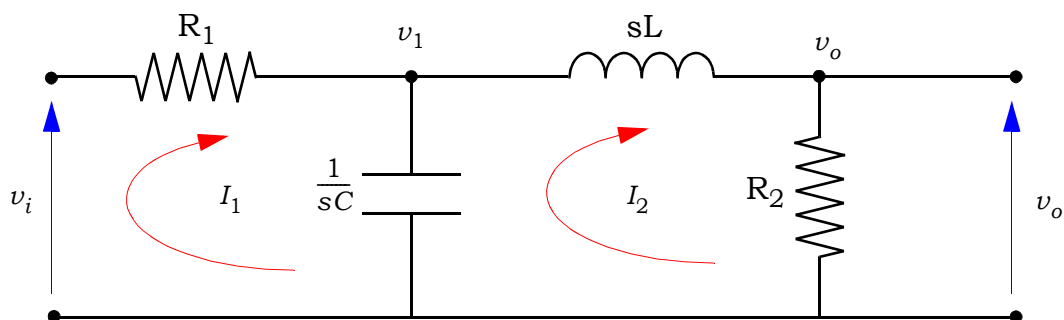


## Tutorial Sheet 3 - Solutions

Q1



Circuit A

(i) **Mesh analysis ...**

$$I_1 R_1 + (I_1 - I_2) \frac{1}{sC} = v_i \text{ and } (I_2 - I_1) \frac{1}{sC} + I_2 sL + I_2 R_2 = 0 \quad (1)$$

Converting these back to the time domain we get:

$$CR_1 I_1 + I_1 - I_2 = C v_i \text{ and } I_2 - I_1 + CL \dot{I}_2 + CR_2 I_2 = 0 \quad (2)$$

Expressing the equations in terms of the highest derivatives we get:

$$\dot{I}_1 - \frac{1}{R_1} v_i = -\frac{1}{CR_1} I_1 + \frac{1}{CR_1} I_2 \text{ and } \dot{I}_2 = -\frac{R_2}{L} I_2 - \frac{1}{CL} I_2 + \frac{1}{CL} I_1 \quad (3)$$

Note that because the derivative of the input  $v_i$  appears in the first equation we bring the highest derivative of  $v_i$  to the LHS as well.

To get a state space model which does not involve the derivative of  $v_i$  we have to define our current state as:

$$x_1 = I_1 - \frac{1}{R_1} v_i, \quad x_2 = I_2 \text{ and } x_3 = \dot{I}_2 \quad (4)$$

Therefore the state equations are:

$$\dot{x}_1 = \dot{I}_1 - \frac{1}{R_1} \dot{v}_i = -\frac{1}{CR_1} I_1 + \frac{1}{CR_1} I_2 = -\frac{1}{CR_1} \left( x_1 + \frac{1}{R_1} v_i \right) + \frac{1}{CR_1} x_2 \quad (5)$$

$$x_2 = \dot{I}_2 = x_3 \quad (6)$$

$$\dot{x}_3 = \dot{I}_2 = -\frac{R_2}{L} \dot{I}_2 - \frac{1}{CL} \dot{I}_2 + \frac{1}{CL} I_1 = -\frac{R_2}{L} x_3 - \frac{1}{CL} x_2 + \frac{1}{CL} \left( x_1 + \frac{1}{R_1} v_i \right) \quad (7)$$

Note that from our definition of the states we have that  $I_1 = x_1 + \frac{1}{R_1} v_i$ .

$$\dot{x}_1 = -\frac{1}{CR_1}x_1 + \frac{1}{CR_1}x_2 - \frac{1}{CR_1^2}v_i \quad (8)$$

$$\dot{x}_2 = x_3 \quad (9)$$

$$\dot{x}_3 = \frac{1}{CL}x_1 - \frac{1}{CL}x_2 - \frac{R_2}{L}x_3 + \frac{1}{CLR_1}v_i \quad (10)$$

or in matrix form:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -\frac{1}{CR_1} & \frac{1}{CR_1} & 0 \\ 0 & 0 & 1 \\ \frac{1}{CL} & -\frac{1}{CL} & -\frac{R_2}{L} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} -\frac{1}{CR_1^2} \\ 0 \\ \frac{1}{CLR_1} \end{bmatrix} v_i \quad (11)$$

Since the output,  $v_o$  is the voltage across  $R_2$  we can write the output equation as:

$$v_o = I_2 R_2 = x_2 R_2 \quad (12)$$

hence the output equation in matrix form is:

$$v_o = \begin{bmatrix} 0 & R_2 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} \quad (13)$$

(ii) Using **nodal analysis** with nodes  $v_1$  and  $v_o$ :

$$\frac{v_1 - v_i}{R_1} + \frac{v_1}{sC} + \frac{v_1 - v_o}{sL} = 0 \text{ and } \frac{v_o - v_1}{sL} + \frac{v_o}{R_2} = 0 \quad (14)$$

$$Lsv_1 - Lsv_i + s^2 LCR_1 v_1 + R_1 v_1 - R_1 v_o = 0 \text{ and } R_2 v_o - R_2 v_1 + sLv_o = 0 \quad (15)$$

Converting these back to the time domain we get:

$$Lv_1 - Lv_i + LCR_1 \ddot{v}_1 + R_1 \dot{v}_1 - R_1 \dot{v}_o = 0 \text{ and } R_2 v_o - R_2 v_1 + L\dot{v}_o = 0 \quad (16)$$

Expressing the equations in terms of the highest derivatives we get:

$$\ddot{v}_1 - \frac{1}{CR_1} \dot{v}_i = -\frac{1}{CR_1} \dot{v}_1 - \frac{1}{LC} v_1 + \frac{1}{LC} v_o \text{ and } v_o = -\frac{R_2}{L} v_o + \frac{R_2}{L} v_1 \quad (17)$$

Note that because the derivative of the input  $v_i$  appears in the first equation we bring the highest derivative of  $v_i$  to the LHS as well.

To get a state space model which does not involve the derivative of  $v_i$  we have to define our node voltage state as:

$$x_1 = v_1, x_2 = v_o \text{ and } x_3 = \dot{v}_1 - \frac{1}{CR_1} v_i \quad (18)$$

Therefore the state equations are:

$$\dot{x}_1 = \dot{v}_1 = x_3 + \frac{1}{CR_1} v_i \quad (19)$$

$$\dot{x}_2 = \dot{v}_o = -\frac{R_2}{L} v_o + \frac{R_2}{L} v_1 = -\frac{R_2}{L} x_2 + \frac{R_2}{L} x_1 \quad (20)$$

$$\dot{x}_3 = \dot{v}_1 - \frac{1}{CR_1} v_i = -\frac{1}{CR_1} v_1 - \frac{1}{LC} v_1 + \frac{1}{LC} v_o = -\frac{1}{CR_1} \left( x_3 + \frac{1}{CR_1} v_i \right) - \frac{1}{LC} x_1 + \frac{1}{LC} x_2 \quad (21)$$

Note that from our definition of the states we have that  $\dot{v}_1 = x_3 + \frac{1}{CR_1} v_i$ .

$$\dot{x}_1 = x_3 + \frac{1}{CR_1} v_i \quad (22)$$

$$\dot{x}_2 = -\frac{R_2}{L} x_2 + \frac{R_2}{L} x_1 \quad (23)$$

$$\dot{x}_3 = -\frac{1}{LC} x_1 + \frac{1}{LC} x_2 - \frac{1}{CR_1} x_3 - \frac{1}{(CR_1)^2} v_i \quad (24)$$

or in matrix form:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 \\ \frac{R_2}{L} & -\frac{R_2}{L} & 0 \\ -\frac{1}{CL} & \frac{1}{CL} & -\frac{1}{CR_1} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} \frac{1}{CR_1} \\ 0 \\ -\frac{1}{(CR_1)^2} \end{bmatrix} v_i \quad (25)$$

Since the output,  $x_2 = v_o$  we can write the output equation in matrix form as:

$$v_o = \begin{bmatrix} 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} \quad (26)$$

(iii) From Q2, Tutorial 2:

$$\frac{v_o}{v_i} = \frac{R_2}{R_1 CLs^2 + (CR_1 R_2 + L)s + (R_1 + R_2)} \quad (27)$$

Notice that the highest power of  $s$  in the denominator is 2. Therefore only 2 states are needed to describe the system (i.e the minimal state realisation).

To derive the minimal state-space model from this transfer function we begin by making the coefficient of the highest power of  $s$  on the denominator equal to 1, i.e.:

$$\frac{v_o}{v_i} = \frac{\left[ \frac{R_2}{R_1 CL} \right]}{s^2 + \left[ \frac{CR_1 R_2 + L}{R_1 CL} \right] s + \left[ \frac{(R_1 + R_2)}{(R_1 CL)} \right]} \quad (28)$$

Converting this into a differential equation gives:

$$\dot{v}_o = -\left[\frac{CR_1R_2 + L}{R_1CL}\right]v_o - \left[\frac{(R_1 + R_2)}{(R_1CL)}\right]v_o + \left[\frac{R_2}{R_1CL}\right]v_i \quad (29)$$

Defining the states as  $x_1 = v_o$  and  $x_2 = \dot{v}_o$  the state equations are:

$$\dot{x}_1 = \dot{v}_o = x_2 \quad (30)$$

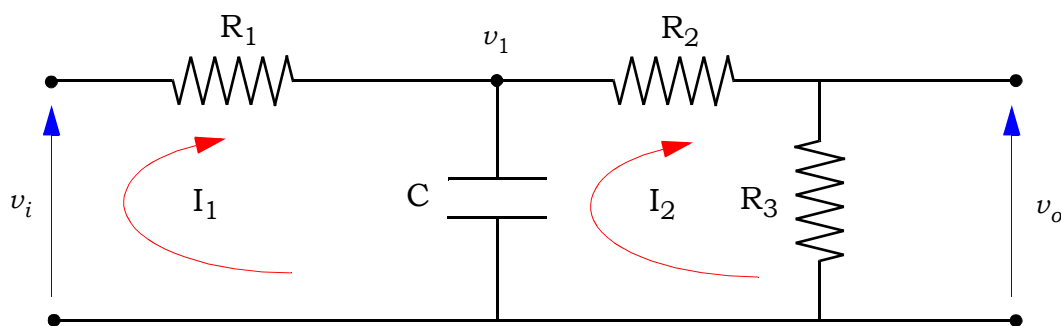
$$\begin{aligned} \dot{x}_2 = \dot{v}_o &= -\left[\frac{CR_1R_2 + L}{R_1CL}\right]v_o - \left[\frac{(R_1 + R_2)}{(R_1CL)}\right]v_o + \left[\frac{R_2}{R_1CL}\right]v_i \\ &= -\left[\frac{CR_1R_2 + L}{R_1CL}\right]x_1 - \left[\frac{(R_1 + R_2)}{(R_1CL)}\right]x_1 + \left[\frac{R_2}{R_1CL}\right]v_i \end{aligned} \quad (31)$$

Writing the state equations in matrix form we get:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\left[\frac{(R_1 + R_2)}{(R_1CL)}\right] & -\left[\frac{CR_1R_2 + L}{R_1CL}\right] \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ \left[\frac{R_2}{R_1CL}\right] \end{bmatrix} v_i \quad (32)$$

and since  $v_o = x_1$  the output equation in matrix form is:

$$v_o = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \quad (33)$$



Circuit B

(i) Using **mesh analysis**:

$$I_1R_1 + (I_1 - I_2)\frac{1}{sC} = v_i \text{ and } (I_2 - I_1)\frac{1}{sC} + I_2(R_2 + R_3) = 0 \quad (34)$$

Converting these back to the time domain we get:

$$CR_1\dot{I}_1 + I_1 - I_2 = C\dot{v}_i \text{ and } I_2 - I_1 + C(R_2 + R_3)\dot{I}_2 = 0 \quad (35)$$

Expressing the equations in terms of the highest derivatives we get:

$$\dot{I}_1 - \frac{1}{R_1} v_i = -\frac{1}{CR_1} I_1 + \frac{1}{CR_1} I_2 \text{ and } \dot{I}_2 = -\frac{1}{C(R_2 + R_3)} I_2 + \frac{1}{C(R_2 + R_3)} I_1 \quad (36)$$

Note that because the derivative of the input  $v_i$  appears in the first equation we bring the highest derivative of  $v_i$  to the LHS as well.

To get a state space model which does not involve the derivative of  $v_i$  we have to define our current state as:

$$x_1 = I_1 - \frac{1}{R_1} v_i \text{ and } x_2 = I_2 \quad (37)$$

Therefore the state equations are:

$$\dot{x}_1 = -\frac{1}{CR_1} \left( x_1 + \frac{1}{R_1} v_i \right) + \frac{1}{CR_1} x_2 \quad (38)$$

$$\dot{x}_2 = -\frac{1}{C(R_2 + R_3)} x_2 + \frac{1}{C(R_2 + R_3)} \left( x_1 + \frac{1}{R_1} v_i \right) \quad (39)$$

Note that from our definition of the states we have that  $I_1 = x_1 + \frac{1}{R_1} v_i$ .

$$\dot{x}_1 = -\frac{1}{CR_1} x_1 + \frac{1}{CR_1} x_2 - \frac{1}{CR_1^2} v_i \quad (40)$$

$$\dot{x}_2 = \frac{1}{C(R_2 + R_3)} x_1 - \frac{1}{C(R_2 + R_3)} x_2 + \frac{1}{CR_1(R_2 + R_3)} v_i \quad (41)$$

or in matrix form:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -\frac{1}{CR_1} & \frac{1}{CR_1} \\ \frac{1}{C(R_2 + R_3)} & -\frac{1}{C(R_2 + R_3)} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} -\frac{1}{CR_1^2} \\ \frac{1}{CR_1(R_2 + R_3)} \end{bmatrix} v_i \quad (42)$$

Since the output,  $v_o$  is the voltage across  $R_3$  we can write the output equation as:

$$v_o = I_2 R_3 = x_2 R_3 \quad (43)$$

hence the output equation in matrix form is:

$$v_o = \begin{bmatrix} 0 & R_3 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \quad (44)$$

(ii) Using **nodal analysis** with nodes  $v_1$  and  $v_o$ :

$$\frac{v_1 - v_i}{R_1} + \frac{v_1}{\frac{1}{sC}} + \frac{v_1 - v_o}{R_2} = 0 \text{ and } \frac{v_o - v_1}{R_2} + \frac{v_o}{R_3} = 0 \quad (45)$$

$$\frac{v_1}{R_1} - \frac{v_i}{R_1} + sCv_1 + \frac{v_1}{R_2} - \frac{v_o}{R_2} = 0 \text{ and } \frac{v_o}{R_2} + \frac{v_o}{R_3} = \frac{v_1}{R_2} \quad (46)$$

Converting these back to the time domain we get:

$$v_1 = -\left[\frac{1}{CR_1} + \frac{1}{CR_2}\right]v_1 + \frac{v_o}{R_2C} + \frac{v_i}{R_1C} \text{ and } v_o = \frac{R_3}{R_2 + R_3}v_1 \quad (47)$$

Note that the second equation has no derivative terms  $\Rightarrow$  it is not a dynamic equation and  $\Rightarrow$  only need one state,  $x_1 = v_1$ .

Therefore the state equations are:

$$\dot{x}_1 = -\left[\frac{1}{CR_1} + \frac{1}{CR_2}\right]x_1 + \frac{v_o}{R_2C} + \frac{v_i}{R_1C} \quad (48)$$

and using the second equation to substitute for  $v_o$ :

$$\dot{x}_1 = -\left[\frac{1}{CR_1} + \frac{1}{CR_2}\right]x_1 + \frac{R_3}{R_2C(R_2 + R_3)}x_1 + \frac{v_i}{R_1C} \quad (49)$$

$$\dot{x}_1 = \left[\frac{R_3}{R_2C(R_2 + R_3)} - \frac{1}{CR_1} - \frac{1}{CR_2}\right]x_1 + \left[\frac{1}{R_1C}\right]v_i \quad (50)$$

Since  $v_o = \frac{R_3}{R_2 + R_3}v_1$  the output equation is simply:

$$v_o = \left[\frac{R_3}{R_2 + R_3}\right]x_1 \quad (51)$$

(iii) From Q2, Tutorial 2:

$$\frac{v_o}{v_i} = \frac{R_3}{(R_1 + R_2 + R_3) + sCR_1(R_2 + R_3)} \quad (52)$$

Notice that the highest power of  $s$  in the denominator is 1. Therefore only 1 state is needed to describe the system (i.e the minimal state realisation). Therefore the state space model obtained using nodal analysis is a minimum state realisation.

To derive a minimal state-space model from the transfer function we begin by making the coefficient of the highest power of  $s$  on the denominator equal to 1, i.e.:

$$\frac{v_o}{v_i} = \frac{\left[\frac{R_3}{CR_1(R_2 + R_3)}\right]}{s + \left[\frac{(R_1 + R_2 + R_3)}{CR_1(R_2 + R_3)}\right]} \quad (53)$$

Converting this into a differential equation gives:

$$\dot{v}_o = -\left[\frac{(R_1 + R_2 + R_3)}{CR_1(R_2 + R_3)}\right]v_o + \left[\frac{R_3}{CR_1(R_2 + R_3)}\right]v_i \quad (54)$$

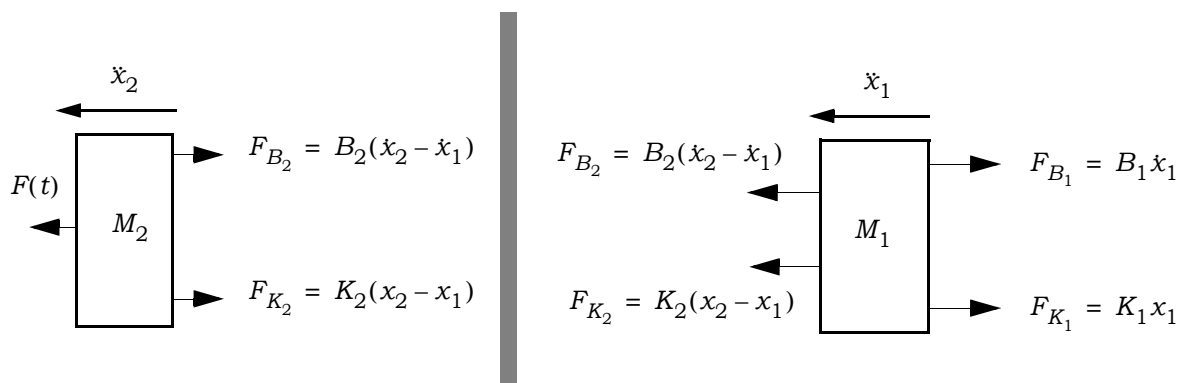
Defining the state as  $x_1 = v_o$  we get the state equation:

$$\dot{x}_1 = -\left[\frac{(R_1 + R_2 + R_3)}{CR_1(R_2 + R_3)}\right]x_1 + \left[\frac{R_3}{CR_1(R_2 + R_3)}\right]v_i \quad (55)$$

and the output equation is:

$$v_o = [1]x_1 \quad (56)$$

Q2 (i) The free body diagram for the masses in the system are as follows.



Therefore for  $M_2$ , Newton's 2nd law gives:

$$F(t) - B_2(x_2 - x_1) - K_2(x_2 - x_1) = M_2\ddot{x}_2 \quad (57)$$

$$\Rightarrow \ddot{x}_2 = -\frac{B_2}{M_2}\dot{x}_2 - \frac{K_2}{M_2}x_2 + \frac{B_2}{M_2}\dot{x}_1 + \frac{K_2}{M_2}x_1 + \frac{F(t)}{M_2} \quad (58)$$

and for  $M_1$ :

$$B_2(x_2 - x_1) + K_2(x_2 - x_1) - B_1x_1 - K_1x_1 = M_1\ddot{x}_1 \quad (59)$$

$$\Rightarrow \ddot{x}_1 = -\frac{B_1 + B_2}{M_1}\dot{x}_1 - \frac{K_1 + K_2}{M_1}x_1 + \frac{B_2}{M_1}\dot{x}_2 + \frac{K_2}{M_1}x_2 \quad (60)$$

Defining the states as:

$$X_1 = x_1, X_2 = \dot{x}_1, X_3 = x_2 \text{ and } X_4 = \dot{x}_2 \quad (61)$$

we obtain the following state equations:

$$\dot{X}_1 = X_2 \quad (62)$$

$$\dot{X}_2 = -\frac{(B_1 + B_2)}{M_1}X_2 - \frac{(K_1 + K_2)}{M_1}X_1 + \frac{B_2}{M_1}X_4 + \frac{K_2}{M_1}X_3 \quad (63)$$

$$\dot{X}_3 = X_4 \quad (64)$$

$$\dot{X}_4 = -\frac{B_2}{M_2}X_4 - \frac{K_2}{M_2}X_3 + \frac{B_2}{M_2}X_2 + \frac{K_2}{M_2}X_1 + \frac{F(t)}{M_2} \quad (65)$$

Writing these in matrix form gives:

$$\begin{bmatrix} \dot{X}_1 \\ \dot{X}_2 \\ \dot{X}_3 \\ \dot{X}_4 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{(K_1 + K_2)}{M_1} & -\frac{(B_1 + B_2)}{M_1} & \frac{K_2}{M_1} & \frac{B_2}{M_1} \\ 0 & 0 & 0 & 1 \\ \frac{K_2}{M_2} & \frac{B_2}{M_2} & -\frac{K_2}{M_2} & -\frac{B_2}{M_2} \end{bmatrix} \begin{bmatrix} X_1 \\ X_2 \\ X_3 \\ X_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ \frac{1}{M_2} \end{bmatrix} F(t) \quad (66)$$

(ii) Outputs are:

$$y_1 = x_1 \text{ and } y_2 = x_2 - x_1 \quad (67)$$

Therefore in terms of the states we have:

$$y_1 = X_1 \text{ and } y_2 = X_4 - X_2 \quad (68)$$

which in matrix form corresponds to:

$$\begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & -1 & 0 & 1 \end{bmatrix} \begin{bmatrix} X_1 \\ X_2 \\ X_3 \\ X_4 \end{bmatrix} \quad (69)$$

Q3 (i) The 2nd order transfer model (Q5, Tutorial 2) is:

$$\frac{y(z)}{u(z)} = \frac{0.65z^{-1} + 2.55z^{-2}}{1 + 3.1z^{-1} - 2.51z^{-2}} \quad (70)$$

Defining intermediate variable  $q(z)$  we can write:

$$\frac{q(z)}{u(z)} = \frac{1}{1 + 3.1z^{-1} - 2.51z^{-2}} \text{ and } \frac{y(z)}{q(z)} = 0.65z^{-1} + 2.55z^{-2} \quad (71)$$

i.e.:

$$u_k = q_k + 3.1q_{k-1} - 2.51q_{k-2} \text{ and } y_k = 0.65q_{k-1} + 2.55q_{k-2} \quad (72)$$

Rearranging the first equation gives:

$$q_k = -3.1q_{k-1} + 2.51q_{k-2} + u_k \quad (73)$$

Defining the states as:

$$x_1(k) = q_{k-1} \text{ and } x_2(k) = q_{k-2} \quad (74)$$

we obtain the states equations:

$$x_1(k+1) = q_k = -3.1x_1(k) + 2.51x_2(k) + u_k \quad (75)$$

$$x_2(k+1) = q_{k-1} = x_1(k) \quad (76)$$



Therefore the state equation in matrix form is :

$$\begin{bmatrix} x_1(k+1) \\ x_2(k+1) \end{bmatrix} = \begin{bmatrix} -3.1 & 2.51 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} x_1(k) \\ x_2(k) \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u_k \quad (77)$$

or

$$\underline{X}(k+1) = \begin{bmatrix} -3.1 & 2.51 \\ 1 & 0 \end{bmatrix} \underline{X}(k) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u_k \quad (78)$$

The output equation is:

$$y_k = 0.65q_{k-1} + 2.55q_{k-2} = 0.65x_1(k) + 2.55x_2(k)$$

In matrix form this is:

$$y_k = \begin{bmatrix} 0.65 & 2.55 \end{bmatrix} \underline{X}(k) \quad (79)$$

(ii)

Taking the z transform of the state equation we get:

$$z\underline{X}(z) = \begin{bmatrix} -3.1 & 2.51 \\ 1 & 0 \end{bmatrix} \underline{X}(z) + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(z) \quad (80)$$

$$\Rightarrow \left\{ \begin{bmatrix} z & 0 \\ 0 & z \end{bmatrix} - \begin{bmatrix} -3.1 & 2.51 \\ 1 & 0 \end{bmatrix} \right\} \underline{X}(z) = \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(z) \quad (81)$$

$$\Rightarrow \begin{bmatrix} z+3.1 & -2.51 \\ -1 & z \end{bmatrix} \underline{X}(z) = \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(z)$$

$$\Rightarrow \underline{X}(z) = \begin{bmatrix} z+3.1 & -2.51 \\ -1 & z \end{bmatrix}^{-1} \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(z) \quad (82)$$

$$\Rightarrow \underline{X}(z) = \frac{1}{z(z+3.1)-2.51} \begin{bmatrix} z & 2.51 \\ 1 & z+3.1 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} u(z) \quad (83)$$

$$\Rightarrow \underline{X}(z) = \frac{1}{z^2 + 3.1z - 2.51} \begin{bmatrix} z \\ 1 \end{bmatrix} u(z) \quad (84)$$

The z transform of the output equation is:

$$y(z) = \begin{bmatrix} 0.65 & 2.55 \end{bmatrix} \underline{X}(z) \quad (85)$$

Substituting for  $\underline{X}(z)$  gives:

$$y(z) = \frac{1}{z^2 + 3.1z - 2.51} \begin{bmatrix} 0.65 & 2.55 \end{bmatrix} \begin{bmatrix} z \\ 1 \end{bmatrix} u(z) \quad (86)$$

$$\Rightarrow y(z) = \frac{0.65z + 2.55}{z^2 + 3.1z - 2.51} u(z) \quad (87)$$

Hence the transfer function is:

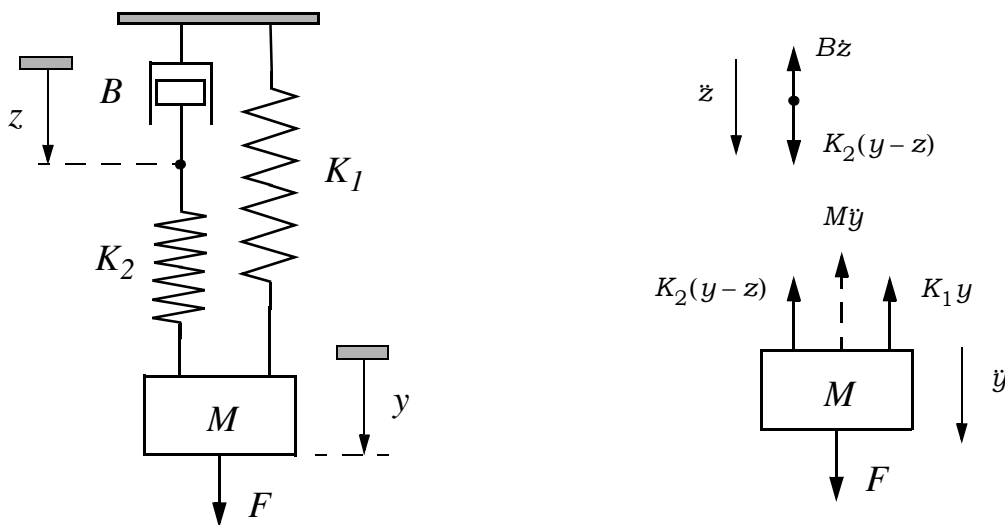
$$\frac{y(z)}{u(z)} = \frac{0.65z + 2.55}{z^2 + 3.1z - 2.51} \quad (88)$$

or

$$\frac{y(z)}{u(z)} = \frac{0.65z^{-1} + 2.55z^{-2}}{1 + 3.1z^{-1} - 2.51z^{-2}} \quad (89)$$

as before!

Q4



(i) Equations of motion for this system are obtained from the free body diagrams shown above, as follows:

$$\begin{aligned} F &= M\ddot{y} + K_1 y + K_2(y - z) \\ K_2(y - z) &= B\dot{z} \end{aligned}$$

Rearranging in terms of the highest derivatives:

$$\ddot{y} = -\frac{(K_1 + K_2)}{M}y + \frac{K_2}{M}z + \frac{F}{M}, \quad \dot{z} = \frac{K_2}{B}(y - z)$$

(ii) Choosing the states as  $x_1 = y$ ,  $x_2 = \dot{y}$  and  $x_3 = z$  and the output as  $y$  gives:

$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= -\frac{(K_1 + K_2)}{M}x_1 + \frac{K_2}{M}x_3 + \frac{F}{M} \\ \dot{x}_3 &= \frac{K_2}{B}x_1 - \frac{K_2}{B}x_3 \end{aligned}$$

or in matrix form:

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 \\ -\frac{(K_1 + K_2)}{M} & 0 & \frac{K_2}{M} \\ \frac{K_2}{B} & 0 & -\frac{K_2}{B} \end{bmatrix} x + \begin{bmatrix} 0 \\ \frac{1}{M} \\ 0 \end{bmatrix} F, \quad y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x$$

Subbing in the numbers gives:

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 \\ -100 & 0 & 40 \\ 10 & 0 & -10 \end{bmatrix} x + \begin{bmatrix} 0 \\ 10 \\ 0 \end{bmatrix} F, \quad y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x$$

(iii)(a)

Taking the Laplace transform of the equations gives:

$$s^2 y(s) = -\frac{(K_1 + K_2)}{M} y(s) + \frac{K_2}{M} z(s) + \frac{F(s)}{M} \Rightarrow \left( s^2 + \frac{(K_1 + K_2)}{M} \right) y(s) = \frac{K_2}{M} z(s) + \frac{F(s)}{M},$$

$$s z(s) = \frac{K_2}{B} (y(s) - z(s)) \Rightarrow \left( s + \frac{K_2}{B} \right) z(s) = \frac{K_2}{B} y(s) \Rightarrow z(s) = \frac{\frac{K_2}{B}}{\left( s + \frac{K_2}{B} \right)} y(s)$$

Substituting for z(s) gives:

$$\begin{aligned} \left[ s^2 + \frac{(K_1 + K_2)}{M} \right] y(s) &= \frac{K_2}{M} \left( \frac{\frac{K_2}{B}}{\left( s + \frac{K_2}{B} \right)} \right) y(s) + \frac{F(s)}{M} \\ \Rightarrow \left( s + \frac{K_2}{B} \right) \left[ s^2 + \frac{(K_1 + K_2)}{M} \right] y(s) &= \frac{K_2 K_2}{M B} y(s) + \frac{F(s)}{M} \left( s + \frac{K_2}{B} \right) \\ \Rightarrow \left[ s^3 + \frac{K_2}{B} s^2 + \frac{(K_1 + K_2)}{M} s + \frac{K_2 (K_1 + K_2)}{B M} \right] y(s) &= \frac{K_2 K_2}{M B} y(s) + \frac{F(s)}{M} \left( s + \frac{K_2}{B} \right) \\ \Rightarrow \left[ s^3 + \frac{K_2}{B} s^2 + \frac{(K_1 + K_2)}{M} s + \frac{K_1 K_2}{B M} \right] y(s) &= \frac{F(s)}{M} \left( s + \frac{K_2}{B} \right) \\ \Rightarrow \frac{y(s)}{F(s)} &= \frac{\frac{1}{M} \left( s + \frac{K_2}{B} \right)}{\left[ s^3 + \frac{K_2}{B} s^2 + \frac{(K_1 + K_2)}{M} s + \frac{K_1 K_2}{B M} \right]} \\ \Rightarrow \frac{y(s)}{F(s)} &= \frac{10(s + 10)}{s^3 + 10s^2 + 100s + 600} \end{aligned}$$

(b)

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 \\ -100 & 0 & 40 \\ 10 & 0 & -10 \end{bmatrix} x + \begin{bmatrix} 0 \\ 10 \\ 0 \end{bmatrix} F, \quad y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} x$$

$$G(s) = C(sI - A)^{-1}B$$

$$\begin{aligned} (sI - A)^{-1} &= \begin{bmatrix} s & -1 & 0 \\ 100 & s & -40 \\ -10 & 0 & s+10 \end{bmatrix}^{-1} = \frac{\begin{bmatrix} \left| \begin{smallmatrix} s & -40 \\ 0 & s+10 \end{smallmatrix} \right| & - & \left| \begin{smallmatrix} 100 & -40 \\ -10 & s+10 \end{smallmatrix} \right| & & \left| \begin{smallmatrix} 100 & s \\ -10 & 0 \end{smallmatrix} \right| \\ - & \left| \begin{smallmatrix} -1 & 0 \\ 0 & s+10 \end{smallmatrix} \right| & & \left| \begin{smallmatrix} s & 0 \\ -10 & s+10 \end{smallmatrix} \right| & - & \left| \begin{smallmatrix} s & -1 \\ -10 & 0 \end{smallmatrix} \right| \\ & \left| \begin{smallmatrix} -1 & 0 \\ s & -40 \end{smallmatrix} \right| & - & \left| \begin{smallmatrix} s & 0 \\ 100 & -40 \end{smallmatrix} \right| & & \left| \begin{smallmatrix} s & -1 \\ 100 & s \end{smallmatrix} \right| \end{bmatrix}}{s \left| \begin{smallmatrix} s & -40 \\ 0 & s+10 \end{smallmatrix} \right| - (-1) \left| \begin{smallmatrix} 100 & -40 \\ -10 & s+10 \end{smallmatrix} \right|} \\ &= \frac{\begin{bmatrix} s^2 + 10s & -100s - 600 & 10s \\ s+10 & s^2 + 10s & 10 \\ 40 & 40s & s^2 + 100 \end{bmatrix}^T}{s^3 + 10s^2 + 100s + 600} = \frac{\begin{bmatrix} s^2 + 10s & s+10 & 40 \\ -100s - 600 & s^2 + 10s & 40s \\ 10s & 10 & s^2 + 100 \end{bmatrix}}{s^3 + 10s^2 + 100s + 600} \end{aligned}$$

Therefore:

$$\begin{aligned} G(s) &= C(sI - A)^{-1}B = \frac{\begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} s^2 + 10s & s+10 & 40 \\ -100s - 600 & s^2 + 10s & 40s \\ 10s & 10 & s^2 + 100 \end{bmatrix} \begin{bmatrix} 0 \\ 10 \\ 0 \end{bmatrix}}{s^3 + 10s^2 + 100s + 600} \\ &= \frac{10(s+10)}{s^3 + 10s^2 + 100s + 600} \end{aligned}$$

(iv)

$$G(s) = \frac{10s + 100}{s^3 + 10s^2 + 100s + 600}$$

Following the method given in the notes we get:

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -600 & -100 & -10 \end{bmatrix} x + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} F, \quad y = \begin{bmatrix} 100 & 10 & 0 \end{bmatrix} x$$

Verifying ...

$$\begin{aligned}
(sI - A)^{-1} &= \begin{bmatrix} s & -1 & 0 \\ 0 & s & -1 \\ 600 & 100 & s+10 \end{bmatrix}^{-1} = \frac{\begin{bmatrix} s^2 + 10s + 100 & -600 & -600s \\ s+10 & s^2 + 10s - 100s - 600 \\ 1 & s & s^2 \end{bmatrix}^T}{s \begin{vmatrix} s & -1 \\ 100 & s+10 \end{vmatrix} - (-1) \begin{vmatrix} 0 & -1 \\ 600 & s+10 \end{vmatrix}} \\
&= \frac{\begin{bmatrix} s^2 + 10s + 100 & s+10 & 1 \\ -600 & s^2 + 10s & s \\ -600s & -100s - 600 & s^2 \end{bmatrix}}{s(s^2 + 10s + 100) + 600} = \frac{\begin{bmatrix} s^2 + 10s + 100 & s+10 & 1 \\ -600 & s^2 + 10s & s \\ -600s & -100s - 600 & s^2 \end{bmatrix}}{s^3 + 10s^2 + 100s + 600} \\
G(s) = C(sI - A)^{-1}B &= \frac{\begin{bmatrix} 100 & 10 & 0 \end{bmatrix} \begin{bmatrix} s^2 + 10s + 100 & s+10 & 1 \\ -600 & s^2 + 10s & s \\ -600s & -100s - 600 & s^2 \end{bmatrix} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}{s^3 + 10s^2 + 100s + 600} \\
&= \frac{\begin{bmatrix} 100 & 10 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ s \\ s^2 \end{bmatrix}}{s^3 + 10s^2 + 100s + 600} \\
&= \frac{10s + 100}{s^3 + 10s^2 + 100s + 600}
\end{aligned}$$