

MODULE - I

Information Theory and Coding Techniques

What is Information Theory? A tool to solve problems related to

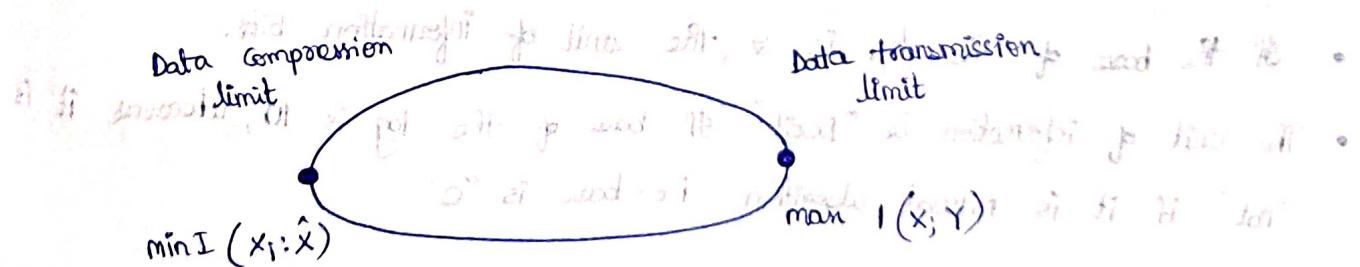
The movements of and transformations of information, just like those of a fluid, are constrained by mathematical and physical laws. These laws have deep connections with:

Probability theory, statistics, and combinatorics.

- 1) Probability theory, statistics, and combinatorics.
- 2) Thermodynamics (statistical physics).
- 3) Spectral Analysis, Fourier (and other) transforms.
- 4) Sampling theory, prediction, estimation theory.
- 5) Electrical engineering (bandwidth; signal-to-noise ratio).
- 6) Complexity theory (minimal description length).
- 7) Signal processing, representation, compressibility.

*. Scope of Information Theory:

- Information theory is concerned with fundamental limitations of communication.
- Entropy of the data, (H) is its compression limit of data compression.
- Ex: How many bits are required to represent music source.
- The ultimate limit of reliable communication over a noisy channel is the channel capacity (C), its rate limit.
- Ex: Bits sent in one second over telephone line.



Information theory are the extreme points of communication theory.

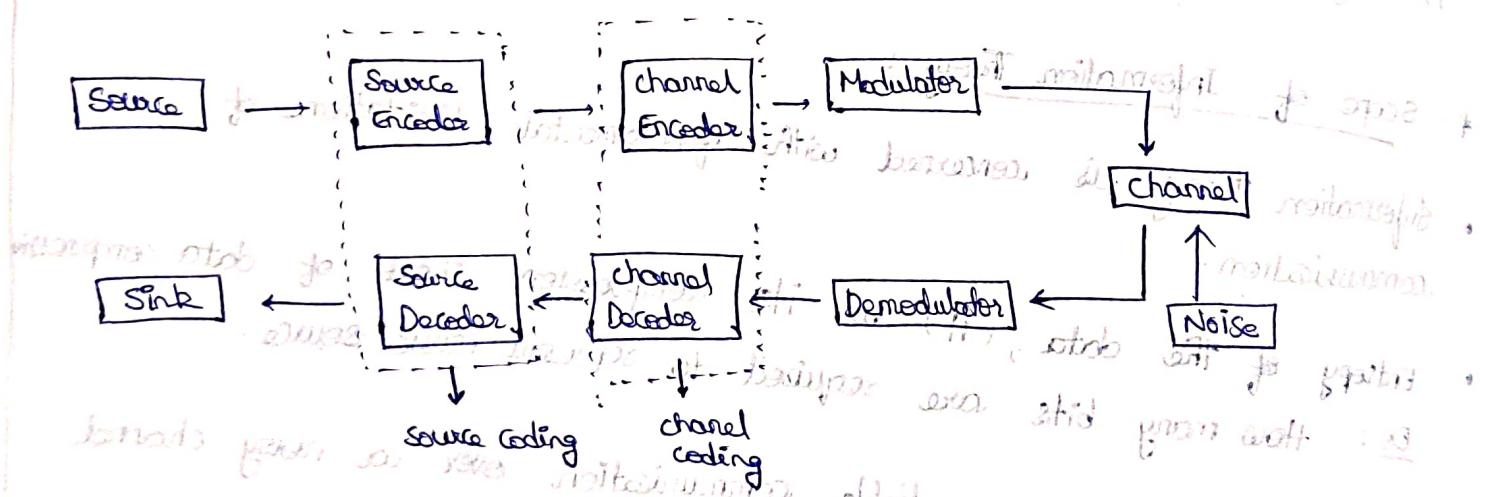
- The data compression minimum $I(X; \hat{X})$ lies at one end of the set of communication ideas. All data compression schemes require description rates at least equal to this minimum.
- At the other extreme is the data transmission maximum $I(X; X)$, known as the channel capacity. Thus all modulation schemes and data compression schemes lie between these limits.

* SCOPE OF CODING THEORY :

- Concerned with practical techniques to realize the limits specified by information theory.

1) Source Coding: converts source output to bits.

2) Channel Coding: adds extra bits to data transmitted over the channel. This redundancy helps combat the errors introduced in transmitted bits due to channel noise.



- Information is measured in terms of its probability.

$$I(X) = -\log_2 P(X)$$

- If the base of the log is 2 , the unit of information is bit.

- The unit of information is "Decit" if base of the log is 10 , whereas it is "nat" if it is natural logarithm i.e. base is "e".

What is the name of a string consisting of n bits?

Ans: sequence

What is entropy?

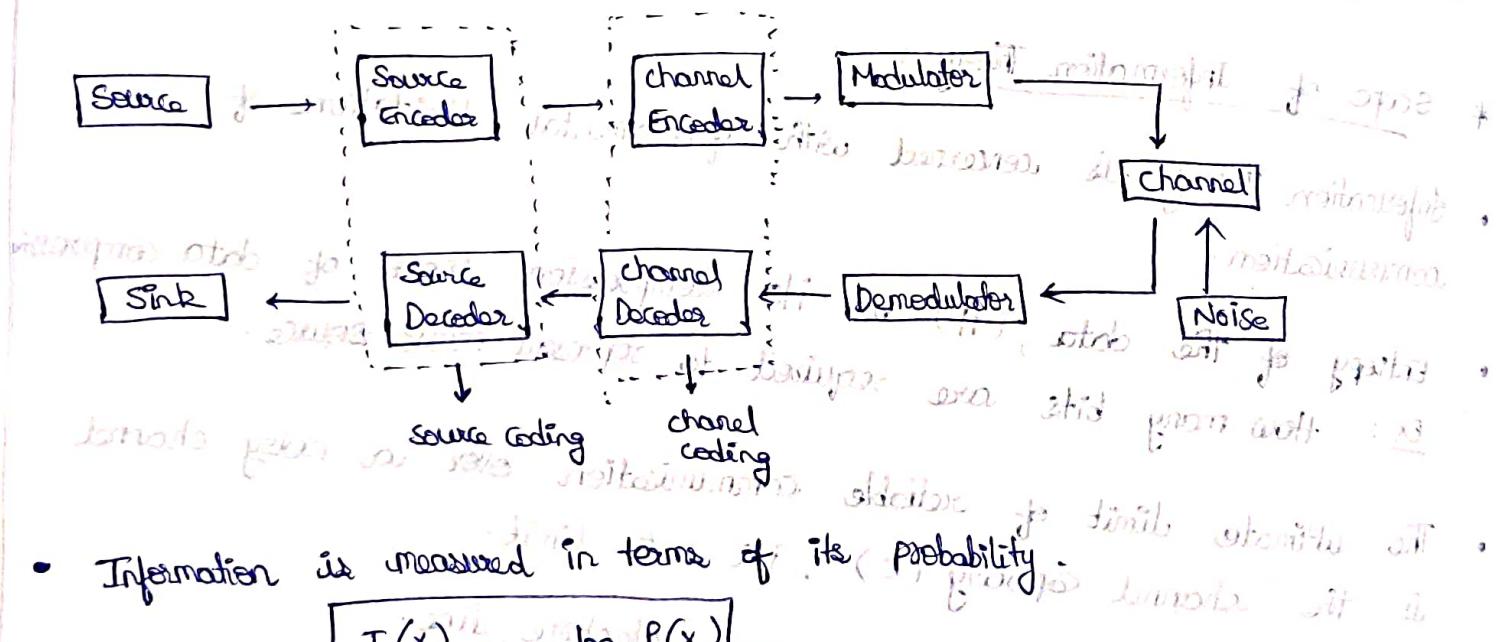
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(Ex: 2 bits)

* RANDOM VARIABLE :

- Random variables are variables that take on values determined by probability distributions. They may be discrete or continuous, in either their domain or their range.

$$(x_1) \text{ cat } (x_2) \text{ dog } \sim (x_3) \text{ car }$$

* ENTROPY :

- Measure of self Information.
- The entropy is a measure of the average uncertainty in the random variable.
- It is the no. of bits on average required to describe the random variable.

$$I(X) = \log_2 \left(\frac{1}{P_X} \right) = -\log_2 P_X$$

- Let X be a discrete random variable (DRV) with alphabet X and probability mass function $p(x) = P_{\{X=x\}}, x \in X$.
- The entropy $H(X)$ of a discrete random variable X is defined by:

$$H(X) = - \sum_{i=1}^n p(i) \log_2 p(x_i)$$

EXAMPLES :

* Find out the entropy of throwing a dice.

$$\Omega = \{\text{One Two Three}\} \text{ events}$$

$$1 = (\text{Absent}) \text{ events}$$

$$P(X) = \left\{ \frac{1}{6}, \frac{1}{6}, \frac{1}{6}, \frac{1}{6}, \frac{1}{6}, \frac{1}{6} \right\}$$

$$H(X) = - \sum_{i=1}^6 p_i \log_2 (p_i) = -\frac{1}{6} \log_2 \left(\frac{1}{6}\right) - \frac{1}{6} \log_2 \left(\frac{1}{6}\right) - \frac{1}{6} \log_2 \left(\frac{1}{6}\right)$$

$$= -\frac{1}{6} \log_2 \left(\frac{1}{6}\right) - \frac{1}{6} \log_2 \left(\frac{1}{6}\right) - \frac{1}{6} \log_2 \left(\frac{1}{6}\right).$$

$$= - \left[\frac{1}{6} \log_2 \left(\frac{1}{6}\right) \right] \times 6 = 2.584 \text{ bits}$$

* If source generates a symbol with their associated probabilities $\{ \frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{16}, \frac{1}{32} \}$. Calculate the entropy.

$$\text{Sol: } H(X) = - \sum_{i=1}^5 P(x_i) \log_2 P(x_i)$$

$$= \sum_{i=1}^5 P(x_i) \log_2 \left(\frac{1}{P(x_i)} \right)$$

$$= \frac{1}{2} \log_2 \left(\frac{1}{\frac{1}{2}} \right) + \frac{1}{4} \log_2 \left(\frac{1}{\frac{1}{4}} \right) + \frac{1}{8} \log_2 \left(\frac{1}{\frac{1}{8}} \right) + \frac{1}{16} \log_2 \left(\frac{1}{\frac{1}{16}} \right) + \frac{1}{32} \log_2 \left(\frac{1}{\frac{1}{32}} \right)$$

$$= \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 + \frac{1}{8} \log_2 8 + \frac{1}{16} \log_2 16 + \frac{1}{32} \log_2 32$$

$$= \frac{1}{2} + \frac{2}{4} + \frac{3}{8} + \frac{4}{16} + \frac{5}{32} = 1.7825 \text{ bits}$$

* Conclusion:

- More Probable — More (Certain) / less uncertain / less entropy
- less Probability — Less chance / More uncertain / More entropy
- More uncertainty — More Entropy
- less uncertainty — less Entropy
- Entropy (source sure event) = 0
- Entropy (impossible) = 1

* Properties of Entropy:

$$(1) H(X) = - \sum P_i \log_2 P_i \text{ for } \sum P_i = 1 \Rightarrow (ii) H(X) = - \sum P_i \log_2 \frac{1}{P_i} = (iii) H(X)$$

$$(ii) H(X) \geq 0; 0 \leq P(x) \leq 1 \Rightarrow \log \frac{1}{P(x)} \geq 0$$

$$(iii) H_b(x) = (\log_b a) H_a(x)$$

$$\text{Ex: } H_2(X) = 2 \cdot \left[\left(\frac{1}{2} \right) \log_2 \frac{1}{2} \right] =$$

- * Consider a random variable that has a uniform distribution over 32 outcomes. Determine its entropy?

$$H(X) = -\sum_{i=1}^{32} \frac{1}{32} \log_2 \frac{1}{32} = 32 \cdot \frac{1}{32} \log_2 32 = 5 \text{ bits/message}$$

Sol:

$$P(X_i) = \frac{1}{32}, \quad i = 1, 2, \dots, 32$$

$$H(X) = \left(\frac{1}{32} \times \log_2 \left(\frac{1}{1/32} \right) \right) \times 32 = \frac{32}{32} \log_2 32 = 5 \text{ bits/message}$$

If all messages are equally likely, then the entropy will be maximum.

Ex: $\frac{1}{16}$ + $(\frac{1}{16})^2$ + $(\frac{1}{16})^3$ + $(\frac{1}{16})^4$ + $(\frac{1}{16})^5$ + $(\frac{1}{16})^6$ + $(\frac{1}{16})^7$ = $\frac{1}{16} \left(1 - \frac{1}{16^8} \right) \approx \frac{1}{16}$ bit/message.

$$\approx \left[\frac{1}{16} \log_2 \left(\frac{1}{\frac{1}{16}} \right) \right] \times 16 = \log_2 2^4 = 4 \log_2 2$$

in horse

- *. The probability of winning for the eight horses are $\left(\frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{16}, \frac{1}{64}, \frac{1}{64}, \frac{1}{64}, \frac{1}{64}\right)$. Calculate the Entropy of the horse race.

$$\text{Sol: } = \frac{1}{2} \log_2 2 + \frac{1}{2} \log_2 4 + \frac{1}{2} \log_2 8 + 9 \log_2 16 + \left(4 \times \frac{1}{64} \log_2 64 \right)$$

$$= \frac{1}{2} + \frac{2}{2} + \frac{3}{2} + \frac{4}{2} + \frac{\cancel{4 \times 6}}{\cancel{64}} =$$

$$\text{and } P = \left(\frac{1}{r\left(\frac{1}{s}\right)}\right)_s \text{Gal} = (\text{augmented}) \text{ Gal}$$

- ## * Information Rate :

A diagram illustrating a source emitting messages. At the top left, there is a box labeled "Source". An arrow originates from this box and points to the right. To the right of the arrow, the text "messages/sec" is written vertically. Above the "Source" box, there is a horizontal line with three points labeled M_1 , M_2 , and M_3 below it. The letter "M" is positioned above the first point M_1 .

$$R = \gamma H \text{ bits/sec}$$

Ex: Let $X = \begin{cases} a & \text{with probability } \frac{1}{2} \\ b & \text{with probability } \frac{1}{4} \\ c & \text{with probability } \frac{1}{8} \\ d & \text{with probability } \frac{1}{8} \end{cases}$. The entropy of X is:

$$\text{Sol: } H(x) = -\frac{1}{2} \log \frac{1}{2} - \frac{1}{4} \log \frac{1}{4} - \frac{1}{8} \log \frac{1}{8} - \frac{1}{8} \log \frac{1}{8} = \frac{7}{4} \text{ bits}$$

* A discrete memory less source has six symbols with probabilities $P_A = \frac{1}{2}$, $P_B = \frac{1}{4}$, $P_C = \frac{1}{8}$, $P_D = P_E = \frac{1}{20}$, $P_F = \frac{1}{40}$. Determine: (i) The entropy of source (ii) The amount of information contained in message ABABBA. (iii) The amount of information contained in the message FDFFDF.

(i) The entropy of source

(ii) The amount of information contained in message ABABBA.

(iii) The amount of information contained in the message FDFFDF.

Sol:

$$H = - \sum_{i=1}^6 P_i \log P_i$$

$$= \frac{1}{2} \log_2 \left(\frac{1}{\frac{1}{2}} \right) + \frac{1}{4} \log_2 \left(\frac{1}{\frac{1}{4}} \right) + \frac{1}{8} \log_2 \left(\frac{1}{\frac{1}{8}} \right) + \frac{1}{20} \log_2 \left(\frac{1}{\frac{1}{20}} \right) + \frac{1}{40} \log_2 \left(\frac{1}{\frac{1}{40}} \right)$$

$$= \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 + \frac{1}{8} \log_2 8 + \frac{1}{20} \log_2 20 + \frac{1}{40} \log_2 40$$

$$= 1.94024 \text{ bits/symbol}$$

ii. $I(X) = \log_2 \left(\frac{1}{P(X)} \right)$

$$P(ABABBA) = P_A \times P_B \times P_A \times P_B \times P_B \times P_A$$

$$= \frac{1}{2} \times \frac{1}{4} \times \frac{1}{2} \times \frac{1}{4} \times \frac{1}{4} \times \frac{1}{2} = \left(\frac{1}{2} \right)^9$$

$$\therefore I(ABABBA) = \log_2 \left(\frac{1}{\left(\frac{1}{2} \right)^9} \right) = 9 \text{ bits}$$

iii. $P_F = \frac{1}{40}$; $P_D = \frac{1}{20}$

$$I(FDDFFD) = 28.93 \text{ bits}$$

$$H = \frac{1}{2} \log \frac{1}{2} + \frac{1}{8} \log \frac{1}{8} + \frac{1}{20} \log \frac{1}{20} + \frac{1}{40} \log \frac{1}{40} = (\times) H$$

* Entropy of Binary Memory less Source

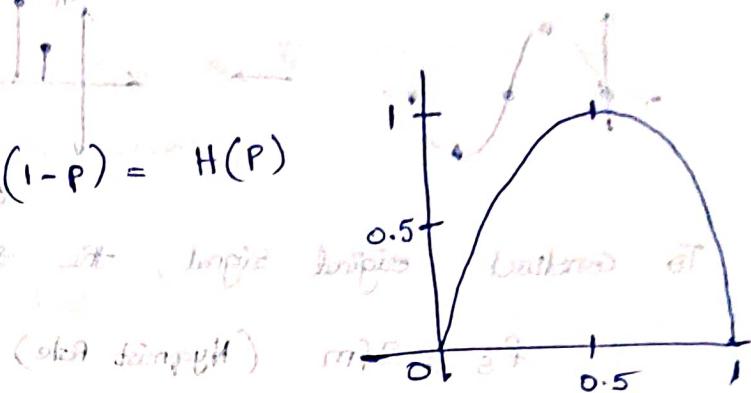
Let $X = \begin{cases} 1 & \text{with probability } p \\ 0 & \text{with probability } 1-p \end{cases}$

Sol:

The Entropy will be:

$$H(X) = -p \log p - (1-p) \log(1-p) = H(p)$$

$$H(X) = 1 \quad \text{when } p = \frac{1}{2}.$$



* An event has 6 possible outcomes with probabilities

$P_1 = \frac{1}{2}, P_2 = \frac{1}{4}, P_3 = \frac{1}{8}, P_4 = \frac{1}{16}, P_5 = \frac{1}{32}, P_6 = \frac{1}{32}$. Find the entropy of the system and the rate of information if there are 16 outcomes per second.

$$H = \left(\frac{1}{2} \log_2 \frac{1}{2} + \frac{1}{4} \log_2 \frac{1}{4} + \frac{1}{8} \log_2 \frac{1}{8} + \frac{1}{16} \log_2 \frac{1}{16} + \frac{1}{32} \log_2 \frac{1}{32} + \frac{1}{32} \log_2 \frac{1}{32} \right) = R$$

$$\text{Sol: } H = -\sum_{i=1}^{16} P_i \log_2 P_i$$

Need to find $R = ?$

Given $\tau = 16$ outcomes/second.

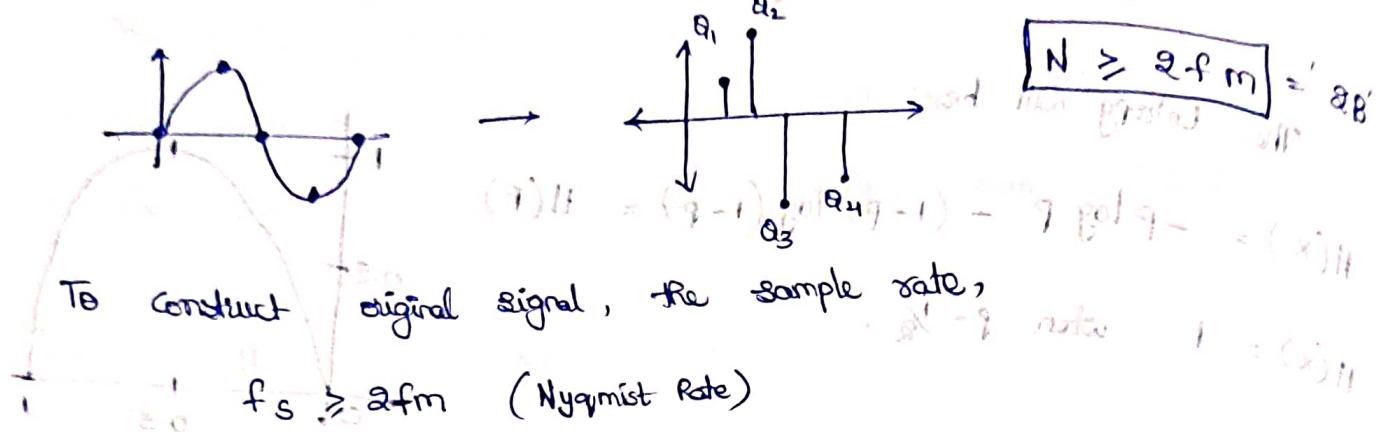
$$H = \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 + \frac{1}{8} \log_2 8 + \frac{1}{16} \log_2 16 + \frac{2}{32} \log_2 32$$

$$= \frac{1}{2} + \frac{2}{4} + \frac{3}{8} + \frac{4}{16} + \left(\frac{5}{32} + \frac{5}{32} \right) = \frac{16 + 16 + 12 + 8 + 10}{16 + 32 + 8 + 8} = \frac{62}{32} = \frac{62}{32} \text{ bits/ outcome}$$

$$R = \tau \cdot H = \frac{16 \text{ outcomes}}{\text{second}} \times \frac{62 \text{ bits}}{32 \text{ outcomes}} = 31 \text{ bits/second.}$$

* A analog signal is bandlimited to B Hz sampled at the Nyquist rate and the samples are quantized into 4 levels. The quantization levels Q_1, Q_2, Q_3, Q_4 (messages) are assumed independent and occur with probabilities $P_1 = P_4 = \frac{1}{8}$ and $P_2 = P_3 = \frac{3}{8}$.

- (a) Find the information rate per second if all the messages are equally likely.



$$f_S = 2B$$

(a)

$$i. P_1 = P_4 = \frac{1}{8}; P_2 = P_3 = \frac{3}{8}$$

$$H = P_1 \log_2\left(\frac{1}{P_1}\right) + P_2 \log_2\left(\frac{1}{P_2}\right) + P_3 \log_2\left(\frac{1}{P_3}\right) + P_4 \log_2\left(\frac{1}{P_4}\right)$$

$$= \sum_{i=1}^4 p_i \log_2 \left(\frac{1}{p_i} \right) = \left(\frac{1}{8} \log_2 8 \right) * 2 + \left(\frac{1}{8} \log_2 2 \right) * \left(\frac{3}{8} \log_2 \left(\frac{8}{3} \right) \right)$$

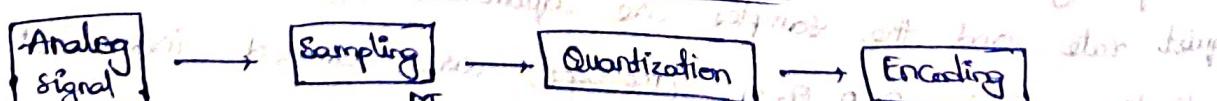
$$H = 1.81127 \text{ Bits/msg}$$

$$\sigma = \frac{1}{2} B \cdot \frac{\partial}{\partial t} \Rightarrow R = \sigma \cdot H = \frac{1}{2} B (1.81128) \approx 3.622 B \text{ bits/sec}$$

$$(b) \quad P_1 = P_2 = P_3 = P_4 = \frac{1}{Q}$$

$$H(X) = \sum_{i=1}^n p_i \log_2 \left(\frac{1}{p_i} \right) = 4 \times \frac{1}{2} \log_2 2 = 2 \text{ Bits/Msg}$$

$$R = \sigma \cdot H = (2B) \left(\frac{2}{2}\right) = '4B' \text{ bits/sec}$$



Q_1	2 0 0	Q_3	1 0 0
Q_2	0 1	Q_4	1 1

Consider a telegraph source having two symbols dot and dash. The dot duration is 0.2 sec and the dash duration is 3 times of the dot duration. The probability of occurrence of dot is twice than that of dash, and time between symbol is 0.2 seconds. Calculate information rate of the telegraph source. (Assume the string consists of 1200 symbols).

Sol: $P_{dot} = 2P_{dash}$ given $P_{dot} + P_{dash} = 1$
 $(I = H)$ since no other symbol is there I = H.

$$P_{dash} = \frac{1}{3}$$

$$P_{dot} = 2\left(\frac{1}{3}\right) = \frac{2}{3}$$

$$H = \frac{1}{3} \log_2\left(\frac{1}{\frac{1}{3}}\right) + \frac{2}{3} \log_2\left(\frac{1}{\frac{2}{3}}\right) = 0.9183 \text{ bits/symbol} = 0.9183$$

$$R = I \cdot H$$

$$T_{dot} = 0.2 \text{ sec} ; T_{dash} = 0.6 \text{ sec}$$

Each string consists of 1200 symbols on the average the dots and dash will appear according to the probability,

$$\text{No. of dots} = \frac{2}{3} \times 1200 = 800$$

$$\text{No. of dash} = \frac{1}{3} \times 1200 = 400$$

Jointly

Total time required,

$$\begin{aligned}
 &= [(No. of dot) (dot duration)] + [(No. of dash) (dash duration)] + [(1200)(0.2)] \\
 &= [(800)(0.2)] + [(400)(0.6)] + [(1200)(0.2)] \\
 &= 640 \text{ sec}
 \end{aligned}$$

Average symbol rate, not present during propagation or transmission

so, average data rate is given as $1.875 \text{ symbols/sec}$

$$r = \frac{1200}{T} = \frac{1200}{640 \text{ sec}} = 1.875 \text{ symbols/sec}$$

Rate of Information, $R = rH$ = $(1.875)(0.9183) = 1.7218 \text{ bits/second}$

(Change 0.281 to ~~0.281~~ 0.281 bits with ~~0.281~~)

- To gain information is to lose uncertainty by the same amount, so I and H differ only in sign ($H = -I$)
- Entropy and Information have units of bits.

$$I_{m,n} = \log_2(P_m P_n) = \log_2 P_m + \log_2 P_n = I_m + I_n$$

$$= \left(\frac{1}{e^{0.0}}\right) \log_2 \frac{1}{e^{0.0}} + \left(\frac{1}{e^{0.0}}\right) \log_2 \frac{1}{e^{0.0}} = H$$

$$= 0.281 + 0.281 = 0.562 = H$$

$$= 0.281 \times 2 = 0.562 = H$$

$$= 0.281 \times \frac{2}{3} = 0.187 = H$$

$$= 0.281 \times \frac{1}{3} = 0.093 = H$$

$$\left[(0.281)(0.281) \right] + \left[(\text{entire data})(\text{data p. 0.0}) \right] + \left[(\text{entire data})(\text{data p. 0.1}) \right] + \left[(\text{entire data})(\text{data p. 0.2}) \right] =$$

$$= [0.281(0.281)] + [0.281(0.281)] + [0.281(0.281)] + [0.281(0.281)] =$$

$$= 0.281^4 = 0.065 = H$$

MODULE-2

(MODULE-2)

* Rule of Probability : Product Rule

$$P(A, B) = \text{Joint Probability of both } A \text{ and } B. \quad \text{If } A \& B \text{ are independent events.}$$

$$= P(A|B) \cdot P(B) \approx P(B|A) \cdot P(A)$$

$$P(A|B) = P(A) \quad \text{and} \quad P(B|A) = P(B) \Rightarrow P(A, B) = P(A) \cdot P(B)$$

Example: Consider two sets $A = \{1, 2, 3, 4, 5, 6\}$ and $B = \{1, 2, 3, 4, 5, 6\}$ representing outcomes of a pair of dice.

$$\text{Ex: If } A = \{1, 2, 3, 4, 5, 6\} \text{ and } B = \{1, 2, 3, 4, 5, 6\} \text{ consider outcome vector of obtained set no.}$$

$$P(A=3, B=4) = P(A=3|B=4) \cdot P(B=4)$$

$$= P(B=4|A=3) \cdot P(A=3)$$

* Rule of Probability : Sum Rule

Summation Method to calculate $\{X\} = \{1, 2, 3, 4, 5, 6\}$

If event A is conditionalized on any one of other events B , then the probability of A is the sum of its joint Probabilities with all B :

$$P(A) = \sum_B P(A, B) = \sum_B P(A|B) \cdot P(B)$$

Marginal
Probability

$$P(B) = \sum_A P(A, B) = \sum_A P(B|A) \cdot P(A)$$

* Rule of Probability : Baye's Theorem!

$$P(B|A) = \frac{P(A|B) \cdot P(B)}{P(A)}$$

It allows to reverse the conditioning of events and to compute $P(B|A)$ from the knowledge of $P(A)$ & $P(B)$.

$$(P(A) \cdot P(B)) + (P(A) \cdot P(B)) = P(A) + P(B) = P(A|B)$$

* . BAYES THEOREM :

$$P(B|A) = \frac{P(A|B) \cdot P(B)}{P(A)} = \frac{P(A,B)}{P(A)}$$

$$P(A|B) = \frac{P(B|A) \cdot P(A)}{P(B)} = \frac{P(A,B)}{(A) \cdot P(B)} = \frac{(A) \cdot (A|B)}{(A) \cdot (A|B)} = 1$$

- Entropy is the uncertainty of a single random variable.
- Conditional entropy $H(x|Y)$, is the entropy of a random variable conditional on the knowledge of another random variable.
- The reduction in uncertainty due to another random variable is called the mutual information.

Ex : $X = \{\text{Probability of Weather Condition}\}$

$Y = \{\text{Probability of raining today}\}$

* JOINT ENTROPY :

- The joint Entropy $H(X,Y)$ of a pair of discrete random variables (X,Y) with a joint distribution $P(X,Y)$ is defined as :

$$H(X,Y) = - \sum_{x \in X} \sum_{y \in Y} P(x,y) \log P(x,y)$$

$$H(X,Y) = \sum_{x,y} P(x,y) \cdot \log \frac{1}{P(x,y)}$$

- Joint Entropy is the additive if X and Y are independent random variables.

$$H(X,Y) = H(X) + H(Y) \text{ iff } P(X,Y) = P(X) \cdot P(Y)$$

* Conditional Entropy of an ensemble X

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- Measures the uncertainty remaining about random variable X after specifying that random variable Y has taken on particular value $y = b_j$.

$$H(X|y=b_j) = \sum_x P(x|y=b_j) \log_2 \frac{1}{P(x|y=b_j)}$$

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Information und Kommunikation

* Conditional Entropy of an ensemble X, given an ensemble Y

$$H(X|Y) = \sum_y P(y) \left[\sum_x P(x|y) \log_2 \frac{1}{P(x|y)} \right]$$

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$$(a) H(X|Y) = \sum_{x,y} P(x,y) \log_2 \frac{1}{P(x|y)} * (Y;X) I$$

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(Y;X) I remains about X.

- This measures the average uncertainty that

when Y is known.

$$H(X|Y) = \sum_{x,y} P(x,y) \cdot \log_2 \frac{1}{P(x|y)} * \text{durchschnittliche Unschärfe}$$

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- Uncertainty of Y given X: $(X;Y) I$ = $(Y;X) I$ conditional uncertainty of

Y given the event $x = x$, averaged over the possible values x of X.

$$H(Y|X) = \sum_{x \in X} P(x) \cdot H(Y|x)$$

(X) H = (X;X) I

$$= - \sum_{x \in X} P(x) \sum_{y \in Y} P(y|x) \log_2 P(y|x)$$

$$= - \sum_{x \in X} \sum_{y \in Y} P(x,y) \cdot \log_2 P(y|x)$$

$$* (X;Y) I = (X) H - (X|Y) H = (Y;X) I$$

$$* \boxed{(X;Y) I = E \log_2 P(Y|X) + (X) H = (Y;X) I}$$

* Mutual Information:

- The mutual information between two random variables is the amount of information that one conveys about the other.
- How much information does the random variable Y give about the random variable X ?
The answer would be the amount by which Y reduces the uncertainty about X , namely:

$$H(X) - H(X|Y)$$

- The mutual information between the discrete random variable X is

is the quantity:

$$I(X;Y) = \boxed{H(X) - H(X|Y)}$$

$$I(X;Y) = \sum_{x,y} P(x,y) \log_2 \frac{P(x,y)}{P(x)P(y)}$$

- If X & Y are independent random variables then the mutual information equals zero.

$$I(X;Y) = I(Y;X)$$

- Non-negativity: Mutual Information is always greater than zero.

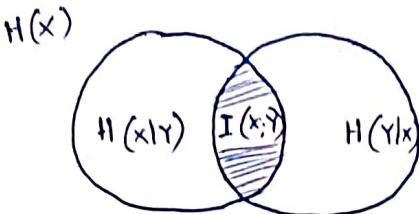
$$I(X;X) = H(X)$$

$$I(Y;Y) = H(Y)$$

$$I(X;Y) = H(X) - H(X|Y)$$

$$I(X;Y) = H(Y) - H(Y|X) = I(Y;X)$$

$$\boxed{I(X;Y) = H(X) + H(Y) - H(X,Y)} *$$



$H(X)$ $H(Y)$

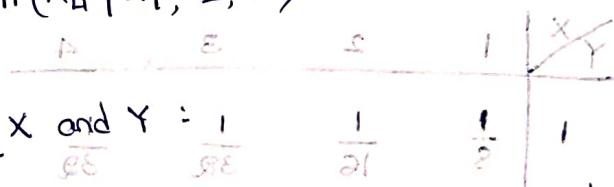
$$H(X) + H(Y) = H(X \cup Y) = H(X) + H(Y|X)$$

$$H(X, Y) = H(X) + H(Y|X)$$

*. chain Rule of Uncertainty :

- It states the uncertainty of random variable vector equals to the uncertainty of its first component plus the uncertainty of its second component when the first is known, ..., plus the uncertainty of its last component when all previous components are known".

$$H(X_1, X_2, X_3, \dots, X_n) = H(X_1) + H(X_2 | X_1) + H(X_3 | X_1, X_2) + \dots + H(X_n | X_1, X_2, \dots, X_{n-1})$$



*. Distance $D(X; Y)$ between X and Y :

- The amount by which the joint entropy of two random variable exceed their mutual information is a measure of the "distance" between them:

$$D(X; Y) = H(X, Y) - I(X; Y)$$

$\frac{1}{2}$	$\frac{1}{2}$	$\frac{1}{2}$	$\frac{1}{2}$	$\frac{1}{2}$
$\frac{1}{2}$	$\frac{1}{2}$	$\frac{1}{2}$	$\frac{1}{2}$	$\frac{1}{2}$

- Note that this quantity satisfies the standard axioms for a distance
- $D(X; Y) = D(Y; X)$
- $D(X; Y) \geq 0$, $D(X; X) = 0$, $D(X; Y) = 0 \Leftrightarrow X \perp\!\!\!\perp Y$
- $D(X; Z) = D(X; Y) + D(Y; Z)$.

*. Kullback - Leibler distance :

$$D_{KL}(P||Q) = \sum_x P(x) \log_2 \frac{P(x)}{Q(x)}$$

$$(X)H(X)$$

$$(Y)H(Y)$$

$$(X|Y)H(X|Y)$$

$$(Y|X)H(Y|X)$$

Summary :

$$1) H(x) = - \sum_{x=x} P(x) \log_2 P(x)$$

$$2) H(x,y) = - \sum_{x=x} \sum_{y=y} P(x,y) \cdot \log_2 P(x,y)$$

$$3) H(x|y) = \sum_{x,y} P(x,y) \log \frac{1}{P(x|y)}$$

: joint probability of x and y

$$4) I(x;y) = \sum_{x,y} P(x,y) \cdot \log_2 \frac{P(x,y)}{P(x)P(y)}$$

: if joint probability of x and y is same as product of individual probabilities of x and y, then information is zero.

Joint and conditional entropy has many properties similar to those of probability distributions.

EXAMPLES :

Let (X,Y) have the following joint distribution.

$X \setminus Y$	1	2	3	4
1	$\frac{1}{8}$	$\frac{1}{16}$	$\frac{1}{32}$	$\frac{1}{32}$
2	$\frac{1}{16}$	$\frac{1}{8}$	$\frac{1}{32}$	$\frac{1}{32}$
3	$\frac{1}{16}$	$\frac{1}{16}$	$\frac{1}{16}$	$\frac{1}{16}$
4	$\frac{1}{4}$	0	0	0

The marginal distribution of X is $(\frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{8})$ and the marginal distribution of Y is $(\frac{1}{4}, \frac{1}{4}, \frac{1}{4}, \frac{1}{4})$. Calculate the following:

a) $H(x)$

b) $H(y)$

c) $H(x|y)$

d) $H(y|x)$

e) $H(x,y)$

$$H(X) = \sum_{i=1}^4 P_i \log_2 \left(\frac{1}{P_i} \right) \cdot (x_i) \text{ H}(x_i)$$

$$P(x) = \left\{ \frac{1}{2}, \frac{1}{4}, \frac{1}{8}, \frac{1}{8} \right\}$$

$$\begin{aligned} a) H(X) &= \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 + \frac{1}{8} \log_2 8 + \frac{1}{8} \log_2 8 \\ &= \frac{1}{4} \text{ bits/symbols} \end{aligned}$$

$$\begin{aligned} b) H(Y) &= \sum_{j=1}^4 P_j \log_2 \left(\frac{1}{P_j} \right) \\ &= \left(\frac{1}{2} \log_2 \frac{1}{2} \right) + \left(\frac{1}{4} \log_2 \frac{1}{4} \right) + \left(\frac{1}{8} \log_2 \frac{1}{8} \right) + \left(\frac{1}{8} \log_2 \frac{1}{8} \right) = H(Y) \text{ H} \\ P(Y) &= \left\{ \frac{1}{4}, \frac{1}{4}, \frac{1}{4}, \frac{1}{4} \right\} \\ &= \left(\frac{1}{2} \log_2 \frac{1}{2} \right) + \left(\frac{1}{2} \log_2 \frac{1}{2} \right) + \left(\frac{1}{2} \log_2 \frac{1}{2} \right) + \left(\frac{1}{2} \log_2 \frac{1}{2} \right) \\ H(Y) &= \frac{1}{4} \log_2 4 + \frac{1}{4} \log_2 4 + \frac{1}{4} \log_2 4 + \frac{1}{4} \log_2 4 \\ &= \left(\frac{1}{2} \log_2 \frac{1}{2} \right) + \dots \\ &= 4 \times \frac{1}{4} \log_2 4 = 2 \text{ bits/symbols} \end{aligned}$$

$$\begin{aligned} c) H(X, Y) &= \text{Joint Entropy} = ? \\ &= \left(\frac{1}{2} \log_2 \frac{1}{2} \right) + \left(\frac{1}{2} \log_2 \frac{1}{2} \times 1 \right) + \left(\frac{1}{4} \log_2 \frac{1}{4} \times 2 \right) + \left(\frac{1}{8} \log_2 \frac{1}{8} \times 2 \right) = \\ &\quad \cancel{H(X+Y)} \end{aligned}$$

$$H(X=1, Y=1) = \frac{1}{16}$$

$$H(X=3, Y=1) = \frac{1}{16} \text{ bits} = \frac{1}{3}$$

$$H(X=1, Y=3) = \frac{1}{32}$$

$$H(X=3, Y=3) = -\frac{1}{16} (Y|X) \text{ H} = 0$$

$$H(X=1, Y=4) = \frac{1}{32}$$

$$H(X=3, Y=4) = \frac{1}{16} (X|Y) \text{ H}$$

$$H(X=2, Y=1) = \frac{1}{16}$$

$$H(X=4, Y=1) = \frac{1}{4} (Y|X) \text{ H}$$

$$H(X=2, Y=2) = \frac{1}{8}$$

$$H(X=3, Y=3) = \frac{1}{32}$$

$$H(X=4, Y=2) = H(X=4, Y=3) = 0$$

$$H(X=3, Y=4) = \frac{1}{32}$$

$$H(X=4, Y=4) = 0$$

$$H(X, Y) = \sum_{x_1} \sum_{y_1} P(x_1, y_1) \cdot \log_2 \left(\frac{1}{P(x_1, y_1)} \right)$$

$$= P(x_1, y_1) \log_2 \frac{1}{P(x_1, y_1)} + P(x_1, y_2) \cdot \log_2 \left(\frac{1}{P(x_1, y_2)} \right) +$$

$$P(x_1, y_3) \log_2 \frac{1}{P(x_1, y_3)} + P(x_1, y_4) \log_2 \frac{1}{P(x_1, y_4)} +$$

$$+ \dots + \dots + P(x_4, y_4) \log_2 \frac{1}{P(x_4, y_4)}$$

$$H(X, Y) = \left(\frac{1}{8} \log_2 8 \right) + \left(\frac{1}{16} \log_2 16 \right) + \left(\frac{1}{32} \log_2 32 \right) + \left(\frac{1}{32} \log_2 32 \right)$$

$$+ \left(\frac{1}{16} \log_2 16 \right) + \left(\frac{1}{8} \log_2 8 \right) + \left(\frac{1}{32} \log_2 32 \right) + \left(\frac{1}{32} \log_2 32 \right)$$

$$+ \left(\frac{1}{16} \log_2 16 \right) + \left(\frac{1}{16} \log_2 16 \right) + \left(\frac{1}{16} \log_2 16 \right) + \left(\frac{1}{16} \log_2 16 \right)$$

$$+ \left(\frac{1}{4} \log_2 4 \right) + 0 + 0 + 0$$

$$= \left(2 \times \frac{1}{8} \log_2 8 \right) + \left(6 \times \frac{1}{16} \log_2 16 \right) + \left(4 \times \frac{1}{32} \log_2 32 \right) + \left(\frac{1}{4} \log_2 4 \right)$$

$$= \frac{27}{8} \text{ bits / symbol}$$

c) As $P(X, Y) = P(X) \cdot P(Y)$ \rightarrow Independent Events

$$\boxed{H(Y|X) = H(X, Y) - H(X)}$$

$$\boxed{H(X|Y) = H(X, Y) - H(Y)}$$

$$\text{Given } \therefore H(X) = \frac{7}{4} H$$

$$\frac{1}{27} \therefore H(Y) = \frac{2}{27} H$$

$$\frac{1}{8} \therefore H(X, Y) = \frac{27}{8} H$$

c) $H(X|Y) = \frac{27}{8} - \frac{7}{4} = \frac{11}{8} H$

$$\frac{1}{27} \therefore H(X, Y) = \frac{11}{8} H$$

d) $H(Y|X) = \frac{27}{8} - \frac{7}{4} = \frac{13}{8} H$

$$\frac{1}{27} \therefore H(X, Y) = \frac{13}{8} H$$

* Calculate the probability that if somebody is "tall" (meaning) taller than 6ft or whatever), that person must be male. Assume that the probability of being male is $P(M) = 0.5$ and so likewise (for) being female $P(F) = 0.5$. Suppose that 80% of males are T (i.e. tall), and that 6% of females are tall. Solve:

- If you knew that somebody is male, how much information do you gain (in bits) by learning that he is also tall?
- 1) How much do you gain by learning that a female is tall?
 - 2) Finally, how much information do you gain from learning that a tall person is female?

Sol: Given, $P(M) = 0.5$ $P(F) = 0.5$

$P(T|M) = 0.2$ $\therefore P(T|F) = 0.06$

1) Information Gain, $I(M|T) = ?$ [Tall Person is Male]

We know that,

$$P(M|T) = \frac{P(M, T)}{P(T)} = \frac{P(T|M) \cdot P(M)}{P(T)}$$

$$I(T|M) = -\log_2 P(T|M)$$

$$= -\log_2 (0.2)$$

Using sum & product rule:

$$P(T) = P(T|M) \cdot P(M) + P(T|F) \cdot P(F)$$

$$= (0.2)(0.5) + (0.06)(0.5) = 0.13$$

$$I(T|M) = 2.319 \text{ bits}$$

$$P(M|T) = \frac{(0.2)(0.5)}{0.13} = 0.7692 \approx 0.77 \approx 0.77$$

$$I(M|T) = -\log_2 P(M|T) = -\log_2 (0.77) \approx 0.3785 \text{ bits}$$

2) $I(T|F) = ?$

$P(T|F) = 0.06$ for following all links (Binary)

first step $I(T|F) = -\log_2 P(T|F) = -\log_2 (0.06) = 4.05889$ bits

3) $I(F|T) = ?$

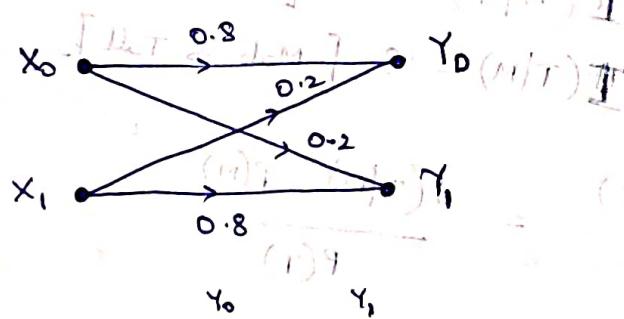
$$P(F|T) = \frac{P(F,T)}{P(T)} = \frac{P(T|F) \cdot P(F)}{P(T)}$$

Not so easy as top channel pt since we have to consider all channels

first point $\frac{(0.06)(0.5)}{0.13} = 0.23076$

$$I(F|T) = -\log_2 P(F|T) = -\log_2 (0.23076) = 2.115477 \text{ bits}$$

* calculate the resultant channel matrix for given Binary Symmetric channel [select or major link] $S = T(M) I$



sol:

$$P(Y|X) = \begin{bmatrix} x_0 & x_1 \end{bmatrix} \begin{bmatrix} 0.8 & 0.2 \\ 0.2 & 0.8 \end{bmatrix}$$

Assume, $[P(x_0) \cdot P(x_1)]^{\alpha} = [0.4 \cdot 0.6]^{\alpha} = 0.064^{\alpha}$

Now,

$$P(Y) = P(X) \cdot P(Y|X) = 0.064^{\alpha} \cdot 0.064^{\alpha} = 0.064^{2\alpha}$$

$$\text{And } [P(Y_0) \cdot P(Y_1)]^{\alpha} = [0.4 \cdot 0.6]^{\alpha} = \begin{bmatrix} 0.8 & 0.2 \\ 0.2 & 0.8 \end{bmatrix}^{\alpha} = \begin{bmatrix} 0.44 & 0.56 \\ 0.56 & 0.44 \end{bmatrix}^{\alpha}$$

Here,

$$I(X;Y) = \sum_{x,y} P(x,y) \log \frac{P(x,y)}{P(x)P(y)}$$

$$= P(x_0, y_0) \log \frac{P(x_0, y_0)}{P(x_0)P(y_0)} + P(x_0, y_1) \log \frac{P(x_0, y_1)}{P(x_0)P(y_1)}$$

$$+ P(x_1, y_0) \log \frac{P(x_1, y_0)}{P(x_1)P(y_0)} + P(x_1, y_1) \log \frac{P(x_1, y_1)}{P(x_1)P(y_1)}$$

First we will have to calculate Joint Probability Matrix:

$$P(Y|X) = \begin{bmatrix} 0.8 & 0.2 \\ 0.2 & 0.8 \end{bmatrix}; \quad P(X) = [P(x_0) \quad P(x_1)] = [0.4 \quad 0.6]$$

$$P(Y_0|x_0) = 0.8 \quad P(Y_1|x_0) = 0.2 \quad P(Y_0|x_1) = 0.2 \quad P(Y_1|x_1) = 0.6$$

$$P(x_0) = 0.4 \quad P(x_1) = 0.6$$

So, Joint Probabilities are:

$$P(x_0, y_0) = P(Y_0|x_0) \cdot P(x_0) = (0.8)(0.4) = 0.32$$

$$P(x_0, y_1) = P(Y_1|x_0) \cdot P(x_0) = (0.2)(0.4) = 0.08$$

$$P(x_1, y_0) = P(Y_0|x_1) \cdot P(x_1) = (0.2)(0.6) = 0.12$$

$$P(x_1, y_1) = P(Y_1|x_1) \cdot P(x_1) = (0.8)(0.6) = 0.48$$

Hence,

$$P(X,Y) = \begin{bmatrix} 0.32 & 0.08 \\ 0.12 & 0.48 \end{bmatrix} = (X|Y)$$

Now,

$$I(X;Y) = \sum_{x,y} P(x,y) \log \frac{P(x,y)}{P(x)P(y)} = (X|S)$$

$$I(X;Y) = \sum_{x,y} P(x,y) \log \frac{P(x,y)}{P(x)P(y)} = (X|S)$$

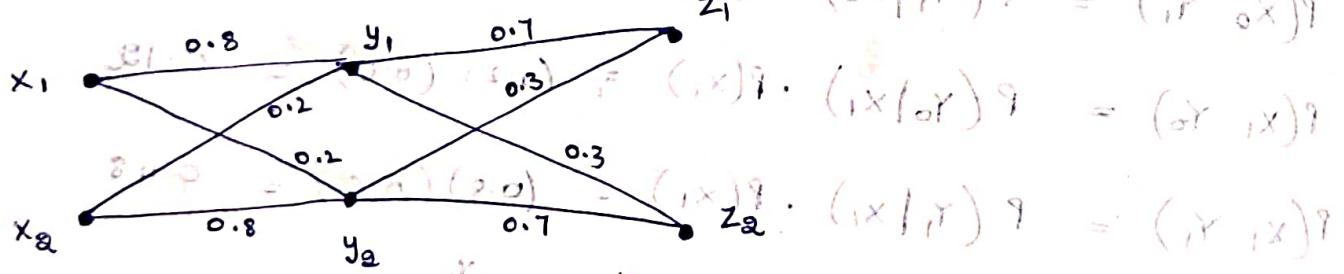
$$\begin{bmatrix} 0.32 & 0.08 \\ 0.12 & 0.48 \end{bmatrix} = (X|Y)$$

And $P(Y_0) = 0.44$

$P(Y_1) = 0.56$

$$\begin{aligned}
 &= P(x_0, y_0) \log \frac{P(x_0, y_0)}{P(x_0)P(y_0)} + P(x_0, y_1) \log \frac{P(x_0, y_1)}{P(x_0)P(y_1)} \\
 &\quad P(x_1, y_0) \log \frac{P(x_1, y_0)}{P(x_1)P(y_0)} + P(x_1, y_1) \log \frac{P(x_1, y_1)}{P(x_1)P(y_1)} \\
 &= (0.32) \log \frac{0.32}{(0.4)(0.44)} + (0.08) \log \frac{0.08}{(0.4)(0.56)} \\
 &\quad (0.12) \log \frac{0.12}{(0.6)(0.44)} + (0.48) \log \frac{0.48}{(0.6)(0.56)} \\
 &\stackrel{\text{Hence}}{=} [(x)_1 (y)_1] = (x,y)_1 \\
 &\stackrel{\text{Hence}}{=} [(x)_1 (y)_1] = (x,y)_1 \\
 &\stackrel{\text{Hence}}{=} (0.275998) + (-0.118834) + (-0.136500) + (0.246995) \\
 &= \underline{\underline{0.2676595 \text{ bits/symbol}}}
 \end{aligned}$$

* Find $P(z_1)$ and $P(z_2)$ if $P(x_1) = 0.6$ and $P(x_2) = 0.4$ for the given cascaded channel.



$$\begin{aligned}
 P(y) &= P(x) \cdot P(Y|x) \\
 P(z) &= P(y) \cdot P(z|Y)
 \end{aligned}$$

$$(a) P(z|x) = P(Y|x) \cdot P(z|Y)$$

$$P(z|Y) = \begin{bmatrix} 0.7 & 0.3 \\ 0.3 & 0.7 \end{bmatrix} \quad \text{and} \quad P(Y|x) = \begin{bmatrix} 0.8 & 0.2 \\ 0.2 & 0.8 \end{bmatrix}$$

Now, resultant channel matrix

$$P(z|x) = P(Y|x) \cdot P(z|Y)$$

$$= \begin{bmatrix} 0.8 & 0.2 \\ 0.2 & 0.8 \end{bmatrix} \left\{ \begin{bmatrix} 0.7 & 0.3 \\ 0.3 & 0.7 \end{bmatrix} \right\} = \begin{bmatrix} 0.62 & 0.38 \\ 0.38 & 0.62 \end{bmatrix} = (2)H$$

$$P(z_1) \text{ and } P(z_2) = ?$$

$$P(z) = P(x) \cdot P(z|x)$$

$$= \begin{bmatrix} 0.6 & 0.4 \end{bmatrix} \begin{bmatrix} 0.62 & 0.38 \\ 0.38 & 0.62 \end{bmatrix}$$

$$P(z) = [0.524 \quad 0.476] \approx [P(z_1) \quad P(z_2)] = (2)H$$

$$\therefore P(z_1) = 0.524 + \frac{1}{2} \cdot 0.476 = 0.524 + 0.476 = 1$$

$$\frac{1}{2} + \frac{1}{2} + \frac{1}{2} + \frac{1}{2} = 1$$

independent, then

where, $n = \text{order of extension}$

Note: If the symbols are statistically

$$H(s^n) = n \cdot H(s)$$

Proof

* Extension of Discrete Memoryless channel:

Given source, $S = \{s_0, s_1, s_2\}$

$$P(S) = \left\{ \frac{1}{4}, \frac{1}{2}, \frac{1}{4} \right\}$$

$$\left(\frac{1}{4} \right) H + \left(\frac{1}{2} \right) H + \left(\frac{1}{4} \right) H = (2)H$$

Extension of source by order 2 i.e. $\frac{1}{4} + \frac{1}{2} + \frac{1}{4}$

$$S^2 = \{s_0, s_1, s_2\} \cdot \{s_0, s_1, s_2\}$$

$$H(S^2) = ? \Rightarrow S^2 = \frac{1}{4} + \frac{1}{2} + \frac{1}{4} = (2)H$$

$$S^2 = \{s_0s_0, s_0s_1, s_0s_2, s_1s_0, s_1s_1, s_1s_2, s_2s_0, s_2s_1, s_2s_2\}$$

$$(and) \frac{1}{4} + \frac{1}{2} + \frac{1}{4} = (2)H$$

$$P(S) = \left\{ \frac{1}{4}, \frac{1}{2}, \frac{1}{4} \right\}$$

$$(1/2)^2, (1/2)^2, (1/2)^2$$

$$P(S^2) = \left\{ \frac{1}{4}, \frac{1}{2}, \frac{1}{4} \right\}, \left\{ \frac{1}{4}, \frac{1}{2}, \frac{1}{4} \right\} \left[\begin{array}{cc} S^2 & S \\ S & S^2 \end{array} \right]$$

$$= \left\{ \frac{1}{4} \times \frac{1}{4}, \frac{1}{4} \times \frac{1}{2}, \frac{1}{4} \times \frac{1}{4}, \frac{1}{2} \times \frac{1}{4}, \frac{1}{2} \times \frac{1}{2}, \frac{1}{2} \times \frac{1}{4}, \frac{1}{4} \times \frac{1}{4}, \frac{1}{4} \times \frac{1}{2} \right\} = \left\{ \frac{1}{16}, \frac{1}{8}, \frac{1}{16}, \frac{1}{16}, \frac{1}{4}, \frac{1}{8}, \frac{1}{16}, \frac{1}{16} \right\}$$

$$H(S^2) = \sum_{i=1}^2 P_i \log_2 \left(\frac{1}{P_i} \right) \quad (2^2)$$

$$= \frac{1}{16} \log_2 16 + \frac{1}{8} \log_2 8 + \frac{1}{16} \log_2 16 + \frac{1}{8} \log_2 8 + \frac{1}{4} \log_2 4$$

$$+ \frac{1}{8} \log_2 8 + \frac{1}{16} \log_2 16 + \frac{1}{8} \log_2 8 + \frac{1}{16} \log_2 16$$

$$= \frac{4}{16} + \frac{3}{8} + \frac{4}{16} + \frac{1}{16} + \frac{3}{8} + \frac{2}{4} + \frac{3}{8} + \frac{4}{16} + \frac{3}{8} + \frac{4}{16}$$

$$= 4\left(\frac{4}{16}\right) + 4\left(\frac{3}{8}\right) + \frac{2}{4}$$

$$H(S^2) = 3$$

$$H(S) = \sum_{i=1}^2 P_i \log_2 \left(\frac{1}{P_i} \right)$$

$$\{1/2, 1/2, 1/2\} = 2$$

$$\left\{ \frac{1}{4}, \frac{1}{2}, \frac{1}{4} \right\} \in (2)^2$$

$$= \frac{1}{4} \log_2 4 + \frac{1}{2} \log_2 2 + \frac{1}{4} \log_2 4 = \frac{2}{4} + \frac{1}{2} + \frac{2}{4} = \frac{2+2+2}{4}$$

$$H(S) = \frac{6}{4} = \frac{3}{2}$$

$$\Rightarrow H(S^2) = 2 \cdot H(S)$$

$$3 = 2 \times \frac{3}{2} \quad (\text{True})$$

$$H(S^2) = 3$$

$$H(S) = 3/2$$

MODULE - 3 : LOSSLESS COMPRESSION

Information transmission over a discrete channel.

- * Rate of information transmission gives average amount of information going into channel:

$$H(x) = \sum_{i=1}^n p_i \log_2 \left(\frac{1}{p_i} \right)$$

- If the source generates 'r' symbols per second. Then the rate of information, $R = r \cdot H$ bits/sec

$$D_{in} = r \cdot H(x)$$

$$\frac{1}{S} = (x)$$

$$\frac{1}{C} = (y)$$

transmissional (and) thus

- Errors are introduced in the channel during transmission.

$$\text{Information Loss} = H(x|Y)$$

$$\therefore \text{Transmitted Information} = H(x) - H(x|Y)$$

- The average rate of information across the channel,

$$\therefore D_t = [H(x) - H(x|Y)] \cdot r \quad \text{bits/sec}$$

- Case-i: When noise is too large, then x and y are statistically independent.

$$H(x|Y) = H(x)$$

As $H(x|Y) = H(x)$; no information is transmitted over the channel

$$\therefore H(x|Y) = H(x)$$

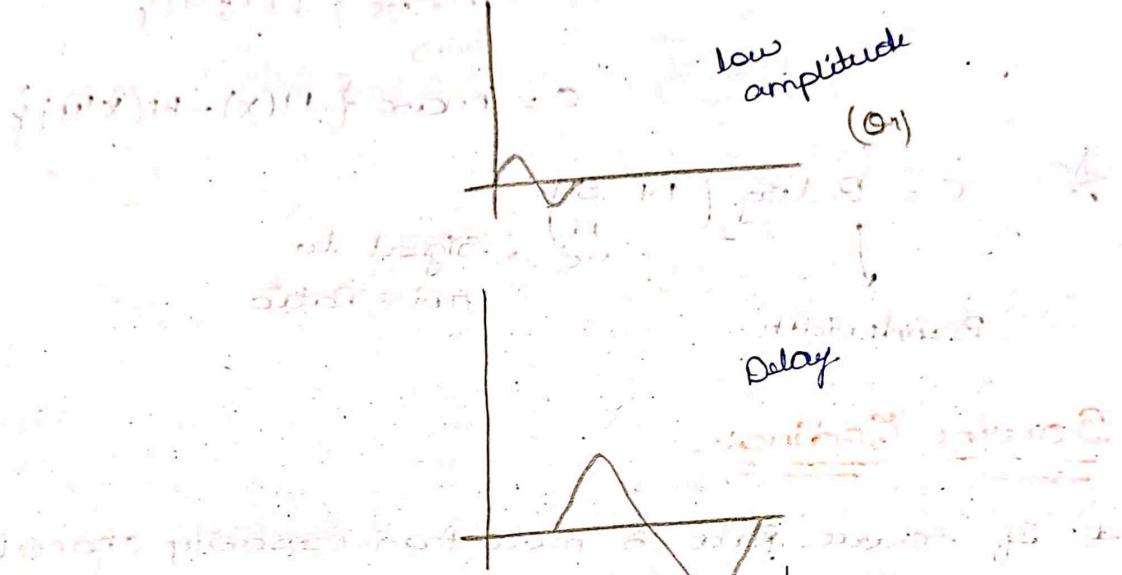
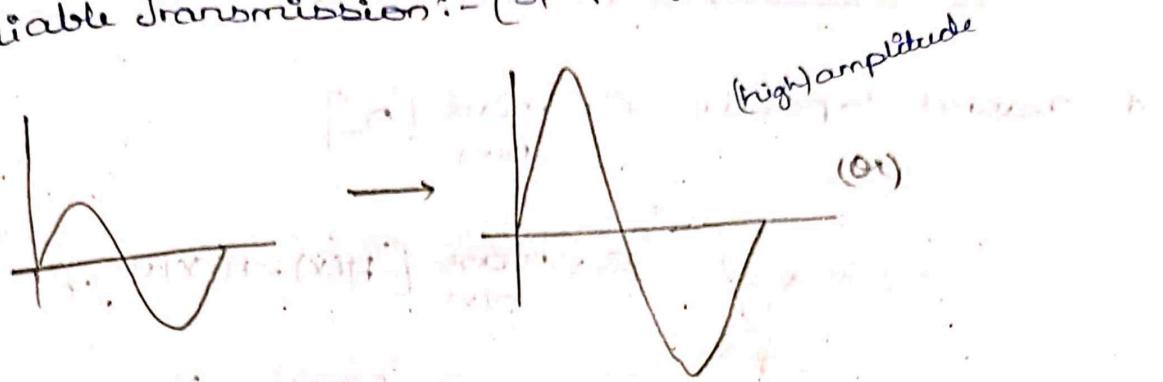
- Case-ii: Errorless Transmission, No information loss

$$D_t = D_{in}$$

$$H(x|Y) = 0$$

Channel Capacity: (Highest Info Rate)

Reliable Transmission:-(shape of signal should be preserved)

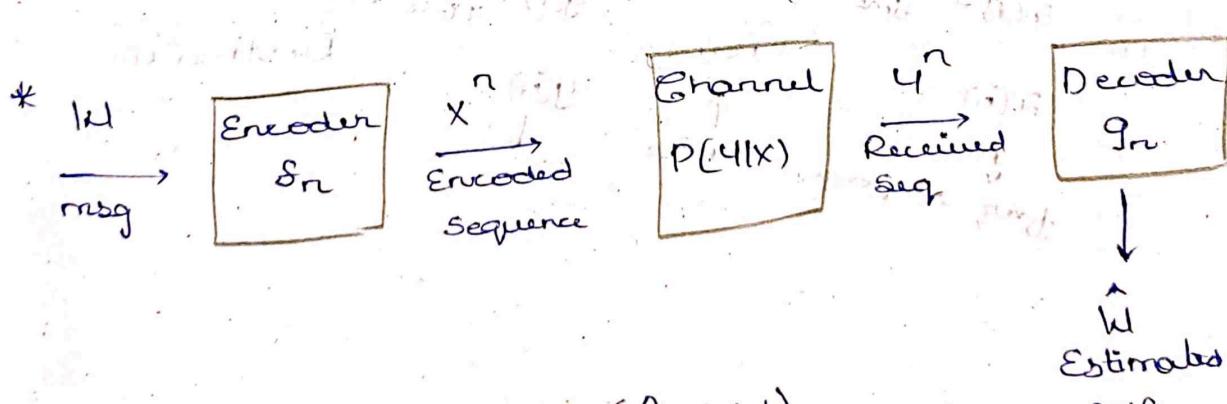


but it should not be like



(Distortion).

* Capacity of channel = max of Mutual Info b/w
msg & O/P of channel



Probability
of error

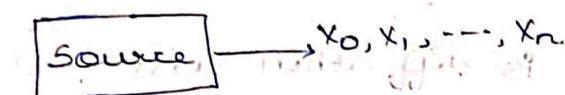
$$P_e = \Pr(\hat{w} \neq w).$$

Types of source coding:-

(i) lossless:- No loss of Info.

(ii) lossy:- There is loss of Info.

Discrete memory less source



$$X = \{x_0, x_1, x_2, \dots, x_n\}$$

$$P(x_0) P(x_1) \dots P(x_n)$$

$$R = rH$$

Entropy

Easiest way of source coding :-

* symbols	prob	codeword	length	Equal length code. So simply write
A	0.5	0 0	2	00, 01, 10, 11
B	0.25	0 1	2	
C	0.125	1 0	2	
D	0.125	1 1	2	

$$\text{Avg length} = \sum 2 \text{ bits/message}$$

* for variable length codeword

[another way]

sym	prob	code	length	codeword
A	0.5	0	1	0
B	0.25	No logic	1	10
C	0.125	simply write	2	110
D	0.125	1 1	2	111

For sample
sym diff
codeword

$$\text{Avg length} = \sum P_i n_i$$

length of
prob particular sym

Entropy of source coding

$$H = \sum P_i \log \frac{1}{P_i}$$

$$\text{Efficiency} = \frac{L_{\text{actual}}}{H} = \eta$$

Properties:-

* Non-Singular:-

$$\text{If } x_i \neq x_j \Rightarrow c(x_i) \neq c(x_j)$$

means if msg is diff then codeword must be diff.

* Uniquely Decodable:-

$$x_1 x_2 \dots x_m \neq x'_1 x'_2 \dots x'_m$$

Ex: $\{0, 010, 01, 10\}$

msg = 01010.

$$\boxed{01} \boxed{010} = \boxed{010} \boxed{10} = \boxed{01} \boxed{0} \boxed{10}$$

Eg: $\{00, 010, 11, 110\}$

* prefix code / Instantaneous code:-

→ No code word should be prefix of

another code word.

Ex: $\{10, 00, 11, 111\}$

Huffman Coding

① Huffman Code:-

Eg:-	Sym	prob	find the code word using Huffman coding
	s ₀	0.1	
	s ₁	0.2	
	s ₂	0.4	
	s ₃	0.2	
	s ₄	0.1	

$$H(S) = \sum_i P_i \log \frac{1}{P_i}$$

$$= 0.4 \log \frac{1}{0.4} + 2(0.2 \log \frac{1}{0.2}) + 2(0.1 \log \frac{1}{0.1})$$

$$\Rightarrow 2.12193$$

∴ Arrange all symbols in Decreasing order of prob
As high as possible (low variance method)

Sol:-

S _k	Stage I	Stage II	Stage III	Stage IV
S ₂	(Decreasing) 0.4	0.4	0.4	0.6
S ₁	0.2	0.2	0.3	0.4
S ₃	0.2	0.2	0.2	
S ₀	0.1	0.2		
S ₄	0.1			

(process:- add least 2 prob & keep as high as u can)

	Start coding	I	II	III	IV	Code word length
S ₂	0.4	0.4	0.4	0.4	0.6	2
S ₁	0.2	0.2	0.2	0.4	0.2	2
S ₃	0.2	0.2	0.2	0.2		3
S ₀	0.1	0.1	0.2	0.1		3
S ₄	0.1					

By tracing code word length

Final $\Rightarrow [00, 10, 11, 010, 011]$

by tracing path.

Average length of codeword :- ~~length of codeword~~

$$\boxed{\sum P_i L_i}$$

$$\Rightarrow (0.4)(5) + (0.2)(5) + (0.2)(5) + (0.1)(3) + (0.1)(3)$$

$$\Rightarrow \underline{2.2}.$$

Efficiency :-

$$\boxed{\eta = \frac{H(\alpha)}{L}}$$

$$= \frac{2.012}{2.2} = \underline{\underline{96\%}}$$

Redundancy :- $r = 1 - \eta = 1 - 0.96 = \underline{\underline{0.04}}$.

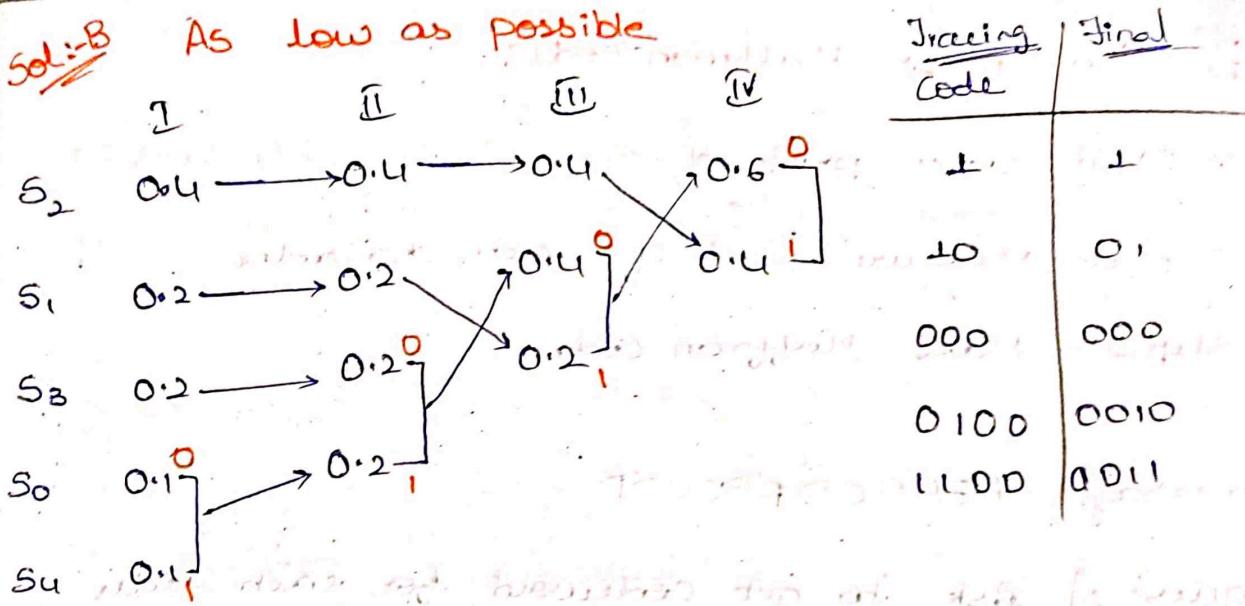
Codeword variance : $\sigma^2 = \sum_{k=1}^K P_k (L_k - \bar{L})^2$

$$\Rightarrow (0.4)(2-2.2)^2 + (0.2)(2-2.2)^2 + (0.2)(2-2.2)^2 + (0.1)(3-2.2)^2 + (0.1)(3-2.2)^2$$

$$\Rightarrow (0.4)(0.2)^2 + (0.2)(0.2)^2 + (0.2)(0.2)^2 + (0.1)(0.8)^2 + (0.1)(0.8)^2$$

$$\Rightarrow 0.016 + 0.008 + 0.008 + 0.064 + 0.064$$

$$\Rightarrow \boxed{0.16}$$



* Find efficiency, redundancy, Avg length for B-1

* More efficiency & low variance \Rightarrow performs well.

* If they didn't mention any method \Rightarrow

use method - I [As high as possible]

b/c we get low variance.

2] Two-pass Huffman Code:-

* Used when prob. of symbol is unknown.

Step-1:- Measure prob. of each character

Step-2:- Make Huffman code.

message: ABAAACDABCCDD-----

ques:-1] Ask to get codeword for each letter

ques-2] They may ask to write seq of letter

for the code like 110010010 \Rightarrow ABDA.

ques-3] Ask to Decode 00101110011 \Rightarrow BCADD

ques-4]

sym	prob
m_1	0.4
m_2	0.6

 Find out the Huffman code for Second-order Extension

\Rightarrow

$m_1 m_1$	0.4×0.4
$m_1 m_2$	0.4×0.6
$m_2 m_1$	0.6×0.4
$m_2 m_2$	0.6×0.6

 perform Huffman for this table

arrange the prob in decrease order & perform Huff

Eg: msg = ABA BABABA BAC CAD AB AC AD A B C A D
apply Huffman code (min variance method) to
find out codeword for "ABCDDCBA".

The parity check bits of $(8,4)$ linear block code are:

$$\begin{aligned}
 C_1 &= m_2 + m_3 + m_4 \\
 C_2 &= m_1 + m_3 + m_4 \\
 C_3 &= m_2 + m_3 + m_4 \\
 C_4 &= m_1 + m_2 + m_3
 \end{aligned}$$

where m_1, m_2, m_3, m_4 are the message bits?

where m_1, m_2, m_3 , determine the Generator & Parity check Matrix for systematic code.

- Determine the Generator & Parity

 - Determine the error detection & error correction capabilities of code.
 - Comment on error detection & error correction capabilities of code.
 - If the received sequence is $\begin{bmatrix} 1 & 0 & 1 & 0 \end{bmatrix}$, determine the transmitted word.

$$\begin{aligned}
 \text{Solut:} \\
 C_1 &= m_2 + m_3 + m_4 \\
 C_2 &= m_1 + m_3 + m_4 \\
 C_3 &= m_2 + m_3 + m_4 \\
 C_4 &= m_1 + m_2 + m_3
 \end{aligned}$$

Given 10110101 , determine the inverse.

$\text{GCD}(10110101, 1101) = 1$

$P^{-1} = \begin{bmatrix} 0 & 1 & 0 & 1 \\ 1 & 0 & 1 & 1 \\ 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 0 \end{bmatrix}$

P = 8

$$k = 4$$

$$av = n - k = 8 - 4 = 4$$

Generator matrix, $G = [I_K \mid P_{K \times q}]$ $K \times n$ block matrix with $q=2^k$

$$G = \left[I_4 \mid P_{4 \times 4} \right]_{4 \times 8} = \left[\begin{array}{cccc|cccc} 1 & 0 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 1 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 & 1 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 0 & 1 & 0 & 0 \end{array} \right]$$

$$\text{iii. Parity check Matrix, } H = [P^T \mid I_{R_k}]$$

$$P = \begin{bmatrix} 0 & 1 & 0 & 1 \\ 1 & 0 & 1 & 1 \\ 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 0 \end{bmatrix}$$

$$P^T = \begin{bmatrix} E & \text{transf} \\ 0 & I \\ 1 & 0 \\ 0 & 1 \\ 1 & 1 \end{bmatrix}$$

$$H = \begin{bmatrix} 0 & 1 & 1 & 1 & 1 & 0 & 0 & 0 \\ 1 & 0 & 1 & 1 & 0 & 1 & 0 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 1 & 0 \\ 1 & 1 & 1 & 0 & 0 & 0 & 0 & 1 \end{bmatrix} \xrightarrow{\text{H}^T} H^T = \begin{bmatrix} 0 & 1 & 1 & 1 & 1 & 0 & 0 & 0 \\ 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 \\ 0 & 1 & 1 & 1 & 1 & 0 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 0 & 0 & 1 \end{bmatrix}$$

$$S = Y \cdot H^T$$

Received Codeword, $Y = 10110101$

$$S = [10110101]$$

$$\begin{bmatrix} 0 & 1 & 0 & 1 \\ 1 & 0 & 1 & 1 \\ 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}_{8 \times 8}$$

$$= \begin{bmatrix} 0 & 0 & 0 & 0 \end{bmatrix}_{1 \times 4}$$

~~error~~ in the $S \Rightarrow X = E \oplus Y$

Received Codeword has ~~error~~ error, The transmitted codeword is the received codeword. $\Rightarrow E = 10000000$

$$\frac{\text{Error detecting capability}}{\begin{bmatrix} 1 & 0 & 1 & 0 & 0 & 0 & 0 & 1 \\ 1 & 1 & 1 & 0 & 1 & 0 & 0 & 0 \end{bmatrix}} = \frac{w_i}{\text{no. of } 1's \text{ in codeword}}$$

$$w_i = \begin{bmatrix} 1 & 0 & 1 & 0 & 0 & 0 & 0 & 1 \\ 1 & 1 & 1 & 0 & 1 & 0 & 0 & 0 \end{bmatrix} = 3$$

No. of errors that can be detected

$$d_{\min} \geq s + 1$$

$$\text{Let } \begin{bmatrix} 1 & 0 & 1 & 0 & 0 & 0 & 0 & 1 \\ 1 & 1 & 1 & 0 & 1 & 0 & 0 & 0 \end{bmatrix} = H \quad (\text{extra code first})$$

$$d_{\min} = w_{\min} = 3$$

\therefore It can detect 2 errors at each codeword.

Correcting capability:

No. of errors that can be corrected:

$$d_{\min} \geq 2t + 1 \text{ i.e. } d_{\min} \text{ is odd number}$$

$$d_{\min} = w_{\min} = 3$$

$$1+2+3 = 6 \text{ (odd)}$$

①

∴ It can correct 1 error at each codeword.

$$\left[(9) \alpha, (1), \beta = (9) \alpha + \frac{\beta}{9} \right]$$

$$x = E \oplus Y$$

$$\begin{array}{r} x = 10100100001 \\ \text{not strand} \\ \hline 10110100 \end{array}$$

Correctly transmitted codeword = 10110100

$$1_1 = (9) \alpha$$

$$1_2 = (9) \beta$$

Weight(9) β

m_1	m_2	m_3	m_4	c_1	c_2	c_3	c_4	Weight(9) β
0	0	0	0	0	0	0	0	0
0	0	0	1	1	1	1	1	4
0	0	1	0	1	1	1	1	3
0	0	1	1	0	0	0	1	4
0	1	0	1	0	0	0	1	4
0	1	1	0	0	1	0	0	3
0	1	1	1	0	1	0	0	5
1	0	0	0	1	0	1	1	5
1	0	0	1	1	0	1	0	4
1	0	1	0	1	0	0	0	4
1	0	1	1	0	1	0	0	5
1	1	0	0	0	0	0	1	3
1	1	0	1	0	0	0	0	4

- * Construct a systematic $(8,4)$ cyclic code with generator polynomial $G(P) = P^3 + P + 1$.
- Determine the Generator Matrix
 - Decide the received codeword is 11011001 Using Syndrome decoding

Sol:

i)

$$G(P) = P^3 + P + 1$$

t th row of Generator Matrix is given by, $R_t(P) \cdot G(P)$

$$P^{n-t} + R_t(P) = Q_t(P) \cdot G(P)$$

$E = n-k = 8-4=4$

$$\text{Here } n=8$$

$$k=4$$

$$q = n-k = 8-4=4$$

For $t=1$:

$$P^{8-1} + R_1(P) = Q_1(P) \cdot G(P) \Rightarrow P^7 + R_1(P) \text{ is polynomial for 1st row of Generator Matrix}$$

$$R_1(P) = \text{rem} \left(\frac{P^{8-1}}{G(P)} \right) = \text{rem} \left(\frac{P^7}{P^3 + P + 1} \right)$$

$$R_1(P) = 1$$

$$\Rightarrow P^7 + 1$$

$$Q_1(P) = P^4 + P^2 + P + 1$$

1st row of Generator Matrix is $P^7 + 1$

$$= 10000001$$

For $t=2$:

$$P^{8-2} + R_2(P) = Q_2(P) \cdot G(P)$$

$$R_2(P) = \text{rem} \left(\frac{P^{8-2}}{G(P)} \right) = \text{rem} \left(\frac{P^6}{P^3 + P + 1} \right)$$

$$R_2(P) = P^2 + P + 1$$

\Rightarrow 2nd row polynomial

$$Q_2(P) = P^6 + P^2 + P + 1$$

2nd row of Generator Matrix = 01000101

$$\begin{array}{r} P^3 + P + 1 \\ P^4 + P^3 + P^2 \\ \hline P^5 + P^4 \end{array}$$

$$\begin{array}{r} P^5 + P^3 + P^2 \\ \hline P^4 + P^3 + P^2 \end{array}$$

$$\begin{array}{r} P^4 + P^2 + P \\ \hline P^3 + P \end{array}$$

$$\begin{array}{r} P^3 + P + 1 \\ \hline 1 \end{array}$$

$$\begin{array}{r} P^3 + P + 1 \\ P^6 \\ \hline P^6 + P^4 + P^3 \end{array}$$

$$\begin{array}{r} P^4 + P^3 \\ \hline P^4 + P^2 + P \end{array}$$

$$\begin{array}{r} P^3 + P^2 + P \\ \hline P^3 + P + 1 \end{array}$$

$$\begin{array}{r} P^3 + P + 1 \\ \hline P^2 + 1 \end{array}$$

$$\text{for } t=3:$$

$$p^{8-3} + R_3(p) = Q_3(p) \cdot G(p)$$

$$R_3(p) = \text{rem}\left(\frac{p^{8-3}}{G(p)}\right) = \text{rem}\left(\frac{p^5 + p^4 + p^3 + p^2}{p^3 + p + 1}\right) \text{ since } \frac{p^5 + p^4 + p^3 + p^2}{p^3 + p + 1}$$

$$R_3(p) = p^2 + p + 1$$

3rd row Generator polynomial
 $(1+p+q+q^2+q^3) \text{ is } p^5 + p^4 + p^3 + p^2 + p + 1 = 1 + q + q^2 + q^3$

$$R_3(p) = p^2 + p + 1$$

3rd row of Generator Matrix = $\begin{matrix} 0 & 0 & 1 & 0 & 0 & 1 & 1 & 1 & 1 \\ & & & & & & 1 & + q & + q^2 + q^3 \end{matrix}$

$$\text{for } t=4:$$

$$p^{8-4} + R_4(p) = Q_4(p) \cdot G(p)$$

$$R_4(p) = \text{rem}\left(\frac{p^{8-4}}{G(p)}\right) = \text{rem}\left(\frac{p^4}{p^3 + p + 1}\right) \text{ since } \frac{p^4}{p^3 + p + 1}$$

$$R_4(p) = \text{rem}\left(\frac{p^{8-4}}{G(p)}\right) = \text{rem}\left(\frac{p^4}{p^3 + p + 1}\right) \text{ since } \frac{p^4}{p^3 + p + 1}$$

4th row polynomial is $p^4 + p^2 + p$

$$R_4(p) = p^2 + p = (1)p \Rightarrow$$

4th row of Generator Matrix = $\begin{matrix} 0 & 0 & 0 & 1 & 1 & 0 & 1 & q & 0 \end{matrix}$

$$Q_4(p) = p$$

4th row of Generator Matrix = $\begin{matrix} 0 & 0 & 0 & 1 & 1 & 0 & 1 & q & 0 \end{matrix}$

Generator Matrix, $G(p) = p \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 & 1 \\ 1 & 1 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 1 & 0 & 1 & 0 \end{bmatrix}$

find basis of space = $\begin{matrix} 0 & 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \end{matrix}$

find basis of space = $\begin{matrix} 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \end{matrix}$

Now, the received codeword is 11011001

find basis of space = $\begin{matrix} 0 & 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \end{matrix}$

Given, $y = \text{find } 11011001$

find basis of space = $\begin{matrix} 0 & 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \end{matrix}$

$$Y(p) = p^7 + p^6 + p^4 + p^3 + 1$$

find basis of space = $\begin{matrix} 0 & 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \end{matrix}$

$$G(p) = p^3 + p + 1$$

(Given in question)

find basis of space = $\begin{matrix} 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 \end{matrix}$

find basis of space = $\begin{matrix} 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 \end{matrix}$

Syndrome Polynomial, $S(P) = \text{rem} \left[\frac{Y(P)}{G(P)} \right]$

$$\rightarrow S(P) = \text{rem} \left[\frac{P^7 + P^6 + P^4 + P^3 + 1}{P^3 + P + 1} \right] = \left(\frac{S-P}{G(P)} \right) \text{rem} \left[\frac{P^7 + P^6 + P^4 + P^3 + 1}{P^3 + P + 1} \right]$$

$$P^3 + P + 1 \quad | \quad P^7 + P^6 + P^4 + P^3 + 1 \quad (P^4 + P^3 + P^2 + P + 1)$$

$$\underline{P^7 + P^5 + P^4}$$

$$P^6 + P^5 + P^3$$

$$\underline{P^5 + P^4 + 1}$$

$$P^5 + P^3 + P^2$$

$$\underline{P^4 + P^3 + P^2 + 1}$$

$$P^5 + P^2 + P$$

$$\underline{P^3 + P + 1}$$

$$P^3 + P + 1 \quad | \quad 0 \quad 0 \quad 1 \quad 0 \quad 0$$

= without余数
for error detection

$$S = 0 \quad 0 \quad 0 \quad 0$$

$$S(P) = 0 \quad 1 \quad 1 \quad 1 \quad 0 \quad 0 \quad 0 \quad 0$$

Error decoding Table for $(8,4)$ cyclic code

$E = 0 \quad 0$	No error
$E = 1 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0$	error in 1st bit
$E = 0 \quad 1 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0$	error in 2nd bit
$E = 0 \quad 0 \quad 1 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0$	error in 3rd bit
$E = 0 \quad 0 \quad 0 \quad 1 \quad 0 \quad 0 \quad 0 \quad 0$	error in 4th bit
$E = 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 0 \quad 0 \quad 0$	error in 5th bit
$E = 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 0 \quad 0$	error in 6th bit
$E = 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 0$	error in 7th bit
$E = 0 \quad 1$	error in 8th bit

Also, we got Generator Matrix:

$$G(P) = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 \end{bmatrix}$$

$$\Rightarrow P = \begin{bmatrix} 0 & 0 & 0 & 1 \\ 0 & 1 & 0 & 1 \\ 0 & 1 & 1 & 1 \\ 0 & 1 & 1 & 0 \end{bmatrix} \xrightarrow{P^T L = I_4} P^T = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 1 \\ 1 & 1 & 1 & 0 \end{bmatrix}$$

$$H = [P^T | I_4] = \begin{bmatrix} 0 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 1 & 0 & 0 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 0 & 1 \end{bmatrix} \quad (2,5)$$

$$H^T = \begin{bmatrix} 0 & 0 & 0 & 1 \\ 0 & 1 & 0 & 1 \\ 0 & 0 & 1 & 1 \\ 0 & 1 & 1 & 0 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \text{Here we got } S = [0 \ 0 \ 0 \ 0]$$

$$\Rightarrow S = [0]$$

$$X = Y$$

No error \Rightarrow Received codeword = Transmitted codeword.

$$R = (1 + \frac{P_d}{P_s})^{-1} = (1 + \frac{1}{2})^{-1} = \frac{1}{2} = 0.5$$

$$[0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0]$$

* Determine the encoded message for the following code using the generator polynomial $G(P) = P^4 + P^3 + 1$ in non-systematic form in cyclic code $(12,8)$

(i) 1100 1100

(ii) 01011111

Sol:

$$\begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

Given $n=12$, $K=8$

$$v = n - K = 12 - 8 = 4$$

$$G(P) = P^4 + P^3 + 1$$

$$\begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

$(12,8)$ is the cyclic block code

1st row of $G(P)$:

$$\text{as } K=8 \\ K-1=7$$

$$P^7 \cdot G(P) = P^7 (P^4 + P^3 + 1) = P^{11} + P^{10} + P^7$$

$$\text{1st row is : } [1 \ 0 \ 0 \ 1 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0]$$

2nd row of $G(P)$:

$$\text{as } K=8 \\ K-2=6$$

$$P^6 \cdot G(P) = P^6 (P^4 + P^3 + 1) = P^{10} + P^9 + P^6$$

$$\text{2nd row is : } [0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0]$$

3rd row of $G(P)$:

$$\text{as } K=8 \\ K-3=5$$

$$P^5 \cdot G(P) = P^5 (P^4 + P^3 + 1) = P^9 + P^8 + P^5$$

$$\text{3rd row is : } [0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 0 \ 0 \ 0 \ 0 \ 0]$$

6th row of $G(P)$:

$$G(P) = 0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 0 \ 0 \ 1 \ 1 \quad \text{is } (7)_m$$

$$\text{as } K=8 \quad P^4 \cdot G(P) = P^4 (P^4 + P^3 + 1) = P^8 + P^7 + P^4$$

$$\text{as } K-4 = 4 \quad 0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0$$

4th row: $[0 \ 0 \ 0 \ 1 \ 0 \ 0 \ 1 \ 0 \ 0 \ 0 \ 0]$

5th row of $G(P)$:

$$G(P) = 0 \ 1 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 0 \ 0 \quad \text{as } K=8$$

$$\text{as } K-5 = 3 \quad 1 \cdot P^3 \cdot G(P) = P^3 (P^4 + P^3 + 1) = P^7 + P^6 + P^3$$

5th row is: $[0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 0 \ 0]$

$$(7)_d \cdot (7)_m = (7)_X$$

6th row of $G(P)$:

$$\text{as } K=8 \quad P^2 \cdot G(P) = P^2 (P^4 + P^3 + 1) = P^6 + P^5 + P^2$$

$$[K-6 = 2 \ 0 \ 0 \ 1 \ 0 \ 1] = (7)_d \cdot [0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 0 \ 0] (7)_X$$

6th row is: $[0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 0]$

7th row of $G(P)$:

$$\text{as } K=8 \quad P \cdot G(P) = (7)_d P (P^4 + P^3 + 1) = P^5 + P^4 + P$$

$$K-7 = 1$$

7th row is: $[0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 0]$

8th row of $G(P)$:

$$\text{as } K=8 \quad P^0 \cdot G(P) = G(P) = P^4 + P^3 + 1$$

$$K-8 = 0$$

8th row is: $[0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1]$

Ansab different Ansab muktab 7th and 8th

$$\text{So, } G(P) = \begin{bmatrix} 1 & 1 & 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 1 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 1 & 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix}$$

Now,

$$X(P) = M(P) \cdot G(P)$$

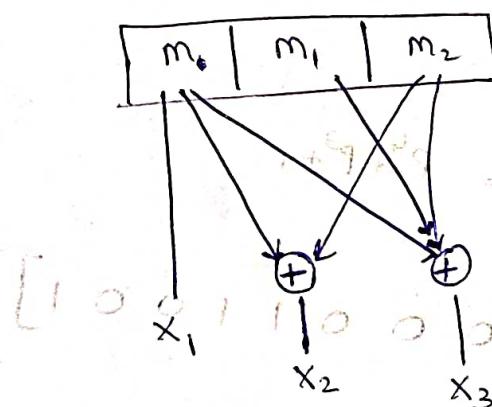
$$\text{i. For } M(P) = 1100111000 \Rightarrow (7)_2 \text{ for } s=4$$

$$X(P) = [11001100] \cdot G(P) = [10100110100]$$

$$\text{ii. For } M(P) = 01011111$$

$$X(P) = [01011111] \cdot G(P) = [0111101010111]$$

* For the given convolutional encoder configuration



- Draw the state diagram and code trellis diagram.
- Encode the sequence 1110 using code tree.

$$\begin{aligned}x_1 &= m \\x_2 &= m \oplus m_2 \\x_3 &= m \oplus m_1 \oplus m_2\end{aligned}$$

$$\text{Code rate, } r = \frac{k}{n} = \frac{1}{3}$$

~~no. of message bit = $k = 1$~~

~~no. of encoded output, $n = 3$~~

- first shift message will enter M .
- second shift message will enter M_1 .
- third shift message will enter M_2 .
- forth shift message will ~~not~~ enter and will be discarded.

$$\begin{aligned}\text{dimension of code} &= (n, k) \\&= (3, 1)\end{aligned}$$

constraint length - no. of shift over which a single message bit influences the encoder output. $K =$

		States	
m_2	m_1	a	b
0	0	0	1
0	1	1	0
1	0	0	1
1	1	1	0

m = current message bit

$m_2 m_1$ = current state

$m_1 m$ = next state

00 - a

01 - b

10 - c

11 - d

<u>State table</u>	<u>Current State</u>	<u>Next State</u>	<u>x_1</u>	<u>x_2</u>	<u>x_3</u>	<u>decimal</u>
m_2	m_1	m	a	b	c	d
0	0	0	a	a	b	0
0	0	1	a	b	1	1
0	1	1	b	d	1	1
0	1	0	b	d	0	2
1	1	1	d	d	0	4
1	1	0	d	c	0	5
1	0	0	a	b	1	6
1	0	1	a	b	0	7

output encoded: 111 110 101 010

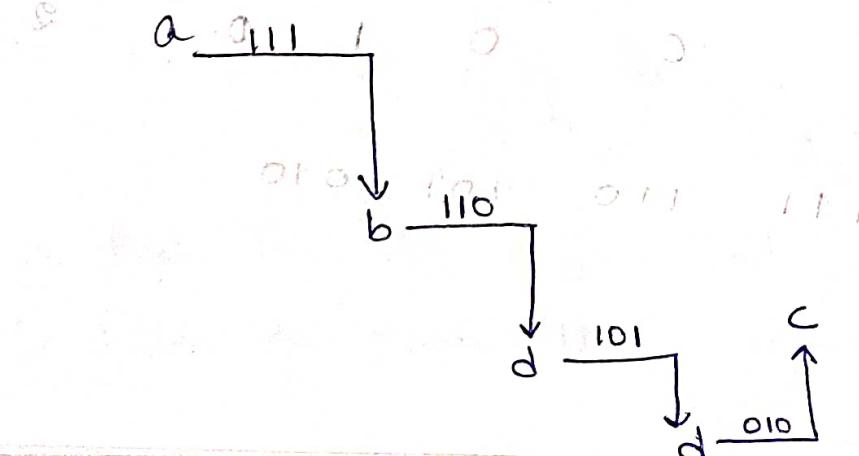
Trick: $\frac{a}{b} = \frac{c}{d} + e$, then find

$$\begin{array}{c|cc} & \frac{0}{a} & \frac{a}{a} \\ \hline a & \frac{0}{a} & a \\ b & \frac{1}{b} & b \\ \hline b & \frac{1}{d} & c \\ b & \frac{1}{d} & d \\ \hline d & \frac{0}{c} & c \end{array}$$

m_2	m_1	m	Current State	next state	x_1	x_2	x_3	decide
0	0	0	a	a	0	0	0	0
0	0	1	a	b	1	1	1	7
0	1	0	b	c	0	0	1	1
0	1	1	b	d	1	1	0	6
1	0	0	c	a	0	1	1	3
1	0	1	c	b	1	0	0	4
1	0	0	d	c	0	1	0	2
1	1	0	d	d	1	0	1	5

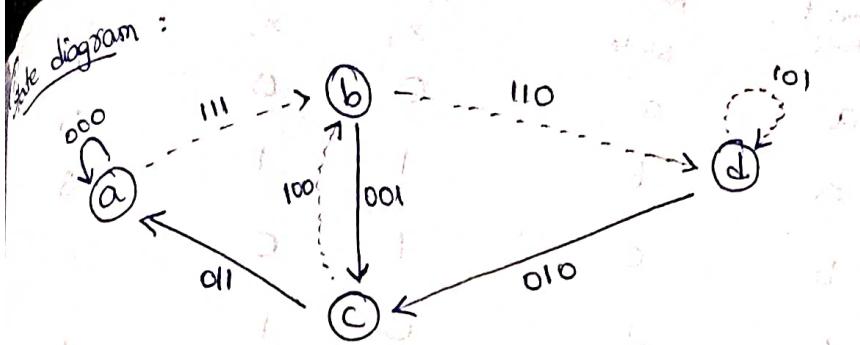
ii) Code tree:

Given sequence, 1110

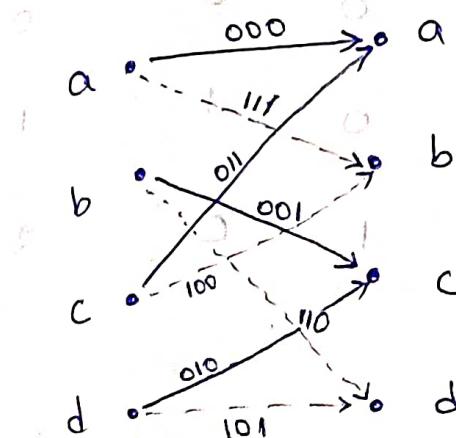


encoded sequence $\frac{111 - 110}{101 - 010}$





Fretti's diagram:



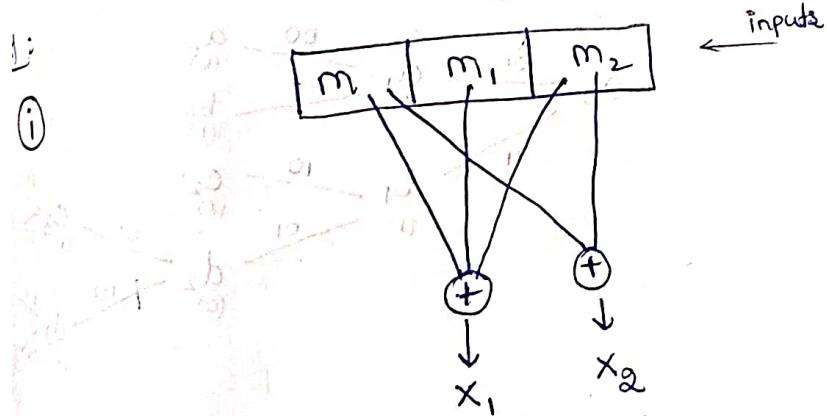
Conclusion :

encoded output : 0111 110 101 010

fall 2010) gabriel idstein

- Ques. 1

 - For the convolutional encoder configuration with code rate $\frac{1}{2}$, constraint length 3 and if the outputs are expressed as $x_1 = m_1 \oplus m_2$ and $x_2 = m_1 \oplus m_2$ draw the diagram of convolutional encoder.
 - Draw the diagram of convolutional encoder using code trellis and viterbi.
 - Decide the sequence 01011110 using viterbi.



$$x_1 = m \oplus m_1 + m_2$$

$$x_2 = m \oplus m_2$$

- Given sequence, 01011110

$$\text{Code rate}, \quad \sigma = \frac{K}{n} = \frac{1}{2}$$

$$\frac{\pi}{2} = 1$$

$$\text{constraint length} = 3$$

$$\text{dimension} = (n, k) = (2, 1)$$

m_2	m_1	m	current state	next state	x_1	x_2	decimal
0	0	0	a	a	0	0	0
0	0	1	a	b	1	1	3
0	1	0	b	c	1	0	2
0	1	1	b	d	0	1	1
1	0	0	c	a	1	1	3
1	0	1	c	b	0	0	0
1	1	0	d	c	0	1	1
1	1	1	d	d	1	0	2

decode sequence : 01 01 11 10

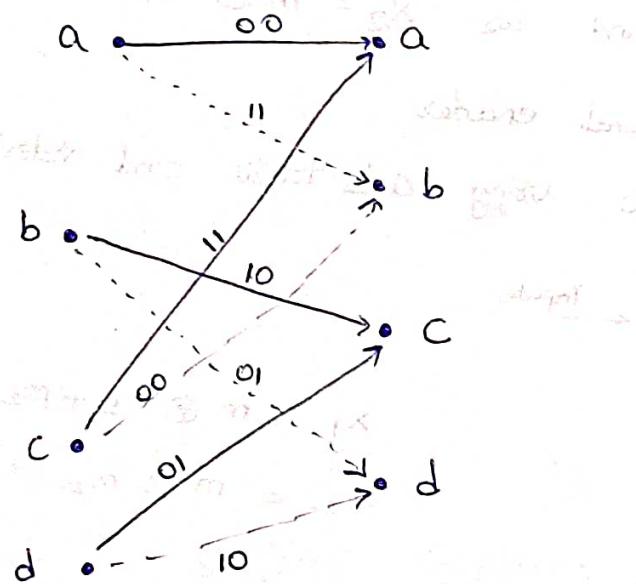
• Viterbi decoding (also called

Toelli's diagram = ~~Maximum Likelihood Decoding~~

Hamming distance :

$$01 \oplus 10 \rightarrow 10$$

output: $\boxed{01}$ among $\boxed{01}$ & $\boxed{11}$, $\boxed{10}$



encoded
actual : 1111

