l_e$ ( $\mu\text{m}$ )	0	10	20	30	40
$\Delta(z_s - z_o)$ (nm)	-15	29	-17	-39	29
$\Delta(z_p - z_o)$ (nm)	2	-1	3	3	-2

**Table 4. Simulation Results for Different Noises at  $l_e = 40 \mu\text{m}$** 

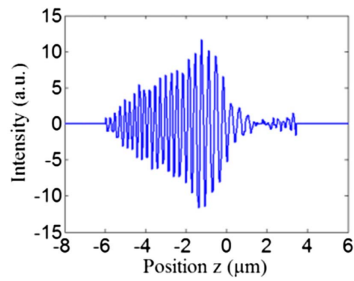
$\varphi_n(z)$	$a_0$ (rad)	$a_1$ (rad $\cdot \mu\text{m}$ )	$z_s - z_o$ ( $\mu\text{m}$ )	$z_p - z_o$ ( $\mu\text{m}$ )
Noise I	-280.721	270.052	-21.490	-21.596
Noise II	-287.446	270.324	-21.512	-21.594
Noise III	-282.214	271.023	-21.567	-21.595

noises produced different values of  $a_0$  and  $a_1$ . The difference in the values of  $z_s - z_o$  among the 3 noises was less than 55 nm. On the other hand, the difference in the values of  $z_p - z_o$  among the 3 noises is less than 2 nm. The conclusions obtained from Table 2 do not change for different noises, and the value  $z_p$  provides high repeatability in surface profile measurement.

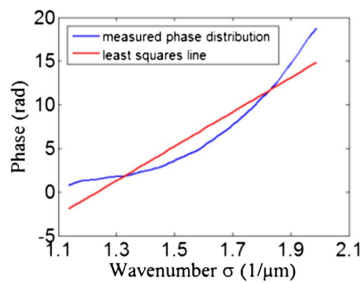
#### 4. EXPERIMENT

A WLSI was constructed as shown in Fig. 1. The light source of interferometer was a supercontinuum light source whose spectral range was 400–1200 nm. The object was a glass with a step of 3  $\mu\text{m}$  width on the surface. The reference surface was a glass with a wedge angle, and it was regarded as one reflecting surface. The two optical fields on the object and reference surfaces were made with the two lenses of L2 and L3 with unity magnification on the camera. The inclination of the reference surface was adjusted so that the phase distribution of the interference pattern on the camera did not indicate a large inclination of the object surface, which was caused by the distribution of the distance difference  $l_e$  as shown in Tables 1 and 2. A high-speed camera was used as the detector. Its pixel size was 20  $\times$  20  $\mu\text{m}$ , and the frame size was 640  $\times$  480 pixels. The measurement points of 56  $\times$  41 pixels with the interval of 200  $\mu\text{m}$  were made by adding up the interference signals detected on one pixel in the region of 10  $\times$  10 pixels. The measuring points were denoted by  $N_x$  and  $N_y$ , where  $N_x$  was from 1 to 56, and  $N_y$  was from 1 to 41. The reference surface was moved by the PZT at a constant velocity of about 80  $\mu\text{m/s}$ , and the sampling interval  $\Delta z$  of the interference signal was 39.6 nm. The data number  $N$  of the detected signal was 1024, and the required interference signal was selected with a rectangular window whose data number was 240. The other data outside the window were zero values. Figure 9 shows the windowed interference signal detected at a measurement point of  $N_x = 18$  and  $N_y = 18$ , and the distance difference  $l_e$  at this measurement point was about 45  $\mu\text{m}$ . The interference signal indicates that the object position was about -1.23  $\mu\text{m}$ . Although  $l_e$  was equal to 88  $\mu\text{m}$  at a measurement point  $N_x = 1$  and  $N_y = 18$ , the object position indicated in the interference signal was -1.35  $\mu\text{m}$ . This fact means that the inclination of the surface profile caused by  $l_e$  did not appear in the detected interference signal. Fourier transform was performed on the windowed





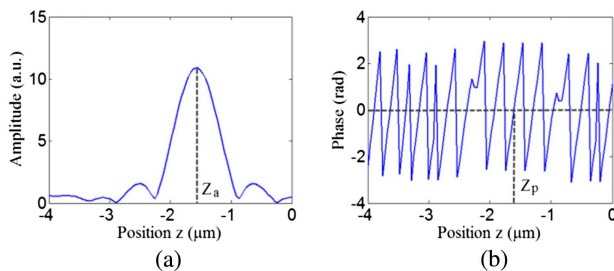
**Fig. 9.** Interference signal detected at measurement point of  $N_x = 18$  and  $N_y = 18$ .



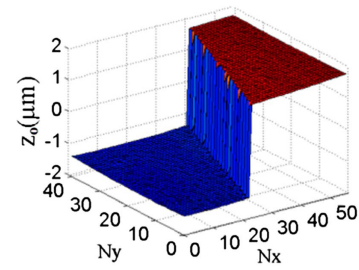
**Fig. 10.** Measured phase distribution of the interference in wave-number domain and the least squares line of the distribution.

interference signal, where the interval of the wavenumber was  $\Delta\sigma = 1/(2N\Delta z) = 0.0123 \mu\text{m}^{-1}$ . Figure 10 shows the phase of Fourier transform of the interference signal with blue color. The non-linear component of this phase distribution is almost similar to that in Fig. 7(b). The least squares line of the phase was shown by the red line. The values of  $a_0$  and  $a_1 - 4\pi z_0$  were  $-24.357 \text{ rad}$  and  $19.715 \text{ rad} \cdot \mu\text{m}$ , respectively. The measurement value  $z_s$  obtained from  $a_1 - 4\pi z_0$  was  $-1.569 \mu\text{m}$ . Figure 11 shows the CVIS obtained by using the least squares line of the phase. The position  $z_a$  was  $-1.571 \text{ nm}$ , and the measurement value  $z_p$  was  $-1.609 \text{ nm}$ .

The signal processing to obtain the two measurement values of  $z_s$  and  $z_p$  was carried out for all of the measurement points to get surface profiles of the glass with the step shape. A surface profile obtained from the measured values of  $z_s$  is shown in Fig. 12, where the tilt and piston components were eliminated.

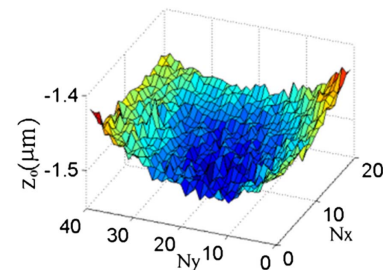


**Fig. 11.** Complex-valued interference signal of Fig. 9 obtained by using the least squares line of the phase shown in Fig. 10.  $z_a = -1.583 \mu\text{m}$  and  $z_p = -1.609 \mu\text{m}$ .

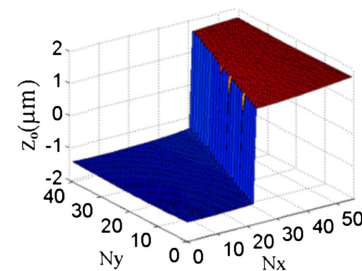


**Fig. 12.** Surface profile obtained from the measured values of  $z_s$ .

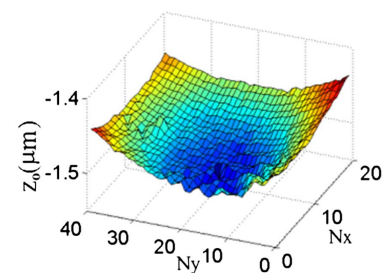
In order to show clearly the small variations in the surface profile, the surface profile on the left part in Fig. 12 is shown in Fig. 13. The magnitude of the small variations was less than  $15 \text{ nm}$ . Figure 14 shows the surface profile obtained from the measured values of  $z_p$ . The surface profile on the left part in Fig. 14 is shown in Fig. 15. The measured surface profile was very smooth, and the magnitude of the small variations was less than  $4 \text{ nm}$ . The repeatability was obtained by calculating a



**Fig. 13.** Surface profile on the left part in Fig. 12.



**Fig. 14.** Surface profile obtained from the measured values of  $z_p$ .



**Fig. 15.** Surface profile on the left side in Fig. 14.

root-mean-square value of the difference between two surface profiles measured in an interval of about 10 min. The repeatability was 1.3 nm in the measurement by  $z_p$ , while it was 13.3 nm in the measurement by  $z_s$ . The step width between  $N_x = 34$  and  $N_x = 37$  along  $N_y = 28$  was  $2.932\ \mu\text{m}$  by the measured values of  $z_p$ . It was made clear that the measurement value  $z_p$  provides a more exact position of a measured reflecting surface than the measurement value  $z_s$ .

## 5. CONCLUSION

The simulation results agreed with the equations described in principle and made clear the characteristics of the proposed signal processing. The existence of the dispersion phase leads to the result that the values of  $a_1$  and  $a_0$  change linearly with increase in the distance difference  $l_e$ . This distribution of the value  $a_1$  produces an inclination of a surface profile in the direction of the  $x$  axis along which the magnitude of the dispersion phase changes linearly. But this inclination of the surface profile can be eliminated by adjusting the inclination of the reference surface in experiments. The changes in the values of  $a_1$  and  $a_0$  caused by the random phase noise generate a large measurement error less than 15 nm in the value  $z_s$ , but the changes do not generate such a large measurement error in the value  $z_p$ . These characteristics of the proposed signal processing without using a measured dispersion phase lead to highly accurate measurement by the value  $z_p$  with an error less than 4 nm. In the future, it is required to investigate how the value of

$a_0$  is made to compensate for the change in  $a_1$ , and also how the phase step of  $2\pi$  appears in the value  $z_p$ .

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