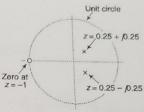
EECE72425 DSP Assignment #6 Combining Systems, Bilinear Transform

<u>Due Monday April 20 at noon. Upload files to eConestoga dropbox</u> <u>The solution will be posted at noon so you have it available for study.</u>

(30 marks, equal weighting for 3 questions on this page)

Given the z-plane pole/zero plot, associated with a 2nd-order IIR digital fiter, in Figure P6–28:

- (a) What is the H(z) transfer function, in terms of z⁻¹ and z⁻², of the Figur P6–28 filter having two poles and a single zero on the z-plane? Show how you arrived at your answer.
- (b) Draw the Direct Form I block diagram of the H(z) filter that implement z = -1 the transfer function arrived at in Part (a) of this problem.



A 1st-order analog highpass filter's s-domain transfer function is

2

$$H(s) = \frac{s}{s + \omega_0}.$$

Determine a digital filter's H(z) z-domain transfer function that simulates H(s) using the bilinear transform process. Given that frequency $\omega_0 = 62.832$ radians/second, assume that the digital filter's sample rate is $f_s = 100$ Hz. Manipulate your final H(z) expression so that it is in the following form:

$$H(z) = \frac{A + Bz^{-1}}{1 + Cz^{-1}}$$

Due to its simplicity, the 1st-order analog lowpass filter shown in Figure P6–47(a) is often used to attenuate high-frequency noise in a $v_{\rm in}(t)$ input signal voltage. This lowpass filter's s-domain transfer function is

$$H(s) = \frac{1}{1 + RCs}.$$

Determine a digital filter's $H_{\rm bt}(z)$ z-domain transfer function that simulates H(s), using the bilinear transform process. Draw the digital filter's Direct Form II block diagram where the coefficients are in terms of R and C. Again, assume that $t_s=1$.

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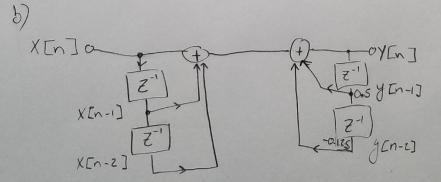
(15 marks, 3 marks for each part a and b, 4 marks part c, 5 marks part d)

Reduce each of the following block diagrams to the **two** forms shown below – **Simplified** Form & **Reduced** form. For each form, express A and B in terms of the G's, H's & K's. Be sure to reduce each fraction to a fraction with a single numerator and denominator.

Reduced form Simplified form A' В a) G_2 G1 H b) G₁ H₂ G1 c) H₁ H₂ K d). G₂ G1

H

Dsp Assingment #6 1) a) Zeroso Z=-1 Poles: Z=0.25+j0.25 Z=0.25-j0.25 (Z+1) (0.25+j0.25-Z)(0.25-j0.25-Z)= (Z+1)



y[n] = X[n-1] + X[n-2] + 0.5y[n-1] -0.125y[n-2]

2)
$$H(s) = \frac{s}{s + \omega_0}$$
 $U_0 = \frac{6z \cdot 63z}{\sqrt{s} + \frac{1}{100}} = 0.01$
 $H(z) = \frac{2}{7} \left(\frac{1-z^{-1}}{1+z^{-1}} \right) + \omega_0 = \frac{2(1-z^{-1})}{2(1-z^{-1})} + \omega_0 = \frac{1-z^{-1}}{2(1-z^{-1})} + \omega_0 = \frac{1-z^{-1}}{2(1-z^{-1})} + \omega_0 = \frac{1-z^{-1}}{2(1-z^{-1})} + \omega_0 = \frac{1-z^{-1}}{2(1-z^{-1})} + \frac{(6z \cdot 83z \times 0.01)}{2} + \frac{(6z \cdot 83z \times 0.01)}{2} = \frac{1-z^{-1}}{2(1-z^{-1})} + \frac{(0.3141)}{2} = \frac{1-z^{-1}}{2(1-z^{-1})} + \frac{(0.3141)}{2} = \frac{1-z^{-1}}{2(1-z^{-1})} + \frac{(0.3141)}{2} = \frac{1-z^{-1}}{2(1-z^{-1})} + \frac{(0.3141)}{2(1-z^{-1})} = \frac{1-z^{-1}}{2(1-z^{-1})} = \frac{1-z^{-1}}{2(1-z^{-1})} + \frac{(0.3141)}{2(1-z^{-1})} = \frac{1-z^{-1}}{2(1-z^{-1})} = \frac{1-z^{-1}}{2(1-z^{-1})} + \frac{(0.3141)}{2(1-z^{-1})} = \frac{(0.3141)}{2(1-z^{-1})} + \frac{(0.3141)}{2(1-z^{-1})} = \frac{(0.3141$

1.314 - 0.522 2

C) Simplified form;
$$A = G_1$$

Reduced form; $A = G_1$
 $A = G_2$
 $A = G_2$
 $A = G_3$
 $A = G_4$
 $A = G_4$

Reduced form
$${}_{\circ}A' = \frac{A}{1+AB} = \frac{G_{1} \cdot \left(\frac{G_{1}}{1+G_{2}K}\right)}{1+H\left(G_{1}\left(\frac{G_{2}N}{1+G_{2}K}\right)\right)} \times \frac{1+G_{2}K}{1+G_{2}K}$$