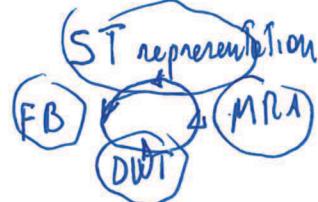


Ondelettes & Filter banks



[1]

Basic idea of the transform-based analysis of signals:
project a signal onto a basis where some characteristics
are easily highlighted.

Eg. CTFP: $\hat{X}(v) = \int X(t) e^{-2\pi v t} dt = \langle X(t), e^{2\pi v t} \rangle$

$$STFT(X(t); w(u)) = \int X(t) W(t) e^{-2\pi v t} dt = \langle X(t), W(t) e^{2\pi v t} \rangle$$

Besides frequency analysis, it could be interesting to have
other kind of analysis, such as a "Time-scale" one:

$$\Psi_{a,b}(t) = \frac{1}{\sqrt{a}} \Psi\left(\frac{t-b}{a}\right)$$

c : scale
b : time-shift

$$CTWT(X(t); a, b) = \langle X(t), \Psi_{a,b}(t) \rangle$$

Continuous function of $a > 0$, $b \in \mathbb{R}$

Ψ : mother wavelet

Remember that, if $\hat{\Psi}(v) = CTFP(\Psi(t))$,

$$\text{Then } CTFP(\Psi\left(\frac{t}{a}\right)) = a \hat{\Psi}(av)$$

Therefore, if $\langle X(t), \Psi(t) \rangle$ tells about the similarity
of X with Ψ at a given scale, changing the scale
implies changing both temporal & frequency resolution,
i.e. where is concentrated most of the energy of Ψ

If a increases, $\Psi_{a,b}(t)$ is more concentrated in time and
more spread in frequency.

So our "dictionary" $\Psi_{a,b}$ makes a Time-scale analysis providing good freq resolution at low freq (low scale) and good time resolution at high frequency. This is particularly suitable for image analysis.

(2)

Moving continuous parameters a & b is "Too much".

We can discretize: $a = 2^{-j}$, $b = a \cdot k$, $k, j \in \mathbb{Z}$

\star
scale: double
The resolution at
each level j

Δ
Time delay:
in scale-related units

Therefore, if $\Psi_{j,n}(t) = 2^{jk} \Psi(2^j t - k)$,

The wavelet coefficients are $c_{j,n} = \langle x(t), \Psi_{j,n}(t) \rangle$

The signal can be reconstructed as $x(t) = \sum_j \sum_n c_{j,n} \tilde{\Psi}_{j,n}(t)$ where $\tilde{\Psi}_{j,n}$ should suitably be found, but we anticipate that it is possible to have $\Psi = \tilde{\Psi}$ (orthogonal wavelets)

$$\Rightarrow x(t) = \sum_{j,n} \langle x(t), \Psi_{j,n}(t) \rangle \Psi_{j,n}(t)$$

Discretization of $x(t)$ brings the DWT.

Target: sparse representation: as many zero as possible

Reason: analysis; modeling; compression

understanding

* page (3) If $f \in V_1$, at least $g \in V_0 : f(t) = g(t)$

but g can be expressed on the basis of V_0 : $g(t) = \sum_k g_k \varphi(t-k) \Rightarrow$

$$f(t) = g(t) = \sum_k g_k \varphi(2^{-1}t - k) \Rightarrow \varphi(2t - n) \text{ is a basis for } V_1$$

Then we can generalize for V_j ; The coeff $2^{j/2}$ is just for normalization

MRA

Now we need to better formalise those concepts. We introduce the idea of MRA

[3]

We consider an MRA over the space $L^2(\mathbb{R})$.

It is defined as a sequence of vector subspaces V_j such that:

A 1) $\forall j \in \mathbb{Z}$, $V_j \subset V_{j+1}$ (increasing)

2) $\cap V_j = \{0\}$ (constant function returning $\phi\}$)
(non redundancy)

3) $\overline{\cup V_j} = L^2$ completeness

B 4) $f(t) \in V_j \Leftrightarrow f(2t) \in V_{j+1}$

5) $f(t) \in V_0 \Leftrightarrow f(t-k) \in V_0 \quad \forall k \in \mathbb{Z}$

C 6) It exists an orthonormal basis generated by a suitable function φ :

$\exists \varphi \in V_0 : \{\varphi(t-k)\}$ generates V_0

φ is called father wavelet

Idea of MRA: The projection of $f \in L^2(\mathbb{R})$ on V_j ,

for j that increases, gives a better & better approx of f
(because of properties A, namely completeness);

moreover, it is a T.S analysis, since 6) implies that

$\varphi_{j,k}(t) = 2^{j/2} \varphi(2^j t - k)$ is a basis of V_j \oplus Example
 $\varphi(t) = \frac{1}{\sqrt{2}}(1+t)$

Note that if $\varphi \in C^n$, the MRA is said to be n -regular

Here it could be nice to show an example with Moon MRA basis

So we consider $f_j = \text{Proj} \{ f \mid V_j \}$ for any $\text{f} \in L^2(\Omega)$

Completeness implies $f_j \rightarrow f$

We define the details $\Delta f_j = f_{j+1} - f_j$

Δf_j belongs to the supplemental space W_j

and $V_j \oplus W_j = V_{j+1}$

Therefore $V_0 \oplus W_0 \oplus \dots \oplus W_j = V_j$

and $V_j = \bigoplus_{l=0}^j W_l$

So we can consider a representation of f as its projection on:

- V_0 and

- W_0, W_1, \dots, W_j

(note that the "starting" resolution 0 is arbitrary. It can be any integer)

projection on V_0 = similarity with φ

projection on W_0 = details w.r.t. V_1 : or details at a first resolution level

: W_j = details from V_j to V_{j+1} , j th res. level

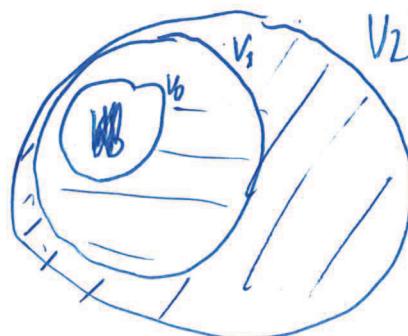
This representation is equivalent to $f_j = \text{Proj} \{ f \mid V_j \}$

Now $f_j(t) = \sum_k c_{j,k} \varphi_{j,k}(t)$ if orthogonal

where $c_{j,k} = \langle f, \tilde{\varphi}_{j,k} \rangle = \langle f, \varphi_{j,k} \rangle$

example: if $\varphi(t) = \mathbb{1}_{(0,1)}(t)$,

f_j is a piecewise constant version of f ,
on intervals of size 2^{-j} .



$$V_0 = \square$$

$$W_0 = \exists \quad V_1 = \square \exists$$

$$W_1 = \square \exists \quad V_2 = \square \exists \square \exists$$

Now, let us consider the relationship between two consecutive subspaces, V_0 & V_1

5/

Since $\varphi \in V_0$ and $V_0 \subset V_1$, then $\varphi \in V_1$, so it can be expressed using the basis of V_1

$$\varphi(t) = \sqrt{2} \sum_k c_k \varphi(2t-k) \quad (1)$$

This is the dilation equation

example: $\varphi(t) = \mathbb{1}_{(0,1)}(t)$

$$\varphi(t) = \varphi(2t) + \varphi(2t+1)$$

$$c_0 = \frac{1}{\sqrt{2}} \quad c_1 = \frac{1}{\sqrt{2}}$$

More in general, the coeffs are computed by projection.

We consider the cos of orthogonal basis:

$$\begin{aligned} c(n) &= \langle \varphi(t), \varphi_{s,n}(t) \rangle = \langle \varphi(t), 2^{\frac{n}{2}} \varphi(2t-n) \rangle \\ &= \sqrt{2} \int \varphi(t) \varphi(2t-n) dt \end{aligned}$$

From (1), if we multiply both by $\varphi(t-m)$ & integrate:

$$\int \varphi(t) \varphi(t-m) dt = \sqrt{2} \sum_k c_k \int \varphi(2t-k) \varphi(t-m) dt$$

$\stackrel{\text{II}}{=} \sqrt{2} \sum_k c_k \int \varphi(2s+2m-k) \varphi(s) ds$

$s = t - m$
 $t = s + m$

$$\sum_k c_k \langle \varphi(s), \varphi_{s,2m-k}(s) \rangle$$

$$\sum_k c_k c_{k-2m} = \delta_m \quad \boxed{\text{Orthogonality}}$$

It can also be proven that $|\hat{c}_w|^2 + |\hat{c}_{w+\frac{1}{2}}|^2 = 2$

See Mallat page 229

So, The sequence $f_j = \text{Proj}(f, V_j)$ is increasing
in resolution, but also redundant, since $V_j \subset V_{j+1}$ [6])

Let us consider the decomposition : $V_j = V_0 \oplus W_0 \oplus \dots \oplus W_j$
meaning that f_j can be represented in terms of f_0 and
of projections in W_i , i.e. $\{0, \dots, j\}$

These projections are the Δf_i

Since let us look for a basis of W_0 .

$$\text{Since } V_0 \oplus W_0 = V_1, \quad W_0 \subset V_1$$

Mallat proved the following. (IR) $\varphi(\cdot)$ is a father wavelet,
i.e., $\varphi(t-k)$ is a basis of V_0 ; and c_n are
the coefficient of the dilation equation, i.e.

$$\varphi(t) = \sqrt{2} \sum_k c_n \varphi(2t-k)$$

Moreover, set $C(z) = \sum_n c_n z^{-n}$, $D(z) = -z^{-(N-1)} C(-z^{-1})$
and $d_n = \text{IFT}(D(z))$

defining $\psi(t) = \sqrt{2} \sum_n d_n \phi(2t-n)$ wavelet
equation

Then $\psi_{j,n}(t) = 2^{-jk} \psi(2^{-j}t-n)$ is a basis for W_j

Proof: see Mallat, Theorem 7.3 page 236.

This is very interesting because, since $L^2(\mathbb{R}) = \overline{\bigcup W_j}$,

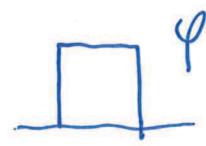
Then The wavelets $\psi_{j,n}$ are a basis for $L^2(\mathbb{R})$

But The $\psi_{j,n}$ are exactly a Time-Scale analysis

So, a link is established between MRF and ST (2)

Example: if $\varphi = \mathbb{1}_{(0,1)}$, $c_0 = \frac{1}{\sqrt{2}}$, $c_1 = \frac{1}{\sqrt{2}}$

$$\Rightarrow d_0 = \frac{1}{\sqrt{2}} \text{ & } d_1 = -\frac{1}{\sqrt{2}}$$



$$\Rightarrow \varphi(t) = \mathbb{1}_{(0,1)}(t) - \mathbb{1}_{(1,2)}(t)$$



$\underline{\epsilon}$ and \underline{d} are just the filters of the Haar F.B.



Summary:

$$\varphi(t) = \sqrt{2} \sum_n c_n \varphi(2t-n)$$

dilation eqn.

$$\psi(t) = \sqrt{2} \sum_n d_n \varphi(2t-n)$$

wavelet eqn

with (2) $\sum c_n c_{n-2m} = \delta_m$ + orthogonality, H.B.
 $D(z) = z^{-(N-1)} C(-z^{-1})$ ← CQF
filter bank

So, given a filter satisfying (2) and $|\hat{c}(v)|^2 + |\hat{c}(v+1_L)|^2 = 1$,

we deduce d_n : Then (difficult in general) we find

φ satisfying the dilation equation:

Then we find ψ .

In conclusion we represent $f_j = \text{Proj}(f|V_j) =$

$$= \text{Proj}(f|V_0) + \sum_{i=0}^j \text{Proj}(f|W_i)$$

"approximation"

"details"

↙ basis of W_j

$$= \sum_n a_{0,k} \varphi(t-k) + \sum_j \sum_k b_{j,k} \psi_{j,k}(t)$$

Now we have to compute $a_{j,n}, b_{j,n}$
in terms of ρ, c, d

8)

The coefficients $a_{j,n}, b_{j,n}$ are the MRA representation
of f in V_j : for $j \rightarrow +\infty$, they fully represent f

It is a recursive relationship. Let us start at level 0.

If $f_1 \in V_1$, also $f_1 \in V_0 \oplus W_0$

So we can use a basis of V_0 , or the 2 basis, V_0' on W_0 :

$$f_1(t) = \sum_n a_{1,n} \varphi_{1,n} = \sum_n a_{0,n} \varphi_{0,n} + \sum_n b_{0,n} \varphi_{0,n} \quad (3)$$

From DE & WE,

$$\begin{aligned} \varphi_{0,m}(t) &= \varphi(t-m) = \sqrt{2} \sum_n c_n \varphi(2t-2m-n) \\ &= \sum_l c(l-2m) \sqrt{2} \varphi(2t-l) \\ &= \sum_l c(l-2m) \varphi_{1,l}(t) \end{aligned} \quad (4)$$

$$\begin{cases} l = 2m+n \\ n = l - 2m \end{cases}$$

$$\text{likewise } \varphi_{0,m}(t) = c_l d(l-2m) \varphi_{1,l}(t) \quad (5)$$

$$\begin{aligned} \text{Now, } a_{0,n} &= \langle f(t), \varphi_{0,n}(t) \rangle = \sum_l c(l-2n) \langle f(t), \varphi_{1,l}(t) \rangle \\ &= \sum_l c(l-2n) a_{1,l} \end{aligned}$$

$$\text{likewise } b_{0,n} = \langle f(t), \varphi_{0,n}(t) \rangle = \sum_l d(l-2n) a_{1,l}$$

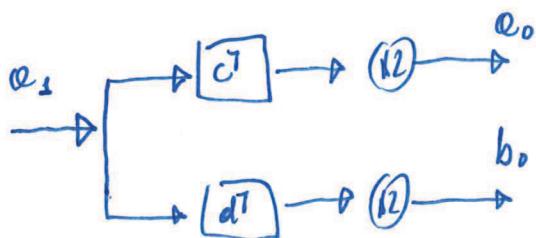
$$\text{Defining } c^T(n) = c(-n)$$

$$d^T(n) = d(-n)$$

$$e_{0,n} = \sum_e c^T (2k-l) e_{1,l} = (c^T * e_1) + 2$$

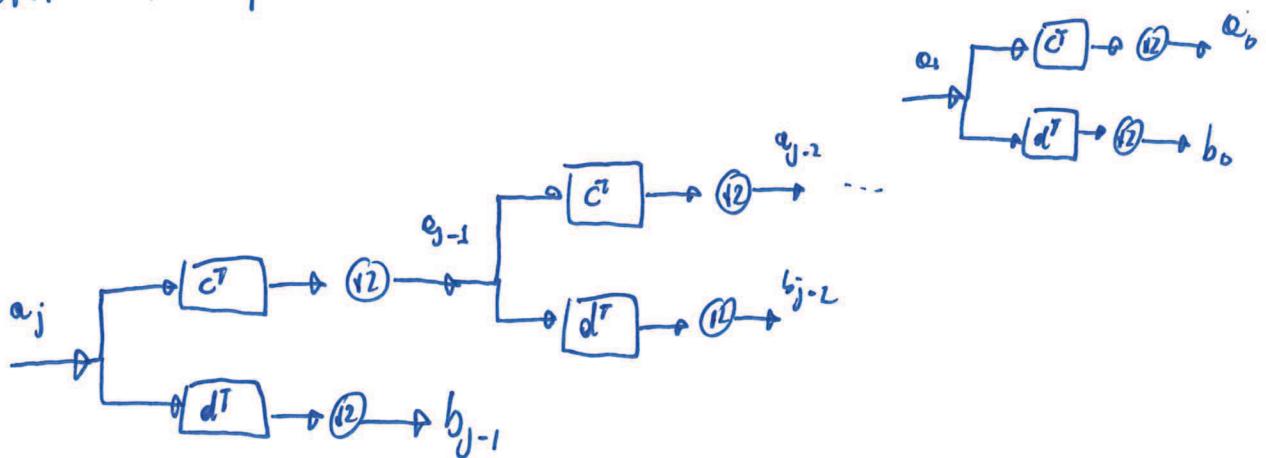
[g]

$$b_{0,n} = \sum_e d^T (2k-l) e_{1,l} = (d^T * e_1) + 2$$



c & d
are CQF!

With multiple levels:



so f_j is represented by the coefficients

$$e_0, b_0, b_1, \dots, b_{j-2}, b_{j-1}$$

↓
lower
resolution
approx.

↑ ↑ ↑
increasing resolution
details

$e_j \rightarrow \{e_0, b_0, \dots, b_{j-2}, b_{j-1}\}$ DWT over j levels

$e_1 \rightarrow \{e_0, b_0\}$ DWT over 1 level \Leftrightarrow two ways FB

Synthesis F.B.

[10]

$$f_1(t) = \sum a_{0,n} \varphi_{0,n}(t) + \sum b_{0,n} \psi_{0,n}(t) \quad (\text{it is } (3))$$

$$\Rightarrow \langle f_1(t), \varphi_{1,l} \rangle = \sum a_{0,n} \langle \varphi_{0,n}(t), \varphi_{1,l}(t) \rangle + \sum b_{0,n} \langle \psi_{0,n}(t), \varphi_{1,l}(t) \rangle$$

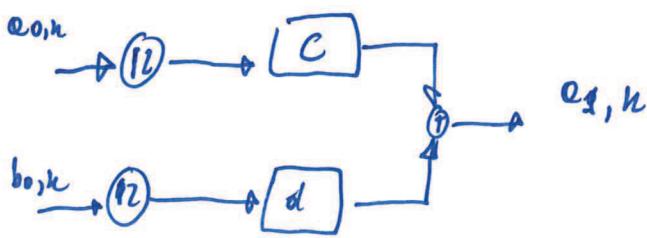
$$a_{1,l} = \sum a_{0,n} \sqrt{2} \int \varphi(t-n) \varphi(2t-l) dt + \dots$$

$$= \sum a_{0,n} \sqrt{2} \int \varphi(s) \varphi(2s+2n-l) dt + \dots \quad \text{like wise}$$

$$= \sum a_{0,n} c_{l-2n} + \sum b_{0,n} d_{l-2n}$$

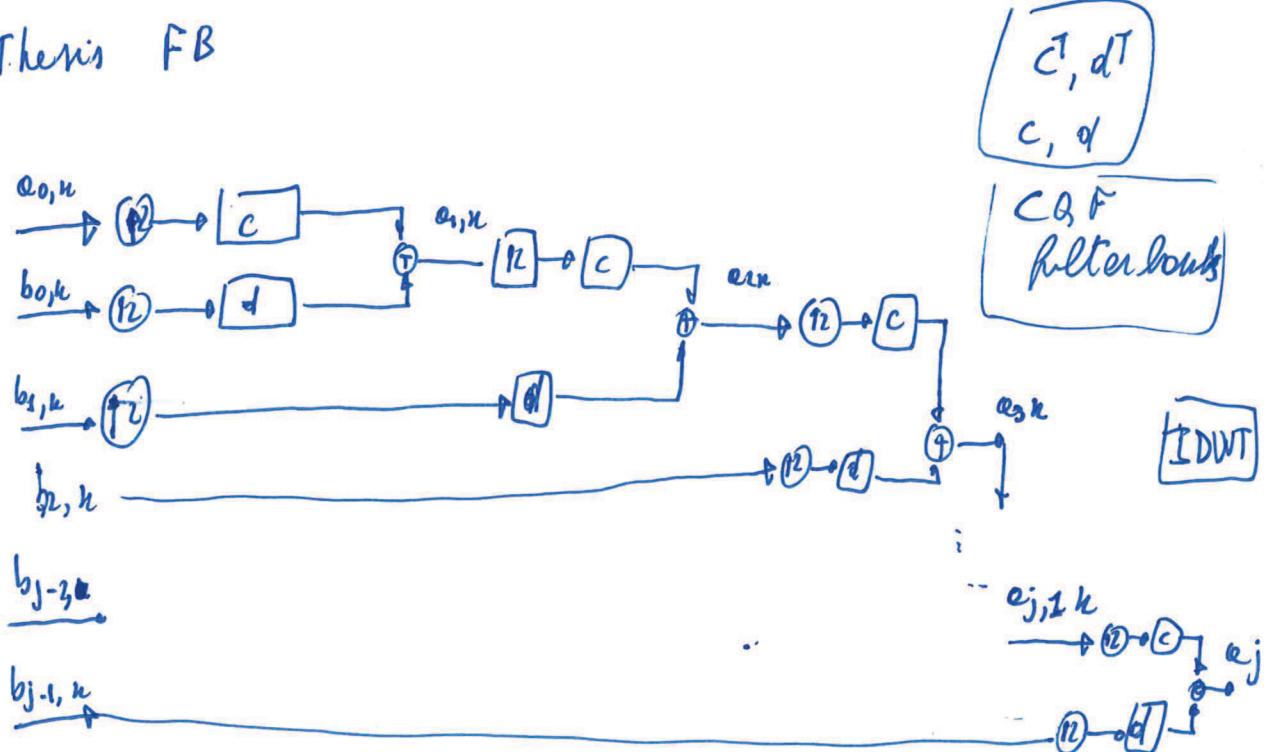
$$\text{Now, } \sum a_{0,n} c_{l-2n} = a_{0,0} c_0 + a_{0,1} c_{l-2} + a_{0,2} c_{l-4} + \dots$$

$$= a_{0,0} c_0 + 0 \cdot c_{l-1} + a_{0,1} c_{l-2} + 0 \cdot c_{l-3} + a_{0,2} c_{l-4} + \dots$$



one level
of wavelet
analysis

Synthesis FB



Other properties of FB for wavelet analysis

11/

Vanishing moments

$$f_j(t) = \sum_n \varphi_{0n} + \sum_{\ell, k} b_{\ell, k} \psi_{\ell, k}$$

It can be proven that the approx error is

$$\| f(t) - f_j(t) \| \approx c (2^{-j})^p \| f^{(p)}(t) \| \quad (6)$$

where p is the multiplicity of the zero of $\hat{h}_0(v)$ in \mathbb{H} :

$$\hat{h}_0(v) = \left(\frac{1+e^{-iv}}{2} \right)^p \cdot Q(v) \Rightarrow H(z) = \left(\frac{1+z^{-1}}{2} \right)^p Q(z)$$

This is equivalent to:

$$\forall j \in \{0, \dots, p-1\}, \sum (-1)^n n^j h_0(n) = 0$$

(6) : polynomials of order up to p are perfectly reconstructed

Also, it can be proved that $h_1 * s$ where s is polynomial is null

So, having a high number of V.M. allows to represent s only with a , since the b will be zero.

If s is piecewise polynomial, b is non zero only "near discontinuities"

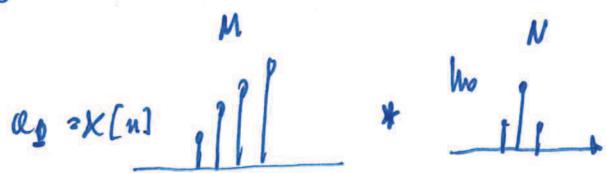
All this is important since we desire a
spars representation of the signal

(2)

Linear phase and borders

What happens when using DWT on finite signals?

- coeff. expansion



$$\text{cond } \{a_0\} = \frac{M+N-1}{2} \xrightarrow{\text{total}} M+N-1$$

$$\text{cond } \{b_0\} = \frac{M+N-1}{2}$$

(expansion)

→ at each level!

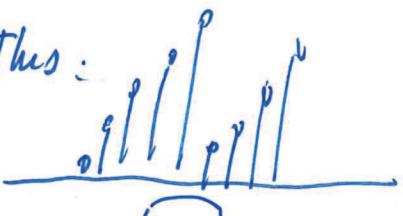
- how to avoid this?

$x \rightarrow \tilde{x}$, periodised version of x

$\begin{cases} a = \tilde{x} * h_o \\ b = x * h_o \end{cases}$ a, b are also periodical with the same period as \tilde{x}

Then, M coeff of a represent perfectly \tilde{x} , and $\neq x$

But This means that we are filtering this:



Periodisation generates
high frequency content, i.e.
more non-null coeffs in b

high frequency
content

Solution: symmetric periodization

(13)

$$\tilde{x} = \begin{cases} x(n) & n: 0 \dots M-1 \\ x(M-1-n) & n: M \dots 2M-1 \end{cases}$$

Then

$$a = (h_0 * \tilde{x}) + 2$$

$$b = (h_1 * \tilde{x}) + 2$$

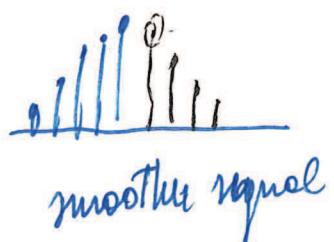
you need $(2M)$
coeffs,

unless h_0, h_1 are linear phase

In that case, a & b are also symmetrical, and
 M samples are enough.

The advantage is that we filter:

less high freq. content



but There is no linear phase CQT
except flocr, which has only 1 VM!

So, for having more VM & linear phase,
we must go beyond CQT, and use
"biorthogonal" filters

Why orthogonality is important!

[16]

because it can be proven that, with ortho. filters,

Core of
finite
signals

$$\|\underline{e}_1\|^2 = \|\underline{a}_0\|^2 + \|\underline{b}_0\|^2 = \|\underline{w}_0\|^2 \quad \underline{w}_0 = \begin{bmatrix} \underline{a}_0 \\ \underline{b}_0 \end{bmatrix}$$

Now, if we perform compression, we reduce the info in \underline{w}_0
• $\tilde{\underline{w}}_0$ is a degradation of \underline{w}_0 :

Now, by orthogonality, $\underline{e}_1 = \underline{W}^T \underline{w}_0$, \underline{W} orthogonal matrix.
and $\underline{w}_0 = \underline{W} \underline{e}_0$

$$\text{and } \tilde{\underline{e}}_1 = \underline{W}^T \tilde{\underline{w}}_0$$

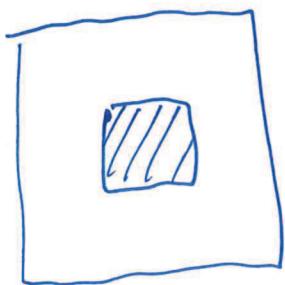
$$\begin{aligned} \|\underline{e}_1 - \tilde{\underline{e}}_1\|^2 &= (\underline{e}_1 - \tilde{\underline{e}}_1)^T (\underline{e}_1 - \tilde{\underline{e}}_1) = (\underline{w}_0 - \tilde{\underline{w}}_0)^T (\underline{W}^T \underline{W} (\underline{w}_0 - \tilde{\underline{w}}_0)) \\ &= \|\underline{w}_0 - \tilde{\underline{w}}_0\|^2 \end{aligned}$$

So, an orthogonal DWT is orthonormal.

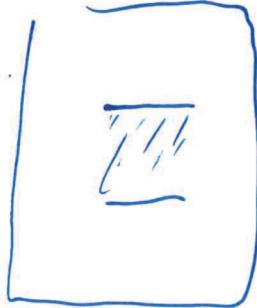
Therefore, we can control the error on \underline{e}_1 by controlling the error on \underline{w}_0 : useful in compression and denoising, since we can optimize our processing by minimizing $\|\underline{w} - \tilde{\underline{w}}\|$, which is simpler than minimizing $\|\underline{W}^T \underline{w} - \underline{W}^T \tilde{\underline{w}}\| = \|\underline{e}_1 - \tilde{\underline{e}}_1\|$

Examples on images

(15)



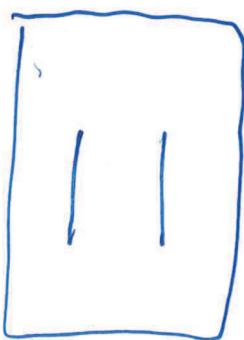
LP on rows



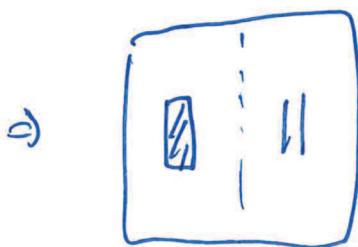
① →



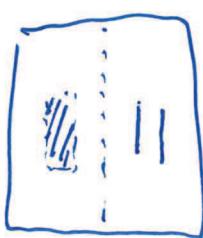
HP on rows



② →



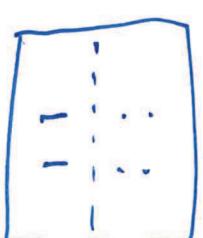
L¹
on cols



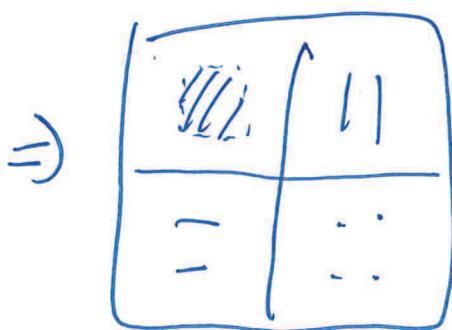
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HP
on cols

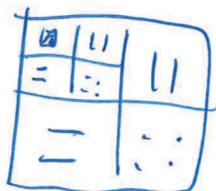


② →



1 level of DWT

2 levels:



⇒ ST analysis