Digital Signal Processing

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1	Software Installation	1
2	Digital Filter	1
3	Difference Equation	2
4	Z-Transform	2
5	Impulse Response	4

1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

wget https://github.com/HARI-donk-EY/ sig pros/blob/main/codes/2/sound files/ Sound Noise.wav

- 2.2 You will find a spectrogram at https://academo. org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play.
 - Observe the spectrogram. What do you find? Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.
- 2.3 Write the python code for removal of out of band noise and execute the code.

Solution:

```
import soundfile as sf
from scipy import signal
#read .wav file
input signal, fs = sf.read('sound files/
   Sound Noise.wav')
#sampling frequency of Input signal
sampl freq=fs
#order of the filter
order=4
#cutoff frquency 4kHz
cutoff freq=4000.0
#digital frequency
Wn=2*cutoff freq/sampl freq
# b and a are numerator and
   denominator polynomials respectively
b, a = signal.butter(order, Wn, 'low')
#filter the input signal with butterworth filter
output signal = signal.filtfilt(b, a,
   input signal)
#output signal = signal.lfilter(b, a, input
   signal)
#write the output signal into .wav file
sf.write('sound files/
    Sound With ReducedNoise.wav',
```

output signal, fs)

2.4 The output of the python script in Problem 2.3 is the audio file Sound With ReducedNoise.wav.

Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 Difference Equation

3.1 Let

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\} \tag{3.1}$$

Sketch x(n).

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The following code yields Fig. 3.3.

wget https://github.com/HARI-donk-EY/sig pros/blob/main/codes/3/xnyn.py

3.3 Repeat the above exercise using C code.

Solution: The following codes can be used, the C-code used to generate y(n) and the plot yields the same Figure as that of 3.3

wget https://github.com/HARI-donk-EY/sig_pros/blob/main/codes/3/xnyn.c

wget https://github.com/HARI-donk-EY/sig pros/blob/main/codes/3/xnyn2.py

4 Z-Transform

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

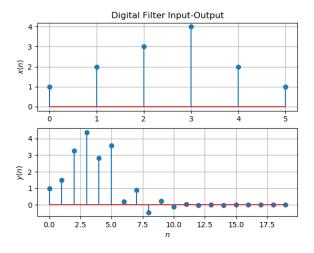


Fig. 3.3

Solution: From (4.1),

$$Z\{x(n-1)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)
$$(4.5)$$

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \tag{4.6}$$

4.2 Obtain X(z) for x(n) defined from 3.1.

Solution: From 3.1,

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\}$$

and from (4.1), and for given x(n) we have,

$$X(n) = \mathcal{Z}\{x(n)\}$$

$$= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + z^{-5}$$
(4.8)

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)} \tag{4.9}$$

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.10)

$$\implies \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \tag{4.11}$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.12)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.13)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.14}$$

Solution: It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.15}$$

and from (4.13),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.16)

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.17}$$

using the fomula for the sum of an infinite geometric progression.

4.5 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a|$$
 (4.18)

Solution: From (4.13), we get,

$$a^{n}u(n) = \begin{cases} a^{n} & n \ge 0\\ 0 & \text{otherwise} \end{cases}$$
 (4.19)

from above,

$$U_a(z) = \sum_{n=0}^{\infty} a^n z^{-n}$$
 (4.20)

$$= \sum_{n=0}^{\infty} (az^{-1})^n = \frac{1}{1 - az^{-1}}, \quad |z| > 1$$
(4.21)

using the fomula for the sum of an infinite

geometric progression.

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.22)

Plot $|H(e^{j\omega})|$. Comment. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of x(n).

Solution: The following code plots Fig. 4.6.

wget https://github.com/HARI-donk-EY/sig_pros/blob/main/codes/4/dtft.py

We can note that the graph in Fig. 4.6 is an even periodic function.

From the equations and we get,

$$|H(e^{j\omega})| = \left| \frac{1 + e^{-2j\omega}}{1 + \frac{1}{2}e^{-j\omega}} \right|$$
(4.23)
$$= \left| \frac{1 + \cos(-2\omega) + j\sin(-2\omega)}{1 + \frac{1}{2}(\cos(-\omega) + j\sin(-\omega))} \right|$$
(4.24)
$$= \sqrt{\frac{(1 + \cos 2\omega)^2 + (\sin 2\omega)^2}{\left(1 + \frac{1}{2}\cos\omega\right)^2 + \left(\frac{1}{2}\sin\omega\right)^2}}$$
(4.25)

$$=\sqrt{\frac{2(1+\cos 2\omega)}{\frac{5}{4}+\cos \omega}}\tag{4.26}$$

$$=\sqrt{\frac{2(2\cos^2\omega)}{\frac{5}{4}+\cos\omega}}\tag{4.27}$$

$$=\frac{4|\cos\omega|}{\sqrt{5+4\cos\omega}}\tag{4.28}$$

and so its fundamental period is 2π .

4.7 Express h(n) in terms of $H(e^{jw})$.

Solution: h(n) is given by the inverse DTFT (IDTFT) of $H(e^{j\omega})$

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) e^{j\omega n} d\omega \qquad (4.29)$$

We can prove this from (4.1),

$$H(e^{jw}) = \sum_{k=-\infty}^{\infty} h(k)e^{-j\omega k}$$
 (4.30)

(4.31)

multiplying both sides with $e^{-j\omega k}$ and

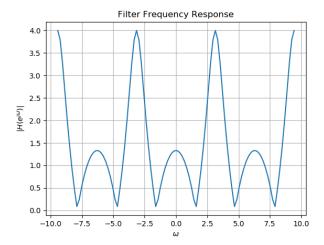


Fig. 4.6: $|H(e^{j\omega})|$

integrating w.r.t. ω .

$$\int_{-\pi}^{\pi} H(e^{jw})e^{j\omega k} d\omega = \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} e^{-j\omega n} e^{j\omega k} d\omega$$

$$(4.32)$$

$$= \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} e^{-j\omega(n-k)} d\omega$$

$$(4.33)$$

we know that $e^{-j\omega(n-k)}$ is an odd function so integration would be zero for all values of $\omega(n-k) \neq 0$, but 1 if $\omega(n-k) = 0$.

$$\int_{-\pi}^{\pi} e^{-j\omega(n-k)} d\omega = \begin{cases} 0 & if n-k \neq 0 \\ 1 & if n-k = 0 \end{cases}$$

From this we can complete the summation.

$$\int_{-\pi}^{\pi} H(e^{jw})e^{j\omega k}d\omega = \sum_{k=-\infty}^{\infty} h(k) \int_{-\pi}^{\pi} e^{-j\omega(n-k)}d\omega$$

$$= h(n) \int_{-\pi}^{\pi} d\omega \qquad (4.34)$$

$$= h(n)2\pi \qquad (4.36)$$

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{jw})e^{j\omega k}d\omega \qquad (4.37)$$

Hence proved.

5 IMPULSE RESPONSE

5.1 Using long division, find

$$h(n), \quad n < 5$$
 (5.1)

for H(z) in (4.6).

Solution:

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.2)

Substitute $z^{-1} = x$

So,

$$H(z) = -4 + 2z^{-1} + \frac{5}{1 + \frac{1}{2}z^{-1}}$$

$$= -4 + 2z^{-1} + 5\sum_{n=0}^{\infty} \left(\frac{-1}{2}\right)^n z^{-n}$$

$$= 1 - \frac{1}{2}z^{-1} + 5\sum_{n=2}^{\infty} \left(\frac{-1}{2}\right)^n z^{-n}$$

$$= \sum_{n=0}^{\infty} \left(\frac{-1}{2}\right)^n z^{-n} + 4\sum_{n=2}^{\infty} \left(\frac{-1}{2}\right)^n z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} u(n) \left(\frac{-1}{2}\right)^n z^{-n} + \sum_{n=-\infty}^{\infty} u(n-2) \left(\frac{-1}{2}\right)^{n-2} z^{-n}$$

$$(5.7)$$

Therefore from (4.1),

$$h(n) = \left(\frac{-1}{2}\right)^n u(n) + \left(\frac{-1}{2}\right)^{n-2} u(n-2)$$
 (5.8)

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z) \tag{5.9}$$

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse response* of the system defined by (3.2).

Take H(z) from (4.6)

Solution: From (4.6),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.10)

$$\implies h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$

using (4.18) and (4.6).

5.3 Sketch h(n). Is it bounded? Justify theoretically.

Solution: The following code plots Fig. 5.3.

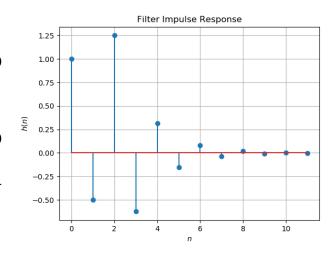


Fig. 5.3: h(n) as the inverse of H(z)

wget https://github.com/HARI-donk-EY/ sig pros/blob/main/codes/5/hn.py

We know,

$$h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$

then,

$$\lim_{n \to \infty} h(n) = \lim_{n \to \infty} \left(-\frac{1}{2} \right)^n u(n) + \lim_{n \to \infty} \left(-\frac{1}{2} \right)^{n-2} u(n-2)$$

(5.12)

$$h(n) = \begin{cases} 5\left(\frac{-1}{2}\right)^n & n \ge 2\\ \left(\frac{-1}{2}\right)^n & 0 \le n < 2\\ 0 & n < 0 \end{cases}$$
 (5.13)

$$L = \lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| \tag{5.15}$$

$$= \lim_{n \to \infty} \left| \frac{\left(-\frac{1}{2}\right)^{n+1} u(n+1) + \left(-\frac{1}{2}\right)^{n-1} u(n-1)}{\left(-\frac{1}{2}\right)^{n} u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)} \right|$$
(5.16)

$$= \left| \frac{-\frac{1}{2} + \left(-\frac{1}{2}\right)^{-1}}{1 + \left(-\frac{1}{2}\right)^{-2}} \right| \tag{5.17}$$

$$= \left| \frac{-\frac{5}{2}}{5} \right| \tag{5.18}$$

$$=\frac{1}{2}$$
 (5.19)

Since L < 1, we can say that h(n) is convergent.

5.4 Is h(n) convergent? Justify using ratio test.

Solution:

$$h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (5.14)$$

If $L = \lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| < 1$, then h(n) is convergent.

5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.20}$$

Is the system defined by (3.2) stable for the impulse response in (5.9)?

Solution:

$$\sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(\left(-\frac{1}{2} \right)^n u(n) + \left(-\frac{1}{2} \right)^{n-2} u(n-2) \right)$$

$$= \sum_{n=0}^{\infty} \left(-\frac{1}{2} \right)^n u(n) + \sum_{(n-2)=0}^{\infty} \left(-\frac{1}{2} \right)^{n-2} u(n-2)$$
(5.22)

$$=2\left(\frac{1}{1+\frac{1}{2}}\right) \tag{5.23}$$

$$=\frac{4}{3}$$
 (5.24)

As $\sum_{n=-\infty}^{\infty} h(n) = \frac{4}{3}$ is less than ∞ , the system defined by (3.2) is stable for the impulse response in (5.9).

5.6 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.25)$$

This is the definition of h(n).

Solution: The following code plots Fig. 5.6. Note that this is the same as Fig. 5.3.

wget https://github.com/HARI-donk-EY/ sig_pros/blob/main/codes/5/hndef.py

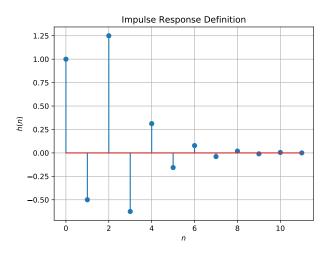


Fig. 5.6: h(n) from the definition

5.7 Compute

$$y(n) = x(n) * h(n) = \sum_{n = -\infty}^{\infty} x(k)h(n - k) \quad (5.26)$$

Comment. The operation in (5.26) is known as

convolution.

Solution: The following code plots Fig. 5.7. Note that this is the same as y(n) in Fig. 3.3.

wget https://github.com/HARI-donk-EY/sig_pros/blob/main/codes/5/ynconv.py

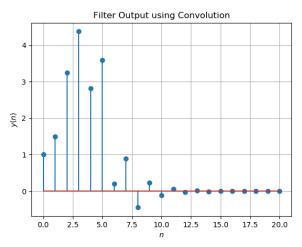


Fig. 5.7: y(n) from the definition of convolution

5.8 Show that

$$y(n) = \sum_{n = -\infty}^{\infty} x(n - k)h(k)$$
 (5.27)

Solution: From (5.26), we substitute k := n - k to get

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$
 (5.28)

$$=\sum_{n-k=-\infty}^{\infty}x\left(n-k\right)h\left(k\right)\tag{5.29}$$

$$=\sum_{k=-\infty}^{\infty}x\left(n-k\right)h\left(k\right)\tag{5.30}$$