8. Duality

$$x(t) = F^{-1}[X(j\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$



$$2\pi x(-t) = \int_{-\infty}^{\infty} X(j\omega)e^{-j\omega t}d\omega$$



$$2\pi x(-j\omega) = \int_{-\infty}^{\infty} X(t)e^{-j\omega t}dt$$



9. Frequency Shifting

$$e^{j\omega_0 t} x(t) \stackrel{F}{\longleftrightarrow} X[j(\omega - \omega_0)]$$

10. Differentiation in Frequency Domain

$$-jtx(t) \stackrel{F}{\longleftrightarrow} \frac{d}{d\omega} X(j\omega)$$

Discussion 4.4 Find the Fourier transform of $\frac{1}{\pi t}$.

Solution:

solution:

$$sgn(t) \stackrel{F}{\longleftrightarrow} \frac{2}{j\omega}$$

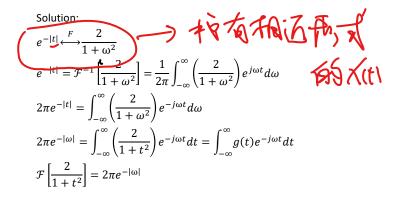
$$sgn(t) = \mathcal{F}^{-1} \left[\frac{2}{j\omega} \right] = \frac{1}{2\pi} \int_{-\infty}^{\infty} \left(\frac{2}{j\omega} \right) e^{j\omega t} d\omega$$
The properties of the properties

$$2\pi \operatorname{sgn}(-t) = -2\pi \operatorname{sgn}(t) = \int_{-\infty}^{\infty} \left(\frac{2}{j\omega}\right) e^{-j\omega t} d\omega$$

$$-2\pi \operatorname{sgn}(\omega) = \int_{-\infty}^{\infty} \left(\frac{2}{jt}\right) e^{-j\omega t} dt = \int_{-\infty}^{\infty} \frac{2\pi}{j} \left(\frac{1}{\pi t}\right) e^{-j\omega t} dt$$

$$\mathcal{F}\left[\frac{1}{\pi t}\right] = j \operatorname{sgn}(\omega)$$

Example 4.6 Find the Fourier transform of $\frac{2}{1+t^2}$



11. Parseval's Relation

$$\int_{-\infty}^{+\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} |X(j\omega)|^2 d\omega$$
$$\int_{-\infty}^{+\infty} x(t)y(t) dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega)Y(-j\omega) d\omega$$

Proof:

$$\int_{-\infty}^{+\infty} x(t)x^{*}(t)dt = \int_{-\infty}^{+\infty} x(t) \left[\frac{1}{2\pi} \int_{-\infty}^{\infty} X^{*}(j\omega)e^{-j\omega t}d\omega \right] dt$$
$$= \frac{1}{2\pi} \int_{-\infty}^{+\infty} X^{*}(j\omega) \left[\int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt \right] d\omega$$
$$= \frac{1}{2\pi} \int_{-\infty}^{+\infty} X^{*}(j\omega)X(j\omega)d\omega$$

Example 4.7 Calculate the following integral $(a = a^* > 0)$.

$$\int_{-\infty}^{\infty} \frac{1}{(a^2 + \omega^2)^2} d\omega$$

Solution:

$$e^{-a|t|} \stackrel{F}{\longleftrightarrow} \frac{2a}{a^2 + \omega^2}$$

Parseval's Relation yields

$$\int_{-\infty}^{\infty} |e^{-a|t|}|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} \left| \frac{2a}{a^2 + \omega^2} \right|^2 d\omega$$

$$\int_{-\infty}^{\infty} \frac{1}{(a^2 + \omega^2)^2} d\omega = \frac{\pi}{2a^2} \int_{-\infty}^{\infty} |e^{-a|t|}|^2 dt$$

$$= \frac{\pi}{a^2} \int_{0}^{\infty} e^{-2at} dt = \frac{\pi}{2a^3}$$

§4.4 The Convolution Property

$$x(t) \stackrel{F}{\longleftrightarrow} X(j\omega) \text{ and } y(t) \stackrel{F}{\longleftrightarrow} Y(j\omega)$$

 $x(t) * y(t) \stackrel{F}{\longleftrightarrow} X(j\omega)Y(j\omega)$

Proof:

$$\begin{split} F[x_1(t) * x_2(t)] &= \int_{-\infty}^{+\infty} \left[\int_{-\infty}^{+\infty} x_1(\tau) x_2(t-\tau) d\tau \right] e^{-j\omega t} dt \\ &= \int_{-\infty}^{+\infty} x_1(\tau) \left[\int_{-\infty}^{+\infty} x_2(t-\tau) e^{-j\omega(t-\tau)} dt \right] e^{-j\omega \tau} d\tau \\ &= X_2(j\omega) \int_{-\infty}^{+\infty} x_1(\tau) e^{-j\omega \tau} d\tau = X_1(j\omega) X_2(j\omega) \end{split}$$

Example 4.8 Calculate the following integral $(a = a^* > 0)$.

$$\int_{-\infty}^{\infty} \frac{\sin^2 a\omega}{\omega^2} d\omega$$

Solution:

$$u(t+a) - u(t-a) \stackrel{F}{\longleftrightarrow} \frac{2sina\omega}{\omega}$$

Parseval's Relation yields

$$\int_{-\infty}^{\infty} |u(t+a) - u(t-a)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} \left| \frac{2\sin a\omega}{\omega} \right|^2 d\omega$$
$$\int_{-\infty}^{\infty} \frac{\sin^2 a\omega}{\omega^2} d\omega = \frac{\pi}{2} \int_{-\infty}^{\infty} |u(t+a) - u(t-a)|^2 dt$$
$$= \frac{\pi}{2} \int_{-a}^{a} 1 dt = a\pi$$

$$y(t) = h(t) * x(t)$$

$$x(t) \longrightarrow h(t) \longrightarrow y(t)$$

$$Y(j\omega) = H(j\omega)X(j\omega)$$

$$X(j\omega) \longrightarrow H(j\omega) \longrightarrow Y(j\omega)$$

 $H(j\omega)$ is known as the frequency response of an LTI system

$$x(t) = \lim_{\omega_0 \to 0} \frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(jk\omega_0) e^{jk\omega_0 t} \omega_0$$

$$y(t) = \lim_{\omega_0 \to 0} \frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} H(jk\omega_0) X(jk\omega_0) e^{jk\omega_0 t} \omega_0$$

(1) Cascade interconnection

$$H_2(j\omega)[H_1(j\omega)X(j\omega)] = [H_1(j\omega)H_2(j\omega)]X(j\omega)$$

$$X(j\omega) \longrightarrow H_1(j\omega) \longrightarrow H_2(j\omega) \longrightarrow Y(j\omega)$$

$$X(j\omega) \longrightarrow H_1(j\omega)H_2(j\omega) \longrightarrow Y(j\omega)$$

(2) Paralleled interconnection

$$H_1(j\omega)X(j\omega) + H_2(j\omega)X(j\omega) = [H_1(j\omega) + H_2(j\omega)]X(j\omega)$$

$$X(j\omega)$$
 $H_1(j\omega)$ $Y(j\omega)$

$$X(j\omega) \longrightarrow H_1(j\omega) + H_2(j\omega) \longrightarrow Y(j\omega)$$

Example 4.10 Determine the response of an ideal low-pass filter to an input signal $x(t) = \frac{\sin \omega_i t}{\pi t}$, when the unit impulse response of the ideal low-pass filter is given by $h(t) = \frac{\sin \omega_c t}{\pi t}$ Solution:

$$X(j\omega) = \mathcal{F}\left[\frac{\sin\omega_{i}t}{\pi t}\right] = \begin{cases} 1 & (|\omega| < \omega_{i}) \\ 0 & (|\omega| > \omega_{i}) \end{cases}$$

$$H(j\omega) = \mathcal{F}\left[\frac{\sin\omega_{c}t}{\pi t}\right] = \begin{cases} 1 & (|\omega| < \omega_{c}) \\ 0 & (|\omega| > \omega_{c}) \end{cases}$$

$$Y(j\omega) = H(j\omega)X(j\omega) = \begin{cases} 1 & (|\omega| < \min\{\omega_{i}, \omega_{c}\}) \\ 0 & (|\omega| > \min\{\omega_{i}, \omega_{c}\}) \end{cases}$$

$$y(t) = \begin{cases} h(t) & (\omega_{c} < \omega_{i}) \\ x(t) & (\omega_{c} > \omega_{i}) \end{cases}$$

Example 4.9 Determine the frequency responses of the differentiator and integrator.

Solution:

$$x(t) \longrightarrow u_1(t) \longrightarrow \frac{d}{dt}x(t)$$

Differentiator
$$h_1(t) = \frac{d}{dt} \delta(t) = u_1(t)$$

$$H_1(j\omega) = \frac{Y(j\omega)}{X(j\omega)} = \frac{j\omega X(j\omega)}{X(j\omega)} = j\omega$$

$$x(t) \longrightarrow u_{-1}(t) \longrightarrow \int_{-\infty}^{t} u(\tau) d\tau$$

Integrator
$$h_2(t) = u(t) = u_{-1}(t)$$

$$H_2(j\omega) = \frac{1}{j\omega} + \pi\delta(\omega)$$

Example 4.11 Determine the response of an LTI real system to a sinusoidal input signal $x(t) = \cos(\omega_0 t + \theta)$.

$$X(j\omega) \longrightarrow H(j\omega) = |H(j\omega)|e^{j4H(j\omega)} \longrightarrow Y(j\omega)$$

Solution:

$$x(t) = \cos(\omega_0 t + \theta) = \frac{1}{2} e^{j\theta} e^{j\omega_0 t} + \frac{1}{2} e^{-j\theta} e^{-j\omega_0 t}$$

$$X(j\omega) = \pi e^{j\theta} \delta(\omega - \omega_0) + \pi e^{-j\theta} \delta(\omega + \omega_0)$$

$$Y(j\omega) = \pi e^{j\theta} H(j\omega) \delta(\omega - \omega_0) + \pi e^{-j\theta} H(j\omega) \delta(\omega + \omega_0)$$

$$y(t) = \frac{1}{2} e^{j\theta} H(j\omega_0) e^{j\omega_0 t} + \frac{1}{2} e^{-j\theta} H(-j\omega_0) e^{-j\omega_0 t}$$

$$- Re\{|H(j\omega_0)|e^{j(\omega_0 t + \theta + 4H(j\omega_0))}\}$$

$$=Re\{|H(j\omega_0)|e^{j[\omega_0t+\theta+\not\sim H(j\omega_0)]}\}$$

$$= |H(j\omega_0)|\cos[\omega_0 t + \theta + \not \Delta H(j\omega_0)]$$