Lecture 6: Frequency Analysis EE213 - Introduction to Signal Processing

Semester 1, 2021

Outline

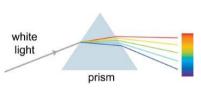
- Frequency analysis of continuous and discrete signals.
 - → Fourier series and Fourier transforms.
- Calculate Fourier series and transforms of elementary signals.

Frequency Analysis

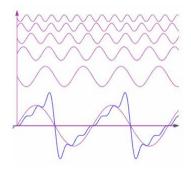
- Frequency analysis
 - → Decomposes a signal into its frequency (sinusoidal) components which are also known as spectrum.
 - →Different signals have different spectra. Thus, spectrum provides an identity or a signature for the signal.
 - →The process of obtaining the spectrum of a given signal using basic mathematical tools is known as frequency or spectral analysis.
- Tools for frequency analysis
 - →Fourier series
 - → Fourier transform

Frequency Analysis...

 A good way to think about frequency analysis of a signal: As a mathematical version of a prism: breaking up a signal into different frequencies, just as a prism (or diffraction grating) breaks up light into different colours (which are light at different frequencies).



Prism breaks up light into different colours



Fourier series and transforms breaks up a signal into different frequencies

 Let x(t) be any real-valued periodic signal having period = T seconds, i.e.,

$$x(t) = x(t+T) \tag{1}$$

• Then x(t) can be expanded as a sum of sinusoids having frequencies that are integer multiples of $f_0 = \frac{1}{T}$

Hertz
$$x(t) = c_0 + c_1 \cos(\omega_0 t - \theta_1) + (c_2 \cos(2\omega_0 t - \theta_2) + c_3 \cos(3\omega_0 t - \theta_3) + c_4 \cos(4\omega_0 t - \theta_4) + \dots$$
 (2)

Recall that

$$\cos(x - y) = \cos(x)\cos(y) + \sin(x)\sin(y) \tag{3}$$

We can rewrite x(t) as

$$x(t) = a_0 + a_1 \cos(\omega_0 t) + a_2 \cos(2\omega_0 t) + a_3 \cos(3\omega_0 t) + \cdots$$
$$+b_1 \sin(\omega_0 t) + b_2 \sin(2\omega_0 t) + b_3 \sin(3\omega_0 t) + \cdots$$
(4)

Further, recall that

$$\cos = \frac{e^{jx} + e^{-jx}}{2} \qquad \sin = \frac{e^{jx} - e^{-jx}}{2j}$$
 (5)

We can also rewrite x(t) as

$$x(t) = x_0 + x_1 e^{j2\pi f_0 t} + x_2 e^{j4\pi f_0 t} + x_3 e^{j6\pi f_0 t} + x_4 e^{j8\pi f_0 t} + \cdots + x_1^* e^{-j2\pi f_0 t} + x_2^* e^{-j4\pi f_0 t} + x_3^* e^{-j6\pi f_0 t} + x_4^* e^{-j8\pi f_0 t} \dots$$
(6)

- Note that xn in the above equation is a complex number and X_n^* is the complex conjugate of xn. That is, if $x_n a_n = + \mathrm{j} b_n$ then $x_n^* = a_n \mathrm{j} b_n$
- We can see that there are three different types of Fourier series, which expand the periodic signal x(t) as a sum of
 - \rightarrow Phase-shifted sinusoids (Equation (2)).
 - \rightarrow Sines plus cosines (Equation (4)).
 - \rightarrow Complex exponentials (Equation (6)).
- They are equivalent but each has a different ease of computation.

Definition (Fourier Series)

We use the FS as sum of complex exponentials

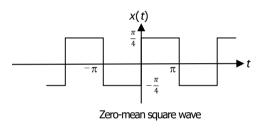
$$x(t) = \sum_{n = -\infty}^{\infty} x_n e^{j2\pi n f_0 t}$$
 (7)

$$x_n = \frac{1}{T} \int_0^T x(t)e^{-j2\pi nf_0 t} dt$$
 (8)

where x_n are the Fourier series coefficient and $n = 0, \pm 1, \pm 2, \ldots$

Example

Find the FS coefficient for the following period signal.



• Fundamental period: $T = 2 \pi$

静下心来好好学习! 争取下午整理完slide!

Example (continued)

DC component :

$$x_0 = rac{1}{T} \int_0^T x(t) dt = rac{1}{2\pi} \left(\int_{-\pi}^0 (-rac{\pi}{4}) + \int_0^\pi (rac{\pi}{4})
ight) dt = 0$$

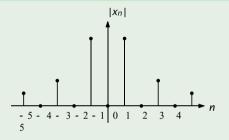
ullet Recall that $\int e^{at}dt=rac{1}{a}e^{at}+c$, where c is a constant. Thus

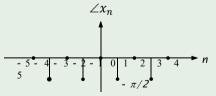
$$x_n = \frac{1}{2\pi} \left(\int_{-\pi}^0 (-\frac{\pi}{4}) e^{-j2\pi n f_0 t} dt + \int_0^{\pi} (\frac{\pi}{4}) e^{-j2\pi n f_0 t} dt \right)$$
 (9)

$$= \frac{1}{2\pi} \left(\int_{-\pi}^{0} (-\frac{\pi}{4}) e^{-jnt} dt + \int_{0}^{\pi} (\frac{\pi}{4}) e^{-jnt} dt \right)$$
 (10)

$$= \begin{cases} 0 & n \text{ even} \\ -\frac{j}{2n} & n \text{ odd} \end{cases}$$
 (11)

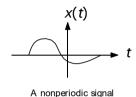
Example (continued)



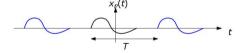


Amplitude and phase representation of the Fourier series of zero-mean square wave

Consider a continuous nonperiodic signal x(t)



• Let us consider a periodic signal $x_p(t)$ obtained by repeating x(t) after a period of T seconds.



The expanded periodic signal

- We can say that when $T \rightarrow \infty$, $x_p(t) \rightarrow x(t)$
- Fourier transform of x(t) can be obtained as a limiting case of Fourier series of x_p (t) as period T $\rightarrow \infty$.

Definition (Fourier Transform)

Fourier Transform of x(t) is given by

$$X(\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt$$
 (12)

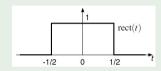
and inverse FT is given by

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$
 (13)

FT is also known as continuous time FT (CTFT).

Example

Find the FT of the unit rectangular signal



FT of unit rectangular signal

$$X(\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt = \int_{-1/2}^{1/2} (1)e^{-j\omega t}dt$$
 (14)

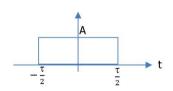
$$= \frac{1}{-j\omega} \left(e^{-j\omega t} \Big|_{-1/2}^{1/2} \right) \tag{15}$$

$$= \frac{\sin(\omega/2)}{\omega/2} = \operatorname{sinc}(\omega/2) \tag{16}$$

Example (continued)

Time and frequency domain representation of the rectangular signal.

Time domain



rectangular signal

$$F(w) = \int_{-\infty}^{\infty} f(t)e^{-j\omega t}dt$$

$$= \frac{A}{-j\omega}e^{-j\omega t} \Big|_{-\tau/2}^{-\tau/2}$$

$$= \frac{A}{-j\omega} \left(e^{-j\frac{\tau}{2}w} - e^{j\frac{\tau}{2}w} \right)$$

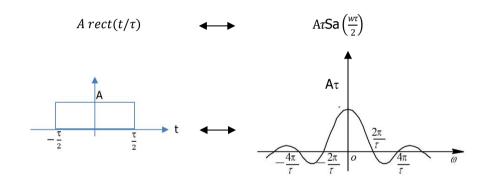
$$= \frac{A}{j\omega} \sin\left(\frac{w\tau}{2}\right) 2j$$

$$= A\tau \sin\left(\frac{w\tau}{2}\right) / \frac{w\tau}{2}$$

 $= A\tau Sa\left(\frac{w\tau}{2}\right)$

 $Sa(x) = \frac{sinx}{}$

Sinc(x)= $\frac{\sin x}{x}$

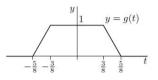


FT of rectangular signal

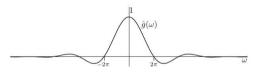
Example

Analyse the following signal in the frequency domain

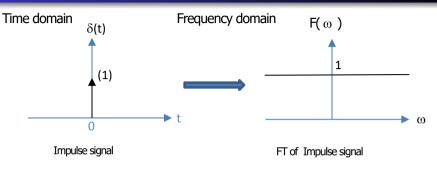
Time domain



Frequency domain (FT)



• The FT of the signal given in this example looks somewhat like the Fourier transform in the previous example but exhibits faster decay for large ω . (what is the intuitive explanation?)



$$F(\omega) = \int_{-\infty}^{\infty} f(t)e^{-j\omega t}dt$$
$$= \int_{-\infty}^{\infty} \delta(t) e^{-j\omega t}dt$$
$$= 1$$

- The Fourier transform is a major cornerstone in the analysis and representation of signals.
- However, computing FT of a signal may be a difficult task, especially when we have a very complex signal.
- Recall that a complex signal can be constructed from elementary signals using proper basic operations. This fact suggests that we can compute FT of a complex signal based on FT of basic signals. Two main ingredients for computing FT
 - → Fourier transform table: show FT of common signals.
 - \rightarrow Fourier transform properties: relates the operations on time domain to the operations on frequency domain.
- From now on, we use notations F() and $F^{-1}()$ to denote the FT and inverse FT, respectively.

Linearity: if

$$x(t) = \alpha_1 x_1(t) + \alpha_2 x_2(t)$$
 (17)

Then the FT of x(t) is

$$F(x(t)) = \alpha_1 F(x_1(t)) + \alpha_2 F(x_2(t))$$
 (18)

Time Shifting:

$$x(t-t_0) \stackrel{\mathcal{F}}{\leftrightarrow} e^{-j\omega t_0} X(\omega)$$
 (19)

Frequency shift:

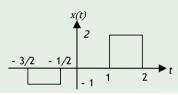
$$X(w-w_0) \stackrel{\mathcal{F}}{\longleftrightarrow} e^{jw_0t} x(t)$$

Time and frequency scaling

$$x(at) \stackrel{\mathcal{F}}{\leftrightarrow} \frac{1}{|a|} X(\frac{\omega}{a})$$
 (20)

Example

Find the FT of the following signal



Solution:

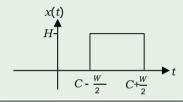
A rect
$$(t/\tau)$$
 \longrightarrow A τ Sa $\left(\frac{\omega\tau}{2}\right)$ $x(t)$ \longleftrightarrow $2e^{-j\frac{3}{2}\omega}$ Sa $\left(\frac{\omega}{2}\right) - e^{j\omega}$ Sa $\left(\frac{\omega}{2}\right)$ $x(t) = 2rect\left(t - \frac{3}{2}\right) - rect(t+1)$

$$2rect\left(t-\frac{3}{2}\right) \longleftrightarrow 2e^{-j\frac{3}{2}\omega}Sa(\frac{\omega}{2})$$

$$rect(t+1) \iff e^{j\omega}Sa(\frac{\omega}{2})$$

Example

Find the FT of the following signal



Solution:

$$A \operatorname{rect}(t/\tau) \longleftrightarrow A \tau \operatorname{Sa}\left(\frac{\omega \tau}{2}\right)$$

$$X(t) = H rect(\frac{t-C}{W})$$

$$H \, rect \, (\frac{t-C}{W}) \longleftrightarrow HW e^{-j\omega C} Sa(\frac{W\omega}{2})$$

Differentiation:

$$\frac{dx(t)}{dt} \stackrel{\mathcal{F}}{\leftrightarrow} j\omega X(\omega) \tag{21}$$

Integration:

$$\int_{-\infty}^{t} x(\tau)d\tau \stackrel{\mathcal{F}}{\leftrightarrow} \frac{1}{j\omega} X(\omega) + \pi X(0)\delta(\omega) \tag{22}$$

 The differentiation and integration properties are particularly useful when we use the Fourier transform to analyse RLC circuits.

Example:

$$\begin{split} \mathbf{u}(\mathbf{t}) &= \int_{-\infty}^t \! \delta(\tau) \; \mathrm{d}\mathbf{t} \\ \delta(\mathbf{t}) & \longleftarrow \quad 1 \\ \mathbf{u}(\mathbf{t}) & \longleftarrow \quad \frac{1}{jw} + \pi \, \delta(w) \end{split}$$

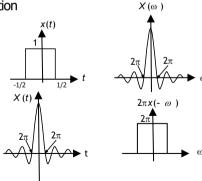
Time/Frequency Duality: If

$$x(t) \stackrel{\mathcal{F}}{\leftrightarrow} X(\omega)$$
 (23)

then

$$X(t) \stackrel{\mathcal{F}}{\leftrightarrow} 2\pi x(-\omega)$$
 (24)

Graphical illustration



FT of DC signal

$$\delta$$
 (t) \longrightarrow 1
$$1 \longrightarrow 2\pi\delta(\omega)$$

2. FT of sinusoidal signal

$$e^{iw_0t} \longrightarrow 2\pi(\omega - \omega_0)$$

$$e^{-jw_0t} \longrightarrow 2\pi(\omega + \omega_0)$$

$$\cos(w_0t) = \frac{1}{2} \left(e^{jw_0t} + e^{-jw_0t} \right)$$

$$\cos(w_0t) \longrightarrow \pi \left[\delta(\omega - \omega_0) + \delta(\omega + \omega_0) \right]$$

$$\sin(w_0t) = \frac{1}{2j} \left(e^{jw_0t} - e^{-jw_0t} \right)$$

$$\sin(w_0t) \longrightarrow j\pi \left[\delta(\omega - \omega_0) - \delta(\omega + \omega_0) \right]$$

Fourier Transform Table

$$f(t) \leftrightarrow F(\omega)$$

$$\delta(t) \leftrightarrow 1$$

$$1 \leftrightarrow 2\pi\delta(\omega)$$

$$e^{\pm j\omega_0 t} \leftrightarrow 2\pi\delta(\omega \mp \omega_0)$$

$$\cos \omega_0 t \leftrightarrow \pi \left[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)\right]$$

$$\sin \omega_0 t \leftrightarrow -j\pi \left[\delta(\omega - \omega_0) - \delta(\omega + \omega_0)\right]$$

$$rect(t/\tau) \leftrightarrow \tau \operatorname{sinc}(\omega \tau/2)$$

$$\frac{W}{\pi} \operatorname{sinc}(Wt) \leftrightarrow rect(\omega/(2W))$$

Fourier transform table.

Discrete-time Fourier Series

DTFS applies to discrete-time periodic signals.

Definition (Discrete-time Fourier Series)

Consider x[n] = x[n + N], where N is the fundamental period. Then the DTFS representation of x[n] is given by

$$X[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-jk\Omega_0 n}, k = 0, 1, \dots, N-1$$
 (25)

$$x[n] = \sum_{k=0}^{N-1} X[k]e^{jn\Omega_0 k}$$
 (26)

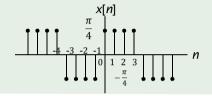
where $\Omega_0 = \frac{2\pi}{N}$ is the fundamental frequency.

- DTFS is similar to CTFS.
- Note that DTFS is a periodic sequence, i.e., X[k] = X[k + N].

Discrete-time Fourier Series...

Example

Determine DTFS of the following discrete periodic signal



Discrete-time Fourier Series...

$$\begin{split} &\mathsf{X}[\text{-}4] = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] \, e^{j4\frac{2\pi}{8}n} = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] \, e^{j\pi n} \\ &= \frac{1}{8} \left(-\frac{\pi}{4} \right) (\; e^{-j4\pi} + e^{-j3\pi} + e^{-j2\pi} + e^{-j\pi} - 1 - e^{j\pi} - \; e^{j2\pi} - \; e^{j3\pi}) \\ &= 0 \\ &\mathsf{X}[\text{-}3] = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] \, e^{j3\frac{2\pi}{8}n} = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] e^{j\frac{3}{4}m} \\ &= \frac{1}{8} \left(-\frac{\pi}{4} \right) (\; e^{-j3\pi} + e^{-j\frac{3}{4}\pi} + e^{-j\frac{3}{4}\pi} + e^{-j\frac{3}{4}\pi} - 1 - e^{j\frac{3}{4}\pi} - e^{j\frac{3}{2}\pi} - e^{j\frac{9}{4}\pi} \right) \\ &= \frac{1}{8} \left(-\frac{\pi}{4} \right) (\; -2 - 2 \mathrm{j} \mathrm{sin} \, \frac{9}{4} \pi - 2 \mathrm{j} \mathrm{sin} \, \frac{3}{2} \pi - 2 \mathrm{j} \mathrm{sin} \, \frac{3}{4} \pi \right) \\ &= \frac{1}{8} \left(-\frac{\pi}{4} \right) (\; -2 + \mathrm{j} 2 - \mathrm{j} 2 \, \sqrt{2} \;) \\ &= \left(-\frac{\pi}{4} \right) \left(-\frac{1+\mathrm{j}(1-\sqrt{2})}{4} \right) \\ &\mathsf{X}[3] = \left(-\frac{\pi}{4} \right) \left(-\frac{1-\mathrm{j}(1-\sqrt{2})}{4} \right) \end{split}$$

Discrete-time Fourier Series...

$$\begin{split} &\mathsf{X}[\text{-}2] = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] \, e^{j2\frac{2\pi}{8}n} = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] \, e^{j\frac{\pi}{2}n} \\ &= \frac{1}{8} \left(-\frac{\pi}{4} \right) (\; e^{-j2\pi} + e^{-j\frac{3}{2}\pi} + e^{-j\pi} + e^{-j\frac{\pi}{2}} - 1 - e^{j\frac{\pi}{2}} - e^{j\pi} - e^{j\frac{3}{2}\pi}) \\ &= 0 \\ &\mathsf{X}[2] = 0 \\ &\mathsf{X}[\text{-}1] = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] \, e^{j\frac{2\pi}{8}n} = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] \, e^{j\frac{\pi}{4}n} \\ &= \frac{1}{8} \left(-\frac{\pi}{4} \right) (\; e^{-j\pi} + e^{-j\frac{3}{4}\pi} + e^{-j\frac{1}{2}\pi} + e^{-j\frac{\pi}{4}} - 1 - e^{j\frac{\pi}{4}} - e^{j\frac{1}{2}\pi} - e^{j\frac{3}{4}\pi}) \\ &= \frac{1}{8} \left(-\frac{\pi}{4} \right) (\; -2 - 2j\sin\frac{3}{4}\pi - 2j\sin\frac{1}{4}\pi - 2j\sin\frac{1}{2}\pi) \\ &= \left(-\frac{\pi}{4} \right) \left(\frac{-1 - j(1 + \sqrt{2})}{4} \right) \\ &\mathsf{X}[1] = \left(-\frac{\pi}{4} \right) \left(\frac{-1 + j(1 + \sqrt{2})}{4} \right) \\ &\mathsf{X}[0] = \frac{1}{8} \; \sum_{n=-4}^{3} \; \mathsf{x}[\mathsf{n}] = 0 \end{split}$$

Discrete-time Fourier Transform

- DTFT applies to discrete aperiodic signals.
- DTFT is similar to FT for continuous signals, except
 - \rightarrow Integral is replaced by summation.
 - \rightarrow x(t) is replaced by sampled values x[n].

Definition (Discrete Time Fourier Transform)

DTFT of x[n] is given by

$$X(\omega) = \sum_{-\infty}^{\infty} x[n]e^{-j\omega n}$$
 (28)

and inverse FT is given by

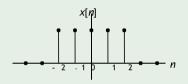
$$x[n] = rac{1}{2\pi} \int_{2\pi} X(\omega) e^{j\omega n} d\omega$$
 (29)

• Note that $X(\omega)$ is continuous and periodic (with a period of 2π , i.e., $X(\omega) = X(\omega + 2\pi)$).

Discrete-time Fourier Transform...

Example

Find the DTFT of the following discrete-time signal



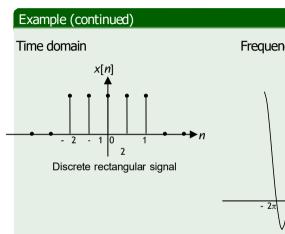
$$X(\omega) = \sum_{-\infty}^{\infty} x[n]e^{-j\omega n} = \sum_{n=-2}^{2} e^{-j\omega n}$$

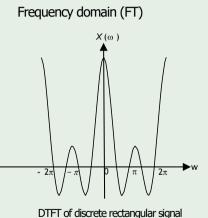
$$= e^{-j2\omega} + e^{-j\omega} + 1 + e^{j2\omega} + e^{j\omega}$$
(30)

$$= e^{j2\omega} \left(1 + e^{-j\omega} + e^{-j2\omega} + e^{-j3\omega} + e^{-j4\omega} \right) \tag{32}$$

$$= e^{j2\omega} \left(\frac{1 - e^{-j5\omega}}{1 - e^{-j\omega}} \right) = \frac{e^{j\frac{5}{2}\omega} - e^{-j\frac{5}{2}\omega}}{e^{j\frac{1}{2}\omega} - e^{-j\frac{1}{2}\omega}} = \frac{\sin(\frac{5}{2}\omega)}{\sin(\frac{1}{2}\omega)}$$
(33)

Discrete-time Fourier Transform...





Properties of DTFT

Linearity: if

$$x[n] = \alpha_1 x_1[n] + \alpha_2 x_2[n]$$
 (34)

Then the FT of x(t) is

$$F(x[n]) = \alpha_1 F(x_1[n]) + \alpha_2 F(x_2[n])$$
(35)

• Time Shifting:

$$x[n-n_0] \stackrel{\mathcal{F}}{\leftrightarrow} e^{-j\omega n_0} X(\omega)$$
 (36)

Frequency shifting:

$$x[n]e^{j\omega_0 n} \overset{\mathcal{F}}{\leftrightarrow} X(\omega - \omega_0)$$
 (37)

This property is also known as modulation property.

Differentiation in frequency

$$\mathcal{F}^{-1}(j\frac{dX(\omega)}{d\omega}) = nx[n] \tag{38}$$

Summary of Frequency Analysis

Time	Periodic	Nonperiodic	
domain			
Continuous	Fourier series $x_n=rac{1}{T}\int_0^T x(t)e^{-j2\pi nf_0t}dt$ $x(t)=\sum_{n=-\infty}^\infty x_ne^{j2\pi nt}$ $x(t)$ has period of T $f_0=1/T$	Fourier transform $X(\omega)=\int_{-\infty}^{\infty}x(t)e^{-j\omega t}dt$ \mathbf{x} (\mathbf{t}) $=\frac{1}{2\pi}\!\!\int_{-\infty}^{\infty}X(\omega)e^{j\omega t}d\omega$	Nonperiodic
Discrete	DT Fourier series $X[k]=rac{1}{N}\sum_{n=0}^{N-1}x[n]e^{-jk\Omega_0n}$ $x[n]=\sum_{k=0}^{N-1}X[k]e^{jn\Omega_0k}$ $x[n]$ and $X[k]$ have period N $\Omega_0=rac{2\pi}{N}$	DT Fourier transform $X(\omega) = \sum_{-\infty}^{\infty} x[n]e^{-j\omega n}$ $x[n] = \frac{1}{2\pi} \int_{2\pi} X(\omega) e^{j\omega n} d\omega$ $X(\omega)$ has period 2π	Periodic
	Discrete	Continuous	Frequency domain